

# Wide-input Voltage CC/CV Step-Down DC/DC Converter

## **FEATURES**

- High Efficiency: Up to 95%@5V
- . 50V Input Voltage Surge
- Wide input voltage: 5V ~ 40V
- Up to 5A Output Current
- Output Voltage up to 12V
- · 200kHz Switching Frequency
- ±6% CC Accuracy
- Compensation of Input /Output Voltage Change
- . Independent of inductance and Inductor DCR
- ±2% Feedback Voltage Accuracy
- . Advanced Feature Set
- . Integrated Soft Start
- . Thermal Shutdown
- · Secondary Cycle-by-Cycle Current Limit
- . EMI Consideration
- ESOP8 Package

### **GENERAL DESCRIPTION**

TMI3494 is a wide input voltage, high efficiency Active CC step-down DC/DC converter that operates in either CV (Constant Output Voltage) mode or CC (Constant Output Current) mode. TMI3494 provides up to 5A output current at 200kHz, Integrated 40mΩ Power MOS, Advanced features include UVLO,

## **APPLICATIONS**

- Car Charger/ Adaptor
- Rechargeable Portable Devices

Thermal Shutdown, Soft Start, OVP.

General-Purpose CC/CV Supply

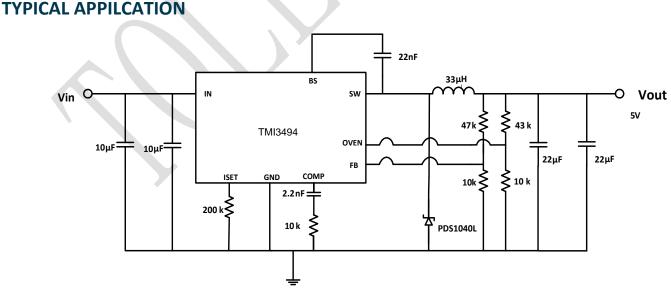


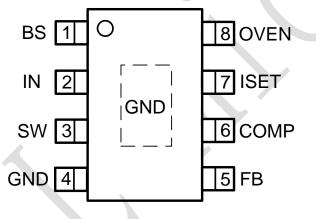
Figure 1. Basic Application Circuit



# ABSOLUTE MAXIMUM RATINGS (Note 1)

Parameter	Value	Unit
Input Supply Voltage, SW	-0.3~50	V
BS to SW Voltage	-0.3~6.0	V
All Pins Voltage	-0.3~6.0	V
Storage Temperature Range	-50~150	°C
Junction Temperature (Note2)	-40~150	°C
Package Thermal Resistance	46	°C/W

# PACKAGE/ORDER INFORMATION



ESOP8

Top Mark: T3494YYXXX (T3494: Device Code, YYXXX: Inside Code)

Part Number	Package	Top mark	Quantity/ Reel
TMI3494	ESOP8	T3494YYXXX	3000





## **PIN FUNCTIONS**

Pin	Name	Function
1	BS	High side Gate Driver Bootstrap pin, Provide supply to high-side LDMOS Gate Driver.
1	CO.	Connect a 22nF capacitor between BS and SW
2	IN	Power Input pin
3	SW	Switch Pin, Connect to external Inductor
4	GND	Ground Pin
5	FB	Feedback Pin
6	COMP	External Compensation Pin
7	ISET	Limit Current Set Pin
		OV (output over voltage) threshold setting pin and enable Input. Must Connect to a
8	OVEN	resistor divider between VOUT and GND to set the output over voltage threshold and
0	OVEN	achieve automatic startup. Drive OVEN logic high to turn off the regulator. Don't leave
		OVEN pin floating.
9	GND	Ground (Exposed pad)

## **ESD RATING**

Items Description		Value	Unit
V <sub>ESD</sub>	Human Body Model for all pins	±2000	V

