# *MPQ18024*



100V, 4A, High-Frequency, Half-Bridge Gate Driver AEC-Q100 Qualified

#### DESCRIPTION

The MPQ18024 is a high-frequency, 100V, halfbridge, N-channel power MOSFET driver. Its low-side and high-side driver channels are controlled independently and matched with less than 5ns of time delay. Under-voltage lockout (UVLO) on the high-side and low-side supplies force their outputs low in the case of an insufficient supply. The integrated bootstrap diode reduces the external component count.

#### **FEATURES**

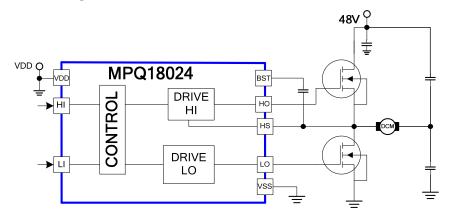
- Guaranteed Industrial / Automotive **Temperature Range Limits**
- Drives an N-Channel MOSFET Half-Bridge
- 100V V<sub>BST</sub> Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Drive Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns of Rise Time and 12ns of Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 160µA
- UVLO for both High-Side and Low-Side
- Available in a SOIC-8E Package
- Available in AEC-Q100 Qualified Grade 1

#### **APPLICATIONS**

- Car DC/DC Power Systems
- Half-Bridge Motor Drivers

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#### TYPICAL APPLICATION





### ORDERING INFORMATION

Part Number*	Package	Top Marking
MPQ18024HN-AEC1	SOIC-8 EP	See Below

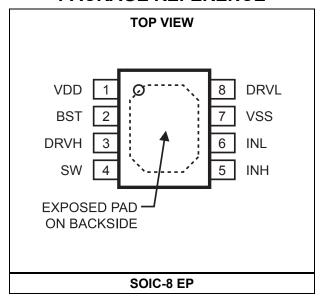
<sup>\*</sup> For Tape & Reel, add suffix -Z (e.g. MPQ18024HN-AEC1-Z).

### **TOP MARKING**

MP18024 LLLLLLL MPSYWW

MP18024: Part number LLLLLLL: Lot number MPS: MPS prefix Y: Year code WW: Week code

### **PACKAGE REFERENCE**





### **PIN FUNCTIONS**

Pin#	Name	Description
1	VDD	<b>Supply input.</b> VDD supplies power to all of the internal circuitries. Place a decoupling capacitor to ground close to VDD to ensure a stable and clean supply.
2	BST	<b>Bootstrap.</b> BST is the positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between BST and SW.
3	DRVH	Floating driver output.
4	SW	Switching node.
5	INH	Control signal input for the floating driver.
6	INL	Control signal input for the low side driver.
7	VSS	Chira arrayand. Connect the avanced and to VCC for avance the array on a setting
Exposed Pad		Chip ground. Connect the exposed pad to VSS for proper thermal operation.
8 DRVL		Low-side driver output.

ABSOLUTE MAXIMUM RATINGS (1) Supply voltage (V <sub>DD</sub> )0.3V to 18V SW voltage (V <sub>SW</sub> )5.0V to 110V BST voltage (V <sub>BST</sub> )0.3V to 110V BST to SW0.3V to 18V DRVH to SW0.3V to (BST - SW) + 0.3V
DRVL to VSS0.3V to (VDD + 0.3V) All other pins0.3V to (V <sub>DD</sub> + 0.3V)
CDM rating (AEC-Q100-011C1) All pins
Lead temperature
<b>Recommended Operating Conditions</b> <sup>(3)</sup> Supply voltage (V <sub>DD</sub> )
(-10V / <100ns) to 100V - VDD SW slew rate<50V/ns

Operating junction temp. (T<sub>J</sub>).....-40°C to 125°C

Thermal Resistance (4)	$oldsymbol{ heta}_{JA}$	$\boldsymbol{\theta}_{JC}$	
SOIC-8 EP	50	12	.°C/W

#### NOTES:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub>(MAX), the junction-to-ambient thermal resistance θ<sub>JA</sub>, and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub>(MAX)=(T<sub>J</sub>(MAX)-T<sub>A</sub>)/θ<sub>JA</sub>. Exceeding the maximum allowable power dissipation produces an excessive die temperature, causing the regulator to go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- circuitry protects the device from permanent damage.

