

#### The Future of Analog IC Technology

# DESCRIPTION

The MPQ1924 is a high-frequency, 100V, halfbridge, N-channel, power MOSFET driver. Its low-side and high-side driver channels are independently controlled and matched with less than 5ns in time delay. Under-voltage lockout on both high-side and low-side supplies force their outputs low in case of insufficient supply. The integrated bootstrap diode reduces external component count.

## FEATURES

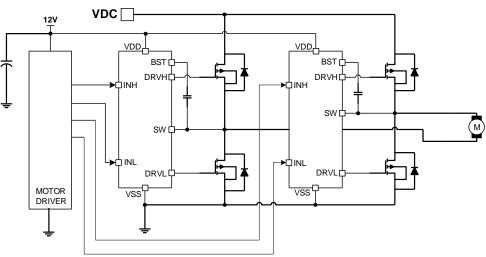
- Drives an N-Channel MOSFET Half Bridge
- 118V V<sub>BST</sub> Voltage Range
- On-Chip Bootstrap Diode
- Typical Propagation Delay of 20ns
- Gate Drive Matching of Less than 5ns
- Drives a 2.2nF Load with 15ns Rise Time and 12ns Fall Time at 12V VDD
- TTL-Compatible Input
- Quiescent Current of Less than 150μA
- UVLO for Both High Side and Low Side
- SOIC-8 Package

### APPLICATIONS

- Motor Drivers
- Telecom Half-Bridge Power Supplies
- Avionics DC-DC Converters
- Two-Switch Forward Converters
- Active-Clamp Forward Converters

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## TYPICAL APPLICATION





### **ORDERING INFORMATION**

Part Number	Package	Top Marking
MPQ1924HS*	SOIC-8	See Below

\* For Tape & Reel, add suffix –Z (e.g. MPQ1924HS–Z) For RoHS compliant packaging, add suffix –LF (e.g. MPQ1924HS–LF–Z)

# **TOP MARKING**

### MP1924

#### LLLLLLL

MPSYWW

MP1924: product code of MPQ1924HS; LLLLLLL: lot number; MPS: MPS prefix; Y: year code; WW: week code;

# PACKAGE REFERENCE

VDD 1 BST 2 DRVH 3 SW 4	0	8 DRVL 7 VSS 6 INL 5 INH
SOIC-8		



### ABSOLUTE MAXIMUM RATINGS <sup>(1)</sup>

Supply Voltage (V <sub>DD</sub> ) SW Voltage (V <sub>SW</sub> )	
BST Voltage (V <sub>BST</sub> )	0.3V to 118V
BST to SW	0.3V to 18V
DRVH to SW0.3V to (B	ST-SW) + 0.3V
DRVL to VSS0.3V to	o (VDD + 0.3V)
All Other Pins0.3V	
Continuous Power Dissipation	$(T_A = 25^{\circ}C)^{(2)}$
SOIC-8	1.3W
Junction Temperature	150°C
Lead Temperature	
Storage Temperature	

# Recommended Operating Conditions <sup>(3)</sup>

Supply Voltage V <sub>DD</sub> .	9.0V to 16.0V
SW Voltage (V <sub>SW</sub> )	
SW Slew Rate	<50V/ns
Operating Junction 1	emp. (T <sub>J</sub> )40°C to 125°C

## Thermal Resistance $^{(4)}$ $\theta_{JA}$

 $\theta_{JC}$ 

#### Notes:

- 1) Exceeding these ratings may damage the device.
- 2) The maximum allowable power dissipation is a function of the maximum junction temperature T<sub>J</sub>(MAX), the junction-to-ambient thermal resistance θ<sub>JA</sub>, and the ambient temperature T<sub>A</sub>. The maximum allowable continuous power dissipation at any ambient temperature is calculated by P<sub>D</sub>(MAX)=(T<sub>J</sub>(MAX)-T<sub>A</sub>)/ θ<sub>JA</sub>. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. Internal thermal shutdown circuitry protects the device from permanent damage.
- The device is not guaranteed to function outside of its operating conditions.
- 4) Measured on JESD51-7, 4-layer PCB.



# **ELECTRICAL CHARACTERISTICS**

 $V_{DD} = V_{BST}-V_{SW} = 12V$ ,  $V_{SS} = V_{SW} = 0V$ , No load at DRVH and DRVL,  $T_A = +25^{\circ}C$ , unless otherwise noted.