### JEDEC specification JS-001

## RECOMMENDED OPERATING CONDITIONS

Items	Description	Min	Max	Unit
Voltage Range	V <sub>IN</sub>	5	40	V



# **ELECTRICAL CHARACTERISTICS (Note 3)**

### (V<sub>IN</sub>=12V, V<sub>OUT</sub>=5V, T<sub>A</sub> = 25°C, unless otherwise noted.)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input Voltage Range	V <sub>IN</sub>		5		40	V
Input Voltage Surge	V <sub>IN</sub>				50	V
Under Voltage Lockout	V <sub>UVLO</sub>	V <sub>IN</sub> rising		4.4		V
UVLO Hysteresis	V <sub>UVLO_HY</sub>			0.3		V
Input Supply Current	l <sub>iN</sub>	No load, V <sub>FB</sub> >0.92V		1		mA
Shutdown Current	I <sub>SD</sub>	OVEN = 0V		15		μA
Feedback Threshold Voltage	V <sub>FBTH</sub>		902	920	938	mV
FB Pin input current	I <sub>FB</sub>		-50		50	nA
OVEN Enable Threshold voltage	Voventh				0.4	V
OVEN Shutdown Threshold voltage	V <sub>OVENSD</sub>		2			V
Input OVP Voltage	VINOVP			41		V
Soft start Time	T <sub>SST</sub>			4.6		ms
Current limit cycle-by-cycle	ILIM_MAX	Duty=0.5		7.5		А
SW leakage	I <sub>SW_LEAK</sub>				10	μΑ
ISET Pull Up Current	V <sub>ISET</sub>			10		μA
Switch On-Resistance (high side)	Rdsonh	By design		40		mΩ
Oscillator Frequency	Fosc			200		kHz
Short circuit Frequency	F <sub>sc</sub>	V <sub>FB</sub> =0V		40		kHz
Minimum Turn-on Time	T <sub>ON_MIN</sub>			200		ns
Maximum Duty-cycle	D <sub>MAX</sub>		85			%
Thermal Shutdown Threshold	T <sub>SDN</sub>			155		°C
Thermal Shutdown Hysteresis	T <sub>SDN_HY</sub>			20		°C

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2:  $T_J$  is calculated from the ambient temperature  $T_A$  and power dissipation  $P_D$  according to the following formula:  $T_J = T_A + (P_D) \times (250^{\circ}C/W)$ .

Note 3: 100% production test at +25°C. Specifications over the temperature range are guaranteed by design and characterization.

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Note 4: Dynamic supply current is higher due to the gate charge being delivered at the switching frequency

## FUNCTIONAL BLOCK DIAGRAM

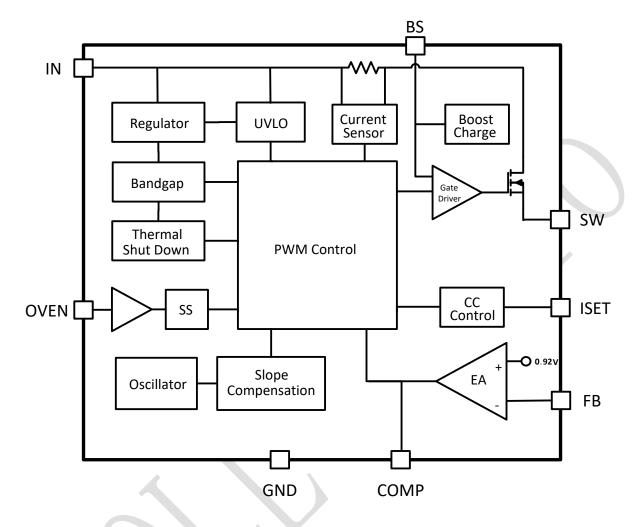


Figure 2. TMI3494 Block Diagram



## **FUNCTION DESCRIPTION**

#### **CV/CC Loop Regulation**

As seen in Functional Block Diagram, theTMI3494 is a peak current mode pulse width modulation (PWM) converter with CC and CV control. The converter operates as follows:

A switching cycle starts when the rising edge of the Oscillator clock output causes the High-Side Power Switch to turn on and the Low-Side Power Switch to turn off. With the SW side of the inductor now connected to IN, the inductor current ramps up to store energy in the magnetic field. The inductor current level is measured by the Current Sense Amplifier and added to the Oscillator ramp signal. If the resulting summation is higher than the COMP voltage, the output of the PWM Comparator goes high. When this happens or when Oscillator clock output goes low, the High-Side Power Switch turns off.

At this point, the SW side of the inductor swings to a diode voltage below ground, causing the inductor current to decrease and magnetic energy to be transferred to output. This state continues until the cycle starts again. The High-Side Power Switch is driven by logic using HSB as the positive rail. This pin is charged to VSW + 5V when the Low-Side Power Switch turns on. The COMP voltage is the integration of the error between FB input and the internal 0.92V reference. If FB is lower than the reference voltage, COMP tends to go higher to increase current to the output. Output current will increase until it reaches the CC limit set by the ISET resistor. At this point, the device will transition from regulating output voltage to regulating output current, and the output voltage will drop with increasing load.

The Oscillator normally switches at 200kHz. However, if FB voltage is less than 0.6V, then the switching frequency decreases until it reaches a typical value of 40kHz at  $V_{FB} = 0.15V$ .

