# Microcomputer Components

8-Bit CMOS Microcontroller

C505 C505C/C505A C505CA

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#### **8-Bit CMOS Microcontroller**

C505 C505C C505A C505CA

#### **Advance Information**

- Fully compatible to standard 8051 microcontroller
- Superset of the 8051 architecture with 8 datapointers
- Up to 20 MHz operating frequency
  - 375 ns instruction cycle time @16 MHz
  - 300 ns instruction cycle time @20 MHz (50 % duty cycle)
- On-chip program memory (with optional memory protection)
  - C505-2R/C505C-2R:
- 16k byte on-chip ROM
- C505A-4E/C505CA-4E: 32k byte on-chip OTP
- alternatively up to 64k byte external program memory
- 256 byte on-chip RAM
- On-chip XRAM
  - C505/C505C :
- 256 byte
- C505A/C505CA: 1K byte
- 32 + 2 digital I/O lines
  - Four 8-bit digital I/O ports
  - One 2-bit digital I/O port (port 4)
  - Port 1 with mixed analog/digital I/O capability

(more features on next page)

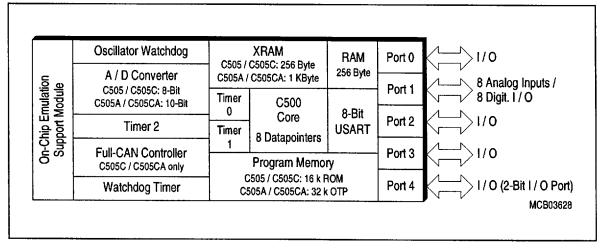


Figure 1 C505 Functional Units

#### Features (cont'd):

- Three 16-bit timers/counters
  - Timer 0 / 1 (C501 compatible)
  - Timer 2 with 4 channels for 16-bit capture/compare operation
- Full duplex serial interface with programmable baudrate generator (USART)
- Full CAN Module, version 2.0 B compliant (C505C and C505CA only)
  - 256 register/data bytes located in external data memory area
  - 1 MBaud CAN baudrate when operating frequency is equal to or above 8 MHz
  - internal CAN clock prescaler when input frequency is over 10 MHz
- On-chip A/D Converter
  - up to 8 analog inputs
  - C505/C505C : 8-bit resolution
  - C505A/C505CA: 10-bit resolution
- Twelve interrupt sources with four priority levels
- On-chip emulation support logic (Enhanced Hooks Technology TM 1)
- Programmable 15-bit watchdog timer
- · Oscillator watchdog
- · Fast power on reset
- · Power Saving Modes
  - Slow-down mode
  - Idle mode (can be combined with slow-down mode)
  - Software power-down mode with wake up capability through P3.2/INTO or P4.1/RXDC pin
- P-MQFP-44 package
- Pin configuration is compatible to C501, C504, C511/C513-family
- Temperature ranges:

SAB-C505 versions  $T_A = 0$  to 70 °C SAF-C505 versions  $T_A = -40$  to 85 °C

SAH-C505 versions  $T_A = -40$  to 110 °C (max. operating frequency: TBD) SAK-C505 versions  $T_A = -40$  to 125 °C (max. operating frequency: 12 MHz

with 50% duty cycle)

Table 1
Differences in Functionality of the C505 MCUs

Device	Internal P	rogram Memory	XRAM Size	A/D Converter	CAN Controller
	ROM	ОТР		Resolution	
C505-2RM	16 KB	_	256 B	8 Bit	_
C505-LM	-	-	256 B	8 Bit	-
C505C-2RM	16 KB	_	256 B	8 Bit	1
C505C-LM	-	_	256 B	8 Bit	1
C505A-4EM	_	32 KB	1 KB	10 Bit	_
C505CA-4EM	_	32 KB	1 KB	10 Bit	1

<sup>1) &</sup>quot;Enhanced Hooks Technology" is a trademark and patent of Metalink Corporation licensed to Siemens.

Table 2
Ordering Information

Туре	Ordering Code	Package	Description (8-Bit CMOS microcontroller)	
SAB-C505-2RM SAB-C505-LM	Q67127-DXXXX Q67127-C2057	P-MQFP-44 P-MQFP-44	with mask-programmable ROM (16K), 20 MHz for external memory (20 MHz)	
SAF-C505-2RM SAF-C505-LM	Q67127-DXXXX Q67127-C2056	P-MQFP-44 P-MQFP-44	7 - 3	
SAB-C505C-2RM SAB-C505C-LM	Q67127-DXXXX Q67127-C2029	P-MQFP-44 P-MQFP-44		
SAF-C505C-2RM SAF-C505C-LM	Q67127-DXXXX Q67127-C2030	P-MQFP-44 P-MQFP-44	Extended temperature. — 40 °C to 85 °C : with mask-progr. ROM (16K) and CAN, 20 MHz for external memory, with CAN (20 MHz)	
SAB-C505A-4EM	Q67127-C2060	P-MQFP-44	with OTP memory (32K), 20 MHz	
SAF-C505A-4EM	Q67127-C2061	P-MQFP-44	Extended temperature. – 40 °C to 85 °C : with OTP memory (32K), 20 MHz	
SAB-C505CA-4EM	Q67127-C1082	P-MQFP-44	with OTP memory (32K) and CAN, 20 MHz	
SAB-C505CA-4EM	Q67127-C2058	P-MQFP-44	Extended temperature. – 40 °C to 85 °C : with OTP memory (32K) and CAN, 20 MHz	

**Note:** The ordering number of the ROM types (DXXXX extension) is defined after program release (verification) of the customer.

Versions for the extended temperature range – 40°C to 110°C (SAH-C505) and – 40°C to 125°C (SAK-C505) are available on request.

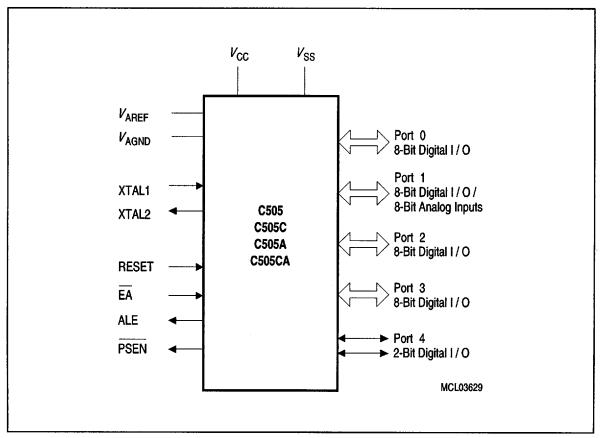


Figure 2 Logic Symbol

#### **Additional Literature**

For further information about the C505/C505C/C505A/C505CA the following literature is available:

Title	Ordering Number
C505 8-Bit CMOS Microcontroller User's Manual	B158-H7116-X-X-7600
C500 Microcontroller Family Architecture and Instruction Set User's Manual	B158-H6987-X-X-7600
C500 Microcontroller Family - Pocket Guide	B158-H6986-X-X-7600

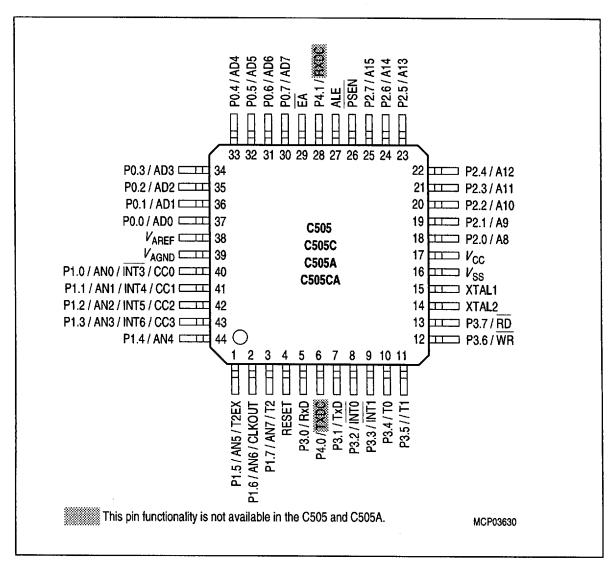


Figure 3
C505 Pin Configuration P-MQFP-44 Package (top view)

Table 3
Pin Definitions and Functions

Symbol	Pin Number	1/O *)	Function	
P1.0-P1.7	40-44,1-3	I/O	rangement. Port 1 pins of or as analog inputs of the have 1's written to them transistors and in that steputs, port 1 pins being extent (I <sub>IL</sub> , in the DC charapullup transistors. Port 1 analog inputs via the regular As secondary digital functimer, clock, capture and corresponding to a segrammed to a one (1) for	tions, port 1 contains the interrupt, d compare pins. The output latch condary function must be prothat function to operate (except for secondary functions are assigned
	40		P1.0 / AN0 / INT3 / CC0	Analog input channel 0 interrupt 3 input / capture/compare channel 0 I/O
	41		P1.1 / AN1 / INT4 / CC1	Analog input channel 1/ interrupt 4 input / capture/compare channel 1 i/O
	42		P1.2 / AN2 / INT5 / CC2	Analog input channel 2 / interrupt 5 input / capture/compare channel 2 I/O
	43		P1.3 / AN3 / INT6 / CC3	Analog input channel 3 interrupt 6 input / capture/compare channel 4 I/O
	44		P1.4 / AN4	Analog input channel 4
			P1.5 / AN5 / T2EX	Analog input channel 5 / Timer 2 external reload / trigger input
	2		P1.6 / AN6 / CLKOUT	Analog input channel 6 / system clock output
	3		P1.7 / AN7 / T2	Analog input channel 7 / counter 2 input
			Port 1 is used for the low-coverification of the C505-2	order address byte during program R and C505C-2R.

<sup>\*)</sup> I = Input O= Output

Table 3
Pin Definitions and Functions (cont'd)

Symbol	Pin Number	1/O *)	Function			
RESET	4	ı	<b>RESET</b> A high level on this pin for one machine cycle while the oscillator is running resets the device. An internal diffused resistor to $V_{\rm SS}$ permits power-on reset using only an external capacitor to $V_{\rm CC}$ .			
P3.0-P3.7	5, 7-13 5 7 8	I/O	Port 3 is an 8-bit qua arrangement. F pulled high by state can be us externally pull characteristics) The output late must be progra (except for Tx	asi-bidirectional port with internal pull-up Port 3 pins that have 1's written to them are the internal pull-up transistors and in that sed as inputs. As inputs, port 3 pins being ed low will source current (I <sub>1L</sub> , in the DC because of the internal pullup transistors. In corresponding to a secondary function mimed to a one (1) for that function to operate D and WR). The secondary functions are pins of port 3 as follows:  Receiver data input (asynch.) or data input/output (synch.) of serial interface Transmitter data output (asynch.) or clock output (synch.) of serial interface External interrupt 0 input / timer 0 gate		
	9		P3.3 / ĪNT1	control input External interrupt 1 input / timer 1 gate control input		
	10		P3.4 / T0	Timer 0 counter input		
	11		P3.5 / T1	Timer 1 counter input		
	12		P3.6 / WR	WR control output; latches the data byte from port 0 into the external data memory		
•	13		P3.7 / RD	RD control output; enables the external data memory		

<sup>\*)</sup> I = Input O= Output

Table 3
Pin Definitions and Functions (cont'd)

Symbol	Pin Number	I/O *)	Function	
P4.0 P4.1	6 28	1/0	is a 2-bit quasi-bidirectional port with internal pull-up arrangement. Port 4 pins that have 1's written to them are pulled high by the internal pull-up transistors and in that state can be used as inputs. As inputs, port 4 pins being externally pulled low will source current ( $I_{IL}$ , in the DC characteristics) because of the internal pullup transistors. The output latch corresponding to the secondary function RXDC must be programmed to a one (1) for that function to operate. The secondary functions are assigned to the two pins of port 4 as follows (C505C and C505CA only): P4.0 / TXDC Transmitter output of CAN controller P4.1 / RXDC Receiver input of CAN controller	
XTAL2	14	0	XTAL2 Output of the inverting oscillator amplifier.	
XTAL1	15	1	Input to the inverting oscillator amplifier and input to the internal clock generator circuits.  To drive the device from an external clock source, XTAL1 should be driven, while XTAL2 is left unconnected. To operate above a frequency of 16 MHz, a duty cycle of the etxernal clock signal of 50 % should be maintained.  Minimum and maximum high and low times as well as rise/fall times specified in the AC characteristics must be observed.	

<sup>\*)</sup> I = Input O= Output

Table 3
Pin Definitions and Functions (cont'd)

Symbol	Pin Number	I/O *)	Function
P2.0-P2.7	18-25	I/O	Port 2 is a an 8-bit quasi-bidirectional I/O port with internal pullup resistors. Port 2 pins that have 1's written to them are pulled high by the internal pullup resistors, and in that state can be used as inputs. As inputs, port 2 pins being externally pulled low will source current ( $I_{\rm IL}$ , in the DC characteristics) because of the internal pullup resistors. Port 2 emits the high-order address byte during fetches from external program memory and during accesses to external data memory that use 16-bit addresses (MOVX @DPTR). In this application it uses strong internal pullup transistors when issuing 1s. During accesses to external data memory that use 8-bit addresses (MOVX @Ri), port 2 issues the contents of the P2 special function register and uses only the internal pullup resistors.
PSEN	26	0	The Program Store Enable output is a control signal that enables the external program memory to the bus during external fetch operations. It is activated every three oscillator periods except during external data memory accesses. Remains high during internal program execution. This pin should not be driven during reset operation.
ALE	27	0	The Address Latch Enable output is used for latching the low-byte of the address into external memory during normal operation. It is activated every three oscillator periods except during an external data memory access. When instructions are executed from internal ROM or OTP (EA=1) the ALE generation can be disabled by bit EALE in SFR SYSCON.  ALE should not be driven during reset operation.

<sup>\*)</sup> I = Input O= Output

Table 3
Pin Definitions and Functions (cont'd)

Symbol	Pin Number	I/O *)	Function
ĒĀ	29		External Access Enable When held at high level, instructions are fetched from the internal ROM or OTP memory when the PC is less than 4000 <sub>H</sub> (C505 and C505C) or less than 8000H (C505A and C505CA). When held at low level, the C505 fetches all instructions from external program memory. EA should not be driven during reset operation. For the C505-L and the C505C-L this pin must be tied low.
P0.0-P0.7	37-30	I/O	Port 0 is an 8-bit open-drain bidirectional I/O port. Port 0 pins that have 1's written to them float, and in that state can be used as high-impendance inputs. Port 0 is also the multiplexed low-order address and data bus during accesses to external program or data memory. In this application it uses strong internal pullup transistors when issuing 1's.  Port 0 also outputs the code bytes during program verification in the C505-2R/C505C-2R. External pullup resistors are required during program verification.
$\overline{V_{AREF}}$	38	-	Reference voltage for the A/D converter.
$\overline{V_{AGND}}$	39		Reference ground for the A/D converter.
$\overline{v_{\mathtt{SS}}}$	16	_	Ground (0 V)
$V_{\sf CC}$	17	_	Power Supply (+ 5 V)

<sup>\*) | =</sup> Input

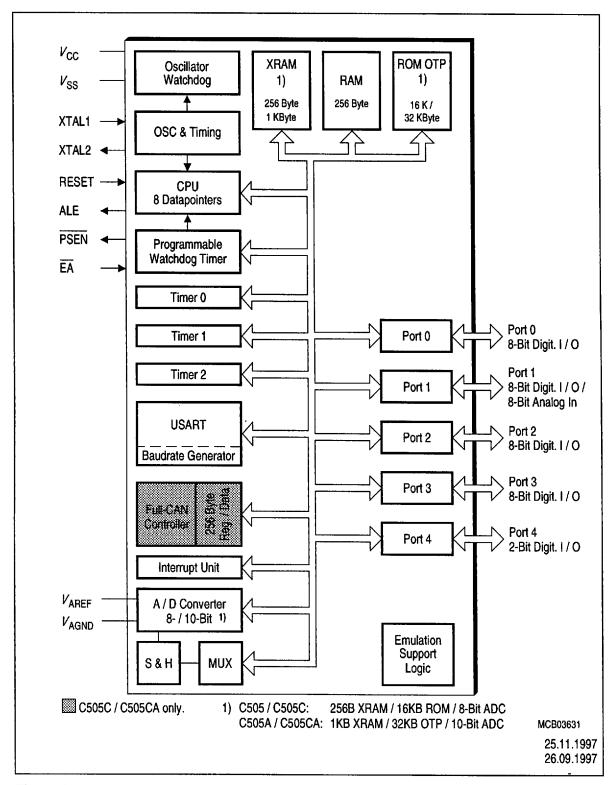


Figure 4
Block Diagram of the C505/C505C/C505A/C505CA

Reset Value: 00H

#### **CPU**

The C505 is efficient both as a controller and as an arithmetic processor. It has extensive facilities for binary and BCD arithmetic and excels in its bit-handling capabilities. Efficient use of program memory results from an instruction set consisting of 44 % one-byte, 41 % two-byte, and 15% three-byte instructions. With a 16 MHz crystal, 58% of the instructions are executed in 375 ns.

### Special Function Register PSW (Address D0<sub>H</sub>)

Bit No.	Bit No. MSB							LSB	
		• •	• •		D3 <sub>H</sub>				
D0 <sub>H</sub>	CY	AC	F0	RS1	RS0	ov	F1	Р	PSW

Bit	Function	Function				
CY	Carry FI Used by	•	instruction.			
AC		y Carry Fla	ag s which execute BCD operations.			
F0	General	Purpose F	Flag			
RS1 RS0			ect Control Bits I to select one of the four register banks.			
	RS1	RS0	Function			
	0	0	Bank 0 selected, data address 00 <sub>H</sub> -07 <sub>H</sub>			
	0	1	Bank 1 selected, data address 08 <sub>H</sub> -0F <sub>H</sub>			
	1	0	Bank 2 selected, data address 10 <sub>H</sub> -17 <sub>H</sub>			
	1	1 Bank 3 selected, data address 18 <sub>H</sub> -1F <sub>H</sub>				
ov	1	Overflow Flag Used by arithmetic instruction.				
F1	General	General Purpose Flag				
P	Set/clea	Parity Flag Set/cleared by hardware after each instruction to indicate an odd/even number of "one" bits in the accumulator, i.e. even parity.				

