SLAS225C-JUNE 1999-REVISED JANUARY 2004







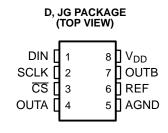
# 2.7-V TO 5.5-V LOW-POWER DUAL 12-BIT DIGITAL-TO-ANALOG CONVERTER WITH INTERNAL REFERENCE AND POWER DOWN

#### **FEATURES**

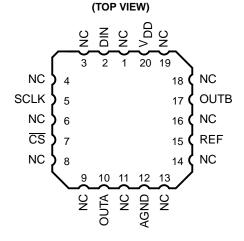
- **Dual 12-Bit Voltage Output DAC**
- **Programmable Internal Reference**
- **Programmable Settling Time:** 
  - 1 µs in Fast Mode,
  - 3.5 µs in Slow Mode
- Compatible With TMS320 and SPI™ Serial **Ports**
- Differential Nonlinearity < 0.5 LSB Typ
- **Monotonic Over Temperature**

#### **APPLICATIONS**

- **Digital Servo Control Loops**
- **Digital Offset and Gain Adjustment**
- **Industrial Process Control**
- **Machine and Motion Control Devices**
- **Mass Storage Devices**



**FK PACKAGE** 



#### **DESCRIPTION**

The TLV5638 is a dual 12-bit voltage output DAC with a flexible 3-wire serial interface. The serial interface allows glueless interface to TMS320, SPI™, QSPI™, and Microwire™ serial ports. It is programmed with a 16-bit serial string containing 4 control and 12 data bits.

The resistor string output voltage is buffered by a x2 gain rail-to-rail output buffer. The buffer features a Class AB output stage to improve stability and reduce settling time. The programmable settling time of the DAC allows the designer to optimize speed vs power dissipation. With its on-chip programmable precision voltage reference, the TLV5638 simplifies overall system design.

Because of its ability to source up to 1 mA, the reference can also be used as a system reference. Implemented with a CMOS process, the device is designed for single supply operation from 2.7 V to 5.5 V. It is available in an 8-pin SOIC package to reduce board space in standard commercial, industrial, and automotive temperature ranges. It is also available in JG and FK packages in the military temperature range.



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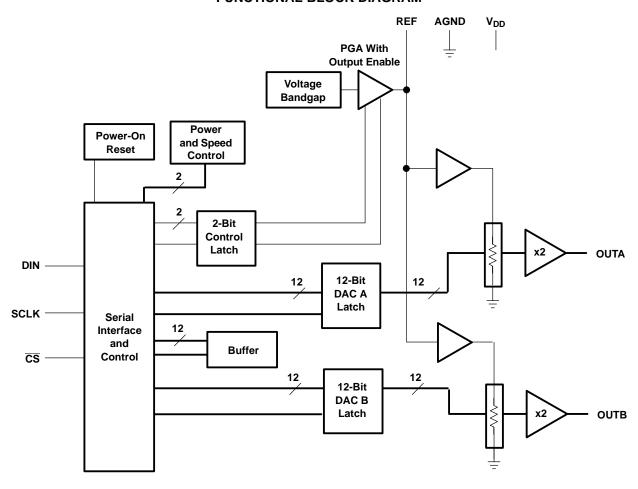
This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

#### **AVAILABLE OPTIONS**

		PACKAGE						
T <sub>A</sub>	SOIC (D)	CERAMIC DIP (JG)	20 PAD LCCC (FK)					
0°C to 70°C	TLV5638CD	_	_					
40°C to 85°C	TLV5638ID	_	_					
40°C to 125°C	TLV5638QD TLV5638QDR	_	_					
55°C to 125°C	_	TLV5638MJG	TLV5638MFK					

#### **FUNCTIONAL BLOCK DIAGRAM**





#### **Terminal Functions**

TERM	IINAL	L/O/D	DESCRIPTION
NAME	NO.	I/O/P	DESCRIPTION
AGND	5	Р	Ground
<del>CS</del>	3	- 1	Chip select. Digital input active low, used to enable/disable inputs
DIN	1	I	Digital serial data input
OUT A	4	0	DAC A analog voltage output
OUT B	7	0	DAC B analog voltage output
REF	6	I/O	Analog reference voltage input/output
SCLK	2	- 1	Digital serial clock input
$V_{DD}$	8	Р	Positive power supply