#### **Over Voltage Protection**

The thresholds of input OVP circuit include are typical 41V. Once the input voltage is higher than the threshold, the high-side MOSFET is turned off. When the input voltage drops lower than the threshold, the high-side MOSFET will be enabled again.

#### **Thermal Shutdown**

The TMI3494 disables switching when its junction temperature exceeds 155°C and resumes when the temperature has dropped by 20°C.

#### **Output Voltage Setting**

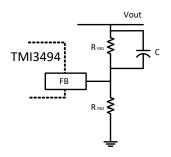


Figure 3. Output Voltage Setting

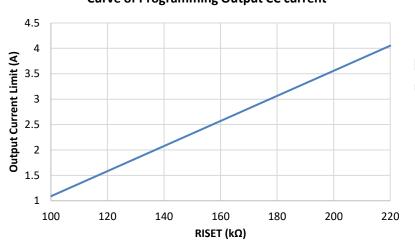


Figure 3 shows the connections for setting the output voltage. Select the proper ratio of the two feedback resistors  $R_{FB1}$  and  $R_{FB2}$  based on the output voltage. Adding a capacitor in parallel with  $R_{FB1}$  helps the system stability. Typically, use  $R_{FB2} \approx 10$ k $\Omega$  and determine  $R_{FB1}$  from the following equation:

$$R_{FB1} = R_{FB2} \left( \frac{V_{OUT}}{0.92V} - 1 \right)$$

#### **CC Current Setting**

TMI3494 constant current value is set by a resistor connected between the ISET pin and GND. The output current threshold is proportional to the ISET pin voltage. To determine the proper resistor value for a desired output current threshold, please refer to Figure 4 below.



#### **Curve of Programming Output CC current**

Figure 4. Curve of Programming Output CC Current

#### CC Loop Stability

The constant-current control loop is internally compensated over the 1500mA -5000mA output range. No additional external compensation is required to stabilize the CC current.

### **EMI Consideration**

Since parasitic inductance and capacitance effects in PCB circuitry would cause a spike voltage on SW node when high-side MOSFET is turned on/off, this spike voltage on SW may impact on EMI performance in the system. In order to enhance EMI performance, there are two methods to suppress the spike voltage. One is to place an RC snubber between SW and GND and make them as close as possible to the high-side MOSFET's source and low-side MOSFET's drain. Another method is to add a resistor in series with the bootstrap capacitor C3. But this method will decrease the driving capability to the high-side MOSFET. It is strongly recommended to reserve the RC snubber during PCB layout for EMI improvement. Moreover, reducing the PHASE trace area and keeping the main power in a small loop will be helpful on EMI performance.





#### PC Board Layout Guidance

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the IC.

1) Arrange the power components to reduce the AC loop size consisting of  $C_{IN}$ , IN pin, SW pin and the schottky diode.

2) Place input decoupling ceramic capacitor  $C_{IN}$  as close to IN pin as possible.  $C_{IN}$  is connected power GND with vias or short and wide path.

3) Return FB, COMP and ISET to signal GND pin, and connect the signal GND to power GND at a single point for best noise immunity. Connect exposed pad to power ground copper area with copper and vias.

4) Use copper plane for power GND for best heat dissipation and noise immunity.

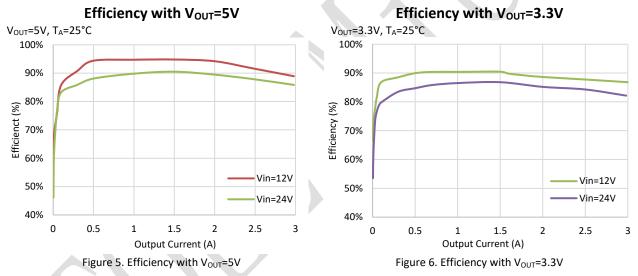
5) Place feedback resistor close to FB pin.

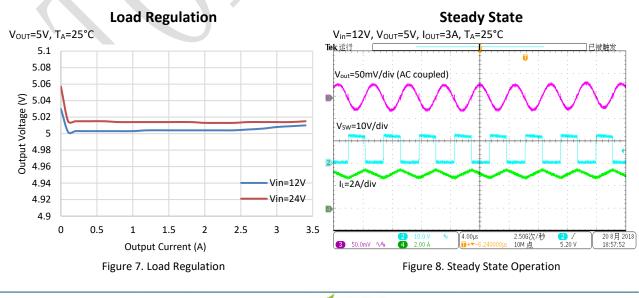
6) Use short trace connecting HSB-CHSB-SW loop

7) Place Schottky diode as close to GND pin as possible.

