

# LMV921/LMV922/LMV924

## Single, Dual and Quad 1.8V, 1MHz, Low Power Operational Amplifiers with Rail-To-Rail Input and Output

### General Description

The LMV921 Single/LMV922 Dual/LMV924 Quad are guaranteed to operate from +1.8V to +5.0V supply voltages and have rail-to-rail input and output. This rail-to-rail operation enables the user to make full use of the entire supply voltage range. The input common mode voltage range extends 300mV beyond the supplies and the output can swing rail-to-rail unloaded and within 100mV from the rail with 600Ω load at 1.8V supply. The LMV921/LMV922/LMV924 are optimized to work at 1.8V which make them ideal for portable two-cell battery-powered systems and single cell Li-Ion systems.

The LMV921/LMV922/LMV924 exhibit excellent speed-power ratio, achieving 1MHz gain bandwidth product at 1.8V supply voltage with very low supply current. The LMV921/LMV922/LMV924 are capable of driving 600Ω load and up to 1000pF capacitive load with minimal ringing. The LMV921/LMV922/LMV924's high DC gain of 100dB makes them suitable for low frequency applications.

The LMV921 (Single) is offered in a space saving SC70-5 and SOT23-5 packages. The SC70-5 package is only 2.0X2.1X1.0mm. These small packages are ideal solutions for area constrained PC boards and portable electronics such as cellphones and PDAs.

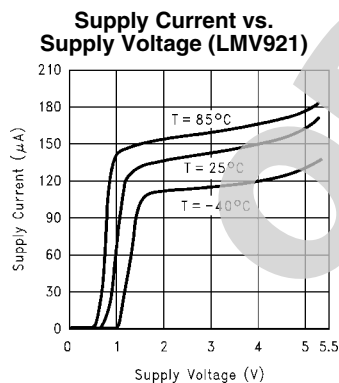
### Features

(Typical 1.8V Supply Values; Unless Otherwise Noted)

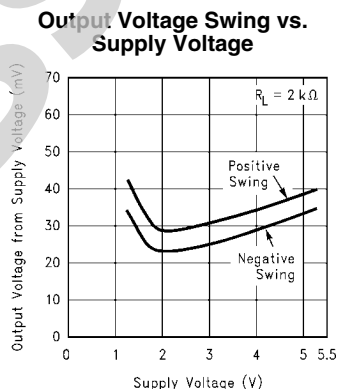
- Guaranteed 1.8V, 2.7V and 5V specifications
- Rail-to-Rail input & output swing
  - w/600Ω load 100 mV from rail
  - w/2kΩ load 30 mV from rail
- $V_{CM}$  300mV beyond rails
- Supply current 145μA/amplifier
- Gain bandwidth product 1MHz
- LMV921 Maximum  $V_{OS}$  6mV
- 90dB gain w/600Ω load
- LMV921 available in Ultra Tiny, SC70-5 package
- LMV922 available in MSOP-8 package
- LMV924 available in TSSOP-14 package

### Applications

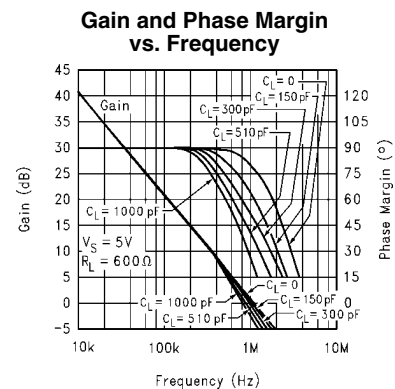
- Cordless/cellular phones
- Laptops
- PDAs
- PCMCIA
- Portable/battery-powered electronic Equipment
- Supply current Monitoring
- Battery monitoring



100979a1



100979a3



100979a7

**Absolute Maximum Ratings** (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

ESD Tolerance (Note 2)	
Machine Model	100V
Human Body Model	2000V
Differential Input Voltage	± Supply Voltage
Supply Voltage (V <sup>+</sup> -V <sup>-</sup> )	5.5V
Output Short Circuit to V <sup>+</sup> (Note 3)	
Output Short Circuit to V <sup>-</sup> (Note 3)	
Storage Temperature Range	-65°C to 150°C
Junction Temperature (Note 4)	150°C
Mounting Temp.	
Infrared or Convection (20 sec)	235°C