Parameter	Symbol	Condition	Min	Тур	Max	Units
Supply Currents						
VDD quiescent current	I <sub>DDQ</sub>	INL = INH = 0		100	150	μA
VDD operating current	I <sub>DDO</sub>	fsw = 500kHz		9		mA
Floating driver quiescent current	I <sub>BSTQ</sub>	INL = INH = 0		60	90	μA
Floating driver operating current	I <sub>BSTO</sub>	fsw = 500kHz		7.5		mA
Leakage current	I <sub>LK</sub>	BST = SW = 100V		0.05	1	μA
Inputs						
INL/INH High				2	2.4	V
INL/INH Low			1	1.4		V
INL/INH internal pull-down resistance	R <sub>IN</sub>			185		kΩ
Under Voltage Protection						
VDD rising threshold	V <sub>DDR</sub>		8.1	8.4	8.8	V
VDD hysteresis	V <sub>DDH</sub>			0.5		V
(BST-SW) rising threshold	V <sub>BSTR</sub>		6.9	7.3	7.7	V
(BST-SW) hysteresis	V <sub>BSTH</sub>			0.55		V
Bootstrap Diode			•			
Bootstrap diode VF @ 100µA	$V_{F1}$			0.5		V
Bootstrap diode VF @ 100mA	$V_{F2}$			0.95		V
Bootstrap diode dynamic R	R <sub>D</sub>	@ 100mA		2		Ω
Low Side Gate Driver						
Low level output voltage	V <sub>OLL</sub>	I <sub>O</sub> = 100mA		0.08		V
High level output voltage to rail	V <sub>OHL</sub>	I <sub>O</sub> = -100mA		0.23		V
Source Current <sup>(5)</sup>	I <sub>OHL</sub>	$V_{DRVL} = 0V, V_{DD} = 12V$		3		А
Source Current		$V_{DRVL} = 0V, V_{DD} = 16V$		4.7		Α
Sink Current <sup>(5)</sup>	I <sub>OLL</sub>	$V_{DRVL} = V_{DD} = 12V$		4.5		Α
		$V_{DRVL} = V_{DD} = 16V$		6		Α
Floating Gate Driver						
Low level output voltage	V <sub>OLH</sub>	I <sub>O</sub> = 100mA		0.08		V
High level output voltage to rail V <sub>OHH</sub>		I <sub>O</sub> = -100mA		0.23		V
Source Current <sup>(5)</sup>	I <sub>ОНН</sub>	$V_{DRVH} = 0V, V_{DD} = 12V$		2.6		Α
		$V_{DRVH} = 0V, V_{DD} = 16V$		4		Α
Sink Current <sup>(5)</sup>	I <sub>OLH</sub>	$V_{DRVH} = V_{DD} = 12V$		4.5		Α
		$V_{DRVH} = V_{DD} = 16V$		5.9		Α



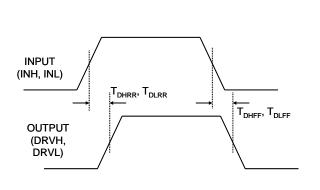
### ELECTRICAL CHARACTERISTICS (continued)

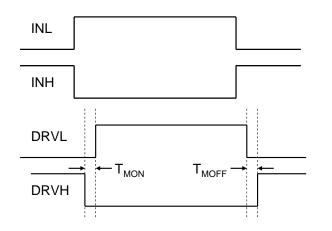
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Parameter	Symbol	Condition	Min	Тур	Max	Units	
Switching Spec Low Side Gate Driver							
Turn-off propagation delay INL falling to DRVL falling	$T_{DLFF}$			20		ns	
Turn-on propagation delay INL rising to DRVL rising	T <sub>DLRR</sub>			20			
DRVL rise time		$C_L = 2.2 nF$		15		ns	
DRVL fall time		$C_L = 2.2 nF$		9		ns	
Switching Spec Floating Gate	e Driver						
Turn-off propagation delay INH falling to DRVH falling	$T_{DHFF}$			20		ns	
Turn-on propagation delay INH rising to DRVH rising	T <sub>DHRR</sub>			20		ns	
DRVH rise time		$C_L = 2.2 nF$		15		ns	
DRVH fall time		$C_L = 2.2 nF$		12		ns	
Switching Spec Matching	Switching Spec Matching						
Floating driver turn-off to low side drive turn-on <sup>(5)</sup>	T <sub>MON</sub>			1	5	ns	
Low side driver turn-off to floating driver turn-on <sup>(5)</sup>	T <sub>MOFF</sub>			1	5	ns	
Minimum input pulse width that changes the output <sup>(5)</sup>	$T_{PW}$				50	ns	
Bootstrap diode turn-on or turn-off time <sup>(5)</sup>	T <sub>BS</sub>			10		ns	
Thermal shutdown				150		°C	
Thermal shutdown hysteresis				25		°C	