## **TYPICAL PERFORMANCE CHARACTERISTICS**

#### (V<sub>IN</sub>=12V, V<sub>OUT</sub>=5V, C<sub>IN</sub>=10μFx2, C<sub>OUT</sub>=22μFx2, L1=33μH, T<sub>A</sub>=25°C, unless otherwise noted.)

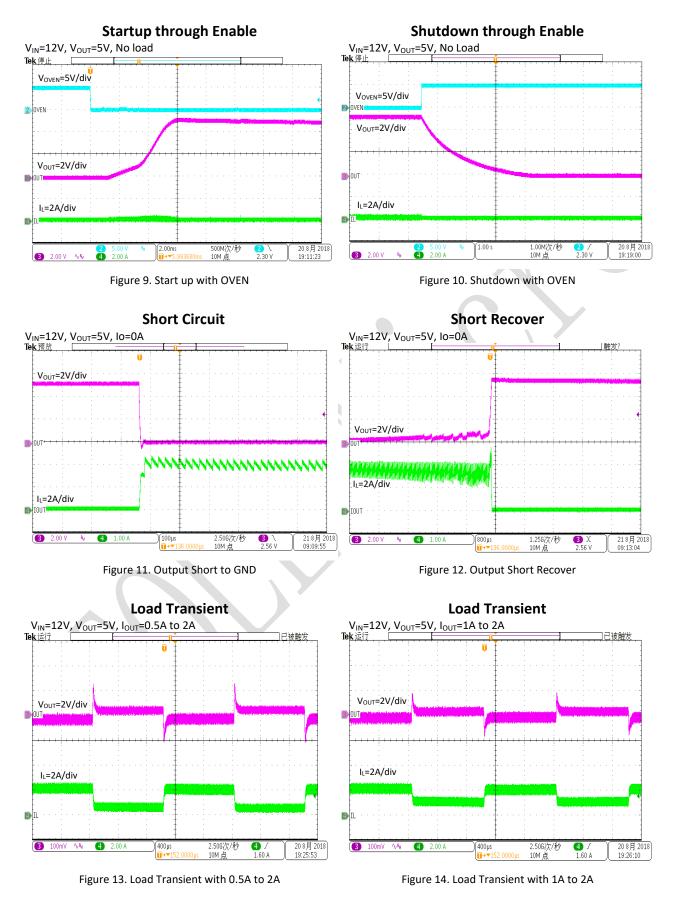




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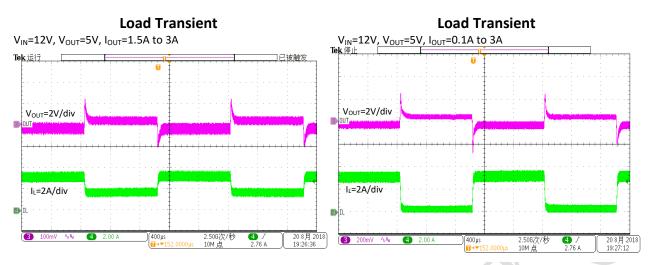


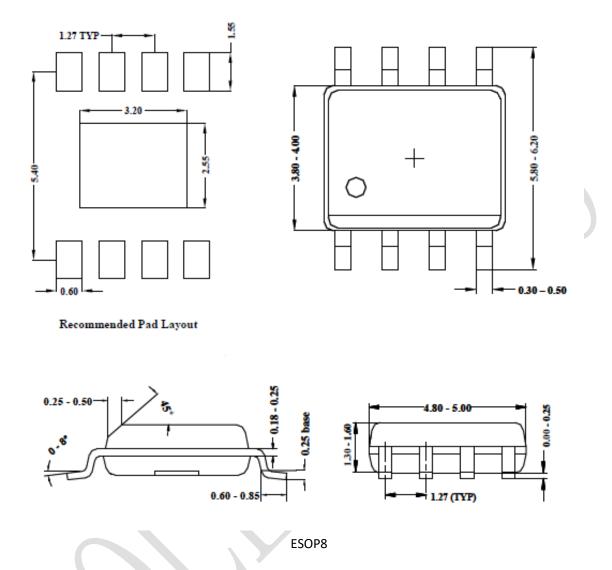
Figure 15. Load Transient with 1.5A to 3A

Figure 16. Load Transient with 0.1A to 3A

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## PACKAGE INFORMATION



#### Note:

- 1) All dimensions are in millimeters.
- 2) Package length does not include mold flash, protrusion or gate burr.
- 3) Package width does not include interlead flash or protrusion.
- 4) Lead popularity (bottom of leads after forming) shall be 0.10 millimeters max.
- 5) Pin 1 is lower left pin when reading top mark from left to right.