  3) The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



# **ELECTRICAL CHARACTERISTICS**

 $V_{DD}$  =  $V_{BST}$  -  $V_{SW}$  = 12V,  $V_{SS}$  =  $V_{SW}$  = 0V, no load at DRVH and DRVL,  $T_J$  = -40°C to +125°C, typical values tested at  $T_J$  = +25°C, unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply Currents	-		•	·	•	•
VDD quiescent current	I <sub>DDQ</sub>	INL = INH = 0		120	160	μA
VDD operating current	I <sub>DDO</sub>	fsw = 500kHz		9		mA
Floating driver quiescent				70	400	
current	I <sub>BSTQ</sub>	INL = INH = 0		70	100	μA
Floating driver operating	Івѕто	fsw = 500kHz		8.5		mA
current Leakage current	I <sub>LK</sub>	BST = SW = 100V		0.05	2.5	μA
Inputs	ILK	BOT		0.00	2.0	μΛ
INL/INH high				2.2	2.6	V
INL/INH low			1	1.5	2.0	V
INL/INH internal pull-down						
resistance	Rin			185		kΩ
Under-Voltage Protection (UV	P)		L.	1	ı	
VDD rising threshold	V <sub>DDR</sub>		8.1	8.5	8.9	V
VDD hysteresis	V <sub>DDH</sub>			0.5		V
(BST - SW) rising threshold	V <sub>BSTR</sub>		6.8	7.4	8	V
(BST - SW) hysteresis	V <sub>BSTH</sub>			0.55		V
Bootstrap Diode			ľ	1	I	· ·
Bootstrap diode VF @ 100µA	$V_{F1}$			0.5		V
Bootstrap diode VF @ 100mA	V <sub>F2</sub>			0.95		V
Bootstrap diode dynamic R	$R_D$	@ 100mA		2.3		Ω
Low-Side Gate Driver						1
Low-level output voltage	Voll	I <sub>O</sub> = 100mA		0.08		V
High-level output voltage to rail	Vohl	I <sub>O</sub> = -100mA		0.23		V
Deals will the autrent (5)	_	$V_{DRVL} = 0V$ , $V_{DD} = 12V$		3		Α
Peak pull-up current (5)	I <sub>OHL</sub>	$V_{DRVL} = 0V$ , $V_{DD} = 16V$		4.7		Α
Dook pull down ourrent (5)	I <sub>OLL</sub>	$V_{DRVL} = V_{DD} = 12V$		4.5		Α
Peak pull-down current (5)		$V_{DRVL} = V_{DD} = 16V$		6		Α
Floating Gate Driver						
Low-level output voltage	$V_{OLH}$	I <sub>O</sub> = 100mA		0.08		V
High-level output voltage to rail	Vонн	I <sub>O</sub> = -100mA		0.23		V
Peak pull-up current (5)		$V_{DRVH} = 0V$ , $V_{DD} = 12V$		2.6		Α
reak pull-up current (9)	Іонн	$V_{DRVH} = 0V$ , $V_{DD} = 16V$		4		Α
Peak pull-down current (5)	Іогн	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
Peak pull-down current		$V_{DRVH} = V_{DD} = 16V$		5.9		Α
Switching spec – low-side gate driver						
Turn-off propagation delay INL falling to DRVL falling	$T_DLFF$			20		ns
Turn-on propagation delay INL rising to DRVL rising	T <sub>DLRR</sub>			20		
DRVL rise time		C <sub>L</sub> = 2.2nF		15		ns
DRVL fall time		C <sub>L</sub> = 2.2nF		9		ns



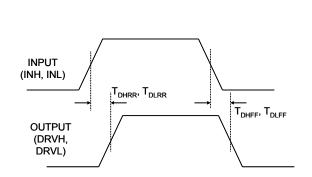
# **ELECTRICAL CHARACTERISTICS** (continued)

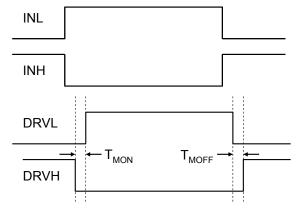
 $V_{DD} = V_{BST} - V_{SW} = 12V$ ,  $V_{SS} = V_{SW} = 0V$ , no load at DRVH and DRVL,  $T_J = -40^{\circ}C$  to +125°C, typical values tested at  $T_J = +25^{\circ}C$ , unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Switching Spec – Floating Gate Driver						
Turn-off propagation delay INL falling to DRVH falling	$T_{DHFF}$			20		ns
Turn-on propagation delay INL rising to DRVH rising	$T_{DHRR}$			20		ns
DRVH rise time		C <sub>L</sub> = 2.2nF		15		ns
DRVH fall time		C <sub>L</sub> = 2.2nF		12		ns
Switching Spec – Matching						
Floating driver turn-off to low- side drive turn-on (5)	T <sub>MON</sub>			1	5	ns
Low-side driver turn-off to floating driver turn-on (5)	T <sub>MOFF</sub>			1	5	ns
Minimum input pulse width that changes the output	T <sub>PW</sub>				50 <sup>(5)</sup>	ns
Bootstrap diode turn-on or turn-off time	T <sub>BS</sub>			10 (5)		ns
Thermal shutdown (5)				170		°C
Thermal shutdown hysteresis				25		°C

#### NOTE:

### **TIMING DIAGRAM**





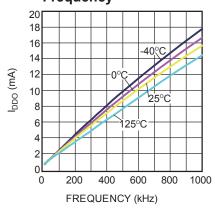
<sup>5)</sup> Guaranteed by design.