#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted)(1)

		UNIT		
Supply voltage (V <sub>DD</sub> to AGND)		7 V		
Reference input voltage range		-0.3 V to V <sub>DD</sub> + 0.3 V		
Digital input voltage range		-0.3 V to V <sub>DD</sub> + 0.3 V		
Operating free-air temperature range, T <sub>A</sub>	TLV5638C	0°C to 70°C		
	TLV5638I	-40°C to 85°C		
	TLV5638Q	-40°C to 125°C		
	TLV5638M	-55°C to 125°C		
Storage temperature range, T <sub>stg</sub>	-65°C to 150°C			
Lead temperature 1,6 mm (1/16 inch) from c	ase for 10 seconds	260°C		

<sup>(1)</sup> Stresses beyond those listed under, absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under, recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

#### **DISSIPATION RATING TABLE**

PACKAGE	T <sub>A</sub> ≤ 25°C POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C <sup>(1)</sup>	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 85°C POWER RATING	T <sub>A</sub> = 125°C POWER RATING
D	635 mW	5.08 mW/°C	407 mW	330 mW	127 mW
FK	1375 mW	11.00 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.40 mW/°C	672 mW	546 mW	210 mW

<sup>(1)</sup> This is the inverse of the traditional Junction-to-Ambient thermal Resistance (Rθ<sub>JA</sub>). Thermal Resistances are not production tested and are for informational purposes only.



#### RECOMMENDED OPERATING CONDITIONS

			MIN	NOM	MAX	UNIT
Cumply valtage V	V <sub>DD</sub> = 5 V		4.5	5	5.5	V
Supply voltage, V <sub>DD</sub>	$V_{DD} = 3 V$		2.7	3	3.3	V
Power on reset, POR			0.55 <sup>(1)</sup>		2 <sup>(1)</sup>	V
High level digital input valtage V	$V_{DD} = 2.7 \text{ V}$		2			V
High-level digital input voltage, V <sub>IH</sub>	$V_{DD} = 5.5 \text{ V}$		2.4			
	V <sub>DD</sub> = 2.7 V		4.5 5 5  2.7 3 3  0.55(1) 2  2 2  2.4  M 00  AGND 2.048 V <sub>DD</sub> -1  AGND 1.024 V <sub>DD</sub> -1  2 10  3 0  40 8  40 12	0.6	V	
Low-level digital input voltage, $V_{\rm IL}$	V	TLV5638C and TLV5638I			1	V
	$V_{DD} = 5.5 \text{ V}$	TLV5638Q and TLV5638M			0.8	٧
Reference voltage, V <sub>ref</sub> to REF terminal	V <sub>DD</sub> = 5 V <sup>(2)</sup>		AGND	2.048	V <sub>DD</sub> -1.5	V
Reference voltage, V <sub>ref</sub> to REF terminal	$V_{DD} = 3 V^{(2)}$		AGND	1.024	V <sub>DD</sub> -1.5	V
Load resistance, R <sub>L</sub>			2			kΩ
Load capacitance, C <sub>L</sub>					100	pF
Clock frequency, f <sub>CLK</sub>					20	MHz
	TLV5638C		0		70	
Operating tree air temperature T	TLV5638I		40		85	°C
Operating free-air temperature, T <sub>A</sub>	TLV5638Q		40		125	٠.
	TLV5638M		55		125	

<sup>(1)</sup> This parameter is not tested for Q and M suffix devices.

## **ELECTRICAL CHARACTERISTICS**

over recommended operating conditions,  $V_{ref} = 2.048 \text{ V}$ ,  $V_{ref} = 1.024 \text{ V}$  (unless otherwise noted)

POWE	R SUPPLY							
	PARAMETER	TEST CO	NDITIONS		TL TI	UNIT		
					MIN	TYP	MAX	
			V <sub>DD</sub> = 5 V,	Fast		4.3	7	mA
			Int. ref.	Slow		2.2	3.6	
	Power supply current	No load, All inputs = AGND or $V_{DD}$ , DAC latch = 0x800	V <sub>DD</sub> = 3 V, Int. ref.	Fast		3.8	6.3	mA
				Slow		1.8	3.0	
I <sub>DD</sub>			V <sub>DD</sub> = 5 V, Ext. ref.	Fast		3.9	6.3	3.0 mA
				Slow		1.8	3.0	
			$V_{DD} = 3 V$	Fast		3.5	5.7	mA
			Ext. ref.	Slow		1.5	2.6	mA
	Power-down supply current					0.01	10	μΑ
PSRR	Dower cumply rejection ratio	Zero scale, (1)		65		٩D		
PORK	Power supply rejection ratio	Full scale, (2)				65		dB