#### **Memory Organization**

The C505 CPU manipulates operands in the following four address spaces:

- On-chip program memory: 16 Kbyte ROM (C505-2R/C505C-2R) or 32 Kbyte OTP (C505A-4E/C505CA-4E)
- Totally up to 64 Kbyte internal/external program memory
- up to 64 Kbyte of external data memory
- 256 bytes of internal data memory
- Internal XRAM data memory: 256 byte (C505/C505C)

1k byte (C505A/C505CA)

- a 128 byte special function register area

Figure 5 illustrates the memory address spaces of the C505 versions.

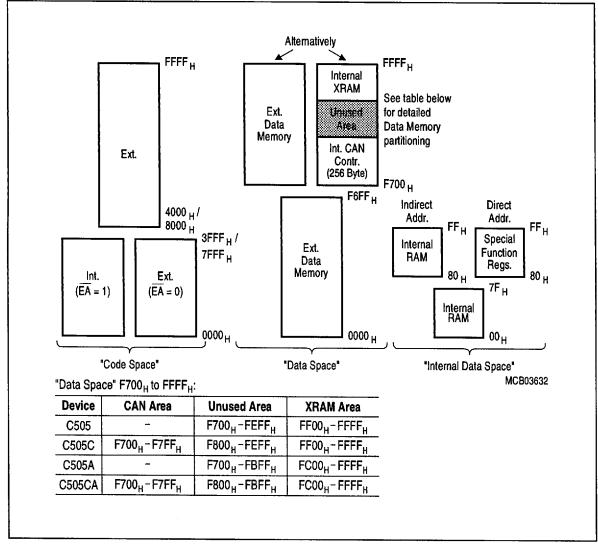


Figure 5 C505 Memory Map Memory Map

#### **Reset and System Clock**

The reset input is an active high input at pin RESET. Since the reset is synchronized internally, the RESET pin must be held high for at least two machine cycles (12 oscillator periods) while the oscillator is running. A pulldown resistor is internally connected to  $V_{\rm SS}$  to allow a power-up reset with an external capacitor only. An automatic reset can be obtained when  $V_{\rm CC}$  is applied by connecting the RESET pin to  $V_{\rm CC}$  via a capacitor. **Figure 6** shows the possible reset circuitries.

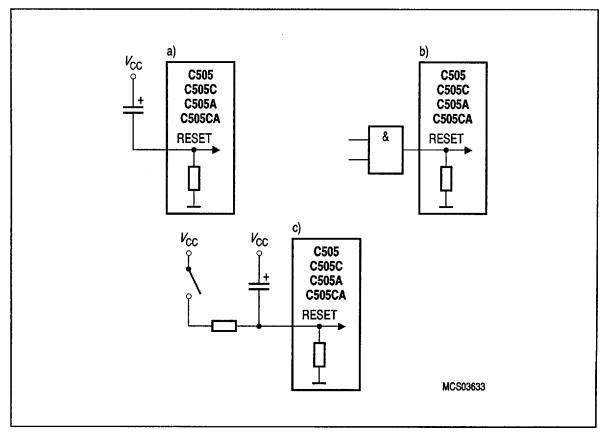


Figure 6
Reset Circuitries

Figure 7 shows the recommended oscillator circuits for crystal and external clock operation.

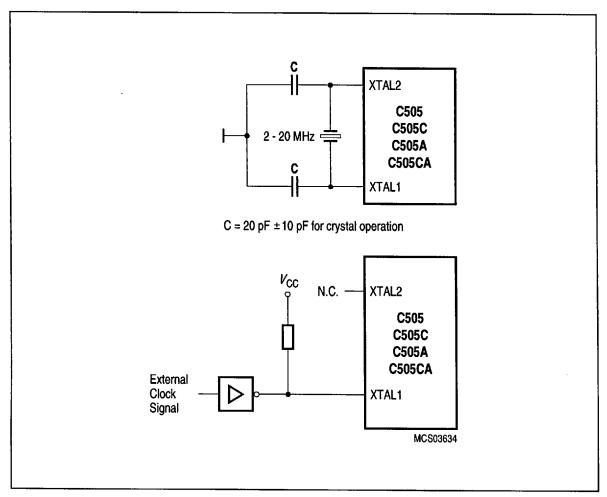


Figure 7
Recommended Oscillator Circuitries

#### **Multiple Datapointers**

As a functional enhancement to the standard 8051 architecture, the C505 contains eight 16-bit datapointers instead of only one datapointer. The instruction set uses just one of these datapointers at a time. The selection of the actual datapointer is done in the special function regsiter DPSEL. **Figure 8** illustrates the datapointer addressing mechanism.

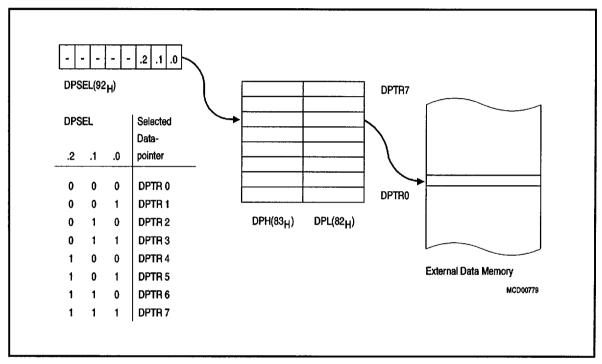


Figure 8
External Data Memory Addressing using Multiple Datapointers

#### **Enhanced Hooks Emulation Concept**

The Enhanced Hooks Emulation Concept of the C500 microcontroller family is a new, innovative way to control the execution of C500 MCUs and to gain extensive information on the internal operation of the controllers. Emulation of on-chip ROM based programs is possible, too.

Each production chip has built-in logic for the supprt of the Enhanced Hooks Emulation Concept. Therefore, no costly bond-out chips are necessary for emulation. This also ensure that emulation and production chips are identical.

The Enhanced Hooks Technology<sup>TM 1)</sup>, which requires embedded logic in the C500 allows the C500 together with an EH-IC to function similar to a bond-out chip. This simplifies the design and reduces costs of an ICE-system. ICE-systems using an EH-IC and a compatible C500 are able to emulate all operating modes of the different versions of the C500 microcontrollers. This includes emulation of ROM, ROM with code rollover and ROMless modes of operation. It is also able to operate in single step mode and to read the SFRs after a break.

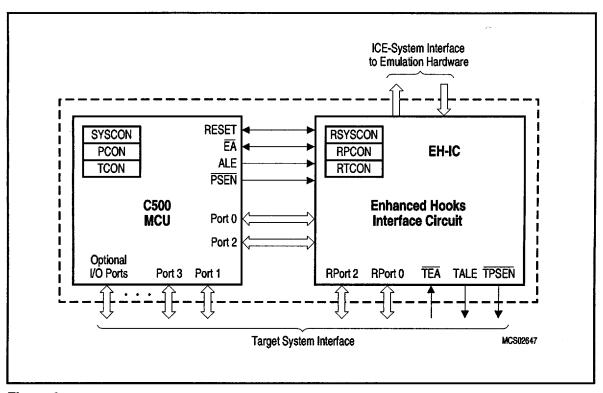


Figure 9
Basic C500 MCU Enhanced Hooks Concept Configuration

Port 0, port 2 and some of the control lines of the C500 based MCU are used by Enhanced Hooks Emulation Concept to control the operation of the device during emulation and to transfer informations about the program execution and data transfer between the external emulation hardware (ICE-system) and the C500 MCU.

<sup>1) &</sup>quot;Enhanced Hooks Technology" is a trademark and patent of Metalink Corporation licensed to Siemens.

#### **Special Function Registers**

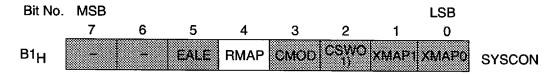
The registers, except the program counter and the four general purpose register banks, reside in the special function register area. The special function register area consists of two portions: the standard special function register area and the mapped special function register area. Five special function register of the C505 (PCON1,P1ANA, VR0, VR1, VR2) are located in the mapped special function register area. For accessing the mapped special function register area, bit RMAP in special function register SYSCON must be set. All other special function registers are located in the standard special function register area which is accessed when RMAP is cleared ("0").

The registers and data locations of the CAN controller (CAN-SFRs) are located in the external data memory area at addresses F700<sub>H</sub> to F7FF<sub>H</sub>..

Special Function Register SYSCON (Address B1H)

Reset Value : XX100X01B

(C505CA only) Reset Value : XX100001B



The functions of the shaded bits are not described here.

1) This bit is only available in the C505CA.

Bit	Function
RMAP	Special function register map bit RMAP = 0: The access to the non-mapped (standard) special function register area is enabled.  RMAP = 1: The access to the mapped special function register area is enabled.

As long as bit RMAP is set, mapped special function register area can be accessed. This bit is not cleared by hardware automatically. Thus, when non-mapped/mapped registers are to be accessed, the bit RMAP must be cleared/set respectively by software.

All SFRs with addresses where address bits 0-2 are 0 (e.g.  $80_{H}$ ,  $88_{H}$ ,  $90_{H}$ ,  $98_{H}$ , ...,  $F8_{H}$ ,  $FF_{H}$ ) are bitaddressable.

The 52 special function registers (SFRs) in the standard and mapped SFR area include pointers and registers that provide an interface between the CPU and the other on-chip peripherals. The SFRs of the C505 are listed in **table 4** and **table 5**. In **table 4** they are organized in groups which refer to the functional blocks of the C505. The CAN-SFRs (applicable for the C505C and C505CA only) are also included in **table 4**. **Table 5** illustrates the contents of the SFRs in numeric order of their addresses. **Table 6** list the CAN-SFRs in numeric order of their addresses.

Table 4 **Special Function Registers - Functional Blocks** 

Block	Symbol	Name	Address	Contents after Reset
CPU	ACC	Accumulator	E0 <sub>H</sub> 1)	00 <sub>H</sub>
	В	B-Register	F0H 1)	00H
	DPH	Data Pointer, High Byte	83H	00H
	DPL	Data Pointer, Low Byte	82 <sub>H</sub>	00H
	DPSEL	Data Pointer Select Register	92 <sub>H</sub>	XXXXX000 <sub>B</sub> 3)
	PSW	Program Status Word Register	D0H 1)	00H
	SP	Stack Pointer	81 <sub>H</sub>	07 <sub>H</sub>
	SYSCON <sup>2)</sup>	System Control Register	B1 <sub>H</sub>	XX100X01B <sup>3) 6)</sup> XX100001B <sup>3) 7)</sup>
	VR0 4)	Version Register 0	FCH	C5 <sub>H</sub>
	VR1 4)	Version Register 1	FD <sub>H</sub>	05H <sup>6)</sup>
			FDH	85H <sup>7)</sup>
	VR2 4)	Version Register 2	FEH	5)
A/D-	ADCON0 <sup>2)</sup>	A/D Converter Control Register 0	D8 <sub>H</sub> 1)	00X00000 <sub>B</sub> <sup>3)</sup>
Converter	ADCON1	A/D Converter Control Register 1	DCH	01XXX000 <sub>B</sub> 3)
	ADDAT	A/D Converter Data Reg. (C505/C505C)	D9H	00H
	ADST	A/D Converter Start Reg. (C505/C505C)	DAH	XXH <sup>3)</sup>
	ADDATH	A/D Converter High Byte Data Register (C505A/C505CA)	D9H	00H
	ADDATL	A/D Converter Low Byte Data Register (C505A/C505CA)	DA <sub>H</sub>	00XXXXXXB <sub>3</sub> )
	P1ANA 2) 4)	Port 1 Analog Input Selection Register	90H 1)	FFH
Interrupt	IENO 2)	Interrupt Enable Register 0	A8 <sub>H</sub> 1)	00 <sub>H</sub>
System	IEN1 2)	Interrupt Enable Register 1	B8 <sub>H</sub> 1)	00H
	IP0 <sup>2)</sup>	Interrupt Priority Register 0	A9 <sub>H</sub>	00H
	IP1	Interrupt Priority Register 1	в9 <sub>Н</sub>	XX000000 <sub>B</sub> 3)
	TCON 2)	Timer Control Register	88H 1)	00H
	T2CON 2)	Timer 2 Control Register	C8 <sub>H</sub> 1)	00X00000 <sub>B</sub>
	SCON 2)	Serial Channel Control Register	98H <sup>1)</sup>	00H
	IRCON	Interrupt Request Control Register	C0H 1)	00H
XRAM	XPAGE	Page Address Register for Extended on-chip XRAM and CAN Controller	91 <sub>H</sub>	00 <sub>H</sub>
	SYSCON <sup>2)</sup>	System Control Register	B1 <sub>H</sub>	XX100X01 <sub>B</sub> <sup>3) 6)</sup> XX100001 <sub>B</sub> <sup>3) 7)</sup>

<sup>1)</sup> Bit-addressable special function registers

<sup>2)</sup> This special function register is listed repeatedly since some bits of it also belong to other functional blocks.

3) "X" means that the value is undefined and the location is reserved

4) This SFR is a mapped SFR. For accessing this SFR, bit RMAP in SFR SYSCON must be set.

5) The content of this SFR varies with the actual step of the C505 (eg. 01<sub>H</sub> for the first step)

<sup>6)</sup> C505 / C505A only 7) C505C / C505CA only

Table 4 Special Function Registers - Functional Blocks (cont'd)

Block	Symbol	Name	Address	Contents after Reset
Ports	P0	Port 0	80H 1)	FFH
	P1	Port 1	90H 1)	FFH
	P1ANA 2) 4)	Port 1 Analog Input Selection Register	90H 1)	FFH
	P2	Port 2	A0H 1)	FFH
	P3	Port 3	B0H 1)	FFH
	P4	Port 4	E8H 1)	XXXXXX11 <sub>B</sub>
Serial	ADCON0 <sup>2)</sup>	A/D Converter Control Register 0	D8 <sub>H</sub> 1)	00X00000 <sub>B</sub> <sup>3)</sup>
Channel	PCON <sup>2)</sup>	Power Control Register	87 <sub>H</sub>	00H
	SBUF	Serial Channel Buffer Register	99 <sub>H</sub>	XXH <sup>3)</sup>
	SCON	Serial Channel Control Register	98H 1)	00H
	SRELL	Serial Channel Reload Register, low byte	AAH	D9H
	SRELH	Serial Channel Reload Register, high byte	BAH	XXXXXX11 <sub>B<sup>3)</sup></sub>
Timer 0/	TCON	Timer 0/1 Control Register	88 <sub>H</sub> 1)	00 <sub>H</sub>
Timer 1	TH0	Timer 0, High Byte	8C <sub>H</sub>	00H
	TH1	Timer 1, High Byte	8D <sub>H</sub>	00H
	TL0	Timer 0, Low Byte	8A <sub>H</sub>	00H
	TL1	Timer 1, Low Byte	8B <sub>H</sub>	00H
	TMOD	Timer Mode Register	89 <sub>H</sub>	00H
Compare/	CCEN	Comp./Capture Enable Reg.	C1 <sub>H</sub>	00H 3)
Capture	CCH1	Comp./Capture Reg. 1, High Byte	C3 <sub>H</sub>	00H
Unit /	CCH2	Comp./Capture Reg. 2, High Byte	C5 <sub>H</sub>	00H
Timer 2	ССНЗ	Comp./Capture Reg. 3, High Byte	C7 <sub>H</sub>	00H
	CCL1	Comp./Capture Reg. 1, Low Byte	C2 <sub>H</sub>	00 <sub>H</sub>
	CCL2	Comp./Capture Reg. 2, Low Byte	C4 <sub>H</sub>	00 <sub>H</sub>
	CCL3	Comp./Capture Reg. 3, Low Byte	C6 <sub>H</sub>	00 <sub>H</sub>
	CRCH	Reload Register High Byte	CBH	00H
	CRCL	Reload Register Low Byte	CAH	00H
	TH2	Timer 2, High Byte	CDH	00H
	TL2	Timer 2, Low Byte	CCH	00H
	T2CON	Timer 2 Control Register	C8H 1)	00X00000B 3)
	IENO 2)	Interrupt Enable Register 0	A8H 1)	00 <sub>H</sub>
	IEN1 <sup>2)</sup>	Interrupt Enable Register 1	B8H 1)	00H
Watchdog	WDTREL	Watchdog Timer Reload Register	86 <sub>H</sub>	00 <sub>H</sub>
	IENO 2)	Interrupt Enable Register 0	A8 <sub>H</sub> 1)	00 <sub>H</sub>
	IEN1 <sup>2)</sup>	Interrupt Enable Register 1	B8 <sub>H</sub> 1)	00 <sub>H</sub>
	IPO <sup>2)</sup>	Interrupt Priority Register 0	A9H	00H
Pow. Save		Power Control Register	87 <sub>H</sub>	00 <sub>H</sub>
Modes	PCON1 4)	Power Control Register 1	88 <sub>H</sub> 1)	0XX0XXXXB3)

<sup>1)</sup> Bit-addressable special function registers
2) This special function register is listed repeatedly since some bits of it also belong to other functional blocks.
3) "X" means that the value is undefined and the location is reserved
4) SFR is located in the mapped SFR area. For accessing this SFR, bit RMAP in SFR SYSCON must be set.