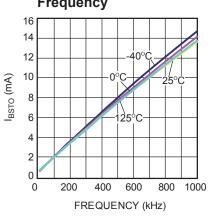
#### TYPICAL PERFORMANCE CHARACTERISTICS

 $V_{DD}$  = 12V,  $V_{SS}$  =  $V_{SW}$  = 0V,  $T_A$  = +25°C, unless otherwise noted.

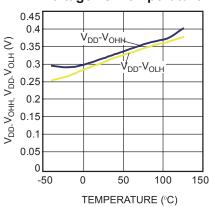
I<sub>DDO</sub> Operation Current vs. Frequency



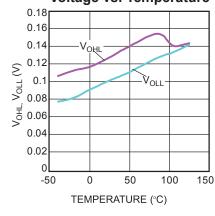
I<sub>BSTO</sub> Operation Current vs. Frequency



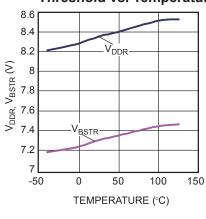
High-Level Output Voltage vs. Temperature



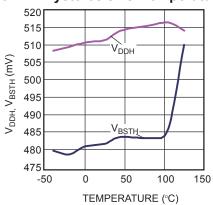
Low-Level Output Voltage vs. Temperature



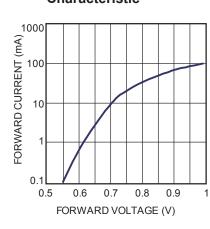
Under-Voltage Lockout
Threshold vs. Temperature



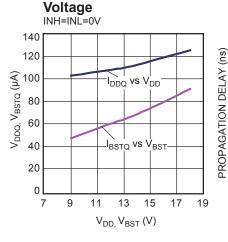
**Under-Voltage Lockout Hysteresis vs. Temperature** 



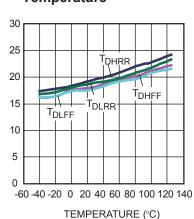
**Bootstrap Diode I-V Characteristic** 



Quiescent Current vs.



Propagation Delay vs. Temperature

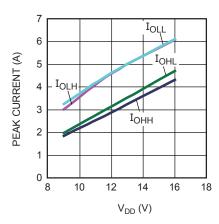




### TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $V_{DD}$  = 12V,  $V_{SS}$  =  $V_{SW}$  = 0V,  $T_A$  = +25°C, unless otherwise noted.

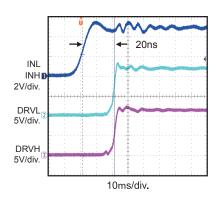
Peak Current vs. V<sub>DD</sub> Voltage

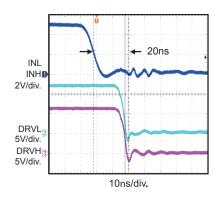


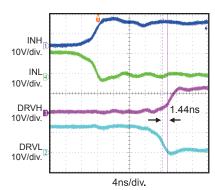
### **Turn-On Propagation Delay**

### **Turn-Off Propagation Delay**

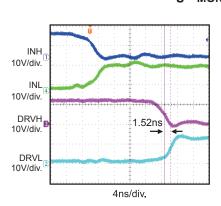
# **Gate Drive Matching T<sub>MOFF</sub>**



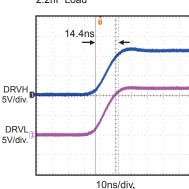




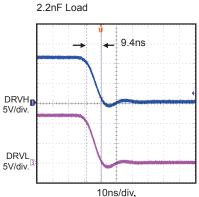
#### Gate Drive Matching T<sub>MON</sub>







#### **Drive Fall Time**





# **BLOCK DIAGRAM**

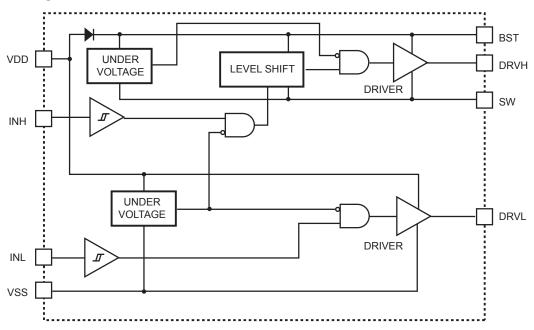


Figure 1: Functional Block Diagram

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### **APPLICATION**

The input signals of INH and INL can be controlled independently. If both INH and INL control the high-side MOSFET (HS-FET) and low-side MOSFET (LS-FET) of the same bridge, shoot through can be prevented by setting a sufficient dead time between INH and INL low, and vice versa (see Figure 2). Dead time is defined as the time interval between INH low and INL low.