<sup>(1)</sup> Power supply rejection ratio at zero scale is measured by varying V<sub>DD</sub> and is given by: PSRR = 20 log [(E<sub>ZS</sub>(V<sub>DD</sub>max) - E<sub>ZS</sub>(V<sub>DD</sub>min))/V<sub>DD</sub>max]

<sup>(2)</sup> Due to the x2 output buffer, a reference input voltage ≥ (V<sub>DD</sub>-0.4 V)/2 causes clipping of the transfer function. The output buffer of the internal reference must be disabled, if an external reference is used.

<sup>(2)</sup> Power supply rejection ratio at full scale is measured by varying V<sub>DD</sub> and is given by: PSRR = 20 log [(E<sub>G</sub>(V<sub>DD</sub>max) - E<sub>G</sub>(V<sub>DD</sub>min))/V<sub>DD</sub>max]



## **ELECTRICAL CHARACTERISTICS (Continued)**

over recommended operating conditions,  $V_{ref} = 2.048 \text{ V}$ ,  $V_{ref} = 1.024 \text{ V}$  (unless otherwise noted)

STATIC DAC	SPECIFICATIONS						
	PARAMETER	TEST CONDITION	S	MIN	TYP	MAX	UNIT
	Resolution			12			bits
INL	Integral nonlinearity, end point ad-	See (1)	C and I suffixes		±1.7	±4	LSB
INL	justed	See W	Q and M suffixes		±1.7	±6	LSB
DNL	Differential nonlinearity	See (2)			±0.4	±1	LSB
E <sub>ZS</sub>	Zero-scale error (offset error at zero scale)	See <sup>(3)</sup>				±24	mV
E <sub>ZS</sub> TC	Zero-scale-error temperature coefficient	See (4)			10		ppm/°C
E <sub>G</sub>	Gain error	See <sup>(5)</sup>				±0.6	% full scale V
$E_GT_C$	Gain error temperature coefficient	See (6)			10		ppm/°C
OUTPUT SPE	ECIFICATIONS						
V <sub>O</sub>	Output voltage	$R_L = 10 \text{ k}\Omega$		0		$V_{DD}$ -0.4	V
	Output load regulation accuracy	$V_{O} = 4.096 \text{ V}, 2.048 \text{ V}, R_{L} = 2 \text{ ks}$	Ω			±0.25	% full scale V
REFERENCE	PIN CONFIGURED AS OUTPUT (RE	F)					
Vref(OUTL)	Low reference voltage			1.003	1.024	1.045	V
Vref(OUTH)	High reference voltage	V <sub>DD</sub> > 4.75 V		2.027	2.048	2.069	V
Iref(source)	Output source current					1	mA
Iref(sink)	Output sink current			-1			mA
	Load capacitance					100	pF
PSRR	Power supply rejection ratio				-65		dB
REFERENCE	PIN CONFIGURED AS INPUT (REF)						
VI	Input voltage			0		V <sub>DD</sub> -1.5	V
R <sub>I</sub>	Input resistance				10		$M\Omega$
C <sub>I</sub>	Input capacitance				5		pF
	Defense as investigate	DEE 0.0 V . 4.004 V de	Fast	1.3		MHz	
	Reference input bandwidth	REF = $0.2 V_{pp} + 1.024 V dc$	Slow		525		kHz
	Reference feedthrough	REF = 1 V <sub>pp</sub> at 1.024 V dc <sup>(7)</sup>		-80		dB	
DIGITAL INP	UTS			*			
I <sub>IH</sub>	HIgh-level digital input current	$V_I = V_{DD}$				1	μA
I <sub>IL</sub>	Low-level digital input current	V <sub>I</sub> = 0 V		-1			μA
C <sub>i</sub>	Input capacitance				8		pF

<sup>(1)</sup> The relative accuracy or integral nonlinearity (INL) sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors. Tested from code 32 to 4095.