Table 4
Special Function Registers - Functional Blocks (cont'd)

Block	Symbol	Name	Address	Contents after
				Reset
CAN	CR	Control Register	F700 <sub>H</sub>	01 <sub>H</sub>
Controller	SR	Status Register	F701 <sub>H</sub>	XXH 3)
	IR	Interrupt Register	F702 <sub>H</sub>	XXH <sup>3)</sup>
(C505C/	BTR0	Bit Timing Register Low	F704 <sub>H</sub>	
C505CA	BTR1	Bit Timing Register High	F705 <sub>H</sub>	0UUUUUUU <sub>B</sub> 3)
only)	GMS0	Global Mask Short Register Low	F706H	UUH3)
	GMS1	Global Mask Short Register High	F707 <sub>H</sub>	UUU11111 <sub>B</sub> 3)
	UGML0	Upper Global Mask Long Register Low	F708 <sub>H</sub>	UUH3)
	UGML1	Upper Global Mask Long Register High	F709 <sub>H</sub>	UUH3)
	LGML0	Lower Global Mask Long Register Low	F70A <sub>H</sub>	UUH <sup>3)</sup>
	LGML1	Lower Global Mask Long Register High	F70B <sub>H</sub>	UUUUU000 <sub>B</sub> 3)
	UMLM0	Upper Mask of Last Message Register Low	F70CH	UUH <sup>3)</sup>
	UMLM1	Upper Mask of Last Message Register High	F70D <sub>H</sub>	UU <sub>H</sub> 3)
	LMLMO	Lower Mask of Last Message Register Low	F70E <sub>H</sub>	UU <sub>H</sub> 3)
	LMLM1	Lower Mask of Last Message Register High	F70F <sub>H</sub>	UUUUU000 <sub>B</sub> 3)
		Message Object Registers :		
	MCR0	Message Control Register Low	F7n0 <sub>H</sub> 5)	UU <sub>H</sub> 3)
	MCR1	Message Control Register High	F7n1 <sub>H</sub> 5)	UU <sub>H</sub> 3)
	UAR0	Upper Arbitration Register Low	F7n2 <sub>H</sub> 5)	UUH <sup>3)</sup>
	UAR1	Upper Arbitration Register High	F7n3 <sub>H</sub> 5)	UUH <sup>3)</sup>
	LAR0	Lower Arbitration Register Low	F7n4 <sub>H</sub> 5)	UUH <sup>3)</sup>
	LAR1	Lower Arbitration Register High	F7n5 <sub>H</sub> 5)	UUUUU000 <sub>B</sub> 3)
	MCFG	Message Configuration Register	F7n6 <sub>H</sub> 5)	UUUUUU00 <sub>B<sup>3)</sup></sub>
	DB0n	Message Data Byte 0	F7n7 <sub>H</sub> 5)	XXH <sup>3)</sup>
	DB1n	Message Data Byte 1	F7n8 <sub>H</sub> 5)	XXH <sup>3)</sup>
	DB2n	Message Data Byte 2	F7n9H 5)	XXH 3)
	DB3n	Message Data Byte 3	F7nA <sub>H</sub> 5)	XXH <sup>3)</sup>
	DB4n	Message Data Byte 4	F7nB <sub>H</sub> 5)	XXH 3)
	DB5n	Message Data Byte 5	F7nCH 5)	XXH <sup>3)</sup>
	DB6n	Message Data Byte 6	F7nDH 5)	XXH <sup>3)</sup>
	DB7n	Message Data Byte 7	F7nE <sub>H</sub> 5)	XXH <sup>3)</sup>

<sup>1)</sup> Bit-addressable special function registers

<sup>2)</sup> This special function register is listed repeatedly since some bits of it also belong to other functional blocks.

<sup>3) &</sup>quot;X" means that the value is undefined and the location is reserved. "U" means that the value is unchanged by a reset operation. "U" values are undefined (as "X") after a power-on reset operation

<sup>4)</sup> SFR is located in the mapped SFR area. For accessing this SFR, bit RMAP in SFR SYSCON must be set.

<sup>5)</sup> The notation "n" (n= 1 to F) in the message object address definition defines the number of the related message object.

Table 5 Contents of the SFRs, SFRs in numeric order of their addresses

Addr	Register	Content after Reset <sup>1)</sup>	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
80H2)	P0	FFH	7	6	.5	4	.3	.2	.1	.0
81 <sub>H</sub>	SP	07 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
82 <sub>H</sub>	DPL	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
83 <sub>H</sub>	DPH	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
86 <sub>H</sub>	WDTREL	00 <sub>H</sub>	WDT PSEL	.6	.5	.4	.3	.2	.1	.0
87 <sub>H</sub>	PCON	00H	SMOD	PDS	IDLS	SD	GF1	GF0	PDE	IDLE
88 <sub>H</sub> 2)	TOON	00H	TF1	TRI	TFO	TRO	IE1	m	IEO	ITO
88H33	PCON1	OXXO- XXXXB	EWPD	-	-	WS	-	-	-	-
89 <sub>H</sub>	TMOD	00H	GATE	C/T	M1	МО	GATE	C/T	M1	МО
8A <sub>H</sub>	TL0	00H	.7	.6	.5	.4	.3	.2	.1	.0
8B <sub>H</sub>	TL1	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
8CH	TH0	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
8D <sub>H</sub>	TH1	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
90H <sup>2)</sup>	P1	FFH	T2	CLK- OUT	TZEX	4	.3	INT5	INT4	D
90 <sub>H</sub> <sup>n</sup>	PIANA	FFH	EAN7	EAN6	EAN5	EAN4	EAN3	EAN2	EAN1	EAN0
91 <sub>H</sub>	XPAGE	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
92 <sub>H</sub>	DPSEL	XXXX-	_	-		-	_	.2	.1	.0
98H2)	SCON	00H	SMO	SM1	SM2	REN	TB8	RB8	TI	RI
99H	SBUF	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
A0H2	P2	FFH	7	6	.5	.4	.3	.2	.1	.0
A8 <sub>H</sub> 2	IEN0	00H	EA	WDT	ET2	ES	ET1	EX1	ETO	EX0
A9 <sub>H</sub>	IP0	00 <sub>H</sub>	OWDS	WDTS	.5	.4	.3	.2	.1	.0
AA <sub>H</sub>	SRELL	D9 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0

X means that the value is undefined and the location is reserved
 Bit-addressable special function registers
 SFR is located in the mapped SFR area. For accessing this SFR, bit RMAP in SFR SYSCON must be set.

Table 5 Contents of the SFRs, SFRs in numeric order of their addresses (cont'd)

Addr	Register	Content after Reset <sup>1)</sup>	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
BOHn	P3	FFH	FID	WR	T1	TO	INTT	INTO	TxD	FIXD
B1 <sub>H</sub>	SYSCON	XX10- 0X01 <sub>B</sub>	_	_	EALE	RMAP	CMOD	_	XMAP1	XMAP0
B1 <sub>H</sub>	SYSCON	XX10- 0001 <sub>B</sub>	-	_	EALE	RMAP	CMOD	CSWO	XMAP1	XMAP0
B8H <sup>a</sup>	IEN1 3)	00H	EXEN2	SWDT	EX6	EX5	EX4	EX3	0	EADC
88 <sub>H</sub> *	IEN1 "	00H	EXEN2	SWDT	EX6	EX5	EX4	EX3	ECAN	EADC
B9 <sub>H</sub>	IP1	XX00- 0000 <sub>B</sub>	-	_	.5	.4	.3	.2	.1	.0
BAH	SRELH	XXXX- XX11 <sub>B</sub>	_	_	_	_	_	_	.1	.0
CoHa	IRCON	00 <sub>H</sub>	EXF2	TF2	IEX6	IEX5	IEX4	IEX3	SWI	IADC
C1 <sub>H</sub>	CCEN	00H	COCA H3	COCAL 3	COCA H2	COCAL 2	COCA H1	COCAL 1	COCA H0	COCAL 0
C2 <sub>H</sub>	CCL1	00H	.7	.6	.5	.4	.3	.2	.1	.0
СЗН	CCH1	00H	.7	.6	.5	.4	.3	.2	.1	.0
C4 <sub>H</sub>	CCL2	00H	.7	.6	.5	.4	.3	.2	.1	.0
C5 <sub>H</sub>	CCH2	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
C6H	CCL3	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
C7 <sub>H</sub>	ССНЗ	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0
C8H29	T2CON	00X0- 0000B	T2PS	ISFR	-	T2F1	T2R0	T2CM	T2I1	T210
CAH	CRCL	00H	.7	.6	.5	.4	.3	.2	.1	.0
СВН	CRCH	00H	.7	.6	.5	.4	.3	.2	.1	.0
CCH	TL2	00H	.7	.6	.5	.4	.3	.2	1.1	.0
CDH	TH2	00H	.7	.6	.5	.4	.3	.2	.1	.0
D0H <sup>2)</sup>	PSW	00H	CY	AC	FO	RS1	RS0	OV	F1	P
D8Ha	ADCON0	<del></del>	BD	CLK		BSY	ADM	MX2	MX1	MX0
D9 <sub>H</sub>	ADDAT 3)	00 <sub>H</sub>	.7	.6	.5	.4	.3	.2	.1	.0

<sup>1)</sup> X means that the value is undefined and the location is reserved
2) Bit-addressable special function registers
3) C505 / C505A only
4) C505C / C505CA only

Table 5 Contents of the SFRs, SFRs in numeric order of their addresses (cont'd)

Addr	Register	Content after Reset <sup>1)</sup>	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
D9 <sub>H</sub>	ADDATH	00 <sub>H</sub>	.9	.8	.7	.6	.5	.4	.3	.2
DA <sub>H</sub>	ADST 6)	XXXX- XXXX <sub>B</sub>	_	_	_	-	_	-	-	_
DA <sub>H</sub>	ADDATL	00XX- XXXX <sub>B</sub>	.1	.0	_	_	-	_	-	_
DCH	ADCON1	01XX- X000 <sub>B</sub>	ADCL1	ADCL0	-	_	_	MX2	MX1	MX0
E0H <sup>27</sup>	P4	OO <sub>H</sub> XXXX- XX11 <sub>B</sub>	7	5	5	4	3	2	.1 RXDC	0 FXDC
FOH <sup>2)</sup> FCH <sup>3)4)</sup>	***************************************	00 <sub>H</sub> C5 <sub>H</sub>	. <b>7</b> 1	.6 1	.5 0	<b>4</b> 0	3 0	.2 1	ੀ ß	.0 1
FDH <sup>3)4)</sup>		05 <sub>H</sub>	0	0	0	0	0	1	0	1
FEH <sup>3)4)</sup>	VR2	01 <sub>H</sub> <sup>5) 6)</sup>	.7	.6	.5	.4	.3	.2	.1	.0

<sup>1)</sup> X means that the value is undefined and the location is reserved

<sup>2)</sup> Bit-addressable special function registers
3) SFR is located in the mapped SFR area. For accessing this SFR, bit RMAP in SFR SYSCON must be set.
4) These are read-only registers

The content of this SFR varies with the actual of the step C505 (eg. 01<sub>H</sub> or 11<sub>H</sub> for the first step)

<sup>6)</sup> C505 / C505C only

<sup>7)</sup> C505A / C505CA only

Table 6 Contents of the CAN Registers in numeric order of their addresses (C505C/C505CA only)

Addr. n=1-F <sub>H</sub> 1)	Register	Content after Reset <sup>2)</sup>	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
F700 <sub>H</sub>	CR	01 <sub>H</sub>	TEST	CCE	0	0	EIE	SIE	IE	INIT
F701 <sub>H</sub>	SR	хх <sub>Н</sub>	BOFF	EWRN	_	RXOK	ТХОК	LEC2	LEC1	LEC0
F702 <sub>H</sub>	IR	хх <sub>Н</sub>				IN'	TID	•		<del>-                                    </del>
F704 <sub>H</sub>	BTR0	UU <sub>H</sub>	S	JW			В	RP		
F705 <sub>H</sub>	BTR1	0บบบ. บบบบ <sub>B</sub>	0		TSEG2			TS	EG1	
F706 <sub>H</sub>	GMS0	UU <sub>H</sub>				ID2	8-21		•	
F707 <sub>H</sub>	GMS1	UUU1. 1111 <sub>B</sub>		ID20-18		1	1	1	1	1
F708 <sub>H</sub>	UGML0	υυ <sub>Η</sub>				ID2	8-21			
F709 <sub>H</sub>	UGML1	UUΗ				ID2	0-13			
F70A <sub>H</sub>	LGML0	υυΗ				ID.	12-5			
F70B <sub>H</sub>	LGML1	บบบบ. บ000 <sub>B</sub>			ID4-0			0	0	0
F70C <sub>H</sub>	UMLMO	υυ <sub>Η</sub>				ID2	8-21			
F70D <sub>H</sub>	UMLM1	υυ <sub>Η</sub>		ID20-18				ID17-13	3	
F70E <sub>H</sub>	LMLMO	υυ <sub>Η</sub>			71 2.55	ID	12-5			
F70F <sub>H</sub>	LMLM1	UUUU. U000 <sub>B</sub>			ID4-0			0	0	0
F7n0 <sub>H</sub>	MCR0	υυ <sub>Η</sub>	MS	GVAL	T)	KIE	R	XIE	IN	TPND
F7n1 <sub>H</sub>	MCR1	υυ <sub>Η</sub>	RM	FPND	TX	RQ	ş	GLST JUPD	NE'	WDAT
F7n2 <sub>H</sub>	UAR0	UU <sub>H</sub>			, ,	ID2	8-21		— •	
F7n3 <sub>H</sub>	UAR1	UU <sub>H</sub>	ID20-18 ID17-13							
F7n4 <sub>H</sub>	LAR0	υυ <sub>Η</sub>	ID12-5							
F7n5 <sub>H</sub>	LAR1	UUUU. U000 <sub>B</sub>		ID4-0 0 0				0		
F7n6 <sub>H</sub>	MCFG	UUUU. UU00B		D	LC		DIR	XTD	0	0

<sup>1)</sup> The notation "n" (n= 1 to F) in the address definition defines the number of the related message object.
2) "X" means that the value is undefined and the location is reserved. "U" means that the value is unchanged by a reset operation. "U" values are undefined (as "X") after a power-on reset operation

Table 6
Contents of the CAN Registers in numeric order of their addresses (cont'd) (C505C/C505CA only)

Addr. n=1-F <sub>H</sub> 1)	Register	Content after Reset <sup>2)</sup>	Ì	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
F7n7 <sub>H</sub>	DB0n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7n8 <sub>H</sub>	DB1n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7n9 <sub>H</sub>	DB2n	XXH	.7	.6	.5	.4	.3	.2	.1	.0
F7nA <sub>H</sub>	DB3n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7nB <sub>H</sub>	DB4n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7nC <sub>H</sub>	DB5n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7nD <sub>H</sub>	DB6n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0
F7nE <sub>H</sub>	DB7n	хх <sub>Н</sub>	.7	.6	.5	.4	.3	.2	.1	.0

The notation "n" (n= 1 to F) in the address definition defines the number of the related message object.

<sup>2) &</sup>quot;X" means that the value is undefined and the location is reserved. "U" means that the value is unchanged by a reset operation. "U" values are undefined (as "X" after a power-on reset operation

#### I/O Ports

The C505 has four 8-bit I/O ports and one 2-bit I/O port. Port 0 is an open-drain bidirectional I/O port, while ports 1 to 4 are quasi-bidirectional I/O ports with internal pullup resistors. That means, when configured as inputs, ports 1 to 4 will be pulled high and will source current when externally pulled low. Port 0 will float when configured as input.

The output drivers of port 0 and 2 and the input buffers of port 0 are also used for accessing external memory. In this application, port 0 outputs the low byte of the external memory address, time multiplexed with the byte being written or read. Port 2 outputs the high byte of the external memory address when the address is 16 bits wide. Otherwise, the port 2 pins continue emitting the P2 SFR contents. In this function, port 0 is not an open-drain port, but uses a strong internal pullup FET.

Port 4 is 2-bit I/O port with CAN controller specific alternate functions. The eight analog input lines are realized as mixed digital/analog inputs. The 8 analog inputs, AN0-AN7, are located at the port 1 pins P1.0 to P1.7. After reset, all analog inputs are disabled and the related pins of port 1 are configured as digital inputs. The analog function of a specific port 1 pin is enabled by bits in the SFR P1ANA. Writing a 0 to a bit position of P1ANA assigns the corresponding pin to operate as analog input.

Note: P1ANA is a mapped SFR and can be only accessed if bit RMAP in SFR SYSCON is set.

#### Timer / Counter 0 and 1

Timer/Counter 0 and 1 can be used in four operating modes as listed in table 7:

Table 7
Timer/Counter 0 and 1 Operating Modes

Mode	Description	TM	IOD	Input Clock		
		M1	MO	internal	external (max)	
0	8-bit timer/counter with a divide-by-32 prescaler	0	0	f <sub>osc</sub> /6x32	f <sub>osc</sub> /12x32	
1	16-bit timer/counter	1	1			
2	8-bit timer/counter with 8-bit autoreload	1	0	£ 10	£ 40	
3	Timer/counter 0 used as one 8-bit timer/counter and one 8-bit timer Timer 1 stops	1	1	f <sub>osc</sub> /6	f <sub>osc</sub> /12	

In the "timer" function (C/ $\overline{T}$  = '0') the register is incremented every machine cycle. Therefore the count rate is  $f_{OSC}/6$ .

In the "counter" function the register is incremented in response to a 1-to-0 transition at its corresponding external input pin (P3.4/T0, P3.5/T1). Since it takes two machine cycles to detect a falling edge the max. count rate is  $f_{\rm OSO}/12$ . External inputs  $\overline{\rm INT0}$  and  $\overline{\rm INT1}$  (P3.2, P3.3) can be programmed to function as a gate to facilitate pulse width measurements. **Figure 10** illustrates the input clock logic.

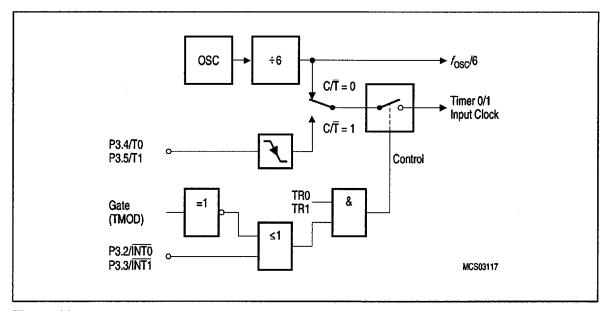


Figure 10
Timer/Counter 0 and 1 Input Clock Logic

#### Timer/Counter 2 with Compare/Capture/Reload

The timer 2 of the C505 provides additional compare/capture/reload features. which allow the selection of the following operating modes:

- Compare : up to 4 PWM signals with 16-bit/300 ns resolution (@ 20 MHz clock)

- Capture : up to 4 high speed capture inputs with 300 ns resolution

- Reload : modulation of timer 2 cycle time

The block diagram in **figure 11** shows the general configuration of timer 2 with the additional compare/capture/reload registers. The I/O pins which can used for timer 2 control are located as multifunctional port functions at port 1.

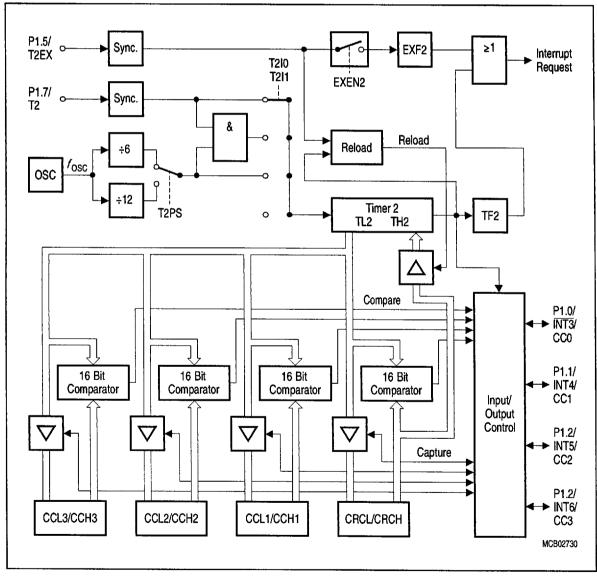


Figure 11
Timer 2 Block Diagram

#### **Timer 2 Operating Modes**

The timer 2, which is a 16-bit-wide register, can operate as timer, event counter, or gated timer. A roll-over of the count value in TL2/TH2 from all 1's to all 0's sets the timer overflow flag TF2 in SFR IRCON, which can generate an interrupt. The bits in register T2CON are used to control the timer 2 operation.