Note:

5) Guaranteed by design.









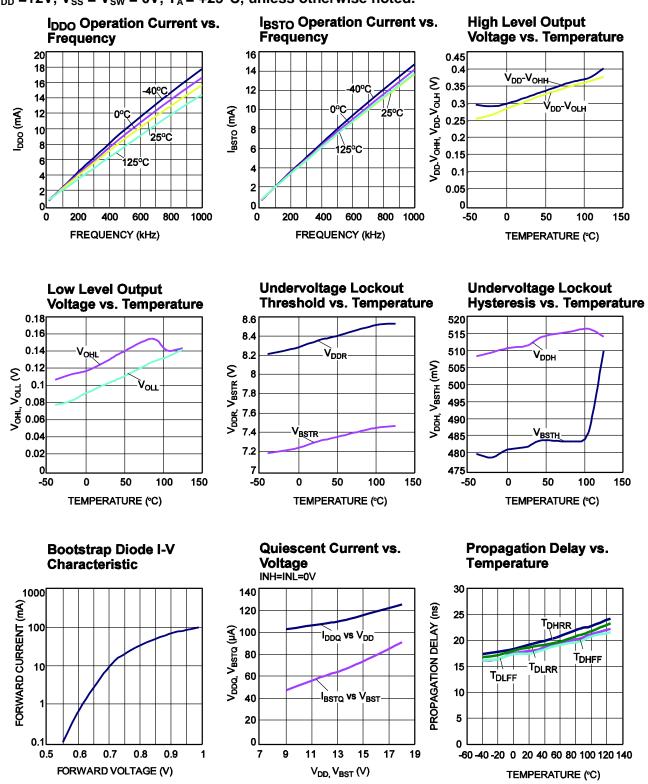
# **PIN FUNCTIONS**

SOIC-8 Pin #	Name	Description
1	VDD	Supply input. This pin supplies power to all the internal circuitry. Place a decoupling capacitor to ground close to this pin to ensure stable and clean supply.
2	BST	Bootstrap. This is the positive power supply for the internal floating high-side MOSFET driver. Connect a bypass capacitor between this pin and SW pin.
3	DRVH	Floating driver output.
4	SW	Switching node.
	NC	No connection.
5	INH	Control signal input for the floating driver.
6	INL	Control signal input for the low side driver.
7	VSS, exposed pad	Chip ground. Connect exposed pad to VSS for proper thermal operation.
8	DRVL	Low side driver output.



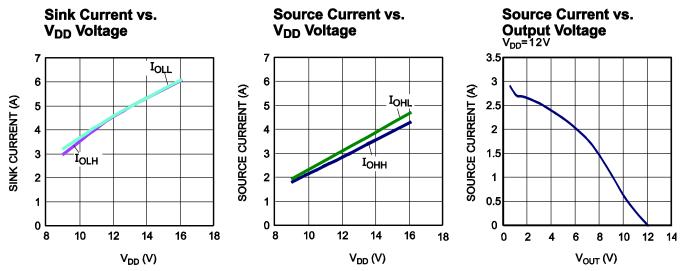
## **TYPICAL PERFORMANCE CHARACTERISTICS**

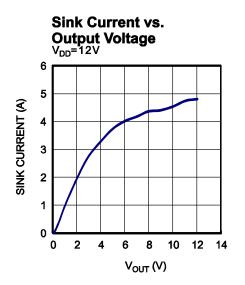
 $V_{DD}$  =12V,  $V_{SS}$  =  $V_{SW}$  = 0V,  $T_A$  = +25°C, unless otherwise noted.