<sup>(2)</sup> The differential nonlinearity (DNL) sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

Zero-scale error is the deviation from zero voltage output when the digital input code is zero.

Zero-scale-error temperature coefficient is given by:  $E_{ZS}$  TC = [ $E_{ZS}$  ( $T_{max}$ ) -  $E_{ZS}$  ( $T_{min}$ )]/ $V_{ref}$ × 10<sup>6</sup>/( $T_{max}$  -  $T_{min}$ ). Gain error is the deviation from the ideal output (2 $V_{ref}$  - 1 LSB) with an output load of 10 k $\Omega$  excluding the effects of the zero-error. Gain temperature coefficient is given by:  $E_{G}$  TC = [ $E_{G}$ ( $T_{max}$ ) -  $E_{G}$  ( $T_{min}$ )]/ $V_{ref}$ × 10<sup>6</sup>/( $T_{max}$  -  $T_{min}$ ). Reference feedthrough is measured at the DAC output with an input code = 0x000.



## **ELECTRICAL CHARACTERISTICS (Continued)**

over recommended operating conditions,  $V_{ref} = 2.048 \text{ V}$ ,  $V_{ref} = 1.024 \text{ V}$  (unless otherwise noted)

ANALOG	OUTPUT DYNAMIC PERFORMANCE							
	PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
+	Output pattling time full apple	$R_1 = 10 \text{ k}\Omega$ , $C_1 = 100 \text{ pF}$ , $See^{(1)}$	Fast		1	3		
t <sub>s(FS)</sub>	Output settling time, full scale	R <sub>L</sub> = 10 kΩ <sub>2</sub> , G <sub>L</sub> = 100 pr, See ···	Slow		3.5	7	μs	
	Output settling time, code to code	$R_1 = 10 \text{ k}\Omega$ , $C_1 = 100 \text{ pF}$ , See <sup>(2)</sup>	Fast		0.5	1.5		
t <sub>s(CC)</sub>	Output settling time, code to code	R <sub>L</sub> = 10 kΩ <sub>2</sub> , G <sub>L</sub> = 100 pr, See (-)	Slow		1	2	μs	
SR	Slew rate	$R_1 = 10 \text{ k}\Omega$ , $C_1 = 100 \text{ pF}$ , See <sup>(3)</sup>	Fast		12		1//	
SK	Siew rate	$R_L = 10 \text{ kgz}$ , $G_L = 100 \text{ pr}$ , See	Slow		1.8		√ V/μs	
	Glitch energy	DIN = 0 to 1, FCLK = 100 kHz, $\overline{CS}$ = $V_{DD}$			5		nV-s	
SNR	Signal-to-noise ratio			69	74			
S/(N+D)	Signal-to-noise + distortion	$f_s = 480 \text{ kSPS}, f_{out} = 1 \text{ kHz}, R_L = 10 \text{ k}\Omega,$		58	67		dB	
THD	Total harmonic distortion	Č <sub>L</sub> = 100 pF			69	57	ub	
	Spurious free dynamic range			57	72			

<sup>(1)</sup> Settling time is the time for the output signal to remain within ±0.5 LSB of the final measured value for a digital input code change of 0x020 to 0xFDF and 0xFDF to 0x020 respectively. Not tested, assured by design.

#### **DIGITAL INPUT TIMING REQUIREMENTS**

		MIN	NOM	MAX	UNIT
t <sub>su(CS-CK)</sub>	Setup time, $\overline{\text{CS}}$ low before first negative SCLK edge	10			ns
t <sub>su(C16-CS)</sub>	Setup time, 16 <sup>th</sup> negative SCLK edge (when D0 is sampled) before $\overline{\text{CS}}$ rising edge	10			ns
$t_{wH}$	SCLK pulse width high	25			ns
t <sub>wL</sub>	SCLK pulse width low	25			ns
t <sub>su(D)</sub>	Setup time, data ready before SCLK falling edge	10			ns
t <sub>h(D)</sub>	Hold time, data held valid after SCLK falling edge	5			ns

