

### LMV225/LMV226/LMV228 RF Power Detector for CDMA and WCDMA

Check for Samples: LMV225, LMV226, LMV228

### **FEATURES**

- 30 dB Linear in dB Power Detection Range
- **Output Voltage Range 0.2 to 2V**
- **Logic Low Shutdown**
- Multi-Band Operation from 450 MHz to 2000
- **Accurate Temperature Compensation**
- Packages:
  - DSBGA Thin 1.0 mm x 1.0 mm x 0.6 mm
  - DSBGA Ultra Thin 1.0 mm x 1.0 mm x 0.35 mm
  - WSON 2.2 mm x 2.5 mm x 0.8 mm
    - (LMV225 and LMV228)

### **APPLICATIONS**

- **CDMA RF Power Control**
- **WCDMA RF Power Control**
- **PA Modules**

CDMA2000 RF Power Control

### **Typical Application**

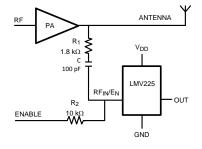


Figure 1. LMV225

### DESCRIPTION

The LMV225/LMV226/LMV228 are 30 dB RF power detectors intended for use in CDMA and WCDMA applications. The device has an RF frequency range from 450 MHz to 2 GHz. It provides an accurate temperature and supply compensated output voltage that relates linearly to the RF input power in dBm. The circuit operates with a single supply from 2.7V to The LMV225/LMV226/LMV228 have an integrated filter for low-ripple average power detection of CDMA signals with 30 dB dynamic range. Additional filtering can be applied using a single external capacitor.

The LMV225 has an RF power detection range from -30 dBm to 0 dBm and is ideally suited for direct use combination with resistive taps. LMV226/LMV228 have a detection range from -15 dBm to 15 dBm and are intended for use in combination with a directional coupler. The LMV226 is equipped with a buffered output which makes it suitable for GSM, EDGE, GPRS and TDMA applications.

The device is active for Enable = HI, otherwise it is in a low power consumption shutdown mode. During shutdown the output will be LOW. The output voltage ranges from 0.2V to 2V and can be scaled down to meet ADC input range requirements.

The LMV225/LMV226/LMV228 power detectors are offered in the thin 1.0 mm x 1.0 mm x 0.6 mm DSBGA package and the ultra thin 1.0 mm x 1.0 mm x 0.35 mm DSBGA package. The LMV225 and the LMV228 are also offered in the 2.2 mm x 2.5 mm x 0.8 mm WSON package.

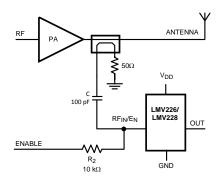


Figure 2. LMV226/LMV228

Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet. All trademarks are the property of their respective owners.





These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### ABSOLUTE MAXIMUM RATINGS (1)(2)

Supply Voltage	
V <sub>DD</sub> - GND	6.0V Max
ESD Tolerance (3)	1
Human Body Model	2000V
Machine Model	200V
Storage Temperature Range	−65°C to 150°C
Junction Temperature (4)	150°C Max
Mounting Temperature, Infrared or convection (20 sec)	
Tin/Lead	235°C
Lead-Free	260°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (3) Human body model: 1.5 k $\Omega$  in series with 100 pF. Machine model,  $0\Omega$  in series with 100 pF.
- (4) The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$  and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)} T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board

### OPERATING RATINGS (1)

Supply Voltage	2.7V to 5.5V
Temperature Range	-40°C to +85°C
RF Frequency Range	450 MHz to 2 GHz

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.



### 2.7 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified all limits are specified to  $V_{DD} = 2.7V$ :  $T_{c} = 25^{\circ}C$ . **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition		Min	Тур	Max	Units
I <sub>DD</sub>	Supply Current	Active Mode: RF <sub>IN</sub> /E <sub>N</sub> = V <sub>DD</sub> (DC), No RF Input Power	LMV225		4.8	7 <b>8</b>	
		Present	LMV226		4.9	6.2 <b>8</b>	mA
			LMV228		4.9	6.2 <b>8</b>	
		Shutdown: $RF_{IN}/E_N = GND$ (D Power Present	C), No RF Input		0.44	4.5	μΑ
$V_{LOW}$	E <sub>N</sub> Logic Low Input Level					0.8	V
V <sub>HIGH</sub>	E <sub>N</sub> Logic High Input Level			1.8			V
t <sub>on</sub>	Turn-on-Time (3)	No RF Input Power Present,	LMV225		2.1		μs
		Output Loaded with 10 pF	LMV226		1.2		
			LMV228		1.7		
t <sub>r</sub>	Rise Time <sup>(4)</sup>	Step from no Power to 0 dBm Applied, Output Loaded with 10 pF	LMV225		4.5		
		Step from no Power to	LMV226		1.8		μs _
		15 dBm Applied, Output Loaded with 10 pF	LMV228		4.8		
I <sub>EN</sub>	Current into RF <sub>IN</sub> /E <sub>N</sub> Pin					1	μΑ
P <sub>IN</sub>	Input Power Range (5)	LMV225			-30 0		dBm
					-43 -13		dBV
		LMV226			-15 15		dBm
					-28 2		dBV
		LMV228			-15 15		dBm
					-28 2		dBV

 <sup>(1)</sup> Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>.
 (2) All limits are specified by design or statistical analysis

Turn-on time is measured by connecting a 10 k $\Omega$  resistor to the RF<sub>IN</sub>/E<sub>N</sub> pin. Be aware that in the actual application on the front page, the RC-time constant of resistor R2 and capacitor C adds an additional delay.

Typical values represent the most likely parametric norm.

Power in dBV = dBm + 13 when the impedance is  $50\Omega$ .



### 2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD}$  = 2.7V;  $T_J$  = 25°C. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition		Min	Тур	Max	Units
	Logarithmic Slope (6)	900 MHz	LMV225		44.0		
			LMV226		44.5		
			LMV228 DSBGA		44.0		
			LMV228 WSON		48.5		
		1800 MHz	LMV225		39.4		
			LMV226		41.6		
			LMV228 DSBGA		41.9		
			LMV228 WSON		47.4		\//dD
		1900 MHz	LMV225		38.5		mV/dB
			LMV226		41.2		
			LMV228 DSBGA		41.6		
			LMV228 WSON		46.6		
		2000 MHz	LMV225		38.5		
			LMV226		41.0		
			LMV228 DSBGA		41.2		
			LMV228 WSON		45.4		
	Logarithmic Intercept (6)	900 MHz	LMV225		-45.5		
			LMV226		-24.5		
			LMV228 DSBGA		-27.2		
			LMV228 WSON		-23.7		
		1800 MHz	LMV225		-46.6		
			LMV226		-25.1		
			LMV228 DSBGA		-28.2		
			LMV228 WSON		-23.8		dD.m
		1900 MHz	LMV225		-46.3		dBm
			LMV226		-24.9		
			LMV228 DSBGA		-28.0		
			LMV228 WSON		-23.7		
		2000 MHz	LMV225		-46.7		
			LMV226		-24.7		
			LMV228 DSBGA		-28.0		_
			LMV228 WSON		-23.6		
$V_{OUT}$	Output Voltage	No RF Input Power Present	LMV225		214	350	1
			LMV226		223	350	mV
				228	350		
I <sub>OUT</sub>	Output Current Sourcing/Sinking	LMV226 Only		4.5	5.3		mA
R <sub>OUT</sub>	Output Impedance	LMV225/LMV228 only, no RF Present	Input Power		19.8	29 <b>34</b>	kΩ

<sup>(6)</sup> Device is set in active mode with a 10 k $\Omega$  resistor from  $V_{DD}$  to RF<sub>IN</sub>/E<sub>N</sub>. RF signal is applied using a 50 $\Omega$  RF signal generator AC coupled to the RF<sub>IN</sub>/E<sub>N</sub> pin using a 100 pF coupling capacitor.