<u>Timer Mode</u>: In timer function, the count rate is derived from the oscillator frequency. A prescaler offers the possibility of selecting a count rate of 1/6 or 1/12 of the oscillator frequency.

<u>Gated Timer Mode</u>: In gated timer function, the external input pin T2 (P1.7) functions as a gate to the input of timer 2. If T2 is high, the internal clock input is gated to the timer. T2 = 0 stops the counting procedure. This facilitates pulse width measurements. The external gate signal is sampled once every machine cycle.

Event Counter Mode: In the event counter function, the timer 2 is incremented in response to a 1-to-0 transition at its corresponding external input pin T2 (P1.7). In this function, the external input is sampled every machine cycle. Since it takes two machine cycles (12 oscillator periods) to recognize a 1-to-0 transition, the maximum count rate is 1/6 of the oscillator frequency. There are no restrictions on the duty cycle of the external input signal, but to ensure that a given level is sampled at least once before it changes, it must be held for at least one full machine cycle.

Reload of Timer 2: Two reload modes are selectable:

In mode 0, when timer 2 rolls over from all 1's to all 0's, it not only sets TF2 but also causes the timer 2 registers to be loaded with the 16-bit value in the CRC register, which is preset by software. In mode 1, a 16-bit reload from the CRC register is caused by a negative transition at the corresponding input pin P1.5/T2EX. This transition will also set flag EXF2 if bit EXEN2 in SFR IEN1 has been set.

#### **Timer 2 Compare Modes**

The compare function of a timer/register combination operates as follows: the 16-bit value stored in a compare or compare/capture register is compared with the contents of the timer register; if the count value in the timer register matches the stored value, an appropriate output signal is generated at a corresponding port pin and an interrupt can be generated.

#### Compare Mode 0

In compare mode 0, upon matching the timer and compare register contents, the output signal changes from low to high. It goes back to a low level on timer overflow. As long as compare mode 0 is enabled, the appropriate output pin is controlled by the timer circuit only and writing to the port will have no effect. **Figure 12** shows a functional diagram of a port circuit when used in compare mode 0. The port latch is directly controlled by the timer overflow and compare match signals. The input line from the internal bus and the write-to-latch line of the port latch are disconnected when compare mode 0 is enabled.

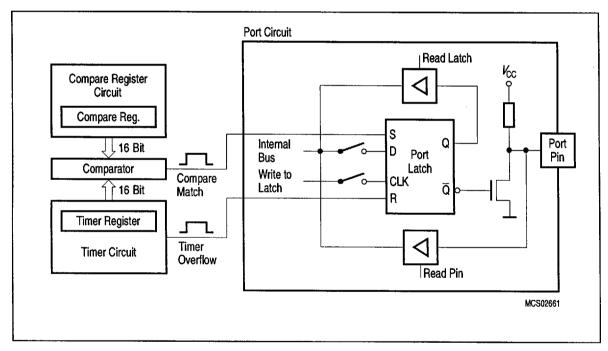


Figure 12
Port Latch in Compare Mode 0

#### Compare Mode 1

If compare mode 1 is enabled and the software writes to the appropriate output latch at the port, the new value will not appear at the output pin until the next compare match occurs. Thus, it can be choosen whether the output signal has to make a new transition (1-to-0 or 0-to-1, depending on the actual pin-level) or should keep its old value at the time when the timer value matches the stored compare value.

In compare mode 1 (see **figure 13**) the port circuit consists of two separate latches. One latch (which acts as a "shadow latch") can be written under software control, but its value will only be transferred to the port latch (and thus to the port pin) when a compare match occurs.

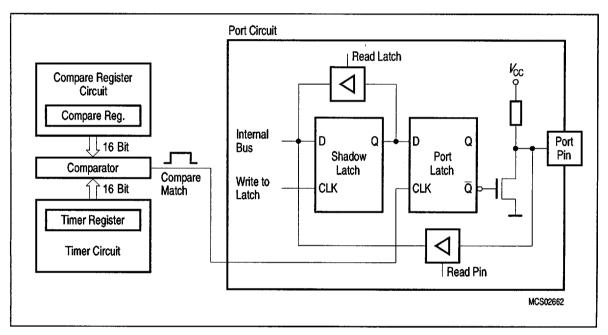


Figure 13
Compare Function in Compare Mode 1

#### **Timer 2 Capture Modes**

Each of the compare/capture registers CC1 to CC3 and the CRC register can be used to latch the current 16-bit value of the timer 2 registers TL2 and TH2. Two different modes are provided for this function.

In mode 0, the external event causing a capture is :

- for CC registers 1 to 3: a positive transition at pins CC1 to CC3 of port 1
- for the CRC register: a positive or negative transition at the corresponding pin, depending on the status of the bit I3FR in SFR T2CON.

In <u>mode 1</u> a capture occurs in response to a write instruction to the low order byte of a capture register. The write-to-register signal (e.g. write-to-CRCL) is used to initiate a capture. The timer 2 contents will be latched into the appropriate capture register in the cycle following the write instruction. In this mode no interrupt request will be generated.

#### Serial Interface (USART)

The serial port is full duplex and can operate in four modes (one synchronous mode, three asynchronous modes) as illustrated in table 8.

Table 8
USART Operating Modes

Mada	SC	ON	Description
Mode	SM0	SM1	
0	0	0	Shift register mode, fixed baud rate Serial data enters and exits through R×D; T×D outputs the shift clock; 8-bit are transmitted/received (LSB first)
1	0	1	8-bit UART, variable baud rate 10 bits are transmitted (through T×D) or received (at R×D)
2	1	0	9-bit UART, fixed baud rate 11 bits are transmitted (through T×D) or received (at R×D)
3	1	1	9-bit UART, variable baud rate Like mode 2

For clarification some terms regarding the difference between "baud rate clock" and "baud rate" should be mentioned. In the <u>asynchronous modes</u> the serial interfaces require a clock rate which is 16 times the baud rate for internal synchronization. Therefore, the baud rate generators/timers have to provide a "baud rate clock" (output signal in **figure 14** to the serial interface which - there divided by 16 - results in the actual "baud rate". Further, the abbrevation f<sub>OSC</sub> refers to the oscillator frequency (crystal or external clock operation).

The variable baud rates for modes 1 and 3 of the serial interface can be derived either from timer 1 or from a decdicated baud rate generator (see **figure 14**).

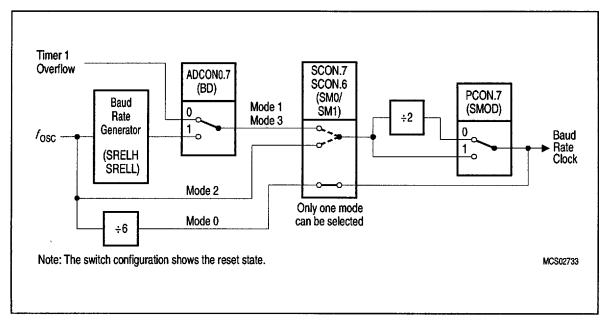


Figure 14
Block Diagram of Baud Rate Generation for the Serial Interface

**Table 9** below lists the values/formulas for the baud rate calculation of the serial interface with its dependencies of the control bits BD and SMOD.

Table 9
Serial Interface - Baud Rate Dependencies

Serial Interface	Active	Control Bits	Baud Rate Calculation		
Operating Modes	BD SMOD				
Mode 0 (Shift Register)	-	_	f <sub>osc</sub> / 6		
Mode 1 (8-bit UART) Mode 3 (9-bit UART)	0	X	Controlled by timer 1 overflow : (2 <sup>SMOD</sup> × timer 1 overflow rate) / 32		
	1	Х	Controlled by baud rate generator $(2^{\text{SMOD}} \times f_{\text{OSC}})$ / $(32 \times \text{baud rate generator overflow rate})$		
Mode 2 (9-bit UART)	_	0	f <sub>osc</sub> / 32 f <sub>osc</sub> / 16		

## CAN Controller (C505C and C505CA only)

The on-chip CAN controller, compliant to version 2.0B, is the functional heart which provides all resources that are required to run the standard CAN protocol (11-bit identifiers) as well as the extended CAN protocol (29-bit identifiers). It provides a sophisticated object layer to relieve the CPU of as much overhead as possible when controlling many different message objects (up to 15). This includes bus arbitration, resending of garbled messages, error handling, interrupt generation, etc. In order to implement the physical layer, external components have to be connected to the C505.

The internal bus interface connects the on-chip CAN controller to the internal bus of the microcontroller. The registers and data locations of the CAN interface are mapped to a specific 256 byte wide address range of the external data memory area (F700<sub>H</sub> to F7FF<sub>H</sub>) and can be accessed using MOVX instructions. **Figure 15** shows a block diagram of the on-chip CAN controller.

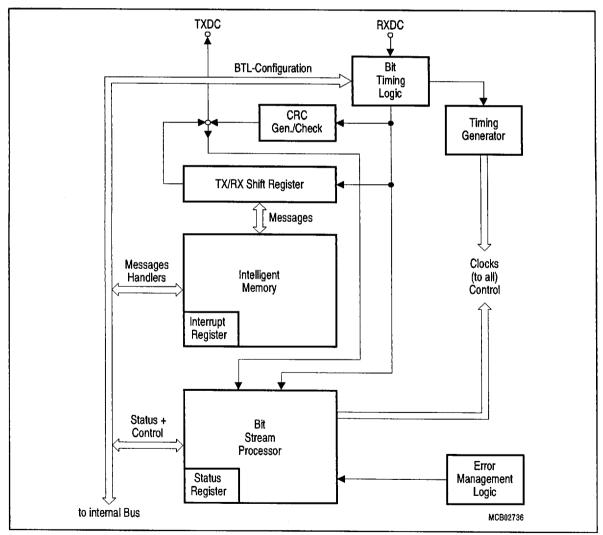


Figure 15
CAN Controller Block Diagram

The **TX/RX Shift Register** holds the destuffed bit stream from the bus line to allow the parallel access to the whole data or remote frame for the acceptance match test and the parallel transfer of the frame to and from the Intelligent Memory.

The **Bit Stream Processor (BSP)** is a sequencer controlling the sequential data stream between the TX/RX Shift Register, the CRC Register, and the bus line. The BSP also controls the EML and the parallel data stream between the TX/RX Shift Register and the Intelligent Memory such that the processes of reception, arbitration, transmission, and error signalling are performed according to the CAN protocol. Note that the automatic retransmission of messages which have been corrupted by noise or other external error conditions on the bus line is handled by the BSP.

The Cyclic Redundancy Check Register (CRC) generates the Cyclic Redundancy Check code to be transmitted after the data bytes and checks the CRC code of incoming messages. This is done by dividing the data stream by the code generator polynomial.

The **Error Management Logic (EML)** is responsible for the fault confinement of the CAN device. Its counters, the Receive Error Counter and the Transmit Error Counter, are incremented and decremented by commands from the Bit Stream Processor. According to the values of the error counters, the CAN controller is set into the states error *active*, error *passive* and busoff.

The **Bit Timing Logic (BTL)** monitors the busline input RXDC and handles the busline related bit timing according to the CAN protocol. The BTL synchronizes on a *recessive* to *dominant* busline transition at *Start of Frame* (hard synchronization) and on any further *recessive* to *dominant* busline transition, if the CAN controller itself does not transmit a *dominant* bit (resynchronization). The BTL also provides programmable time segments to compensate for the propagation delay time and for phase shifts and to define the position of the Sample Point in the bit time. The programming of the BTL depends on the baudrate and on external physical delay times.

The Intelligent Memory (CAM/RAM array) provides storage for up to 15 message objects of maximum 8 data bytes length. Each of these objects has a unique identifier and its own set of control and status bits. After the initial configuration, the Intelligent Memory can handle the reception and transmission of data without further microcontroller actions.

#### **CAN Controller Software Initialization**

The very first step of the initialization is the CAN controller input clock selection. A divide-by-2 prescaler is enabled by default after reset (**figure 16**). Setting bit CMOD (SYSCON.3) disables the prescaler. The purpose of the prescaler selection is:

- to ensure that the CAN controller is operable when  $f_{
  m osc}$  is over 10 MHz (bit CMOD =0)
- to achieve the maximum CAN baudrate of 1 Mbaud when  $f_{\rm osc}$  is 8 MHz (bit CMOD=1)

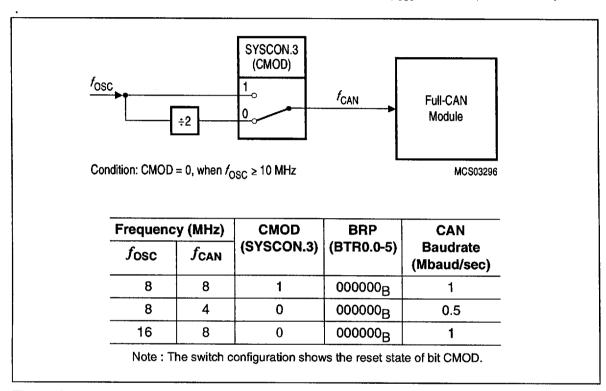


Figure 16
CAN Controller Input Clock Selection

#### 8-Bit A/D Converter (C505 and C505C only)

The C505/C505C includes a high performance / high speed 8-bit A/D converter (ADC) with 8 analog input channels. It operates with a successive approximation technique and provides the following features:

- 8 multiplexed input channels (port 1), which can also be used as digital outputs/inputs
- 8-bit resolution
- Internal start-of-conversion trigger
- Interrupt request generation after each conversion
- Single or continuous conversion mode

The 8-bit ADC uses two clock signals for operation: the conversion clock  $f_{\rm ADC}$  (=1/ $t_{\rm ADC}$ ) and the input clock  $f_{\rm IN}$  (1/ $t_{\rm IN}$ ).  $f_{\rm ADC}$  is derived from the C505 system clock  $f_{\rm OSC}$  which is applied at the XTAL pins via the ADC clock prescaler as shown in **figure 17**. The input clock is equal to  $f_{\rm OSC}$ . The conversion clock  $f_{\rm ADC}$  is limited to a maximum frequency of 1.25 MHz. Therefore, the ADC clock prescaler must be programmed to a value which assures that the conversion clock does not exceed 1.25 MHz. The prescaler ratio is selected by the bits ADCL1 and ADCL0 of SFR ADCON1.

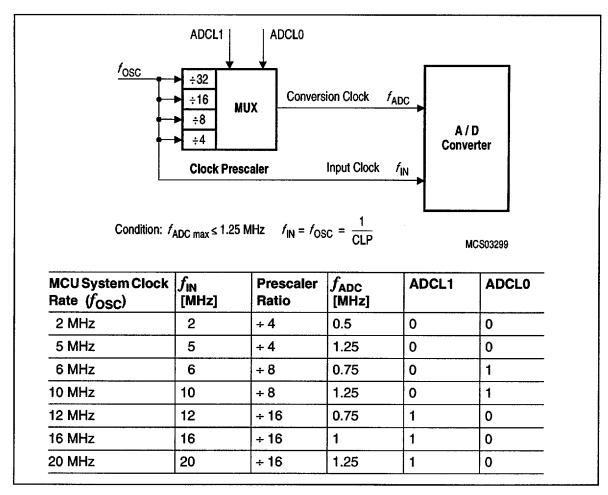


Figure 17
8-Bit A/D Converter Clock Selection

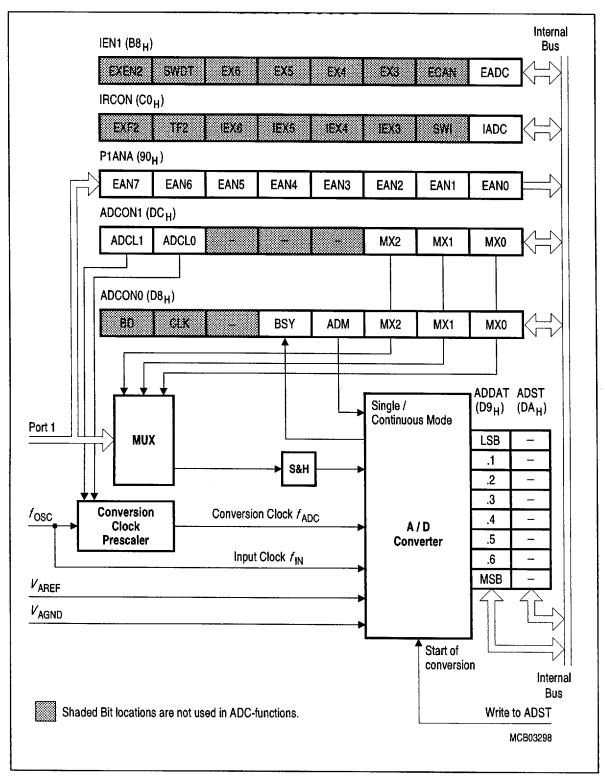


Figure 18
Block Diagram of the 8-Bit A/D Converter

### 10-Bit A/D Converter (C505A and C505CA only)

The C505 includes a high performance / high speed 10-bit A/D-Converter (ADC) with 8 analog input channels. It operates with a successive approximation technique and uses self calibration mechanisms for reduction and compensation of offset and linearity errors. The A/D converter provides the following features:

- 8 multiplexed input channels (port 1), which can also be used as digital inputs/outputs
- 10-bit resolution
- Single or continuous conversion mode
- Internal start-of-conversion trigger capability
- Interrupt request generation after each conversion
- Using successive approximation conversion technique via a capacitor array
- Built-in hidden calibration of offset and linearity errors

The 10-bit ADC uses two clock signals for operation: the conversion clock  $f_{\rm ADC}$  (=1/ $t_{\rm ADC}$ ) and the input clock  $f_{\rm IN}$  (=1/ $t_{\rm IN}$ ).  $f_{\rm ADC}$  is derived from the C505 system clock  $f_{\rm OSC}$  which is applied at the XTAL pins. The input clock  $f_{\rm IN}$  is equal to  $f_{\rm OSC}$  The conversion  $f_{\rm ADC}$  clock is limited to a maximum frequency of 2 MHz. Therefore, the ADC clock prescaler must be programmed to a value which assures that the conversion clock does not exceed 2 MHz. The prescaler ratio is selected by the bits ADCL1 and ADCL0 of SFR ADCON1.