## **TYPICAL PERFORMANCE CHARACTERISTICS** (continued)

 $V_{DD}$  =12V,  $V_{SS}$  =  $V_{SW}$  = 0V,  $T_A$  = +25°C, unless otherwise noted.







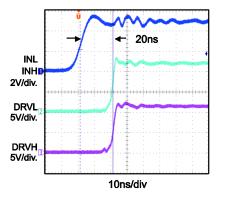
### **TYPICAL PERFORMANCE CHARACTERISTICS** (continued)

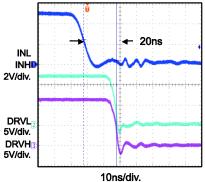
 $V_{DD}$  =12V,  $V_{SS}$  =  $V_{SW}$  = 0V,  $T_A$  = +25°C, unless otherwise noted.

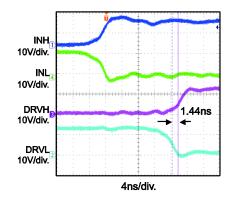
#### **Turn-on Propagation Delay**

y Turn-off Propagation Delay

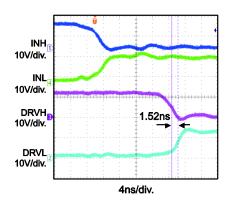
Gate Drive Matching T<sub>MOFF</sub>





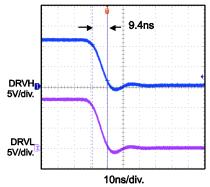


Gate Drive Matching T<sub>MON</sub>



DRVH DRVH SV/div. DRVL 14.4ns 14.4ns DRVL DRVL 10ns/div.







# **BLOCK DIAGRAM**

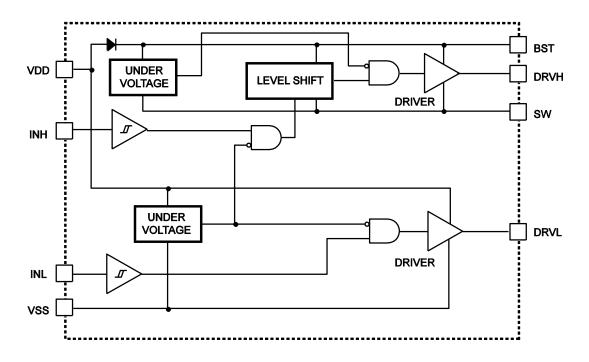
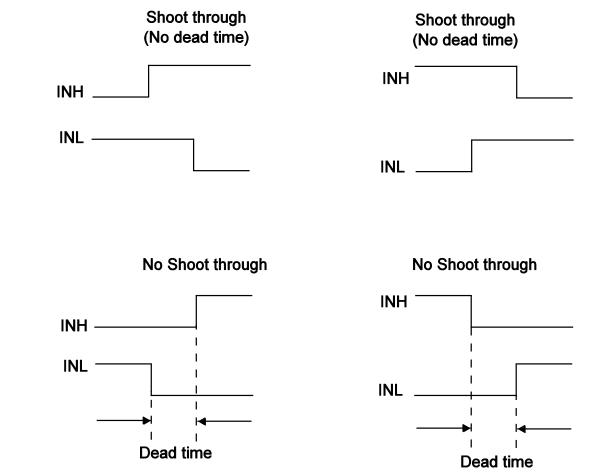


Figure 2: Function Block Diagram



## **APPLICATION**

The input signals of INH and INL can be controlled independently. If both INH and INL control the high-side MOSFET and low-side MOSFET of the same bridge, then users must avoid shoot through by setting sufficient dead time between INH and INL low, and vice versa. See Figure 3 below. Dead time is defined as the time interval between INH low and INL low.







## **REFERENCE DESIGN CIRCUITS**

#### Half Bridge Converter

The MPQ1924 drives the MOSFETS with alternating signals (with dead time) in half-bridge converter topology. Therefore, from the PWM

controller drives INH and INL with alternating signals the input voltage can go up to 100V.