#### PARAMETER MEASURMENT INFORMATION

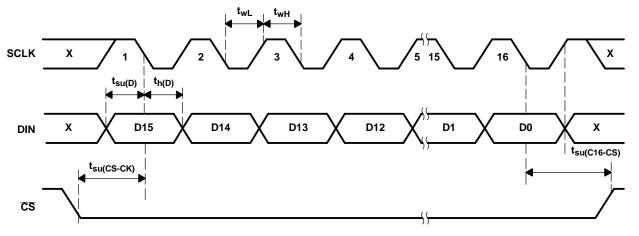


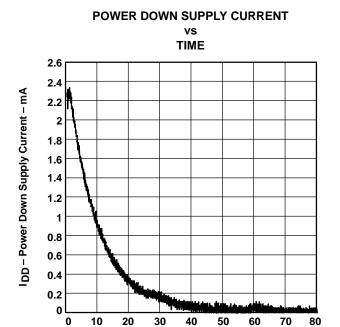
Figure 1. Timing Diagram

<sup>(2)</sup> Settling time is the time for the output signal to remain within ± 0.5 LSB of the final measured value for a digital input code change of one count. Not tested, assured by design.

<sup>(3)</sup> Slew rate determines the time it takes for a change of the DAC output from 10% to 90% full-scale voltage.



#### TYPICAL CHARACTERISTICS



# t – Time – $\mu$ s Figure 2.

# SUPPLY CURRENT vs FREE-AIR TEMPERATURE

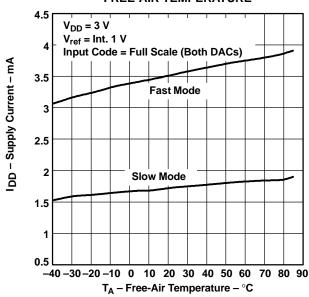


Figure 4.

# SUPPLY CURRENT

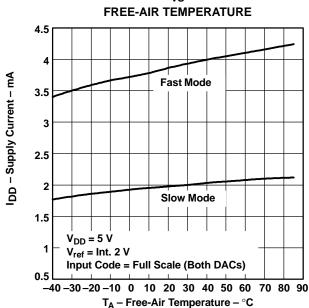


Figure 3.

# OUTPUT VOLTAGE vs

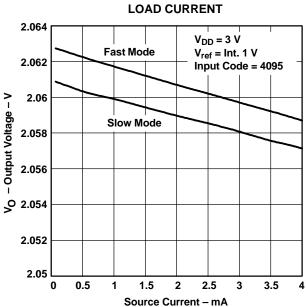
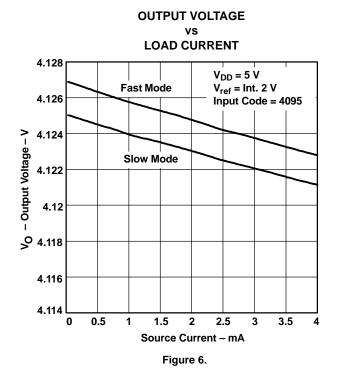


Figure 5.



## **TYPICAL CHARACTERISTICS (continued)**



#### OUTPUT VOLTAGE vs LOAD CURRENT

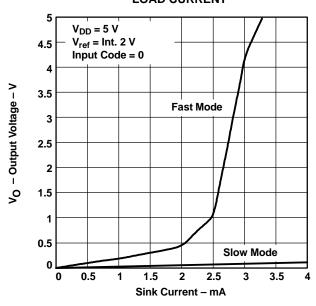


Figure 8.

# OUTPUT VOLTAGE vs

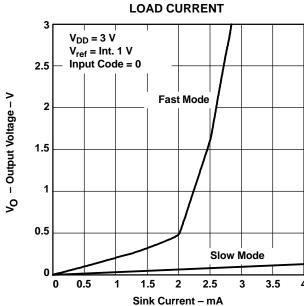


Figure 7.

# TOTAL HARMONIC DISTORTION AND NOISE vs FREQUENCY

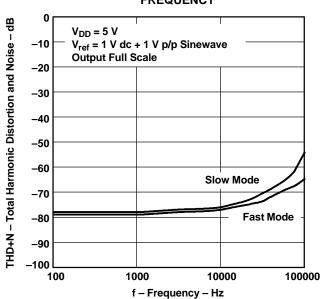
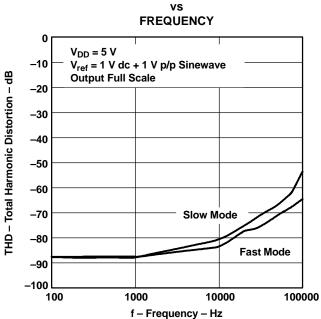


Figure 9.