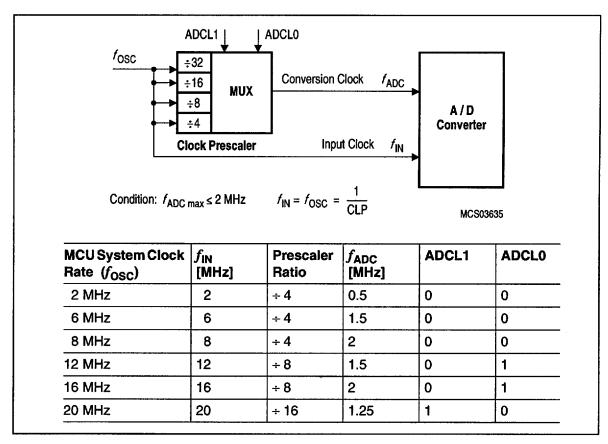


Figure 19
10-Bit A/D Converter Clock Selection

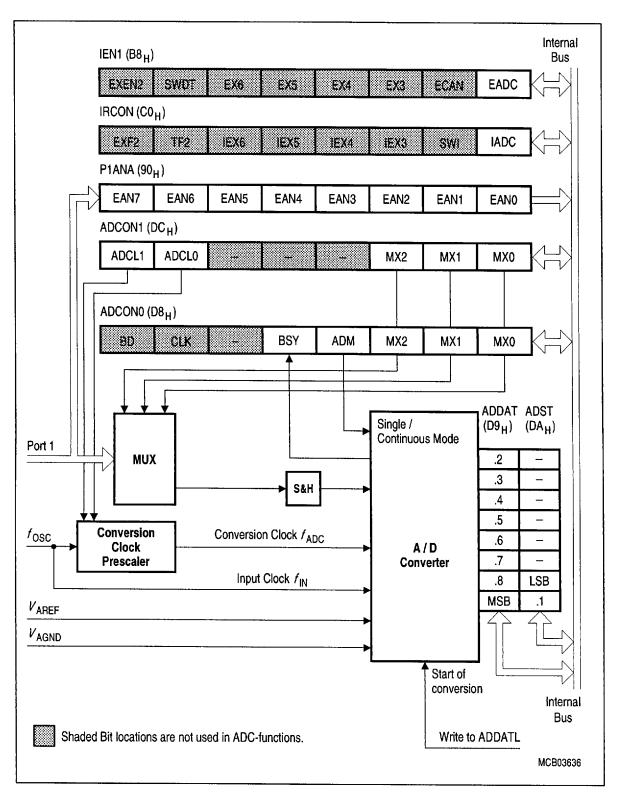


Figure 20
Block Diagram of the 10-Bit A/D Converter

### **Interrupt System**

The C505 provides 12 interrupt vectors with four priority levels. Five interrupt requests can be generated by the on-chip peripherals (timer 0, timer 1, timer 2, serial interface, A/D converter). One interrupt can be generated by the CAN controller (C505C and C505CA only) or by a software setting and in this case the interrupt vector is the same. Six interrupts may be triggered externally (P3.2/INT0, P3.3/INT1, P1.0/AN0/INT3/CC0, P1.1/AN1/INT4/CC1, P1.2/AN2/INT5/CC2, P1.3/AN3/INT6/CC3). Additionally, the P1.5/AN5/T2EX can trigger an interrupt. The wake-up from power-down mode interrupt has a special functionality which allows to exit from the software power-down mode by a short low pulse at either pin P3.2/INT0 or the pin P4.1/RXDC.

Figure 21 to 23 give a general overview of the interrupt sources and illustrate the request and the control flags which are described in the next sections. Table 10 lists all interrupt sources with their request flags and interrupt vectior addresses.

Table 10 Interrupt Source and Vectors

Interrupt Source	Interrupt Vector Address	Interrupt Request Flags
External Interrupt 0	0003 <sub>H</sub>	IE0
Timer 0 Overflow	000B <sub>H</sub>	TF0
External Interrupt 1	0013 <sub>H</sub>	IE1
Timer 1 Overflow	001B <sub>H</sub>	TF1
Serial Channel	0023 <sub>H</sub>	RI/TI
Timer 2 Overflow / Ext. Reload	002B <sub>H</sub>	TF2 / EXF2
A/D Converter	0043 <sub>H</sub>	IADC
CAN Controller / Software Interrupt	004B <sub>H</sub>	-/SWI
External interrupt 3	0053 <sub>H</sub>	IEX3
External Interrupt 4	005B <sub>H</sub>	IEX4
External Interrupt 5	0063 <sub>H</sub>	IEX5
External interrupt 6	006B <sub>H</sub>	IEX6
Wake-up from power-down mode	007B <sub>Н</sub>	_

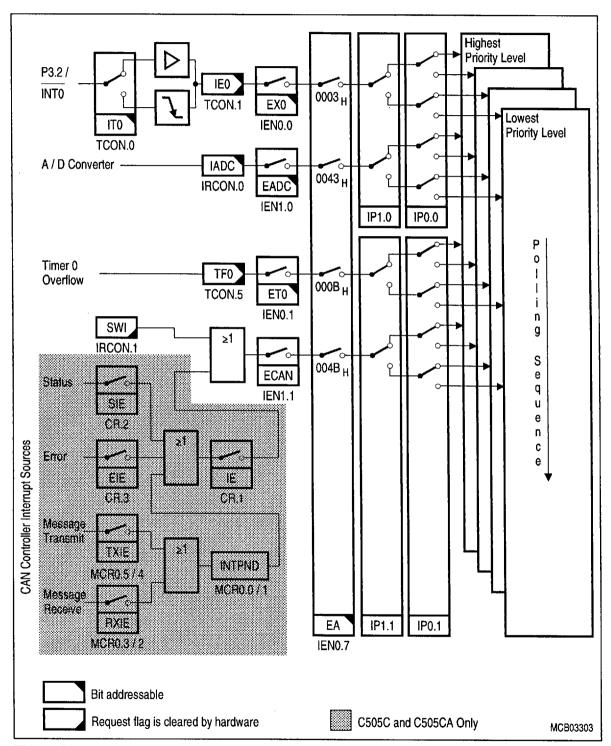


Figure 21 Interrupt Structure, Overview Part 1

**Note:** Each of the 15 CAN controller message objects (C505C and C505CA only), shown in the shaded area of **Figure 21** provides the bits/flags.

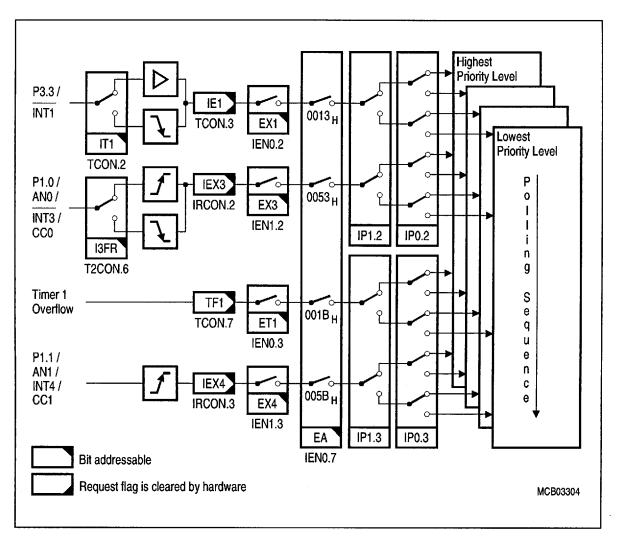


Figure 22 Interrupt Structure, Overview Part 2

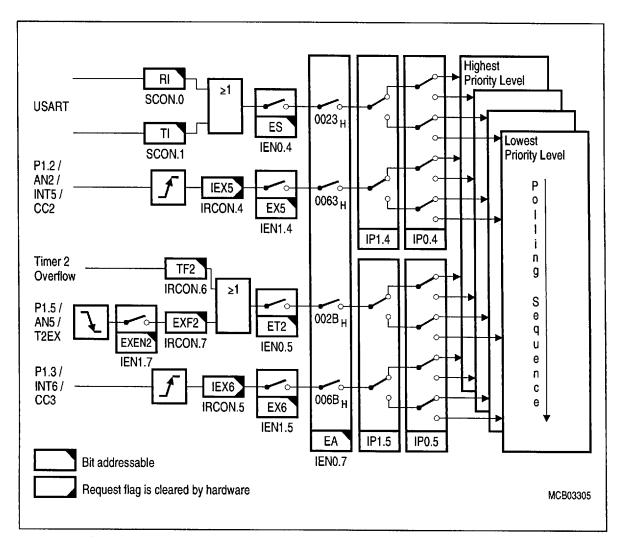


Figure 23 Interrupt Structure, Overview Part 3

#### **Fail Save Mechanisms**

The C505 offers enhanced fail safe mechanisms, which allow an automatic recovery from software upset or hardware failure:

- a programmable watchdog timer (WDT), with variable time-out period from 192 μs up to approx. 412.5 ms at 16 MHz.
- an oscillator watchdog (OWD) which monitors the on-chip oscillator and forces the microcontroller into reset state in case the on-chip oscillator fails; it also provides the clock for a fast internal reset after power-on.

The watchdog timer in the C505 is a 15-bit timer, which is incremented by a count rate of  $f_{\rm osc}/12$  upto  $f_{\rm osc}/192$ . The system clock of the C505 is divided by two prescalers, a divide-by-two and a divide-by-16 prescaler. For programming of the watchdog timer overflow rate, the upper 7 bits of the watchdog timer can be written. **Figure 24** shows the block diagram of the watchdog timer unit.

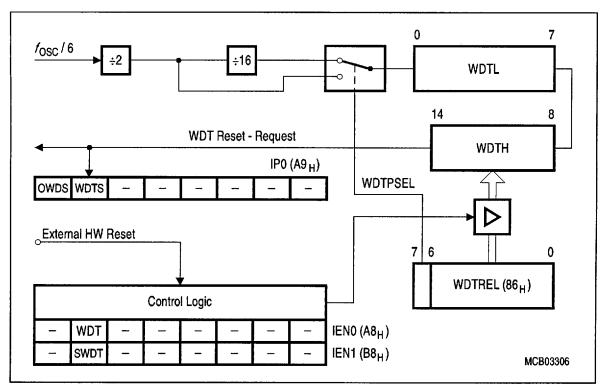


Figure 24
Block Diagram of the Programmable Watchdog Timer

The watchdog timer can be started by software (bit SWDT in SFR IEN1) but it cannot be stopped during active mode of the device. If the software fails to refresh the running watchdog timer an internal reset will be initiated on watchdog timer overflow. For refreshing of the watchdog timer the content of the SFR WDTREL is transferred to the upper 7-bit of the watchdog timer. The refresh sequence consists of two consequtive instructions which set the bits WDT and SWDT each. The reset cause (external reset or reset caused by the watchdog) can be examined by software (flag WDTS). It must be noted, however, that the watchdog timer is halted during the idle mode and power down mode of the processor.

# **Oscillator Watchdog**

The oscillator watchdog unit serves for three functions:

- Monitoring of the on-chip oscillator's function
  - The watchdog supervises the on-chip oscillator's frequency; if it is lower than the frequency of the auxiliary RC oscillator in the watchdog unit, the internal clock is supplied by the RC oscillator and the device is brought into reset; if the failure condition disappears (i.e. the on-chip oscillator has a higher frequency than the RC oscillator), the part, in order to allow the oscillator to stabilize, executes a final reset phase of typ. 1 ms; then the oscillator watchdog reset is released and the part starts program execution from address 0000H again.
- Fast internal reset after power-on
   The oscillator watchdog unit provides a clock supply for the reset before the on-chip oscillator has started. The oscillator watchdog unit also works identically to the monitoring function.
- Control of external wake-up from software power-down mode
  When the power-down mode is left by a low level at the P3.2/INTO pin or the P4.1/RXDC pin, the oscillator watchdog unit assures that the microcontroller resumes operation (execution of the power-down wake-up interrupt) with the nominal clock rate. In the power-down mode the RC oscillator and the on-chip oscillator are stopped. Both oscillators are started again when power-down mode is released. When the on-chip oscillator has a higher frequency than the RC oscillator, the microcontroller starts program execution by processing a power down interrupt after a final delay of typ. 1 ms in order to allow the on-chip oscillator to stabilize.

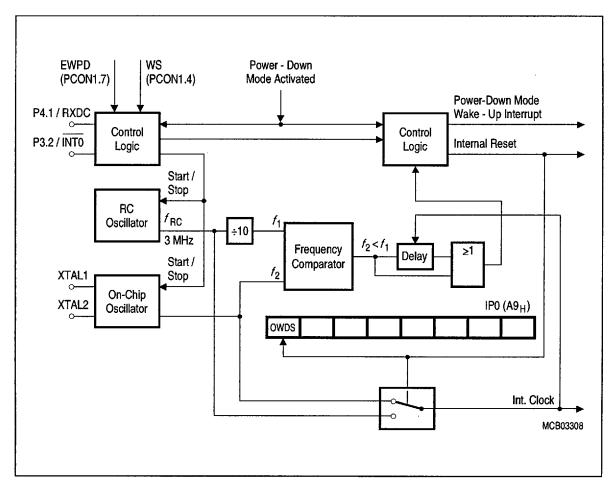


Figure 25
Functional Block Diagram of the Oscillator Watchdog

### **Power Saving Modes**

The C505 provides two basic power saving modes, the idle mode and the power down mode. Additionally, a slow down mode is available. This power saving mode reduces the internal clock rate in normal operating mode and it can be also used for further power reduction in idle mode.

#### - Idle mode

In the idle mode the main oscillator of the C505 continues to run, but the CPU is gated off from the clock signal. All peripheral units are further provided with the clock. The CPU status is preserved in its entirety. The idle mode can be terminated by any enabled interrupt of a peripheral unit or by a hardware reset.

#### - Power down mode

The operation of the C505 is completely stopped and the oscillator is turned off. This mode is used to save the contents of the internal RAM with a very low standby current. Power down mode is entered by software and can be left by reset or by a short low pulse at pin P3.2/INTO.or P4.1/RXDC.

#### - Slow down mode

The controller keeps up the full operating functionality, but its normal clock frequency is internally divided by 32. This slows down all parts of the controller, the CPU and all peripherals, to 1/32-th of their normal operating frequency. Slowing down the frequency significantly reduces power consumption.

In the power down mode of operation,  $V_{\rm CC}$  can be reduced to minimize power consumption. It must be ensured, however, that  $V_{\rm CC}$  is not reduced before the power down mode is invoked, and that  $V_{\rm CC}$  is restored to its normal operating level, before the power down mode is terminated. **Table 11** gives a general overview of the entry and exit procedures of the power saving modes.

Table 11
Power Saving Modes Overview

Mode	Entering (Instruction Example)	Leaving by	Remarks
Idle Mode	ORL PCON, #01H ORL PCON, #20H	Ocurrence of an interrupt from a peripheral unit	CPU clock is stopped; CPU maintains their data; peripheral units are active (if
		Hardware Reset	enabled) and provided with clock
Power Down Mode	ORL PCON, #02H	Hardware Reset	Oscillator is stopped;
	ORL PCON, #40H Short low pulse pin P3.2/INTO o P4.1/RXDC		contents of on-chip RAM and SFR's are maintained;
Slow Down Mode	ORL PCON,#10H	ANL PCON,#0EFH or Hardware Reset	Oscillator frequency is reduced to 1/32 of its nominal frequency

### OTP Memory Operation (C505A and C505CA only)

The C505A/C505CA contains a 32k byte one-time programmable (OTP) program memory. With the C505A/C505CA fast programming cycles are achieved (1 byte in 100  $\mu$ sec). Also several levels of OTP memory protection can be selected.

For programming of the device, the C505A/C505CA must be put into the programming mode. This typically is done not in-system but in a special programming hardware. In the programming mode the C505A/C505CA operates as a slave device similar as an EPROM standalone memory device and must be controlled with address/data information, control lines, and an external 11.5V programming voltage. **Figure 26** shows the pins of the C505A/C505CA which are required for controlling of the OTP programming mode.

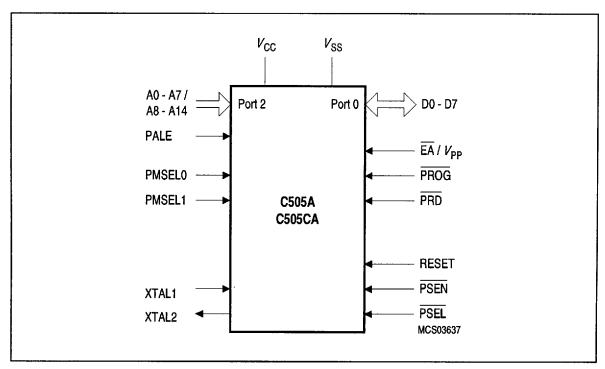


Figure 26
Programming Mode Configuration

# Pin Configuration in Programming Mode

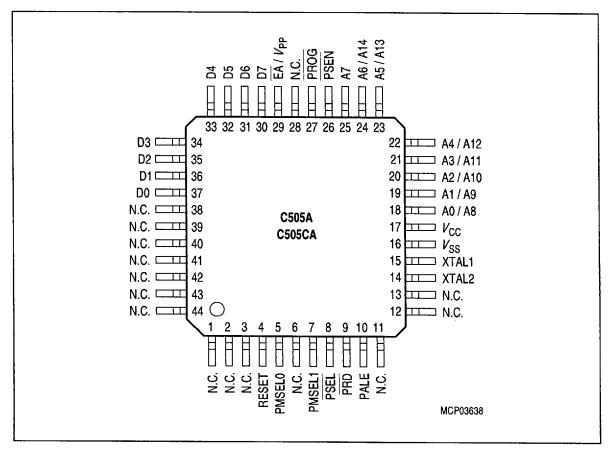


Figure 27
P-MQFP-44 Pin Configuration of the C505A/C505CA in Programming Mode (Top View)

The following table 12 contains the functional description of all C505A/C505CA pins which are required for OTP memory programming.