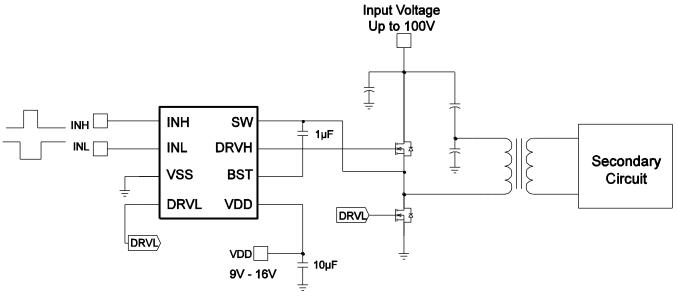


Figure 4: Half Bridge Converter

#### **Two-Switch Forward Converter**

In two-switch forward converter topology, both MOSFETs are turned on and off simultaneously. The input signal (INH and INL) comes from a PWM controller that senses the output voltage (and output current during current-mode control).

The Schottky diodes clamp the reverse swing of the power transformer and must be rated for the input voltage. The input voltage can go up to 100V.

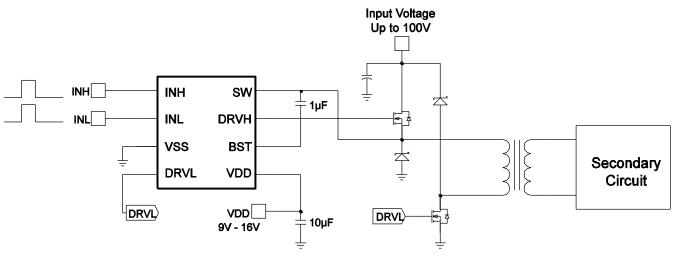


Figure 5: Two-Switch Forward Converter



#### **Active-Clamp Forward Converter**

In active-clamp forward converter topology, the MPQ1924 drives the MOSFETs with alternating signals. The high-side MOSFET, in conjunction with  $C_{reset}$ , is used to reset the power transformer in a lossless manner.

This topology lends itself well to run at duty cycles exceeding 50%. The device may not be able to run at 100V under this topology.

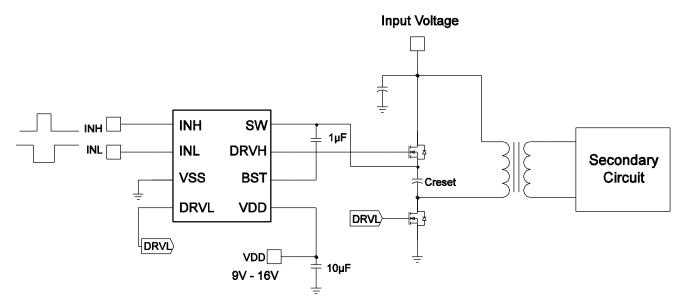
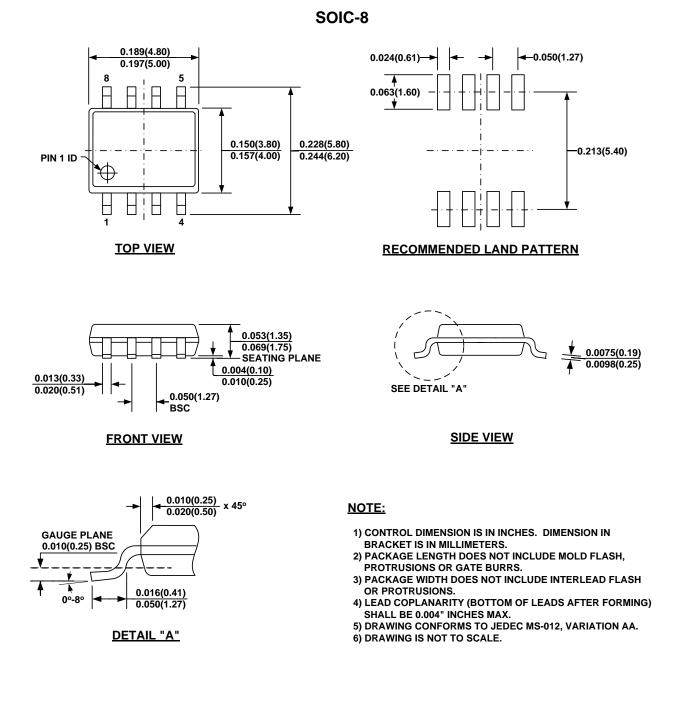


Figure 6 Active-Clamp Forward Converter



## **PACKAGE INFORMATION**



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