### 2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD}$  = 2.7V;  $T_J$  = 25°C. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition	า	Min	Тур	Max	Units
e <sub>n</sub>	Output Referred Noise	RF Input = 1800 MHz, -10 d and 5 dBm for LMV226/LMV 10 kHz			700		nV/√Hz
	Variation Due to Temperature	900 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0.64 -1.07		
		900 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.05 -0.02		
			LMV228 DSBGA		+0.22 -0.36		
			LMV228 WSON		+0.87 -0.87		
			LMV225		+0.09 -0.86		
		1800 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.07 -0.10		
			LMV228 DSBGA		+0.29 -0.57		
			LMV228 WSON		+1.04 -1.23		40
		1900 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0 -0.69		dB
		1900 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0 -0.10		
			LMV228 DSBGA		+0.23 -0.64		
			LMV228 WSON		+1.05 -1.45		
		2000 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0 -0.86		
		2000 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0 -0.29		
			LMV228 DSBGA		+0.27 -0.65		
			LMV228 WSON		+1.04 -2.02		



### 5.0 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified to V<sub>DD</sub> = 5.0V; T<sub>J</sub> = 25°C. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition		Min	Тур	Max	Units
I <sub>DD</sub>	Supply Current	Active Mode: $RF_{IN}/E_N = V_{DD}$ (DC), no RF Input Power	LMV225		5.3	7.5 <b>9</b>	
		Present.	LMV226		5.3	6.8 <b>9</b>	mA
			LMV228		5.4	6.8 <b>9</b>	
		Shutdown: $RF_{IN}/E_N = GND$ (D Power Present.	C), no RF Input		0.32	4.5	μΑ
$V_{LOW}$	E <sub>N</sub> Logic Low Input Level					0.8	V
V <sub>HIGH</sub>	E <sub>N</sub> Logic High Input Level			1.8			V
t <sub>on</sub>	Turn-on-Time (3)	No RF Input Power Present,	LMV225		2.1		
		Output Loaded with 10 pF	LMV226		1.0		μs
			LMV228		1.7		
t <sub>r</sub>	Rise Time (4)	Step from no Power to 0 dBm Applied, Output Loaded with 10 pF	LMV225		4.5		
		Step from no Power to			1.4		μs
		15 dBm Applied, Output Loaded with 10 pF	LMV228		4.8		
I <sub>EN</sub>	Current Into RF <sub>IN</sub> /E <sub>N</sub> Pin		•			1	μΑ
P <sub>IN</sub>	Input Power Range (5)	LMV225			-30 0		dBm
					-43 -13		dBV
		LMV226			−15 15		dBm
					-28 2		dBV
		LMV228			-15 15		dBm
					-28 2		dBV

<sup>(1)</sup> Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ . All limits are specified by design or statistical analysis

Turn-on time is measured by connecting a 10 k $\Omega$  resistor to the RF<sub>IN</sub>/E<sub>N</sub> pin. Be aware that in the actual application on the front page, the RC-time constant of resistor  $R_2$  and capacitor C adds an additional delay.

Typical values represent the most likely parametric norm.

Power in dBV = dBm + 13 when the impedance is  $50\Omega$ . (5)



### 5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD}$  = 5.0V;  $T_J$  = 25°C. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition		Min	Тур	Max	Units
	Logarithmic Slope (6)	900 MHz	LMV225		44.6		
			LMV226		44.6		
			LMV228 DSBGA		44.2		
			LMV228 WSON		48.4		
		1800 MHz	LMV225		40.6		
			LMV226		42.2		
			LMV228 DSBGA		42.4		
			LMV228 WSON		48.3		~~\//dF
		1900 MHz	LMV225		39.6		mV/dE
			LMV226		41.8		
			LMV228 DSBGA		42.2		
			LMV228 WSON		47.8		
		2000 MHz	LMV225		39.7		
			LMV226		41.6		
			LMV228 DSBGA		41.8		
			LMV228 WSON		47.2		
	Logarithmic Intercept (6)	900 MHz	LMV225		-47.0		
			LMV226		-25.0		
			LMV228 DSBGA		-27.7		
			LMV228 WSON		-23.9		-
		1800 MHz	LMV225		-48.5		
			LMV226		-25.7		
			LMV228 DSBGA		-28.9		
			LMV228 WSON		-23.6		dBm
		1900 MHz	LMV225		-48.2		ubili
			LMV226		-25.6		
			LMV228 DSBGA		-28.7		
			LMV228 WSON		-23.1		
		2000 MHz	LMV225		-48.9		
			LMV226		-25.5		
			LMV228 DSBGA		-28.7		
			LMV228 WSON		-23.0		
OUT	Output Voltage	No RF Input Power Present	LMV225		222	400	
			LMV226		231	400	mV
			LMV228		244	400	
OUT	Output Current Sourcing/Sinking	LMV226 Only		4.5	5.3		mA
OUT	Output Impedance	No RF Input Power Present			23.7	29 <b>31</b>	kΩ

<sup>(6)</sup> Device is set in active mode with a 10 k $\Omega$  resistor from  $V_{DD}$  to RF<sub>IN</sub>/E<sub>N</sub>. RF signal is applied using a 50 $\Omega$  RF signal generator AC coupled to the RF<sub>IN</sub>/E<sub>N</sub> pin using a 100 pF coupling capacitor.

Product Folder Links: LMV225 LMV226 LMV228



### 5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD}$  = 5.0V;  $T_J$  = 25°C. **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Condition	1	Min	Тур	Max	Units
e <sub>n</sub>	Output Referred Noise	RF Input = 1800 MHz, -10 d and 5 dBm for LMV226/LMV: 10 kHz			700		nV/√Hz
	Variation Due to Temperature	900 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0.89 -1.16		
		900 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.25 -0.16		
			LMV228 DSBGA		+0.46 -0.62		
			LMV228 WSON		+1.39 -1.19		
			LMV225		+0.3 -0.82		
		1800 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.21 -0.09		
			LMV228 DSBGA		+0.55 -0.78		
			LMV228 WSON		+1.39 -1.43		dB
		1900 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0.34 -0.63		uв
		1900 MHz, RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.21 -0.19		
			LMV228 DSBGA		+0.55 -0.93		
			LMV228 WSON		+1.54 -1.64		
		2000 MHz, RF <sub>IN</sub> = 0 dBm Referred to 25°C	LMV225		+0.22 -0.75		
		2000 MHz RF <sub>IN</sub> = 15 dBm Referred to 25°C	LMV226		+0.25 -0.34		
			LMV228 DSBGA		+0.61- 0.91		
			LMV228 WSON		+0.89 -0.99		



### **CONNECTION DIAGRAM**

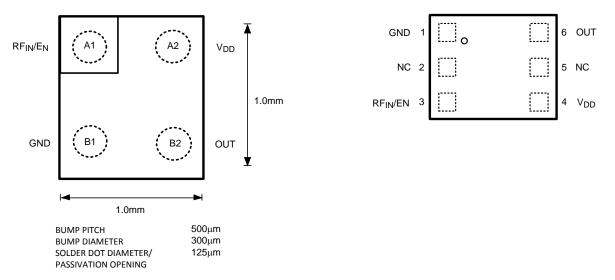


Figure 3. 4-Bump DSBGA – Top View See Package Number YZR0004 or YPD0004

Figure 4. 6-pin WSON – Top View See Package Number NGF0006A

### **PIN DESCRIPTIONS**

	Р	in	Name	Description
	DSBGA	WSON6		
Power Supply	A2	4	$V_{DD}$	Positive Supply Voltage
	B1	1	GND	Power Ground
	A1	3	RF <sub>IN</sub> /E <sub>N</sub>	DC voltage determines enable state of the device (HIGH = device active). AC voltage is the RF input signal to the detector (beyond 450 MHz). The RF $_{\text{IN}}$ /E $_{\text{N}}$ pin is internally terminated with 50 $\Omega$ in series with 45 pF.
Output	B2	6	Out	Ground referenced detector output voltage (linear in dBm)