Table 12
Pin Definitions and Functions in Programming Mode

Symbol	Pin Number	I/O *)	Function				
RESET	4	ı	Reset This input must be at static "1" (active) level during the whole programming mode.				
PMSEL0 PMSEL1	5 7	1	Programming mode selection pins These pins are used to select the different access modes in programming mode. PMSEL1,0 must satisfy a setup time to the rising edge of PALE. When the logic level of PMSEL1,0 is changed, PALE must be at low level.				
			PMSEL1	PMSEL0	Access Mode		
			0	0	Reserved		
			0	1	Read version bytes		
			1	0	Program/read lock bits		
			1	1	Program/read OTP memory byte		
PSEL	8		Basic programming mode select  This input is used for the basic programming mode selection and must be switched according figure 3-1.				
PRD	9		This input is	Programming mode read strobe This input is used for read access control for OTP memory read, Version Register read, and lock bit read operations.			
PALE	10	1	Programming address latch enable PALE is used to latch the high address lines. The high address lines must satisfy a setup and hold time to/from the falling edge of PALE. PALE must be at low level when the logic level of PMSEL1,0 is changed.				
XTAL2	14	0	XTAL2 Output of the inverting oscillator amplifier.				
XTAL1	15	ı	XTAL1 Input to the oscillator amplifier.				
$\overline{V_{\rm SS}}$	16	-	Circuit ground potential must be applied in programming mode.				
$V_{\sf GC}$	17	-	Power supply terminal must be applied in programming mode.				

<sup>\*)</sup> I = Input

O= Output

Table 12
Pin Definitions and Functions in Programming Mode (cont'd)

Symbol	Pin Number	I/O *)	Function			
P2.0-7	18-25	I	Address lines P2.0-7 are used as multiplexed address input lines A0-A7 and A8-A14. A8-A14 must be latched with PALE.			
PSEN	26	]	Program store enable This input must be at static "0" level during the whole programming mode.			
PROG	27		Programming mode write strobe This input is used in programming mode as a write strobe for OTP memory program, and lock bit write operations During basic programming mode selection a low level must be applied to PROG.			
EA/V <sub>PP</sub>	29		External Access / Programming voltage This pin must be at 11.5V (V <sub>PP</sub> ) voltage level during programming of an OTP memory byte or lock bit. During an OTP memory read operation this pin must be at V <sub>IH</sub> high level. This pin is also used for basic programming mode selection. At basic programming mode selection a low level must be applied to EAV <sub>PP</sub> .			
D7-0	30-37	I/O	Data lines 0-7 During programming mode, data bytes are transferred via the bidirectional port 0 data lines.			
N.C.	1-3, 6, 11-13, 28, 38-44	-	Not Connected These pins should not be connected in programming mode.			

<sup>\*)</sup> I = Input O= Output

# **Basic Programming Mode Selection**

The basic programming mode selection scheme is shown in figure 28.

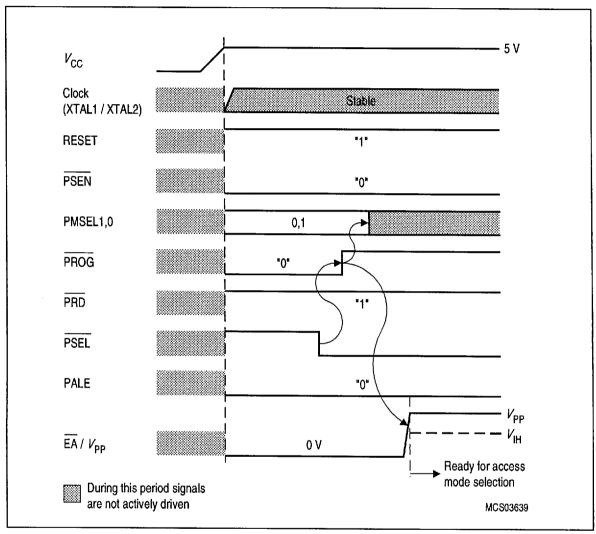


Figure 28
Basic Programming Mode Selection

Table 13
Access Modes Selection

Access Made	EA/	<u> </u>	PRD	P		MSEL	Address	Data (Port 0)
Access Mode	V <sub>PP</sub>	PROG		1	0	(Port 2)		
Program OTP memory byte	V <sub>PP</sub>	T.	Н	Н	Н	A0-7	D0-7	
Read OTP memory byte	V <sub>iH</sub>	Н	T	1		A8-14		
Program OTP lock bits	V <sub>PP</sub>	T	Н	Н	L	_	D1,D0 see	
Read OTP lock bits	V <sub>IH</sub>	Н	T	1			table 14	
Read OTP version byte	V <sub>IH</sub>	Н	T	L	Н	Byte addr. of version byte	D0-7	

# Lock Bits Programming / Read

The C505A/C505CA has two programmable lock bits which, when programmed according **table 14**, provide four levels of protection for the on-chip OTP code memory. The state of the lock bits can also be read.

Table 14 Lock Bit Protection Types

Lock B	Bits at D1,D0	Protection	Protection Type
D1	D0	Level	
1	1	Level 0	The OTP lock feature is disabled. During normal operation of the C505A/C505CA, the state of the $\overline{\text{EA}}$ pin is not latched on reset.
1	0	Level 1	During normal operation of the C505A/C505CA, MOVC instructions executed from external program memory are disabled from fetching code bytes from internal memory. $\overline{EA}$ is sampled and latched on reset. An OTP memory read operation is only possible using the ROM/OTP verification mode 2 for protection level 1. Further programming of the OTP memory is disabled (reprogramming security).
0	1	Level 2	Same as level 1, but also OTP memory read operation using OTP verification mode is disabled.
0	0	Level 3	Same as level 2; but additionally external code execution by setting EA=low during normal operation of the C505A/C505CA is no more possible.  External code execution, which is initiated by an internal program (e.g. by an internal jump instruction above the ROM boundary), is still possible.

# **Absolute Maximum Ratings**

Ambient temperature under bias (T <sub>A</sub> )	- 40 °C to 125 °C
Storage temperature ( $T_{\text{stq}}$ )	- 65 ℃ to 150 ℃
Voltage on $V_{\rm cc}$ pins with respect to ground ( $V_{\rm ss}$ )	- 0.5 V to 6.5 V
Voltage on any pin with respect to ground $(V_{ss})$	$-$ 0.5 V to $V_{\rm CC}$ +0.5 V
Input current on any pin during overload condition	- 10 mA to 10 mA
Absolute sum of all input currents during overload condition	I 100 mA I
Power dissipation	TBD

**Note:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage of the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for longer periods may affect device reliability. During overload conditions  $(V_{IN} > V_{CC} \text{ or } V_{IN} < V_{SS})$  the Voltage on  $V_{CC}$  pins with respect to ground  $(V_{SS})$  must not exceed the values defined by the absolute maximum ratings.

## **DC Characteristics**

 $V_{\rm CC}$  = 5 V + 10%, - 15%;  $V_{\rm SS}$  = 0 V

 $T_{\rm A}$  = 0 to 70 °C for the SAB- versions  $T_{\rm A}$  = -40 to 85 °C for the SAF- versions

 $T_A = -40$  to 110 °C for the SAH- versions  $T_A = -40$  to 125 °C for the SAK- versions

Parameter	Symbol	Symbol Limit Values			<b>Test Condition</b>
		min.	max.	1	
Input low voltages all except EA, RESET EA pin RESET pin	$egin{array}{c} V_{IL} \ V_{IL1} \ V_{IL2} \end{array}$	- 0.5 - 0.5 - 0.5	0.2 V <sub>CC</sub> - 0.1 0.2 V <sub>CC</sub> - 0.3 0.2 V <sub>CC</sub> + 0.1	V V	_ _ _
Input high voltages all except XTAL1, RESET XTAL1 pin RESET pin	$V_{1H}$ $V_{1H1}$ $V_{1H2}$	0.2 V <sub>cc</sub> + 0.9 0.7 V <sub>cc</sub> 0.6 V <sub>cc</sub>	$V_{\rm cc} + 0.5$ $V_{\rm cc} + 0.5$ $V_{\rm cc} + 0.5$	V	_ _ _ _
Output low voltages Ports 1, 2, 3, 4 Port 0, ALE, PSEN	$V_{ extsf{OL}}$	_	0.45 0.45	V	$I_{\rm OL}$ = 1.6 mA <sup>1)</sup> $I_{\rm OL}$ = 3.2 mA <sup>1)</sup>
Output high voltages Ports 1, 2, 3, 4	$V_{OH}$	2.4 0.9 V <sub>cc</sub>	_	V	$I_{OH} = -80 \mu\text{A}$ $I_{OH} = -10 \mu\text{A}$
Port 0 in external bus mode, ALE, PSEN	$V_{OH2}$	2.4 0.9 V <sub>cc</sub>		V	$I_{OH} = -800 \mu\text{A}$ $I_{OH} = -80 \mu\text{A}^{2}$
Logic 0 input current Ports 1, 2, 3, 4	$I_{IL}$	- 10	- 70	μΑ	$V_{\rm IN} = 0.45 \ { m V}$
Logical 0-to-1 transition current Ports 1, 2, 3, 4	$I_{TL}$	<b>– 65</b>	- 650	μА	$V_{IN}$ = 2 V
Input leakage current Port 0, AN0-7 (Port 1), EA	ILI	_	±1	μА	$0.45 < V_{IN} < V_{CC}$
Pin capacitance	$C_{10}$	-	10	pF	$f_c$ 1 MHz, $T_A = 25$ °C
Overload current	$I_{\sf OV}$	_	±5	mA	3) 4)
Programming voltage	$V_{\sf PP}$	10.9	12.1	V	11.5 V ±5% <sup>5)</sup>
Supply current at $\overline{\text{EA}}/V_{\text{CC}}$			30	mA	5) 6)

Notes see next but one page 61

# **Power Supply Currents**

	Parameter		Symbol	Limit '	Values	Unit	Test Condition
				typ. 12)	max. <sup>13)</sup>		
C505 / C505C	Active Mode	12 MHz 20 MHz	I <sub>CC</sub>	19.7 32	TBD TBD	mA	7)
	Idle Mode	12 MHz 20 MHz	I <sub>cc</sub>	11.7 17.8	TBD TBD	mA	8)
	Active Mode with slow-down enabled	12 MHz 20 MHz	I <sub>cc</sub>	4.4 4.9	TBD TBD	mA	9)
	Idle Mode with slow-down enabled	12 MHz 20 MHz	I <sub>cc</sub> I <sub>cc</sub>	3.6 4.0	TBD TBD	mA	10)
	Power down current		$I_{PD}$	7	TBD	μА	$V_{\rm CC}$ = 25.5 V <sup>11)</sup>
C505A C505CA	Active Mode	12 MHz 20 MHz	$I_{\rm cc}$ $I_{\rm cc}$	18.2 28.8	TBD TBD	mA	7)
	Idle Mode	12 MHz 20 MHz	I <sub>cc</sub> I <sub>cc</sub>	9.4 14.1	TBD TBD	mA	8)
	Active Mode with slow-down enabled	12 MHz 20 MHz	$I_{\rm cc}$ $I_{\rm cc}$	3.5 4.2	TBD TBD	mA	9)
	Idle Mode with slow-down enabled	12 MHz 20 MHz	$I_{\text{CC}}$	3.0 3.4	TBD TBD	mA	10)
	Power down current		$I_{PD}$	40	TBD	μΑ	$V_{\rm CC}$ = 25.5 V <sup>11)</sup>

Notes see next page 61

#### Notes:

- 1) Capacitive loading on ports 0 and 2 may cause spurious noise pulses to be superimposed on the V<sub>OL</sub> of ALE and port 3. The noise is due to external bus capacitance discharging into the port 0 and port 2 pins when these pins make 1-to-0 transitions during bus operation. In the worst case (capacitive loading > 100 pF), the noise pulse on ALE line may exceed 0.8 V. In such cases it may be desirable to qualify ALE with a schmitt-trigger, or use an address latch with a schmitt-trigger strobe input.
- 2) Capacitive loading on ports 0 and 2 may cause the  $V_{\rm OH}$  on ALE and  $\overline{\rm PSEN}$  to momentarily fall below the 0.9  $V_{\rm CC}$  specification when the address lines are stabilizing.
- 3) Overload conditions occur if the standard operating conditions are exceeded, ie. the voltage on any pin exceeds the specified range (i.e.  $V_{\rm OV} > V_{\rm CC} + 0.5$  V or  $V_{\rm OV} < V_{\rm SS}$  0.5 V). The supply voltage  $V_{\rm CC}$  and  $V_{\rm SS}$  must remain within the specified limits. The absolute sum of input currents on all port pins may not exceed 50 mA.
- 4) Not 100% tested, guaranteed by design characterization.
- 5) Only valid for C505A and C505CA.
- 6) Only valid for C505A and C505CA in programming mode.
- 7)  $I_{\rm CC}$  (active mode) is measured with: XTAL1 driven with  $t_{\rm R}$ ,  $t_{\rm F}$  = 5 ns, 50% duty cycle ,  $V_{\rm IL}$  =  $V_{\rm SS}$  + 0.5 V,  $V_{\rm IH}$  =  $V_{\rm CC}$  0.5 V; XTAL2 = N.C.; EA = Port 0 = RESET =  $V_{\rm CC}$ ; all other pins are disconnected.
- 8)  $I_{\rm CC}$  (idle mode) is measured with all output pins disconnected and with all peripherals disabled; XTAL1 driven with  $t_{\rm R}$ ,  $t_{\rm F}$  = 5 ns, 50% duty cycle,  $V_{\rm IL}$  =  $V_{\rm SS}$  + 0.5 V,  $V_{\rm IH}$  =  $V_{\rm CC}$  0.5 V; XTAL2 = N.C.; RESET =  $\overline{\rm EA}$  =  $V_{\rm SS}$ ; Port0 =  $V_{\rm CC}$ ; all other pins are disconnected;
- 9)  $I_{CC}$  (active mode with slow-down mode) is measured: TBD
- 10)  $I_{CC}$  (idle mode with slow-down mode) is measured : TBD
- 11)  $I_{\rm PD}$  (power-down mode) is measured under following conditions:  $\overline{\rm EA} = {\rm Port}~0 = V_{\rm CC}$ ; RESET = $V_{\rm SS}$ ; XTAL2 = N.C.; XTAL1 =  $V_{\rm SS}$ ;  $V_{\rm AGND} = V_{\rm SS}$ ;  $V_{\rm AREF} = V_{\rm CC}$ ; all other pins are disconnected.
- 12) The typical  $I_{\rm CC}$  values are periodically measured at  $T_{\rm A}$  = + 25 °C but not 100% tested.
- 13) The maximum  $I_{\rm CC}$  values are measured under worst case conditions ( $T_{\rm A}$  = 0 °C or 40 °C and  $V_{\rm CC}$  = 5.5 V)

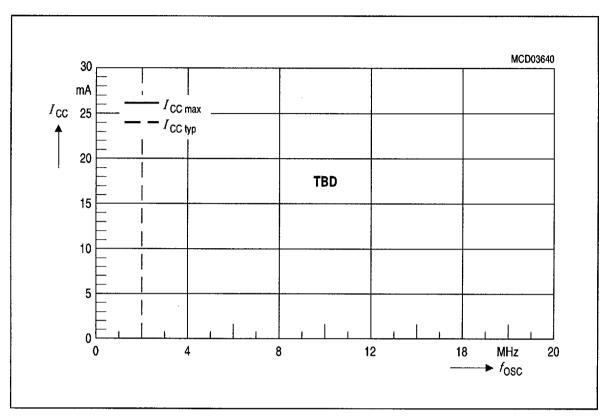


Figure 29 ICC Diagram of C505 and C505C

C505/C505C: Power Supply Current Calculation Formulas

Parameter	Symbol	Formula
Active mode	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD
Idle mode	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD
Active mode with slow-down enabled	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD
Idle mode with slow-down enabled	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD

Note:  $f_{osc}$  is the oscillator frequency in MHz.  $I_{CC}$  values are given in mA.

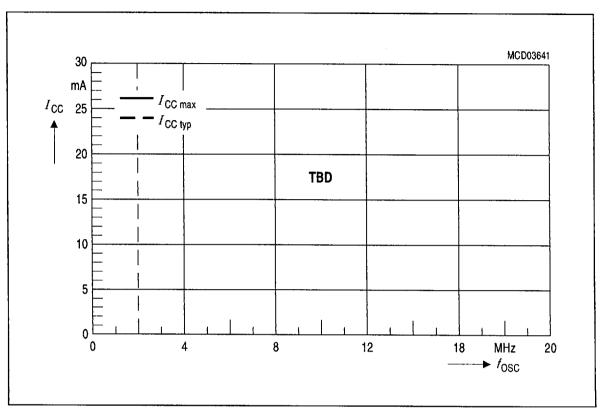


Figure 30 ICC Diagram of C505A and C505CA

C505A: Power Supply Current Calculation Formulas

Parameter	Symbol	Formula
Active mode	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD
Idle mode	$I_{ ext{CC typ}}$ $I_{ ext{CC max}}$	TBD TBD
Active mode with slow-down enabled	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD
Idle mode with slow-down enabled	$I_{ m CC\ typ}$ $I_{ m CC\ max}$	TBD TBD

Note:  $f_{osc}$  is the oscillator frequency in MHz.  $I_{CC}$  values are given in mA.

#### A/D Converter Characteristics of C505 and C505C

 $V_{\rm CC}$  = 5 V + 10%, - 15%;  $V_{\rm SS}$  = 0 V

 $T_{\rm A}$  = 0 to 70 °C for the SAB- versions

 $T_{\rm A}$  = -40 to 85 °C for the SAF- versions

 $T_{\rm A}$  = -40 to 110 °C for the SAH- versions

 $T_{\rm A}$  = -40 to 125 °C for the SAK- versions

4 V  $\leq V_{\mathsf{AREF}} \leq V_{\mathsf{CC}}$  +0.1 V;  $V_{\mathsf{SS}}$  - 0.1 V  $\leq V_{\mathsf{AGND}} \leq V_{\mathsf{SS}}$  + 0.2 V

Parameter	Symbol	Limit	Values	Unit	Test Condition	
		min. max.				
Analog input voltage	$V_{AIN}$	V <sub>AGND</sub> - 0.2	V <sub>AREF</sub> + 0.2	V	1)	
Sample time	t <sub>s</sub>	_	$64 \times t_{\rm IN}$ $32 \times t_{\rm IN}$ $16 \times t_{\rm IN}$ $8 \times t_{\rm IN}$	ns	Prescaler ÷ 32 Prescaler ÷ 16 Prescaler ÷ 8 Prescaler ÷ 4 <sup>2)</sup>	
Conversion cycle time	t <sub>ADCC</sub>	_	$320 \times t_{\rm IN}$ $160 \times t_{\rm IN}$ $80 \times t_{\rm IN}$ $40 \times t_{\rm IN}$	ns	Prescaler ÷ 32 Prescaler ÷ 16 Prescaler ÷ 8 Prescaler ÷ 4 <sup>3)</sup>	
Total unadjusted error	$T_{UE}$	_	±2	LSB	$V_{\rm SS}$ + 0.5 V $\leq$ $V_{\rm AIN}$ $\leq$ $V_{\rm CC}$ - 0.5 V	
Internal resistance of reference voltage source	R <sub>AREF</sub>	_	t <sub>ADC</sub> / 500	kΩ	t <sub>ADC</sub> in [ns] <sup>5) 6)</sup>	
Internal resistance of analog source	R <sub>ASRC</sub>	_	t <sub>s</sub> / 500	kΩ	<i>t</i> <sub>s</sub> in [ns] <sup>2) 6)</sup>	
ADC input capacitance	C <sub>AIN</sub>	_	50	pF	6)	

Notes see next page.