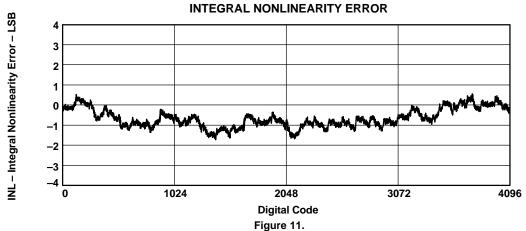


## **TYPICAL CHARACTERISTICS (continued)**

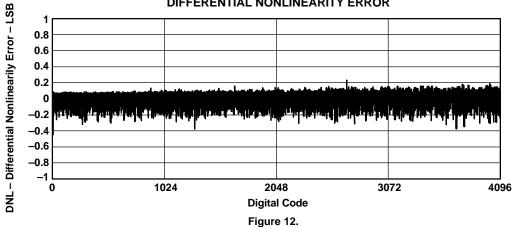
# TOTAL HARMONIC DISTORTION













#### **APPLICATION INFORMATION**

#### **GENERAL FUNCTION**

The TLV5638 is a dual 12-bit, single supply DAC, based on a resistor string architecture. It consists of a serial interface, a speed and power-down control logic, a programmable internal reference, a resistor string, and a rail-to-rail output buffer.

The output voltage (full scale determined by reference) is given by:

$$2 \text{ REF } \frac{\text{CODE}}{0 \text{x} 1000} \text{ [V]}$$

Where REF is the reference voltage and CODE is the digital input value in the range 0x000 to 0xFFF. A power on reset initially puts the internal latches to a defined state (all bits zero).

#### **SERIAL INTERFACE**

A falling edge of  $\overline{CS}$  starts shifting the data bit-per-bit (starting with the MSB) to the internal register on the falling edges of SCLK. After 16 bits have been transferred or  $\overline{CS}$  rises, the content of the shift register is moved to the target latches (DAC A, DAC B, BUFFER, CONTROL), depending on the control bits within the data word.

Figure 13 shows examples of how to connect the TLV5638 to TMS320, SPI™, and Microwire™.

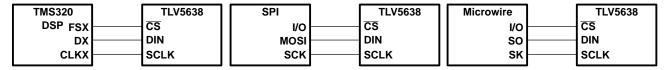


Figure 13. Three-Wire Interface

Notes on SPI<sup>™</sup> and Microwire<sup>™</sup>: Before the controller starts the data transfer, the software has to generate a falling edge on the pin connected to  $\overline{CS}$ . If the word width is 8 bits (SPI<sup>™</sup> and Microwire<sup>™</sup>), two write operations must be performed to program the TLV5638. After the write operation(s), the holding registers or the control register are updated automatically on the 16<sup>th</sup> positive clock edge.

#### SERIAL CLOCK FREQUENCY AND UPDATE RATE

The maximum serial clock frequency is given by:

$$f_{\text{sclkmax}} = \frac{1}{t_{\text{whmin}} + t_{\text{wlmin}}} = 20 \text{ MHz}$$

The maximum update rate is:

$$f_{updatemax} = \frac{1}{16(t_{whmin} + t_{wlmin})} = 1.25 \text{ MHz}$$

Note, that the maximum update rate is just a theoretical value for the serial interface, as the settling time of the TLV5638 has to be considered, too.



## **APPLICATION INFORMATION (continued)**

#### **DATA FORMAT**

The 16-bit data word for the TLV5638 consists of two parts:

- Program bits (D15..D12)
- New data (D11..D0)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
R1	SPD	PWR	R0						12 Da	ta bits					

SPD: Speed control bit  $1 \rightarrow \text{fast mode}$   $0 \rightarrow \text{slow mode}$ 

PWR: Power control bit  $1 \rightarrow$  power down  $0 \rightarrow$  normal operation

The following table lists the possible combination of the register select bits:

#### **REGISTERED SELECT BITS**

R1	R0	REGISTER
0	0	Write data to DAC B and BUFFER
0	1	Write data to BUFFER
1	0	Write data to DAC A and update DAC B with BUFFER content
1	1	Write data to control register

The meaning of the 12 data bits depends on the register. If one of the DAC registers or the BUFFER is selected, then the 12 data bits determine the new DAC value:

#### DATA BITS: DAC A, DAC B and BUFFER

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
	New DAC Value										

If control is selected, then D1, D0 of the 12 data bits are used to program the reference voltage:

#### **DATA BITS: CONTROL**

D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Χ	Χ	Χ	Х	Х	Х	Χ	Х	Х	REF1	REF0

REF1 and REF0 determine the reference source and, if internal reference is selected, the reference voltage.