Product Folder Links: LMV225 LMV226 LMV228



### **Block Diagrams**

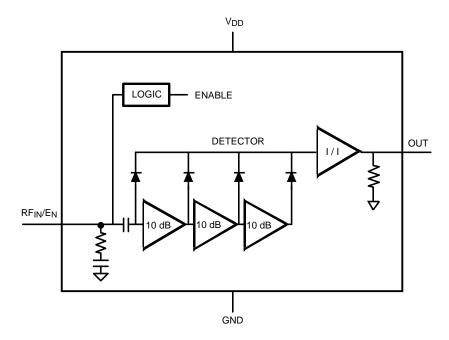


Figure 5. LMV225

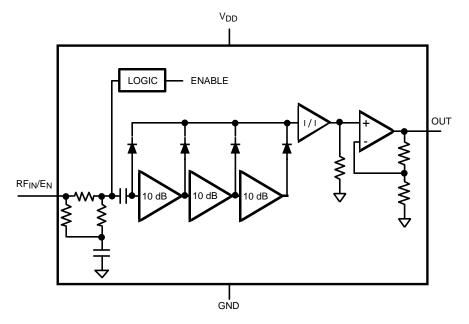


Figure 6. LMV226



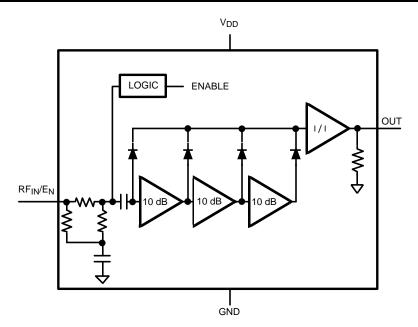
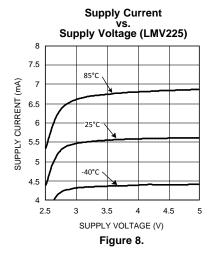


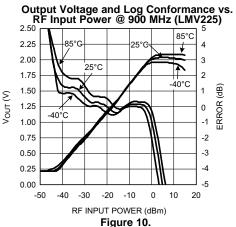
Figure 7. LMV228

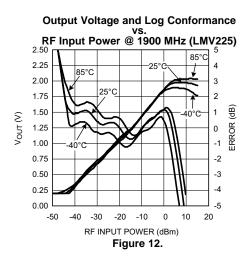


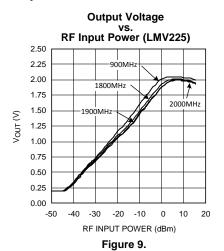
### **TYPICAL PERFORMANCE CHARACTERISTICS LMV225**

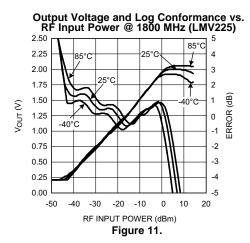
Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

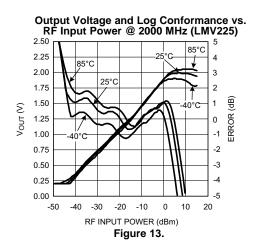














## TYPICAL PERFORMANCE CHARACTERISTICS LMV225 (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

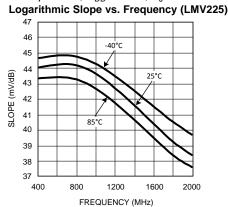


Figure 14.

## Output Variation vs. RF Input Power Normalized to 25°C @ 900 MHz (LMV225)

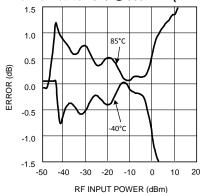


Figure 16.

### Output Variation vs. RF Input Power Normalized to 25°C @ 1900 MHz (LMV225)

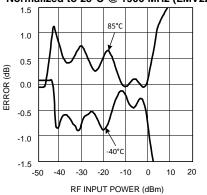


Figure 18.

#### Logarithmic Intercept vs. Frequency (LMV225)

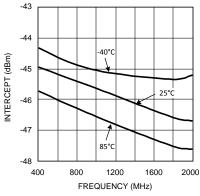


Figure 15.

## Output Variation vs. RF Input Power Normalized to 25°C @ 1800 MHz (LMV225)

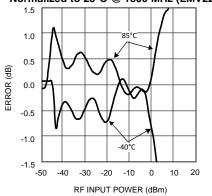


Figure 17.

### Output Variation vs. RF Input Power Normalized to 25°C @ 2000 MHz (LMV225)

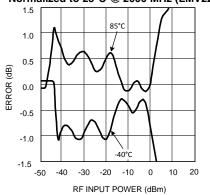
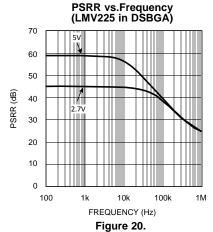


Figure 19.



## TYPICAL PERFORMANCE CHARACTERISTICS LMV225 (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.



PSRR vs. Frequency (LMV225 in WSON)

70

60

50

40

22.7v

20

10

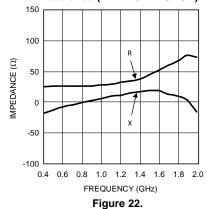
10

1 k 10k 100k 1M

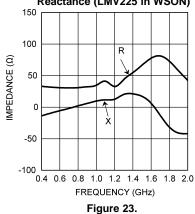
FREQUENCY (Hz)

Figure 21.

RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV225 in DSBGA)



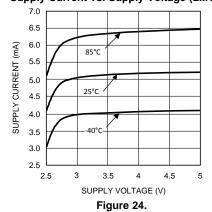
RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV225 in WSON)



**TYPICAL PERFORMANCE CHARACTERISTICS LMV226** 

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

Supply Current vs. Supply Voltage (LMV226)



Output Voltage vs. RF Input Power (LMV226)

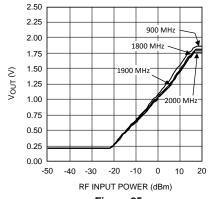


Figure 25.



## TYPICAL PERFORMANCE CHARACTERISTICS LMV226 (continued)

Unless otherwise specified,  $V_{DD}$  = 2.7V,  $T_{J}$ = 25°C.

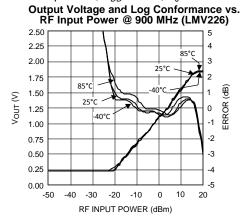


Figure 26.

### Output Voltage and Log Conformance vs. RF Input Power @ 1900 MHz (LMV226)

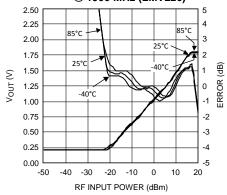


Figure 28.

#### Logarithmic Slope vs. Frequency (LMV226)

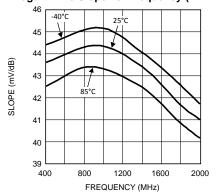


Figure 30.

### Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV226)

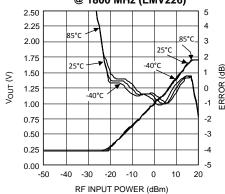


Figure 27.

### Output Voltage and Log Conformance vs. RF Input Power @ 2000 MHz (LMV226)

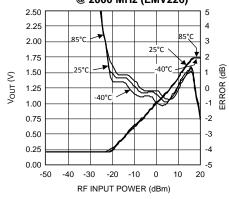


Figure 29.

#### Logarithmic Intercept vs. Frequency (LMV226)

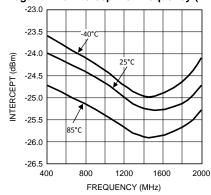
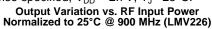


Figure 31.



## TYPICAL PERFORMANCE CHARACTERISTICS LMV226 (continued)

Unless otherwise specified,  $V_{DD}$  = 2.7V,  $T_{J}$ = 25°C.



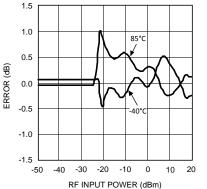


Figure 32.

## Output Variation vs. RF Input Power Normalized to 25°C @ 1900 MHz (LMV226)

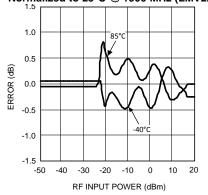
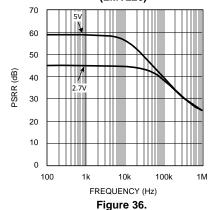


Figure 34.