## Clock calculation table:

Clock Prescaler Ratio	ADCL	.1, 0	t <sub>ADC</sub>	$t_{S}$	t <sub>ADCC</sub>
÷ 32	1	1	$32 \times t_{\text{IN}}$	$64 \times t_{\text{IN}}$	$320 \times t_{\text{IN}}$
÷ 16	1	0	$16 \times t_{\text{IN}}$	$32 \times t_{1N}$	$160 \times t_{\text{IN}}$
÷ 8	0	1	$8 \times t_{\text{IN}}$	$16 \times t_{\text{IN}}$	$80 \times t_{IN}$
÷ 4	0	0	$4 \times t_{\text{IN}}$	$8 \times t_{\text{IN}}$	$40 \times t_{IN}$

Further timing conditions :  $t_{ADC}$  min = 800 ns

 $t_{\rm IN} = 1 / f_{\rm OSC} = t_{\rm CLP}$ 

#### Notes:

- 1)  $V_{\rm AIN}$  may exceed  $V_{\rm AGND}$  or  $V_{\rm AREF}$  up to the absolute maximum ratings. However, the conversion result in these cases will be 00<sub>H</sub> or FF<sub>H</sub>, respectively.
- 2) During the sample time the input capacitance  $C_{\text{AIN}}$  must be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach their final voltage level within  $t_{\text{S}}$ . After the end of the sample time  $t_{\text{S}}$ , changes of the analog input voltage have no effect on the conversion result.
- 3) This parameter includes the sample time  $t_{\rm S}$ , the time for determining the digital result. Values for the conversion clock  $t_{\rm ADC}$  depend on programming and can be taken from the table on the previous page.
- 4)  $T_{UE}$  (max.) is tested at  $-40 \le T_A \le 125$  °C;  $V_{CC} \le 5.5$  V;  $V_{AREF} \le V_{CC} + 0.1$  V and  $V_{SS} \le V_{AGND}$ . It is guaranteed by design characterization for all other voltages within the defined voltage range. If an overload condition occurs on maximum 2 unused analog input pins and the absolute sum of input overload currents on all analog input pins does not exceed 10 mA, an additional conversion error of 1/2 LSB is permissible.
- 5) During the conversion the ADC's capacitance must be repeatedly charged or discharged. The internal resistance of the reference source must allow the capacitance to reach their final voltage level within the indicated time. The maximum internal resistance results from the programmed conversion timing.
- 6) Not 100% tested, but guaranteed by design characterization.

#### A/D Converter Characteristics of C505A and C505CA

 $V_{\rm CC} = 5 \text{ V} + 10\%, -15\%; V_{\rm SS} = 0 \text{ V}$ 

 $T_A = 0$  to 70 °C for the SAB- versions

 $T_{\rm A}$  = -40 to 85 °C for the SAF- versions

 $T_{\rm A}$  = -40 to 110 °C for the SAH- versions

 $T_{\rm A}$  = -40 to 125 °C for the SAK- versions

4 V  $\leq V_{\mathsf{AREF}} \leq V_{\mathsf{CC}}$  +0.1 V;  $V_{\mathsf{SS}}$  - 0.1 V  $\leq V_{\mathsf{AGND}} \leq V_{\mathsf{SS}}$  + 0.2 V

Parameter	Symbol	Limit	Values	Unit	Test Condition	
		min.	max.	1		
Analog input voltage	$V_{AIN}$	$V_{AGND}$	$V_{AREF}$	٧	1)	
Sample time	t <sub>S</sub>	_	$64 \times t_{\text{IN}}$ $32 \times t_{\text{IN}}$ $16 \times t_{\text{IN}}$ $8 \times t_{\text{IN}}$	ns	Prescaler ÷ 32 Prescaler ÷ 16 Prescaler ÷ 8 Prescaler ÷ 4 2)	
Conversion cycle time	t <sub>ADCC</sub>	_	$384 \times t_{\rm IN}$ $192 \times t_{\rm IN}$ $96 \times t_{\rm IN}$ $48 \times t_{\rm IN}$	ns	Prescaler ÷ 32 Prescaler ÷ 16 Prescaler ÷ 8 Prescaler ÷ 4	
Total unadjusted error	$T_{\sf UE}$	_	± 2	LSB	$V_{\rm SS}$ + 0.5 V $\leq$ $V_{\rm AIN}$ $\leq$ $V_{\rm CC}$ -0.5 V	
		_	± 4	LSB	$V_{\rm SS} < V_{\rm AIN} < V_{\rm CC} + 0.5 \text{ V} \ V_{\rm CC} - 0.5 \text{ V} < V_{\rm AIN} < V_{\rm CC}^{-4)}$	
Internal resistance of reference voltage source	$R_{AREF}$	_	t <sub>ADC</sub> / 250 - 0.25	kΩ	t <sub>ADC</sub> in [ns] <sup>5) 6)</sup>	
Internal resistance of analog source	R <sub>ASRC</sub>	_	t <sub>s</sub> / 500 - 0.25	kΩ	t <sub>s</sub> in [ns] <sup>2) 6)</sup>	
ADC input capacitance	C <sub>AIN</sub>	-	50	pF	6)	

Notes see next page.

#### Clock calculation table:

Clock Prescaler Ratio	ADCL	_1, 0	t <sub>ADC</sub>	$t_{\mathbb{S}}$	t <sub>ADCC</sub>
÷ 32	1	1	$32 \times t_{\text{IN}}$	$64 \times t_{\text{IN}}$	$384 \times t_{\text{IN}}$
÷ 16	1	0	$16 \times t_{\text{IN}}$	$32 \times t_{\text{IN}}$	$192 \times t_{\text{IN}}$
÷ 8	0	1	8 x t <sub>IN</sub>	$16 \times t_{\rm IN}$	$96 \times t_{\text{IN}}$
÷ 4	0	0	$4 \times t_{\text{IN}}$	$8 \times t_{\text{IN}}$	$48 \times t_{\text{IN}}$

Further timing conditions :  $t_{ADC}$  min = 500 ns

 $t_{\rm IN} = 1 / f_{\rm OSC} = t_{\rm CLP}$ 

#### Notes:

- 1)  $V_{
  m AIN}$  may exeed  $V_{
  m AGND}$  or  $V_{
  m AREF}$  up to the absolute maximum ratings. However, the conversion result in these cases will be  $X000_{
  m H}$  or  $X3FF_{
  m H}$ , respectively.
- 2) During the sample time the input capacitance  $C_{\text{AIN}}$  must be charged/discharged by the external source. The internal resistance of the analog source must allow the capacitance to reach their final voltage level within  $t_{\text{S}}$ . After the end of the sample time  $t_{\text{S}}$ , changes of the analog input voltage have no effect on the conversion result.
- 3) This parameter includes the sample time t<sub>S</sub>, the time for determining the digital result and the time for the calibration. Values for the conversion clock t<sub>ADC</sub> depend on programming and can be taken from the table on the previous page.
- 4) T<sub>UE</sub> is tested at V<sub>AREF</sub> = 5.0 V, V<sub>AGND</sub> = 0 V, V<sub>CC</sub> = 4.9 V. It is guaranteed by design characterization for all other voltages within the defined voltage range.
  If an overload condition occurs on maximum 2 unused analog input pins and the absolute sum of input overload currents on all analog input pins does not exceed 10 mA, an additional conversion error of 1/2 LSB is permissible.
- 5) During the conversion the ADC's capacitance must be repeatedly charged or discharged. The internal resistance of the reference source must allow the capacitance to reach their final voltage level within the indicated time. The maximum internal resistance results from the programmed conversion timing.
- 6) Not 100% tested, but guaranteed by design characterization.

### AC Characteristics (12 MHz, 0.5 Duty Cycle)

 $V_{\rm CC} = 5 \text{ V} + 10\%, -15\%; V_{\rm SS} = 0 \text{ V}$ 

 $T_{\rm A}$  = 0 to 70 °C for the SAB- versions

 $T_{\rm A}$  = -40 to 85 °C for the SAF- versions

 $T_{\rm A}$  = -40 to 110 °C for the SAH- versions

 $T_{\rm A} = -40$  to 125 °C for the SAK- versions

( $C_L$  for port 0, ALE and  $\overrightarrow{PSEN}$  outputs = 100 pF;  $C_L$  for all other outputs = 80 pF)

## **Program Memory Characteristics**

Parameter	Symbol	Limit Values				
		12 MHz clock 0.5 Duty Cycle		Variable Clock 1/CLP = 2 MHz to 12 MHz		
		min.	max.	min.	max.	
ALE pulse width	t <sub>LHLL</sub>	43	-	CLP - 40	-	ns
Address setup to ALE	t <sub>AVLL</sub>	17	-	CLP/2 - 25	_	ns
Address hold after ALE	t <sub>LLAX</sub>	17	_	CLP/2 - 25	_	ns
ALE to valid instruction in	t <sub>LLIV</sub>	_	80	_	2 CLP - 87	ns
ALE to PSEN	t <sub>LLPL</sub>	22	-	CLP/2 - 20		ns
PSEN pulse width	t <sub>PLPH</sub>	95	_	3/2 CLP - 30	-	ns
PSEN to valid instruction in	t <sub>PLIV</sub>	_	60	_	3/2 CLP - 65	ns
Input instruction hold after PSEN	t <sub>PXIX</sub>	0	_	0	-	ns
Input instruction float after PSEN	t <sub>PXIZ</sub> *)	_	32	_	CLP/2 - 10	ns
Address valid after PSEN	t <sub>PXAV</sub> *)	37	_	CLP/2 - 5	_	ns
Address to valid instruction in	t <sub>AVIV</sub>	_	148	_	5/2 CLP - 60	ns
Address float to PSEN	t <sub>AZPL</sub>	0	_	0	_	ns

<sup>1)</sup> Interfacing the C505 to devices with float times up to 37 ns is permissible. This limited bus contention will not cause any damage to port 0 drivers.

# AC Characteristics (12 MHz, 0.5 Duty Cycle, cont'd)

# **External Data Memory Characteristics**

Parameter	Symbol	Limit Values					
		1	Hz clock Ity Cycle	Variable Clock 1/CLP = 2 MHz to 12 MHz			
		min.	max.	min.	max.		
RD pulse width	t <sub>RLRH</sub>	180	_	3 CLP - 70	_	ns	
WR pulse width	t <sub>WLWH</sub>	180	_	3 CLP - 70	_	ns	
Address hold after ALE	t <sub>LLAX2</sub>	56	_	CLP - 27	_	ns	
RD to valid data in	t <sub>RLDV</sub>	-	118	_	5/2 CLP- 90	ns	
Data hold after RD	$t_{RHDX}$	0	_	0	_	ns	
Data float after RD	t <sub>RHDZ</sub>	-	63	_	CLP - 20	ns	
ALE to valid data in	t <sub>LLDV</sub>	_	200	_	4 CLP - 133	ns	
Address to valid data in	t <sub>AVDV</sub>	-	220	_	9/2 CLP - 155	ns	
ALE to WR or RD	t <sub>LLWL</sub>	75	175	3/2 CLP - 50	3/2 CLP + 50	ns	
Address valid to WR	t <sub>AVWL</sub>	70	_	2 CLP - 97	_	ns	
WR or RD high to ALE high	t <sub>WHLH</sub>	17	67	CLP/2 - 25	CLP/2 + 25	ns	
Data valid to WR transition	t <sub>QVWX</sub>	5	_	CLP/2 - 37	_	ns	
Data setup before WR	t <sub>QVWH</sub>	170	_	7/2 CLP - 122	_	ns	
Data hold after WR	t <sub>WHQX</sub>	15	_	CLP/2 - 27	_	ns	
Address float after RD	t <sub>RLAZ</sub>	_	0	_	0	ns	

## **External Clock Drive Characteristics**

Parameter	Symbol		Limit Values		
		Freq			
		min.	max.		
Oscillator period	CLP	83.3	500	ns	
High time	TCLH	20	CLP-TCLL	ns	
Low time	TCLL	20	CLP-TCLH	ns	
Rise time	t <sub>R</sub>	_	12	ns	
Fall time	$t_{F}$	_	12	ns	
Oscillator duty cycle	DC	0.5	0.5	-	

AC Characteristics (16 MHz, 0.4 to 0.6 Duty Cycle)

 $V_{\rm CC} = 5 \text{ V} + 10\%, -15\%; V_{\rm SS} = 0 \text{ V}$ 

 $T_{\rm A}$  = 0 to 70 °C for the SAB- versions

 $T_A = 40$  to 85 °C for the SAF- versions

( $C_L$  for port 0, ALE and  $\overline{\text{PSEN}}$  outputs = 100 pF;  $C_L$  for all other outputs = 80 pF)

# **Program Memory Characteristics**

Parameter	Symbol	Limit Values				
		16-MHz clock Duty Cycle 0.4 to 0.6		Variable Clock 1/CLP= 2 MHz to 16 MHz		
		min.	max.	min.	max.	1
ALE pulse width	t <sub>LHLL</sub>	48		CLP - 15	_	ns
Address setup to ALE	t <sub>AVLL</sub>	10	_	TCL <sub>Hmin</sub> -15	_	ns
Address hold after ALE	t <sub>LLAX</sub>	10	_	TCL <sub>Hmin</sub> -15	_	ns
ALE to valid instruction in	t <sub>LLIV</sub>	_	75	_	2 CLP - 50	ns
ALE to PSEN	t <sub>LLPL</sub>	10	_	TCL <sub>Lmin</sub> -15	-	ns
PSEN pulse width	t <sub>PLPH</sub>	73		CLP+ TCL <sub>Hmin</sub> -15	_	ns
PSEN to valid instruction in	t <sub>PLIV</sub>	_	38	_	CLP+ TCL <sub>Hmin</sub> - 50	ns
Input instruction hold after PSEN	t <sub>PXIX</sub>	0	_	0	_	ns
Input instruction float after PSEN	t <sub>PXIZ</sub> *)	_	15	_	TCL <sub>Lmin</sub> -10	ns
Address valid after PSEN	t <sub>PXAV</sub> *)	20	_	TCL <sub>Lmin</sub> - 5	_	ns
Address to valid instruction in	t <sub>AVIV</sub>	_	95		2 CLP + TCL <sub>Hmin</sub> -55	ns
Address float to PSEN	t <sub>AZPL</sub>	-5	_	-5	_	ns

<sup>1)</sup> Interfacing the C505 to devices with float times up to 20 ns is permissible. This limited bus contention will not cause any damage to port 0 drivers.

# AC Characteristics (16 MHz, 0.4 to 0.6 Duty Cycle, cont'd)

# **External Data Memory Characteristics**

Parameter	Symbol	Limit Values					
		16-MHz clock Duty Cycle 0.4 to 0.6		Variable Clock 1/CLP= 2 MHz to 16 MHz			
		min.	max.	min.	max.		
RD pulse width	t <sub>RLRH</sub>	158	-	3 CLP - 30	_	ns	
WR pulse width	t <sub>WLWH</sub>	158	_	3 CLP - 30	_	ns	
Address hold after ALE	t <sub>LLAX2</sub>	48	-	CLP - 15	_	ns	
RD to valid data in	t <sub>RLDV</sub>	_	100	_	2 CLP+ TCL <sub>Hmin</sub> - 50	ns	
Data hold after RD	t <sub>RHDX</sub>	0	-	0	_	ns	
Data float after RD	t <sub>RHDZ</sub>	-	51	_	CLP - 12	ns	
ALE to valid data in	t <sub>LLDV</sub>	-	200	_	4 CLP - 50	ns	
Address to valid data in	$t_{AVDV}$	-	200	_	4 CLP + TCL <sub>Hmin</sub> -75	ns	
ALE to WR or RD	t <sub>LLWL</sub>	73	103	CLP + TCL <sub>Lmin</sub> - 15	CLP+ TCL <sub>Lmin</sub> + 15	ns	
Address valid to WR	t <sub>AVWL</sub>	95	_	2 CLP - 30	_	ns	
WR or RD high to ALE high	t <sub>WHLH</sub>	10	40	TCL <sub>Hmin</sub> - 15	TCL <sub>Hmin</sub> + 15	ns	
Data valid to WR transition	t <sub>QVWX</sub>	5	<u> </u>	TCL <sub>Lmin</sub> - 20	_	ns	
Data setup before WR	t <sub>QVWH</sub>	163	_	3 CLP + TCL <sub>Lmin</sub> - 50	_	ns	
Data hold after WR	t <sub>WHQX</sub>	5	_	TCL <sub>Hmin</sub> - 20		ns	
Address float after RD	t <sub>RLAZ</sub>	-	0	_	0	ns	

# AC Characteristics (16 MHz, 0.4 to 0.6 Duty Cycle, cont'd)

## **External Clock Drive Characteristics**

Parameter	Symbol	1	ock = 16 MHz ycle 0.4 to 0.6	Variable 1/CLP = 2	Unit	
		min.	max.	min.	max.	
Oscillator period	CLP	62.5	62.5	62.5	500	ns
High time	TCL <sub>H</sub>	25	_	25	CLP - TCLL	ns
Low time	TCL	25	_	25	CLP - TCL <sub>H</sub>	ns
Rise time	t <sub>R</sub>	_	10	_	10	ns
Fall time	t <sub>F</sub>	_	10	_	10	ns
Oscillator duty cycle	DC	0.4	0.6	25 / CLP	1 - 25 / CLP	_
Clock cycle	TCL	25	37.5	CLP * DC <sub>min</sub>	CLP * DC <sub>max</sub>	ns

**Note:** The 16 MHz values in the tables are given as an example for a typical duty cycle variation of the oscillator clock from 0.4 to 0.6.