#### **REFERENCE BITS**

REF1	REF0	REFERENCE
0	0	External
0	1	1.024 V
1	0	2.048 V
1	1	External

#### **CAUTION:**

If external reference voltage is applied to the REF pin, external reference MUST be selected.



#### **EXAMPLES OF OPERATION:**

- 1. Set DAC A output, select fast mode, select internal reference at 2.048 V:
  - a. Set reference voltage to 2.048 V (CONTROL register)

D	15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
	1	1	0	1	0	0	0	0	0	0	0	0	0	0	1	0

b. Write new DAC A value and update DAC A output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	0					New	DAC A c	utput va	lue				

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- 2. Set DAC B output, select fast mode, select external reference:
  - a. Select external reference (CONTROL register):

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	1	0	1	0	0	0	0	0	0	0	0	0	0	0	0

b. Write new DAC B value to BUFFER and update DAC B output:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	1	0	0				1	New BUFF	ER conter	nt and DA	C B outpu	t value			

The DAC A output is updated on the rising clock edge after D0 is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

- 3. Set DAC A value, set DAC B value, update both simultaneously, select slow mode, select internal reference at 1.024 V:
  - a. Set reference voltage to 1.024 V (CONTROL register)

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	1	0	0	0	0	0	0	0	0	0	0	0	1

b. Write data for DAC B to BUFFER:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
0	0	0	1						New D	AC B valu					

c. Write new DAC A value and update DAC A and B simultaneously:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
1	0	0	0						New D	AC A valu	e				

Both outputs are updated on the rising clock edge after D0 from the DAC A data word is sampled.

To output data consecutively using the same DAC configuration, it is not necessary to program the CONTROL register again.

1. Set power-down mode:

D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Χ	Χ	1	Х	Х	Х	Х	Х	Χ	Χ	Х	Х	Х	Х	Χ	Х

X = Don't care



#### LINEARITY, OFFSET, AND GAIN ERROR USING SINGLE ENDED SUPPLIES

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset, the output voltage changes on the first code change. With a negative offset, the output voltage may not change with the first code, depending on the magnitude of the offset voltage.

The output amplifier attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground and clamps the output at 0 V.

The output voltage then remains at zero until the input code value produces a sufficient positive output voltage to overcome the negative offset voltage, resulting in the transfer function shown in Figure 14.

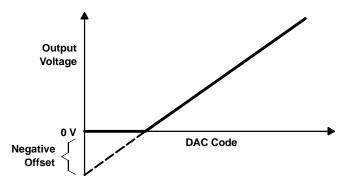


Figure 14. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces this breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below the ground rail.

For a DAC, linearity is measured between zero-input code (all inputs 0) and full-scale code (all inputs 1) after offset and full scale are adjusted out or accounted for in some way. However, single supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity is measured between full-scale code and the lowest code that produces a positive output voltage.



#### **DEFINITIONS OF SPECIFICATIONS AND TERMINOLOGY**

#### Integral Nonlinearity (INL)

The relative accuracy or integral nonlinearity (INL), sometimes referred to as linearity error, is the maximum deviation of the output from the line between zero and full scale excluding the effects of zero code and full-scale errors.

#### **Differential Nonlinearity (DNL)**

The differential nonlinearity (DNL), sometimes referred to as differential error, is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.

#### Zero-Scale Error (E<sub>ZS</sub>)

Zero-scale error is defined as the deviation of the output from 0 V at a digital input value of 0.

#### Gain Error (E<sub>G</sub>)

Gain error is the error in slope of the DAC transfer function.

#### **Total Harmonic Distortion (THD)**

THD is the ratio of the rms value of the first six harmonic components to the value of the fundamental signal. The value for THD is expressed in decibels.