#### PSRR vs. Frequency (LMV226)



### Output Variation vs. RF Input Power Normalized to 25°C @ 1800 MHz (LMV226)

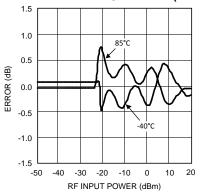


Figure 33.

## Output Variation vs. RF Input Power Normalized to 25°C @ 2000 MHz (LMV226)

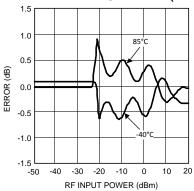


Figure 35.

### RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV226)

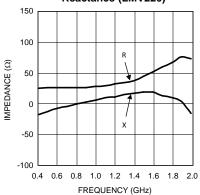


Figure 37.



### TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

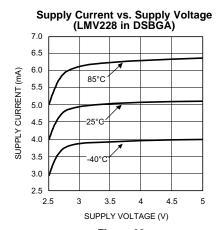


Figure 38.

### Output Voltage and Log Conformance vs. RF Input Power @ 900 MHz (LMV228 in DSBGA)

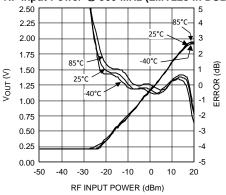


Figure 40.

### Output Voltage and Log Conformance vs. RF Input Power @ 1900 MHz (LMV228 in DSBGA)

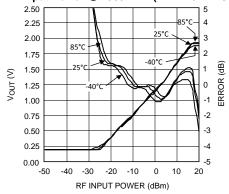


Figure 42.

### Output Voltage vs. RF Input Power (LMV228 in DSBGA)

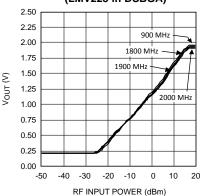


Figure 39.

### Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV228 in DSBGA)

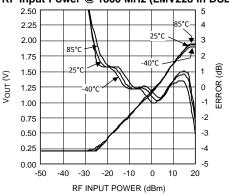


Figure 41.

### Output Voltage and Log Conformance vs. RF Input Power @ 2000 MHz (LMV228 in DSBGA)

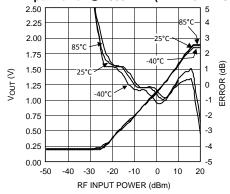


Figure 43.



## TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

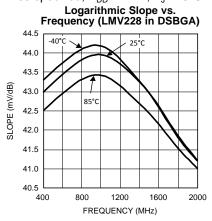
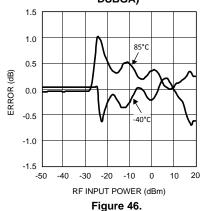


Figure 44.

Output Variation vs.
RF Input Power Normalized to 25°C @ 900 MHz (LMV228 in DSBGA)



Output Variation vs.
RF Input Power Normalized to 25°C @ 1900 MHz (LMV228 in DSBGA)

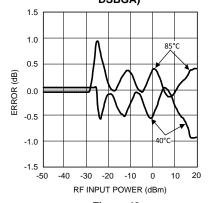


Figure 48.

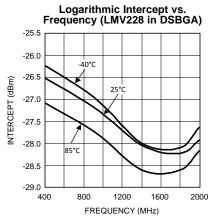


Figure 45.

Output Variation vs.
RF Input Power Normalized to 25°C @ 1800 MHz (LMV228 in DSBGA)

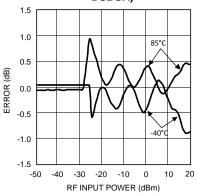


Figure 47.

## Output Variation vs. RF Input Power Normalized to 25°C @ 2000 MHz (LMV228 in DSBGA)

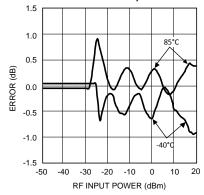
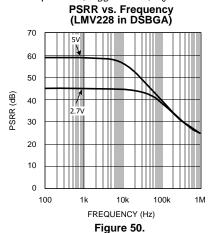


Figure 49.



### TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.



### RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV228 in DSBGA)

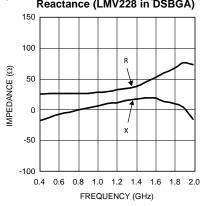
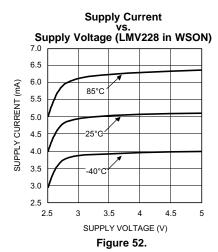
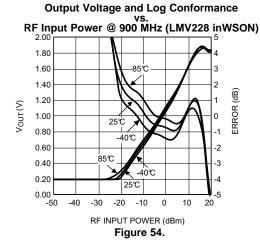


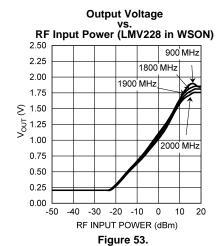
Figure 51.

### TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON

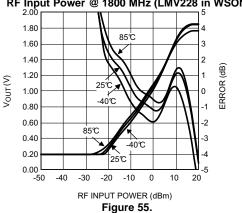
Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.







Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV228 in WSON) 1.80





### TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_{J} = 25$ °C.

### **Output Voltage and Log Conformance**

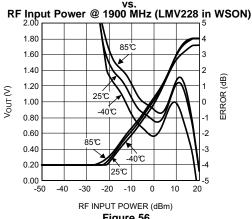
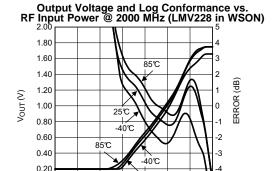


Figure 56.



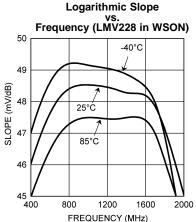
RF INPUT POWER (dBm) Figure 57.

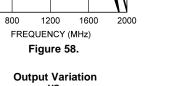
0.00

-50 -40 -30 -20

25℃

0 10 20

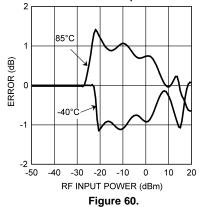




Logarithmic Intercept vs. Frequency (LMV228 in WSON) -22 -23 INTERCEPT (dBm) 25°C -25 85°C -26 400 800 1200 1600

**Output Variation** 

# vs. RF Input Power Normalized to 25°C @ 900 MHz (LMV228 in WSON)



# vs. RF Input Power Normalized to 25°C @ 1800 MHz (LMV228 in WSON)

FREQUENCY (MHz)

Figure 59.

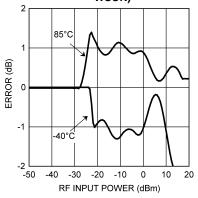


Figure 61.



### TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON (continued)

Unless otherwise specified,  $V_{DD}$  = 2.7V,  $T_J$ = 25°C.

### Output Variation vs. RF Input Power Normalized to 25°C @ 1900 MHz (LMV228 in WSON)

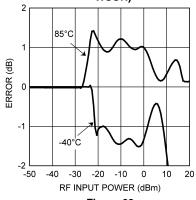
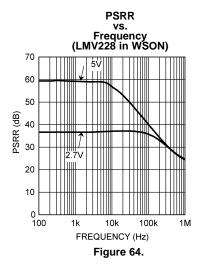


Figure 62.



**Output Variation** vs.
RF Input Power Normalized to 25°C @ 2000 MHz (LMV228 in WSON)

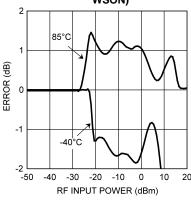


Figure 63.

**RF Input Impedance** Frequency @ Resistance and Reactance (LMV228 in WSON)

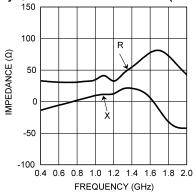


Figure 65.



#### **APPLICATION NOTES**

#### **CONFIGURING A TYPICAL APPLICATION**

The LMV225/LMV226/LMV228 are power detectors intended for CDMA and WCDMA applications. Power applied at its input translates to a DC voltage on the output through a linear-in-dB response. The LMV225 detector is especially suited for power measurements via a high-resistive tap, while the LMV226/LMV228 are designed to be used in combination with a directional coupler. The LMV226 has an additional output voltage buffer and therefore a low output impedance. The key features of the devices are shown in .