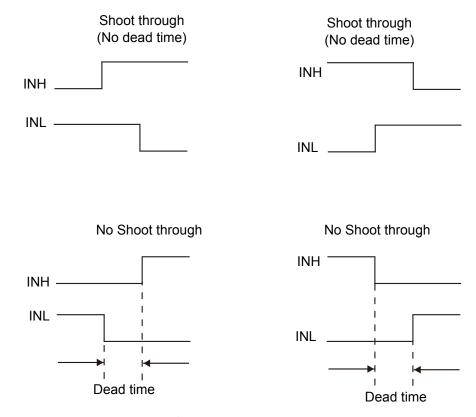


Figure 2: Shoot-Through Timing Diagram



### REFERENCE DESIGN CIRCUITS

### Half-Bridge Converter

The MPQ18024 drives the MOSFETs with alternating signals (with dead time) in a halfbridge converter topology. Because the pulsewidth modulation (PWM) controller drives INH and INL with alternating signals, the input voltage can rise as high as 100V (see Figure 3 through Figure 5).

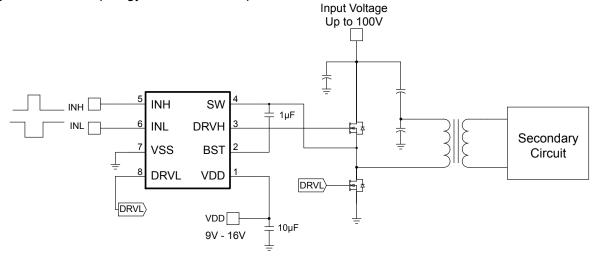


Figure 3: Half-Bridge Converter

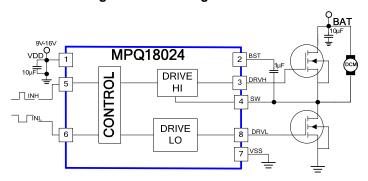


Figure 4: Half-Bridge for Unidirectional Motor

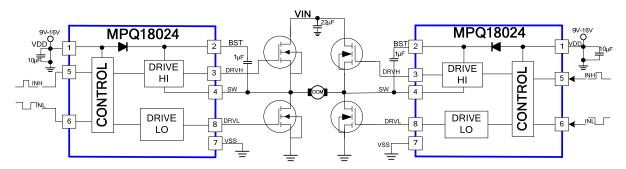
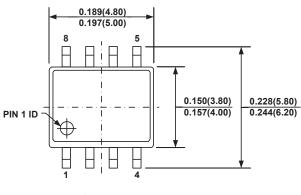


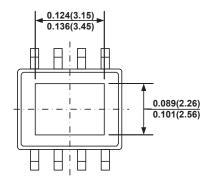
Figure 5: 2x MPQ18024 for One Bidirectional DC Motor



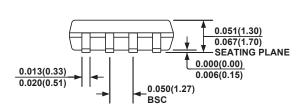
### PACKAGE INFORMATION

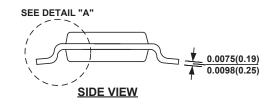
#### SOIC-8E



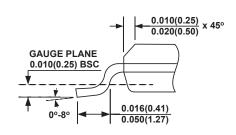


**BOTTOM VIEW TOP VIEW** 

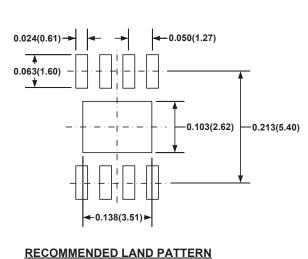




#### **FRONT VIEW**



**DETAIL "A"** 



#### NOTE:

- 1) CONTROL DIMENSION IS IN INCHES. DIMENSION IN BRACKET IS IN MILLIMETERS.
- 2) PACKAGE LENGTH DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.
- 3) PACKAGE WIDTH DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSIONS.
- 4) LEAD COPLANARITY (BOTTOM OF LEADS AFTER FORMING) SHALL BE 0.004" INCHES MAX.
- 5) DRAWING CONFORMS TO JEDEC MS-012, VARIATION BA.
- 6) DRAWING IS NOT TO SCALE.

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