#### Signal-to-Noise Ratio + Distortion (S/N+D)

S/N+D is the ratio of the rms value of the output signal to the rms sum of all other spectral components below the Nyquist frequency, including harmonics but excluding dc. The value for S/N+D is expressed in decibels.

#### Spurious Free Dynamic Range (SFDR)

Spurious free dynamic range is the difference between the rms value of the output signal and the rms value of the largest spurious signal within a specified bandwidth. The value for SFDR is expressed in decibels.





25-Sep-2013

#### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package Drawing	Pins		Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
5962-9957601Q2A	ACTIVE	LCCC	FK	20	Qty 1	(2) TBD	POST-PLATE	N / A for Pkg Type	-55 to 125	(4/5) 5962- 9957601Q2A TLV5638 MFKB	Sample
5962-9957601QPA	ACTIVE	CDIP	JG	8	1	TBD	A42	N / A for Pkg Type	-55 to 125	9957601QPA TLV5638M	Sample
TLV5638CD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C	Sample
TLV5638CDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C	Sample
TLV5638CDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C	Sample
TLV5638CDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	0 to 70	5638C	Sample
TLV5638ID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I	Sample
TLV5638IDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I	Sample
TLV5638IDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I	Sample
TLV5638IDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	5638I	Sample
TLV5638MFKB	ACTIVE	LCCC	FK	20	1	TBD	POST-PLATE	N / A for Pkg Type	-55 to 125	5962- 9957601Q2A TLV5638 MFKB	Sample
TLV5638MJGB	ACTIVE	CDIP	JG	8	1	TBD	A42	N / A for Pkg Type	-55 to 125	9957601QPA TLV5638M	Sample
TLV5638QD	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	V5638	Sample
TLV5638QDG4	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V5638	Sample
TLV5638QDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	V5638	Sample



## PACKAGE OPTION ADDENDUM

25-Sep-2013

Orderable Device	Status	Package Type	_	Pins	_	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)		(3)		(4/5)	
TLV5638QDRG4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM		V5638	Samples

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

- (3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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#### OTHER QUALIFIED VERSIONS OF TLV5638, TLV5638M:

Catalog: TLV5638



# **PACKAGE OPTION ADDENDUM**

25-Sep-2013

● Enhanced Product: TLV5638-EP, TLV5638-EP

Military: TLV5638M

NOTE: Qualified Version Definitions:

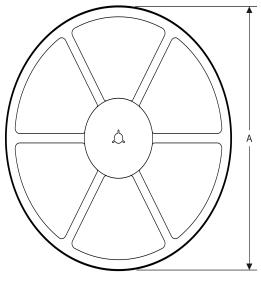
- Catalog TI's standard catalog product
- Enhanced Product Supports Defense, Aerospace and Medical Applications
- Military QML certified for Military and Defense Applications

# PACKAGE MATERIALS INFORMATION

www.ti.com 14-Jul-2012

## TAPE AND REEL INFORMATION

#### **REEL DIMENSIONS**





#### **TAPE DIMENSIONS**



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### TAPE AND REEL INFORMATION

#### \*All dimensions are nominal

Device Device	Package Type	Package Drawing		SPQ	Reel Diameter	Reel Width	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
	1,750	- · · · · · · · · · · · · · · · · · ·				W1 (mm)	(,	()	()	()	()	
TLV5638CDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5638IDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TLV5638QDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1

www.ti.com 14-Jul-2012



\*All dimensions are nominal

_								
	Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
	TLV5638CDR	SOIC	D	8	2500	367.0	367.0	35.0
	TLV5638IDR	SOIC	D	8	2500	367.0	367.0	35.0
	TLV5638QDR	SOIC	D	8	2500	367.0	367.0	35.0

## JG (R-GDIP-T8)

#### **CERAMIC DUAL-IN-LINE**



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. This package can be hermetically sealed with a ceramic lid using glass frit.
- D. Index point is provided on cap for terminal identification.
- E. Falls within MIL STD 1835 GDIP1-T8

# FK (S-CQCC-N\*\*)

# LEADLESS CERAMIC CHIP CARRIER

28 TERMINAL SHOWN



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. This package can be hermetically sealed with a metal lid.
- D. Falls within JEDEC MS-004



# D (R-PDSO-G8)

## PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
- E. Reference JEDEC MS-012 variation AA.



# D (R-PDSO-G8)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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