**Table 1. DEVICE CHARACTERISTICS** 

	Input Range (dBm)	Output Buffer	Application
LMV225	-30 / 0	No	High Resistive Tap
LMV226	-15 / 15	Yes	Directional Coupler
LMV228	-15 / 15	No	Directional Coupler

In order to match the output power range of the power amplifier (PA) with the range of the LMV225's input, the high resistive tap needs to be configured correctly. In case of the LMV226/LMV228 the coupling factor of the directional coupler needs to be chosen correctly.

#### HIGH RESISTIVE TAP APPLICATION

The constant input impedance of the device enables the realization of a frequency independent input attenuation to adjust the LMV225's range to the range of the PA. Resistor  $R_1$  and the  $50\Omega$  input resistance ( $R_{IN}$ ) of the device realize this attenuation (Figure 66). To minimize insertion loss, resistor  $R_1$  needs to be sufficiently large. The following example demonstrates how to determine the proper value for  $R_1$ .

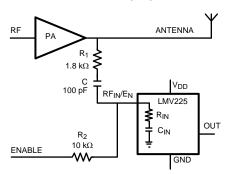


Figure 66. Typical LMV225 Application with High Resistive Tap

Suppose the useful output power of the PA ranges up to +31 dBm. As the LMV225 can handle input power levels up to 0 dBm.  $R_1$  should realize a minimum attenuation of 31 - 0 = 31 dB. The attenuation realized by  $R_1$  and the effective input resistance  $R_{\text{IN}}$  of the detector equals:

$$A_{dB} = 20 \cdot LOG \left[ 1 + \frac{R_1}{R_{IN}} \right] = 31dB \tag{1}$$

Solving this expression for  $R_1$ , using that  $R_{IN} = 50\Omega$ , yields:

$$R_{1} = \begin{bmatrix} \frac{A_{dB}}{20} & 1 \\ 10^{20} & 1 \end{bmatrix} \cdot R_{IN} = \begin{bmatrix} \frac{31}{20} & 1 \\ 10^{20} & 1 \end{bmatrix} \cdot 50 = 1724\Omega$$
 (2)

In Figure 66,  $R_1$  is set to 1800 $\Omega$  resulting in an attenuation of 31.4 dB.



#### DIRECTIONAL COUPLER APPLICATION

The LMV226/LMV228 also has a 50Ω input resistance. However, its input range differs compared to the LMV225, i.e. –15 dBm to +15 dBm. If a typical attenuation of a directional coupler is 20 dB, the LMV226/LMV228 can be directly connected via the directional coupler to the PA without the need of additional external attenuator (Figure 67). Different PA ranges can be configured using couplers with other coupling factors.

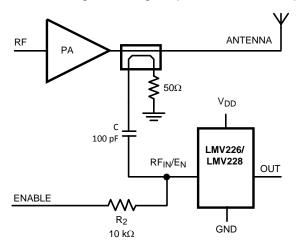


Figure 67. Typical LMV226/LMV228 Application with Directional Coupler

#### SHUTDOWN FUNCTIONALITY

The LMV225/LMV226/LMV228 RF<sub>IN</sub>/E<sub>N</sub> pins have 2 functions combined:

- Enable/Shutdown
- Power input

The capacitor C and the resistor  $R_2$  (Figure 66 and Figure 67) separate the DC shutdown functionality from the AC power measurement. The device is active when Enable = HI, otherwise it is in a low power consumption shutdown mode. During shutdown the output will be LOW.

Capacitor C should be chosen sufficiently large to ensure a corner frequency far below the lowest input frequency to be measured. In case of the LMV225 the corner frequency can be calculated using:

$$f = \frac{1}{2 \pi (R_1 + R_{IN}) \frac{C \cdot C_{IN}}{C + C_{IN}}}$$

where

• 
$$R_{IN} = 50\Omega$$
,  $C_{IN} = 45$  pF typical (3)

With  $R_1$  = 1800 $\Omega$  and C = 100 pF, this results in a corner frequency of 2.8 MHz. This corner frequency is an indicative number. The goal is to have a magnitude transfer, which is sufficiently flat in the used frequency range; capacitor C should be chosen significantly larger than capacitor  $C_{\text{IN}}$  to assure a proper performance of the high resistive tap. Capacitor C shouldn't be chosen excessively large since the RC-time, it introduces in combination with resistor  $R_2$ , adds to the turn-on time of the device.

The LMV226/LMV228 do not use a resistor  $R_1$  like the LMV225. Though a resistor is seen on the coupler side ( $R_{COUPLER}$ ). Therefore a similar equation holds for the LMV226/LMV228 LF corner frequency, where  $R_1$  is replaced with the coupler output impedance ( $R_{COUPLER}$ ).

With  $R_{COUPLER} = 50\Omega$  and C = 100 pF, the resulting corner frequency is 50 MHz.

The output voltage is proportional to the logarithm of the input power, often called "linear-in-dB". Figure 68 shows the typical output voltage versus PA output power of the LMV225 setup as depicted in Figure 66.

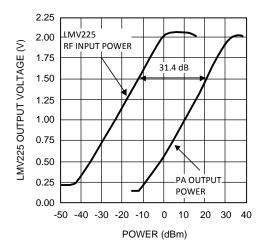


Figure 68. Typical power detector response, V<sub>OUT</sub> vs. PA output Power

### **OUTPUT RIPPLE DUE TO AM MODULATION**

A CDMA modulated carrier wave generally contains some amplitude modulation that might disturb the RF power measurement used for controlling the PA. This section explains the relation between amplitude modulation in the RF signal and the ripple on the output of the LMV225/LMV228. Expressions are provided to estimate this ripple on the output. The ripple can be further reduced by lowpass filtering at the output. This is realized by connecting an capacitor from the output of the LMV225/LMV228 to ground.

### **Estimating Output Ripple**

The CDMA modulated RF input signal of Figure 68 can be described as:

$$V_{\text{IN}}(t) = V_{\text{IN}} [1 + \mu(t)] \cos (2 \cdot \pi \cdot f \cdot t)$$

where

- V<sub>IN</sub> is the amplitude of the carrier frequency
- Amplitude modulation μ(t) can be between -1 and 1

 $V_{IN} = V_{IN} = V$ 

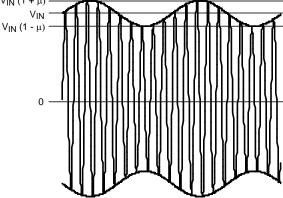


Figure 69. AM Modulated RF Signal



The ripple observed at the output of the detector equals the detectors response to the power variation at the input due to AM modulation (Figure 69). This signal has a maximum amplitude  $V_{IN}$  • (1+ $\mu$ ) and a minimum amplitude  $V_{IN}$  • (1- $\mu$ ), where 1+ $\mu$  can be maximum 2 and 1- $\mu$  can be minimum 0. The amplitude of the ripple can be described with the formula:

$$V_{RIPPLE} = V_{Y} \left[ 10 \text{ LOG} \left[ \frac{V_{IN}^{2} (1 + \mu)^{2}}{2R_{IN}} \right] + 30 \right] - V_{Y} \left[ 10 \text{ LOG} \left[ \frac{V_{IN}^{2} (1 - \mu)^{2}}{2R_{IN}} \right] + 30 \right]$$

$$P_{INMAX} \text{ IN dBm} \qquad P_{INMIN} \text{ IN dBm}$$

#### where

- V<sub>Y</sub> is the slope of the detection curve (Figure 70)
- µ is the modulation index (5)

#### Equation 5 can be reduced to:

$$V_{RIPPLE} = V_{Y} \cdot 20 LOG \left[ \frac{1 + \mu}{1 - \mu} \right]$$
 (6)

Consequently, the ripple is independent of the average input power of the RF input signal and only depends on the logarithmic slope  $V_Y$  and the ratio of the maximum and the minimum input signal amplitude.