# AC Characteristics (20 MHz, 0.5 Duty Cycle)

 $V_{\rm CC} = 5 \text{ V} + 10\%, -15\%; V_{\rm SS} = 0 \text{ V}$ 

 $T_{\rm A}$  = 0 to 70 °C for the SAB- versions

 $T_A = -40$  to 85 °C for the SAF- versions

( $C_{\rm L}$  for port 0, ALE and  $\overline{\rm PSEN}$  outputs = 100 pF;  $C_{\rm L}$  for all other outputs = 80 pF)

### **Program Memory Characteristics**

Parameter	Symbol	Limit Values				
		20 MHz clock 0.5 Duty Cycle		Variable Clock 1/CLP = 2 MHz to 20 MHz		
		min.	max.	min.	max.	
ALE pulse width	t <sub>LHLL</sub>	35	_	CLP - 15	_	ns
Address setup to ALE	t <sub>AVLL</sub>	10	_	CLP/2 - 15	_	ns
Address hold after ALE	t <sub>LLAX</sub>	10	_	CLP/2 - 15	_	ns
ALE to valid instruction in	t <sub>LLIV</sub>	-	55	_	2 CLP - 45	ns
ALE to PSEN	t <sub>LLPL</sub>	10	_	CLP/2 - 15	_	ns
PSEN pulse width	t <sub>PLPH</sub>	60	_	3/2 CLP - 15	_	ns
PSEN to valid instruction in	t <sub>PLIV</sub>	_	25	_	3/2 CLP - 50	ns
Input instruction hold after PSEN	$t_{PXIX}$	0	_	0	_	ns
Input instruction float after PSEN	t <sub>PXIZ</sub> *)	_	20	_	CLP/2 - 5	ns
Address valid after PSEN	t <sub>PXAV</sub> *)	20	_	CLP/2 - 5	_	ns
Address to valid instruction in	t <sub>AVIV</sub>	-	65	_	5/2 CLP - 60	ns
Address float to PSEN	t <sub>AZPL</sub>	- 5	1-	- 5	-	ns

<sup>&</sup>lt;sup>\*)</sup> Interfacing the C505 to devices with float times up to 20 ns is permissible. This limited bus contention will not cause any damage to port 0 drivers.

## AC Characteristics (20 MHz, 0.5 Duty Cycle, cont'd)

### **External Data Memory Characteristics**

Parameter	Symbol	Limit Values				
		20 MHz clock 0.5 Duty Cycle		Variable Clock 1/CLP = 2 MHz to 20 MHz		
		min.	max.	min.	max.	
RD pulse width	t <sub>RLRH</sub>	120	_	3 CLP - 30	_	ns
WR pulse width	t <sub>WLWH</sub>	120	_	3 CLP - 30	-	ns
Address hold after ALE	t <sub>LLAX2</sub>	35	_	CLP - 15	-	ns
RD to valid data in	t <sub>RLDV</sub>	_	75	_	5/2 CLP- 50	ns
Data hold after RD	t <sub>RHDX</sub>	0	_	0	_	ns
Data float after RD	t <sub>RHDZ</sub>	_	38	_	CLP - 12	ns
ALE to valid data in	t <sub>LLDV</sub>	_	150	_	4 CLP - 50	ns
Address to valid data in	t <sub>AVDV</sub>	_	150	_	9/2 CLP - 75	ns
ALE to WR or RD	t <sub>LLWL</sub>	60	90	3/2 CLP - 15	3/2 CLP + 15	ns
Address valid to WR	t <sub>AVWL</sub>	70	-	2 CLP - 30	_	ns
WR or RD high to ALE high	t <sub>whlh</sub>	10	40	CLP/2 - 15	CLP/2 + 15	ns
Data valid to WR transition	tavwx	5	_	CLP/2 - 20	_	ns
Data setup before WR	t <sub>QVWH</sub>	125	_	7/2 CLP - 50	-	ns
Data hold after WR	t <sub>WHQX</sub>	5	_	CLP/2 - 20	_	ns
Address float after RD	t <sub>RLAZ</sub>	-	0	_	0	ns

### **External Clock Drive Characteristics**

Parameter	Symbol		Unit	
		Freq		
		min.	max.	
Oscillator period	CLP	50	500	ns
High time	TCL <sub>H</sub>	15	CLP-TCL <sub>L</sub>	n
Low time	TCLL	15	CLP-TCL <sub>H</sub>	n
Rise time	t <sub>R</sub>	_	10	ns
Fall time	t <sub>F</sub>	_	10	ns
Oscillator duty cycle	DC	0.5	0.5	_

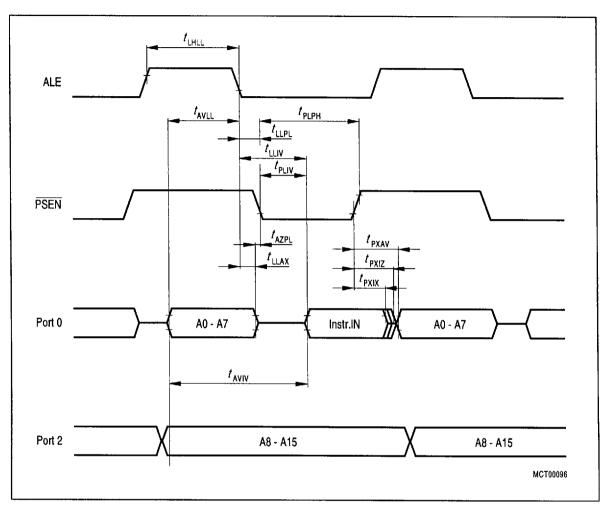


Figure 31 Program Memory Read Cycle

C505 / C505C C505A / C505CA

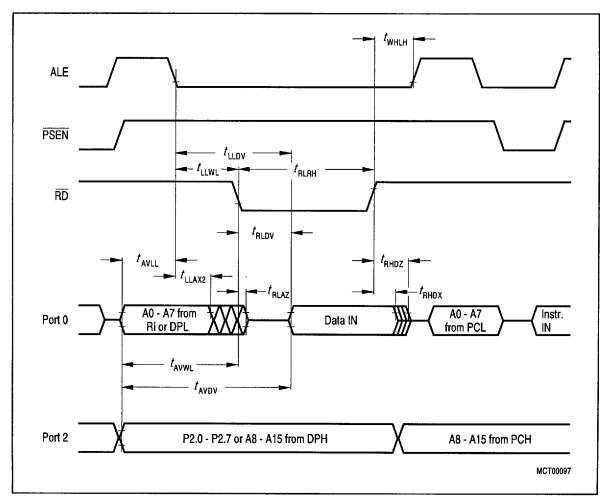


Figure 32 **Data Memory Read Cycle** 

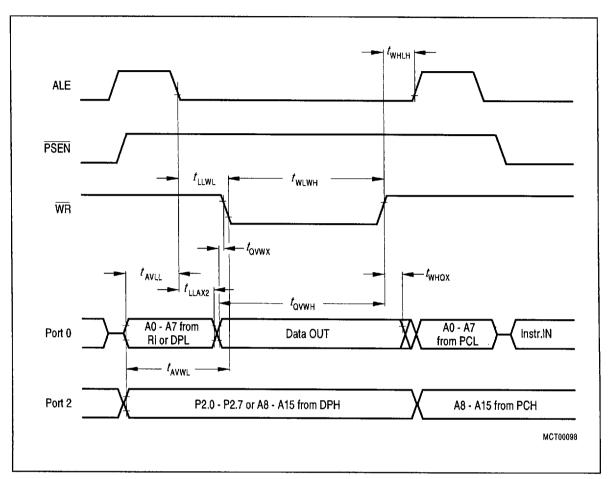


Figure 33
Data Memory Write Cycle

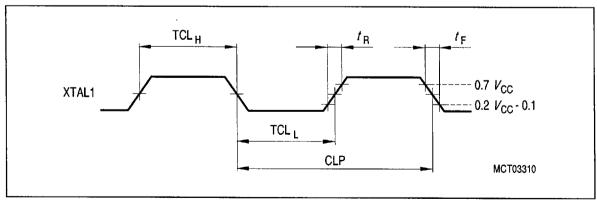


Figure 34
External Clock Drive on XTAL1

# **SIEMENS**

# AC Characteristics of Programming Mode (C505A and C505CA only)

 $V_{\rm CC}$  = 5 V ± 10 %;  $V_{\rm PP}$  = 11.5 V ± 5 %;  $T_{\rm A}$  = 25 °C ± 10 °C

Parameter	Symbol	L	imit Values	Unit
		min.	max.	
ALE pulse width	t <sub>PAW</sub>	35	-	ns
PMSEL setup to ALE rising edge	$t_{PMS}$	10	-	
Address setup to ALE, PROG, or PRD falling edge	t <sub>PAS</sub>	10	_	ns
Address hold after ALE, PROG, or PRD falling edge	t <sub>PAH</sub>	10	_	ns
Address, data setup to PROG or PRD	t <sub>PCS</sub>	100	_	ns
Address, data hold after PROG or PRD	$t_{PCH}$	0	_	ns
PMSEL setup to PROG or PRD	t <sub>PMS</sub>	10	_	ns
PMSEL hold after PROG or PRD	t <sub>PMH</sub>	10	_	ns
PROG pulse width	t <sub>PWW</sub>	100	_	μs
PRD pulse width	t <sub>PRW</sub>	100	_	ns
Address to valid data out	t <sub>PAD</sub>	_	75	ns
PRD to valid data out	t <sub>PRD</sub>	_	20	ns
Data hold after PRD	t <sub>PDH</sub>	0	_	ns
Data float after PRD	$t_{PDF}$	-	20	ns
PROG high between two consecutive PROG low pulses	t <sub>PWH1</sub>	1	-	μs
PRD high between two consecutive PRD low pulses	t <sub>PWH2</sub>	100		ns
XTAL clock period	t <sub>CLKP</sub>	83.3	500	ns

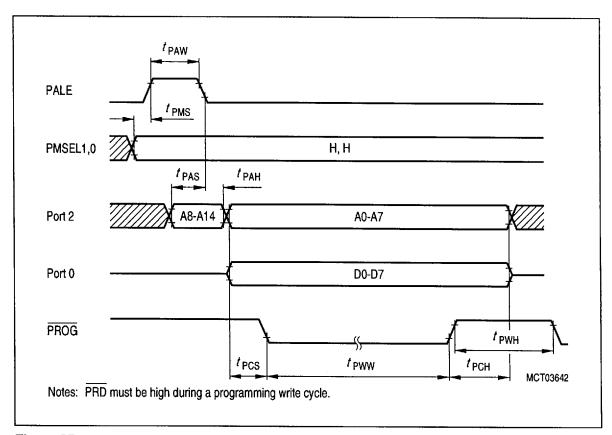


Figure 35
Programming Code Byte - Write Cycle Timing

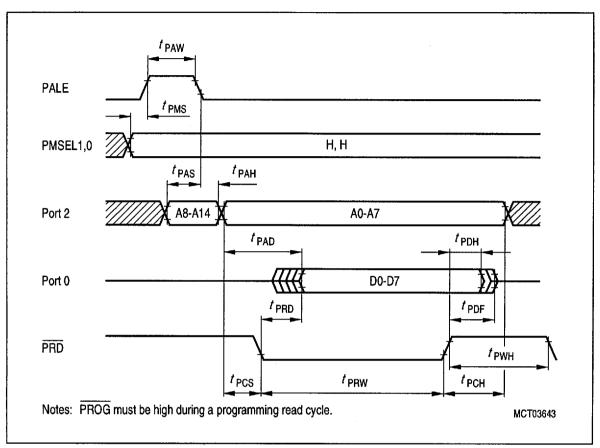


Figure 36 Verify Code Byte - Read Cycle Timing

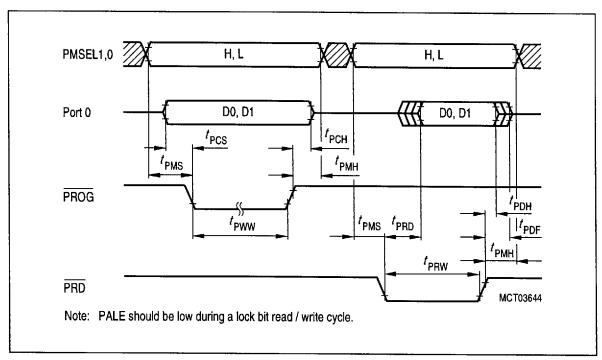


Figure 37 Lock Bit Access Timing

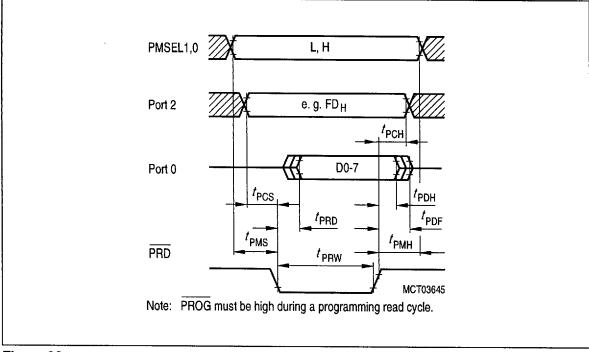


Figure 38 Version Byte Read Timing

#### **ROM/OTP Verification Characteristics for C505**

### ROM Verification Mode 1 (C505-2R and C505C-2R only)

Parameter	Symbol		Limit Values				
		min.		max.	•		
Address to valid data	t <sub>AVQV</sub>	_		5 CLP		ns	

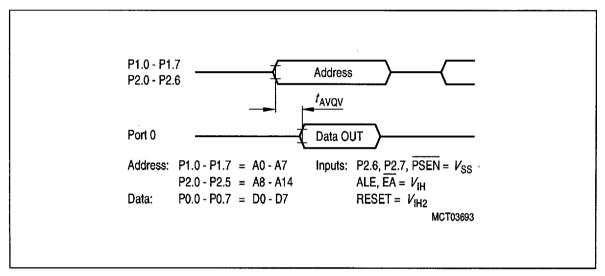


Figure 39 ROM Verification Mode 1

# **ROM/OTP Verification Characteristics for C505 (cont'd)**

#### **ROM/OTP Verification Mode 2**

Parameter	Symbol		Unit		
		min.	typ	max.	
ALE pulse width	t <sub>AWD</sub>	_	CLP	_	ns
ALE period	t <sub>ACY</sub>	_	6 CLP	-	ns
Data valid after ALE	$t_{DVA}$	-	_	2 CLP	ns
Data stable after ALE	t <sub>DSA</sub>	4 CLP		-	ns
P3.5 setup to ALE low	t <sub>AS</sub>	_	$t_{ m CL}$	_	ns
Oscillator frequency	1/ CLP	4	_	6	MHz

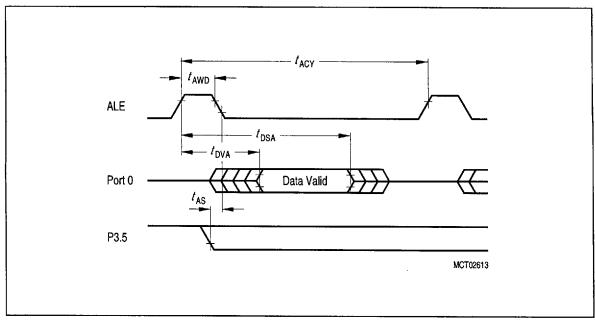
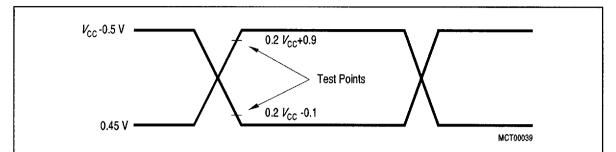


Figure 40 ROM/OTP Verification Mode 2

# **SIEMENS**



AC Inputs during testing are driven at  $V_{\rm CC}$  - 0.5 V for a logic '1' and 0.45 V for a logic '0'. Timing measurements are made at  $V_{\rm IHmin}$  for a logic '1' and  $V_{\rm ILmax}$  for a logic '0'.

Figure 41
AC Testing: Input, Output Waveforms

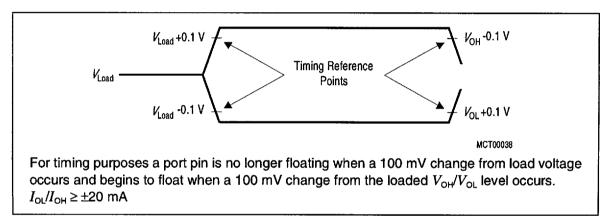


Figure 42 AC Testing : Float Waveforms

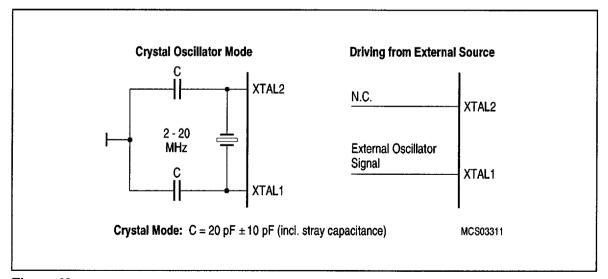


Figure 43
Recommended Oscillator Circuits for Crystal Oscillator

Semiconductor Group

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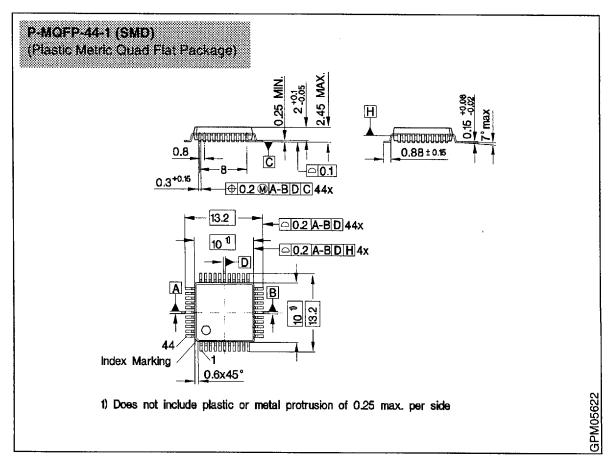


Figure 44 P-MQFP-44 Package Outline

### **Sorts of Packing**

Package outlines for tubes, trays etc. are contained in our Data Book "Package Information"

SMD # Surface Mounted Device

Dimensions in mm