For CDMA, the ratio of the maximum and the minimum input signal amplitude modulation is typically in the order of 5 to 6 dB, which is equivalent to a modulation index  $\mu$  of 0.28 to 0.33.

A further understanding of the equation above can be achieved via the knowledge that the output voltage  $V_{OUT}$  of the LMV225/LMV228 is linear in dB, or proportional to the input power  $P_{IN}$  in dBm. As discussed earlier, CDMA has a modulation in the order of 5 to 6 dB. Since the transfer is linear in dB, the output voltage  $V_{OUT}$  will vary linearly over about 5 to 6 dB in the curve (Figure 70).

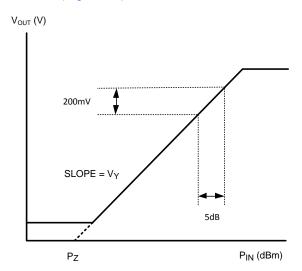


Figure 70. V<sub>OUT</sub> vs. RF Input Power P<sub>IN</sub>

The output voltage variation  $\Delta V_{OUT}$  is thus identical for RF input signals that fall within the linear range (in dB) of the detector. In other words, the output variation is independent of the absolute RF input signal:

$$\Delta V_{O} = V_{Y} \cdot \Delta P_{IN} \tag{7}$$

In which  $V_Y$  is the slope of the curve. The log-conformance error is usually much smaller than the ripple due to AM modulation. In case of the LMV225/LMV228,  $V_Y$  = 40 mV/dB. With  $\Delta P_{IN}$  = 5 dB for CDMA,  $\Delta V_{OUT}$  = 200 mV<sub>PP</sub>. This is valid for all  $V_{OUT}$ .



### Output Ripple with Additional Filtering

The calculated result above is for an unfiltered configuration. When a low pass filter is used by shunting a capacitor of e.g.  $C_{OUT} = 1.5$  nF at the output of the LMV225/LMV228 to ground, this ripple is further attenuated. The cut-off frequency follows from:

$$f_C = \frac{1}{2 \pi C_{OUT} R_O}$$
 (8)

With the output resistance of the LMV225/LMV228  $R_O$  = 19.8 k $\Omega$  typical and  $C_{OUT}$  = 1.5 nF, the cut-off frequency equals  $f_C$  = 5.36 kHz. A 100 kHz AM signal then gets attenuated by 5.36/100 or 25.4 dB. The remaining ripple will be less than 20 mV. With a slope of 40 mV/dB this translates into an error of less than ±0.5 dB. Since the LMV226 has a low output impedance buffer, a capacitor to reduce the ripple will not be effective.

### **Output Ripple Measurement**

Figure 71 shows the ripple reduction that can be achieved by adding additional capacitance at the output of the LMV225/LMV228. The RF signal of 900 MHz is AM modulated with a 100 kHz sinewave and a modulation index of 0.3. The RF input power is swept while the modulation index remains unchanged. Without the output capacitor the ripple is about 200 mV<sub>PP</sub>. Connecting a capacitor of 1.5 nF at the output to ground, results in a ripple of 12 mV<sub>PP</sub>. The attenuation with a 1.5 nF capacitor is then 20 • log (200/12) = 24.4 dB. This is very close to the calculated number of the previous paragraph.

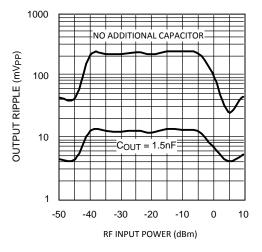


Figure 71. Output Ripple vs. RF Input Power

#### PRINCIPLE OF OPERATION

The logarithmic response of the LMV225/LMV226/LMV228 is implemented by a logarithmic amplifier as shown in Figure 72. The logarithmic amplifier consists of a number of cascaded linear gain cells. With these gain cells, a piecewise approximation of the logarithmic function is constructed.

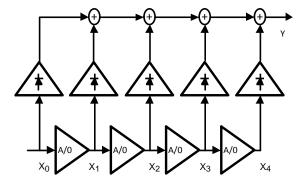


Figure 72. Logarithmic Amplifier



Every gain cell has a response according to Figure 73. At a certain threshold (E<sub>K</sub>), the gain cell starts to saturate, which means that the gain drops to zero. The output of gain cell 1 is connected to the input of gain cell 2 and so

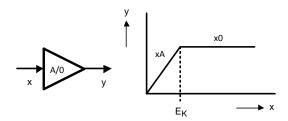


Figure 73. Gain Cell

All gain cell outputs are AM-demodulated with a peak detector and summed together. This results in a logarithmic function. The logarithmic range is about:

where

function.

- n = number of gain cells
- A = gain per gaincell

(9)Figure 74 shows a logarithmic function on a linear scale and the piecewise approximation of the logarithmic

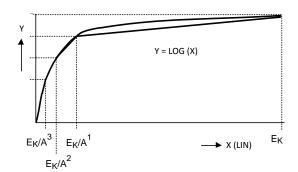


Figure 74. Log-Function on Lin Scale

Figure 75 shows a logarithmic function on a logarithmic scale and the piecewise approximation of the logarithmic function.

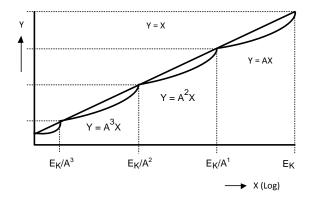


Figure 75. Log-Function on Log Scale



The maximum error for this approximation occurs at the geometric mean of a gain section, which is e.g. for the third segment:

$$\sqrt{\frac{E_K}{A^2} \cdot \frac{E_K}{A^1}} = \frac{E_K}{A\sqrt{A}} \tag{10}$$

The size of the error increases with distance between the thresholds.

#### LAYOUT CONSIDERATIONS

For a proper functioning part a good board layout is necessary. Special care should be taken for the series resistance  $R_1$  (Figure 66) that determines the attenuation. For high resistor values the parasitic capacitance of the resistor may significantly impact the realized attenuation. The effective attenuation will be lower than intended. To reduce the parasitic capacitance across resistor  $R_1$ , this resistor can be composed of several components in series instead of using a single component.





### **REVISION HISTORY**

Cł	nanges from Revision K (March 2013) to Revision L	Pag	ge
•	Changed layout of National Data Sheet to TI format		28





6-Nov-2017

### **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
LMV225SD/NOPB	NRND	WSON	NGF	6	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	A90	
LMV225SDX/NOPB	NRND	WSON	NGF	6	4500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	A90	
LMV225TL/NOPB	NRND	DSBGA	YZR	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV225TLX/NOPB	NRND	DSBGA	YZR	4	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV225UR/NOPB	NRND	DSBGA	YPD	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM			
LMV225URX/NOPB	NRND	DSBGA	YPD	4	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM			
LMV226TL/NOPB	NRND	DSBGA	YZR	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV226TLX/NOPB	NRND	DSBGA	YZR	4	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV226UR/NOPB	NRND	DSBGA	YPD	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM			
LMV228SD/NOPB	NRND	WSON	NGF	6	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM		A89	
LMV228TL/NOPB	NRND	DSBGA	YZR	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV228TLX/NOPB	NRND	DSBGA	YZR	4	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		
LMV228UR/NOPB	NRND	DSBGA	YPD	4	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM	-40 to 85		

<sup>(1)</sup> The marketing status values are defined as follows: **ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.



### **PACKAGE OPTION ADDENDUM**

6-Nov-2017

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

### PACKAGE MATERIALS INFORMATION

www.ti.com 2-Sep-2015

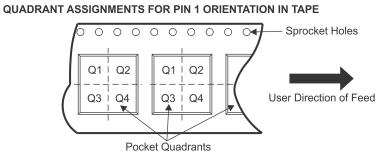
### TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

- Reel Width (WT)



### \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMV225SD/NOPB	WSON	NGF	6	1000	178.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV225SDX/NOPB	WSON	NGF	6	4500	330.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV225TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV225TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV225UR/NOPB	DSBGA	YPD	4	250	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV225URX/NOPB	DSBGA	YPD	4	3000	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV226TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV226TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV226UR/NOPB	DSBGA	YPD	4	250	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV228SD/NOPB	WSON	NGF	6	1000	178.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV228TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV228TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV228UR/NOPB	DSBGA	YPD	4	250	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1

www.ti.com 2-Sep-2015

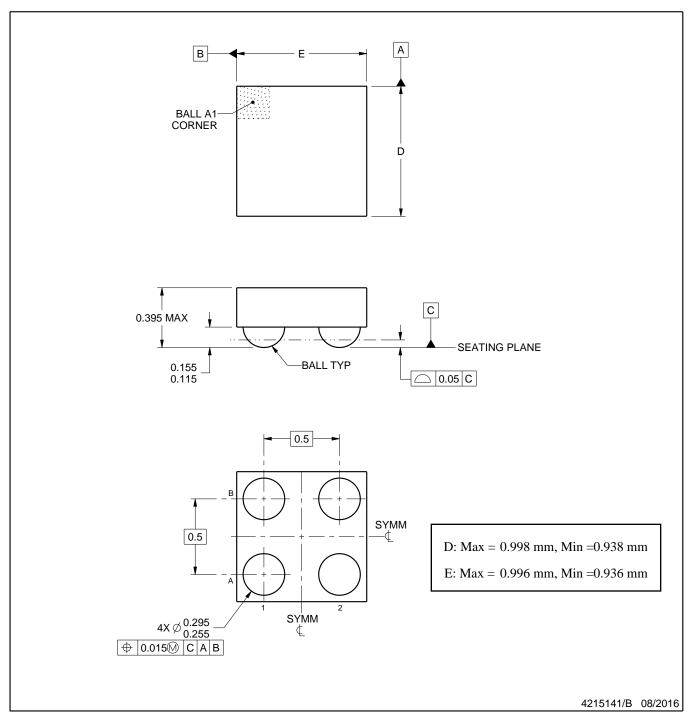


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMV225SD/NOPB	WSON	NGF	6	1000	210.0	185.0	35.0
LMV225SDX/NOPB	WSON	NGF	6	4500	367.0	367.0	35.0
LMV225TL/NOPB	DSBGA	YZR	4	250	210.0	185.0	35.0
LMV225TLX/NOPB	DSBGA	YZR	4	3000	210.0	185.0	35.0
LMV225UR/NOPB	DSBGA	YPD	4	250	210.0	185.0	35.0
LMV225URX/NOPB	DSBGA	YPD	4	3000	210.0	185.0	35.0
LMV226TL/NOPB	DSBGA	YZR	4	250	210.0	185.0	35.0
LMV226TLX/NOPB	DSBGA	YZR	4	3000	210.0	185.0	35.0
LMV226UR/NOPB	DSBGA	YPD	4	250	210.0	185.0	35.0
LMV228SD/NOPB	WSON	NGF	6	1000	210.0	185.0	35.0
LMV228TL/NOPB	DSBGA	YZR	4	250	210.0	185.0	35.0
LMV228TLX/NOPB	DSBGA	YZR	4	3000	210.0	185.0	35.0
LMV228UR/NOPB	DSBGA	YPD	4	250	210.0	185.0	35.0



DIE SIZE BALL GRID ARRAY



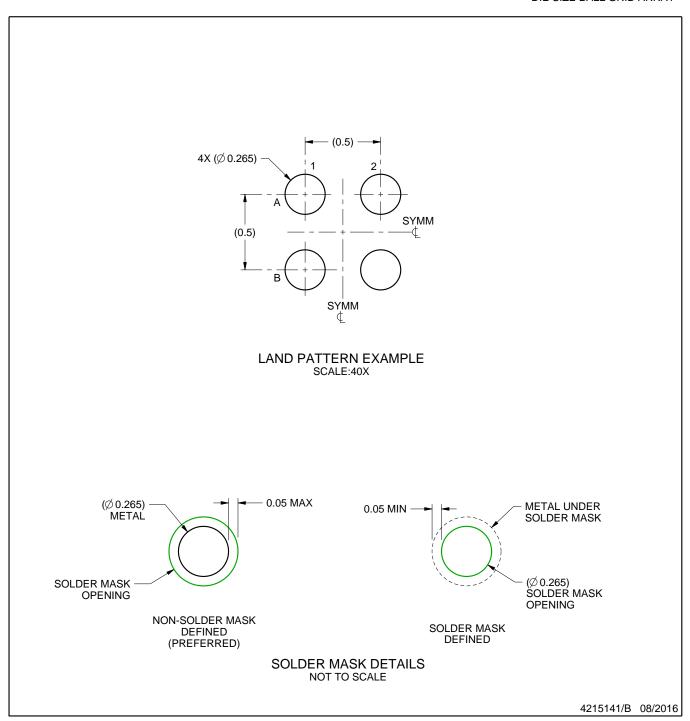
### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.



DIE SIZE BALL GRID ARRAY

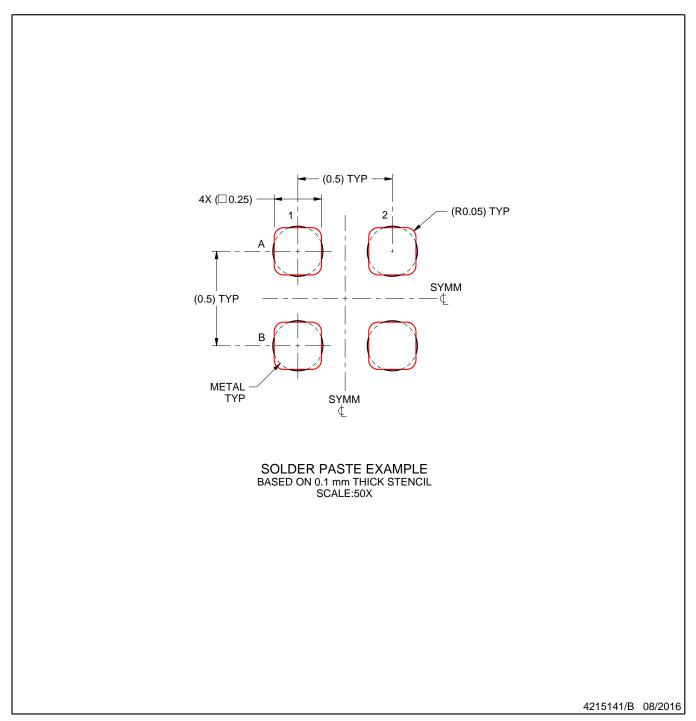


NOTES: (continued)

3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. See Texas Instruments Literature No. SNVA009 (www.ti.com/lit/snva009).



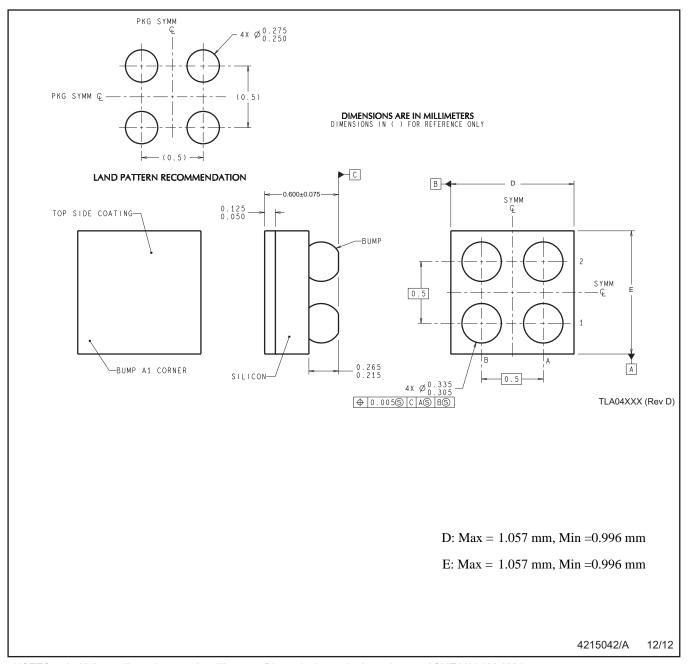
DIE SIZE BALL GRID ARRAY



NOTES: (continued)

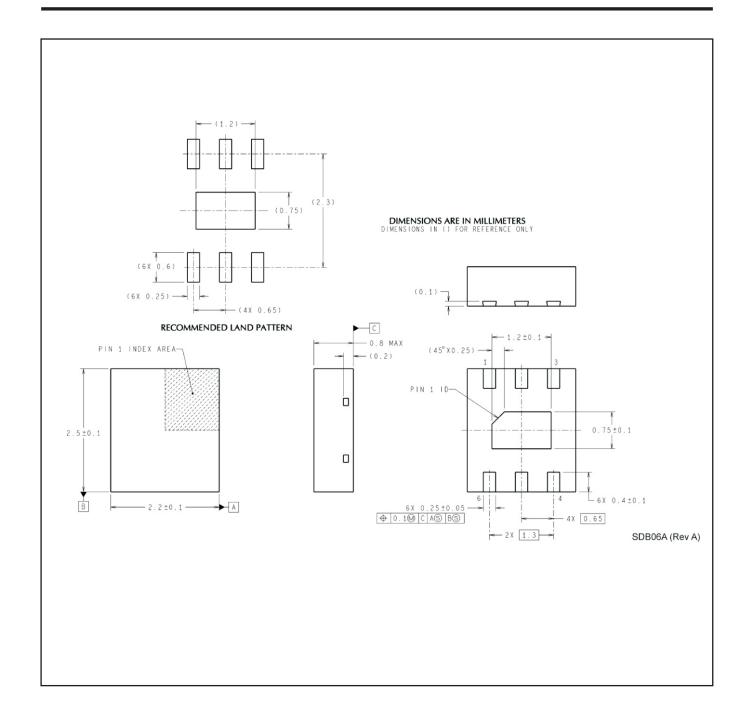
4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.





NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

B. This drawing is subject to change without notice.



#### IMPORTANT NOTICE

Texas Instruments Incorporated (TI) reserves the right to make corrections, enhancements, improvements and other changes to its semiconductor products and services per JESD46, latest issue, and to discontinue any product or service per JESD48, latest issue. Buyers should obtain the latest relevant information before placing orders and should verify that such information is current and complete.

TI's published terms of sale for semiconductor products (http://www.ti.com/sc/docs/stdterms.htm) apply to the sale of packaged integrated circuit products that TI has qualified and released to market. Additional terms may apply to the use or sale of other types of TI products and services.

Reproduction of significant portions of TI information in TI data sheets is permissible only if reproduction is without alteration and is accompanied by all associated warranties, conditions, limitations, and notices. TI is not responsible or liable for such reproduced documentation. Information of third parties may be subject to additional restrictions. Resale of TI products or services with statements different from or beyond the parameters stated by TI for that product or service voids all express and any implied warranties for the associated TI product or service and is an unfair and deceptive business practice. TI is not responsible or liable for any such statements.

Buyers and others who are developing systems that incorporate TI products (collectively, "Designers") understand and agree that Designers remain responsible for using their independent analysis, evaluation and judgment in designing their applications and that Designers have full and exclusive responsibility to assure the safety of Designers' applications and compliance of their applications (and of all TI products used in or for Designers' applications) with all applicable regulations, laws and other applicable requirements. Designer represents that, with respect to their applications, Designer has all the necessary expertise to create and implement safeguards that (1) anticipate dangerous consequences of failures, (2) monitor failures and their consequences, and (3) lessen the likelihood of failures that might cause harm and take appropriate actions. Designer agrees that prior to using or distributing any applications that include TI products, Designer will thoroughly test such applications and the functionality of such TI products as used in such applications.

TI's provision of technical, application or other design advice, quality characterization, reliability data or other services or information, including, but not limited to, reference designs and materials relating to evaluation modules, (collectively, "TI Resources") are intended to assist designers who are developing applications that incorporate TI products; by downloading, accessing or using TI Resources in any way, Designer (individually or, if Designer is acting on behalf of a company, Designer's company) agrees to use any particular TI Resource solely for this purpose and subject to the terms of this Notice.

TI's provision of TI Resources does not expand or otherwise alter TI's applicable published warranties or warranty disclaimers for TI products, and no additional obligations or liabilities arise from TI providing such TI Resources. TI reserves the right to make corrections, enhancements, improvements and other changes to its TI Resources. TI has not conducted any testing other than that specifically described in the published documentation for a particular TI Resource.

Designer is authorized to use, copy and modify any individual TI Resource only in connection with the development of applications that include the TI product(s) identified in such TI Resource. NO OTHER LICENSE, EXPRESS OR IMPLIED, BY ESTOPPEL OR OTHERWISE TO ANY OTHER TI INTELLECTUAL PROPERTY RIGHT, AND NO LICENSE TO ANY TECHNOLOGY OR INTELLECTUAL PROPERTY RIGHT OF TI OR ANY THIRD PARTY IS GRANTED HEREIN, including but not limited to any patent right, copyright, mask work right, or other intellectual property right relating to any combination, machine, or process in which TI products or services are used. Information regarding or referencing third-party products or services does not constitute a license to use such products or services, or a warranty or endorsement thereof. Use of TI Resources may require a license from a third party under the patents or other intellectual property of the third party, or a license from TI under the patents or other intellectual property of TI.

TI RESOURCES ARE PROVIDED "AS IS" AND WITH ALL FAULTS. TI DISCLAIMS ALL OTHER WARRANTIES OR REPRESENTATIONS, EXPRESS OR IMPLIED, REGARDING RESOURCES OR USE THEREOF, INCLUDING BUT NOT LIMITED TO ACCURACY OR COMPLETENESS, TITLE, ANY EPIDEMIC FAILURE WARRANTY AND ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, AND NON-INFRINGEMENT OF ANY THIRD PARTY INTELLECTUAL PROPERTY RIGHTS. TI SHALL NOT BE LIABLE FOR AND SHALL NOT DEFEND OR INDEMNIFY DESIGNER AGAINST ANY CLAIM, INCLUDING BUT NOT LIMITED TO ANY INFRINGEMENT CLAIM THAT RELATES TO OR IS BASED ON ANY COMBINATION OF PRODUCTS EVEN IF DESCRIBED IN TI RESOURCES OR OTHERWISE. IN NO EVENT SHALL TI BE LIABLE FOR ANY ACTUAL, DIRECT, SPECIAL, COLLATERAL, INDIRECT, PUNITIVE, INCIDENTAL, CONSEQUENTIAL OR EXEMPLARY DAMAGES IN CONNECTION WITH OR ARISING OUT OF TI RESOURCES OR USE THEREOF, AND REGARDLESS OF WHETHER TI HAS BEEN ADVISED OF THE POSSIBILITY OF SUCH DAMAGES.

Unless TI has explicitly designated an individual product as meeting the requirements of a particular industry standard (e.g., ISO/TS 16949 and ISO 26262), TI is not responsible for any failure to meet such industry standard requirements.

Where TI specifically promotes products as facilitating functional safety or as compliant with industry functional safety standards, such products are intended to help enable customers to design and create their own applications that meet applicable functional safety standards and requirements. Using products in an application does not by itself establish any safety features in the application. Designers must ensure compliance with safety-related requirements and standards applicable to their applications. Designer may not use any TI products in life-critical medical equipment unless authorized officers of the parties have executed a special contract specifically governing such use. Life-critical medical equipment is medical equipment where failure of such equipment would cause serious bodily injury or death (e.g., life support, pacemakers, defibrillators, heart pumps, neurostimulators, and implantables). Such equipment includes, without limitation, all medical devices identified by the U.S. Food and Drug Administration as Class III devices and equivalent classifications outside the U.S.

TI may expressly designate certain products as completing a particular qualification (e.g., Q100, Military Grade, or Enhanced Product). Designers agree that it has the necessary expertise to select the product with the appropriate qualification designation for their applications and that proper product selection is at Designers' own risk. Designers are solely responsible for compliance with all legal and regulatory requirements in connection with such selection.

Designer will fully indemnify TI and its representatives against any damages, costs, losses, and/or liabilities arising out of Designer's non-compliance with the terms and provisions of this Notice.