**Operating Ratings** (Note 1)

Supply Voltage	1.5V to 5.0V
Temperature Range	-40°C ≤ T <sub>J</sub> ≤ 85°C
Thermal Resistance (θ <sub>JA</sub> )	
Ultra Tiny SC70-5 Package 5-Pin Surface Mount	440 °C/W
Tiny SOT23-5 Package 5-Pin Surface Mount	265 °C/W
MSOP Package 8-Pin Surface Mount	235°C/W
TSSOP Package 14-Pin Surface Mount	155°C/W
SOIC Package 8-Pin Surface Mount	175°C/W
14-Pin Surface Mount	127°C/W

**1.8V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for T<sub>J</sub> = 25°C. V<sup>+</sup> = 1.8V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sup>+</sup>/2, V<sub>O</sub> = V<sup>+</sup>/2 and R<sub>L</sub> > 1 MΩ. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
V <sub>OS</sub>	Input Offset Voltage	LMV921 (Single)	-1.8	6 <b>8</b>	mV max
		LMV922 (Dual)	-1.8	8	mV max
		LMV924 (Quad)		<b>9.5</b>	
TCV <sub>OS</sub>	Input Offset Voltage Average Drift		1		μV/°C
I <sub>B</sub>	Input Bias Current		12	35 <b>50</b>	nA max
I <sub>OS</sub>	Input Offset Current		2	25 <b>40</b>	nA max
I <sub>S</sub>	Supply Current	LMV921 (Single)	145	185 <b>205</b>	μA max
		LMV922 (Dual)	330	400 <b>550</b>	
		LMV924 (Quad)	560	700 <b>850</b>	
CMRR	Common Mode Rejection Ratio	0 ≤ V <sub>CM</sub> ≤ 0.6V	82	62 <b>60</b>	dB min
		-0.2V ≤ V <sub>CM</sub> ≤ 0V	74	50	
		1.8V ≤ V <sub>CM</sub> ≤ 2.0V			
PSRR	Power Supply Rejection Ratio	1.8V ≤ V <sup>+</sup> ≤ 5V, V <sub>CM</sub> = 0.5V	78	67 <b>62</b>	dB min
V <sub>CM</sub>	Input Common-Mode Voltage Range	For CMRR ≥ 50dB	-0.3	-0.2 <b>0</b>	V min
			2.15	2.0 <b>1.8</b>	V max

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
$A_V$	Large Signal Voltage Gain LMV921 (Single)	$R_L = 600\Omega$ to 0.9V, $V_O = 0.2V$ to 1.6V, $V_{CM} = 0.5V$	91	77 <b>73</b>	dB min
		$R_L = 2k\Omega$ to 0.9V, $V_O = 0.2V$ to 1.6V, $V_{CM} = 0.5V$	95	80 <b>75</b>	
	Large Signal Voltage Gain LMV922 (Dual) LMV924 (Quad)	$R_L = 600\Omega$ to 0.9V, $V_O = 0.2V$ to 1.6V, $V_{CM} = 0.5V$	79	65 <b>61</b>	dB min
		$R_L = 2k\Omega$ to 0.9V, $V_O = 0.2V$ to 1.6V, $V_{CM} = 0.5V$	83	68 <b>63</b>	
$V_O$	Output Swing	$R_L = 600\Omega$ to 0.9V $V_{IN} = \pm 100mV$	1.7	1.65 <b>1.63</b>	V min
			0.075	0.090 <b>0.105</b>	V max
		$R_L = 2k\Omega$ to 0.9V $V_{IN} = \pm 100mV$	1.77	1.75 <b>1.74</b>	V min
			0.025	0.035 <b>0.040</b>	V max
$I_O$	Output Short Circuit Current	Sourcing, $V_O = 0V$ $V_{IN} = 100mV$	6	4 <b>3.3</b>	mA min
		Sinking, $V_O = 1.8V$ $V_{IN} = -100mV$	10	7 <b>5</b>	mA min

## 1.8V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ C$ .  $V^+ = 1.8V$ ,  $V^- = 0V$ ,  $V_{CM} = V^+/2$ ,  $V_O = V^+/2$  and  $R_L > 1 M\Omega$ .

**Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Units
SR	Slew Rate	(Note 7)	0.39	V/ $\mu s$
GBW	Gain-Bandwidth Product		1	MHz
$\Phi_m$	Phase Margin		60	Deg
$G_m$	Gain Margin		10	dB
$e_n$	Input-Referred Voltage Noise	$f = 1 kHz$ , $V_{CM} = 0.5V$	45	$\frac{nV}{\sqrt{Hz}}$
$i_n$	Input-Referred Current Noise	$f = 1 kHz$	0.1	$\frac{pA}{\sqrt{Hz}}$
THD	Total Harmonic Distortion	$f = 1kHz$ , $A_V = +1$ $R_L = 600k\Omega$ , $V_{IN} = 1 V_{PP}$	0.089	%
	Amp-to-Amp Isolation	(Note 8)	140	dB

## 2.7V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ C$ .  $V^+ = 2.7V$ ,  $V^- = 0V$ ,  $V_{CM} = V^+/2$ ,  $V_O = V^+/2$  and

$R_L > 1 M\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
$V_{OS}$	Input Offset Voltage	LMV921 (Single)	-1.6	6 <b>8</b>	mV max
		LMV922 (Dual)	-1.6	8	mV
		LMV924 (Quad)		<b>9.5</b>	mV max
$TCV_{OS}$	Input Offset Voltage Average Drift		1		$\mu V/^\circ C$
$I_B$	Input Bias Current		12	35 <b>50</b>	nA max

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
$I_{OS}$	Input Offset Current		2	25 <b>40</b>	nA max
$I_S$	Supply Current	LMV921 (Single)	147	190 <b>210</b>	uA max
		LMV922 (Dual)	380	450 <b>600</b>	
		LMV924 (Quad)	580	750 <b>900</b>	
CMRR	Common Mode Rejection Ratio	$0V \leq V_{CM} \leq 1.5V$	84	62 <b>60</b>	dB min
		$-0.2V \leq V_{CM} \leq 0V$	73	50	
		$2.7V \leq V_{CM} < 2.9V$			
PSRR	Power Supply Rejection Ratio	$1.8V \leq V^+ \leq 5V$ , $V_{CM} = 0.5V$	78	67 <b>62</b>	dB min
$V_{CM}$	Input Common-Mode Voltage Range	For CMRR $\geq 50$ dB	-0.3	-0.2 <b>0</b>	V min
			3.050	2.9 <b>2.7</b>	V max
$A_V$	Large Signal Voltage Gain LMV921 (Single)	$R_L = 600\Omega$ to 1.35V, $V_O = 0.2V$ to 2.5V	98	80 <b>75</b>	dB min
		$R_L = 2k\Omega$ to 1.35V, $V_O = 0.2V$ to 2.5V	103	83 <b>77</b>	
	Large Signal Voltage Gain LMV922 (Dual) LMV924 (Quad)	$R_L = 600\Omega$ to 1.35V, $V_O = 0.2V$ to 2.5V	86	68 <b>63</b>	dB min
		$R_L = 2k\Omega$ to 1.35V, $V_O = 0.2V$ to 2.5V	91	71 <b>65</b>	
$V_O$	Output Swing	$R_L = 600\Omega$ to 1.35V $V_{IN} = \pm 100mV$	2.62	2.550 <b>2.530</b>	V min
			0.075	0.095 <b>0.115</b>	V max
		$R_L = 2k\Omega$ to 1.35V $V_{IN} = \pm 100mV$	2.675	2.650 <b>2.640</b>	V min
			0.025	0.040 <b>0.045</b>	V max
$I_O$	Output Short Circuit Current	Sourcing, $V_O = 0V$ $V_{IN} = 100mV$	27	20 <b>15</b>	mA min
		Sinking, $V_O = 2.7V$ $V_{IN} = -100mV$	28	22 <b>16</b>	mA min

## 2.7V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ C$ .  $V^+ = 2.7V$ ,  $V^- = 0V$ ,  $V_{CM} = 1.0V$ ,  $V_O = 1.35V$  and  $R_L > 1 M\Omega$ .

**Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Units
SR	Slew Rate	(Note 7)	0.41	V/ $\mu$ s
GBW	Gain-Bandwidth Product		1	MHz
$\Phi_m$	Phase Margin		65	Deg.
$G_m$	Gain Margin		10	dB
$e_n$	Input-Referred Voltage Noise	$f = 1 \text{ kHz}$ , $V_{CM} = 0.5V$	45	$\frac{nV}{\sqrt{Hz}}$

Symbol	Parameter	Conditions	Typ (Note 5)	Units
$i_n$	Input-Referred Current Noise	$f = 1 \text{ kHz}$	0.1	$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
THD	Total Harmonic Distortion	$f = 1 \text{ kHz}, A_V = +1$ $R_L = 600\text{k}\Omega, V_{IN} = 1 V_{PP}$	0.077	%
	Amp-to-Amp Isolation	(Note 8)	140	dB

## 5V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ .  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{CM} = V^+/2$ ,  $V_O = V^+/2$  and  $R_L > 1 \text{ M}\Omega$ . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
$V_{OS}$	Input Offset Voltage	LMV921 (Single)	-1.5	6 <b>8</b>	mV max
		LMV922 (Dual)	-1.5	8	mV
		LMV924 (Quad)		<b>9.5</b>	max
$TCV_{OS}$	Input Offset Voltage Average Drift		1		$\mu\text{V}/^\circ\text{C}$
$I_B$	Input Bias Current		12	35 <b>50</b>	nA max
$I_{OS}$	Input Offset Current		2	25 <b>40</b>	nA max
$I_S$	Supply Current	LMV921 (Single)	160	210 <b>230</b>	$\mu\text{A}$ max
		LMV922 (Dual)	400	500 <b>700</b>	
		LMV924 (Quad)	750	850 <b>980</b>	
CMRR	Common Mode Rejection Ratio	$0\text{V} \leq V_{CM} \leq 3.8\text{V}$	86	62 <b>61</b>	dB min
		$-0.2\text{V} \leq V_{CM} \leq 0\text{V}$	72	50	
		$5.0\text{V} \leq V_{CM} \leq 5.2\text{V}$			
PSRR	Power Supply Rejection Ratio	$1.8\text{V} \leq V^+ \leq 5\text{V}$ $V_{CM} = 0.5\text{V}$	78	67 <b>62</b>	dB min
$V_{CM}$	Input Common-Mode Voltage Range	For CMRR $\geq 50\text{dB}$	-0.3	-0.2 <b>0</b>	V min
			5.350	5.2 <b>5.0</b>	V max
$A_V$	Voltage Gain LMV921 (Single)	$R_L = 600\Omega$ to 2.5V $V_O = 0.2\text{V}$ to 4.8V	104	86 <b>82</b>	dB min
		$R_L = 2\text{k}\Omega$ to 2.5V $V_O = 0.2\text{V}$ to 4.8V	108	89 <b>85</b>	
	Voltage Gain LMV922 (Dual) LMV924 (Quad)	$R_L = 600\Omega$ to 2.5V $V_O = 0.2\text{V}$ to 4.8V	90	72 68	dB min
		$R_L = 2\text{k}\Omega$ to 2.5V $V_O = 0.2\text{V}$ to 4.8V	96	77 73	

Symbol	Parameter	Condition	Typ (Note 5)	Limits (Note 6)	Units
$V_O$	Output Swing	$R_L = 600\Omega$ to 2.5V $V_{IN} = \pm 100\text{mV}$	4.895	4.865 <b>4.840</b>	V min
			0.1	0.135 <b>0.160</b>	V max
		$R_L = 2\text{k}\Omega$ to 2.5V $V_{IN} = \pm 100\text{mV}$	4.965	4.945 <b>4.935</b>	V min
			0.035	0.065 <b>0.075</b>	V max
$I_O$	Output Short Circuit Current	LMV921 Sourcing, $V_O = 0\text{V}$ $V_{IN} = 100\text{mV}$	98	85 <b>68</b>	mA min
		LMV922, LMV924 Sourcing, $V_O = 0\text{V}$ $V_{IN} = 100\text{mV}$	60	35	
		Sinking, $V_O = 5\text{V}$ $V_{IN} = -100\text{mV}$	75	65 <b>45</b>	mA min

## 5V AC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J = 25^\circ\text{C}$ .  $V^+ = 5\text{V}$ ,  $V^- = 0\text{V}$ ,  $V_{CM} = V^+/2$ ,  $V_O = 2.5\text{V}$  and  $R_L > 1\text{M}\Omega$ .  
**Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	Units
SR	Slew Rate	(Note 7)	0.45	V/ $\mu\text{s}$
GBW	Gain-Bandwidth Product		1	MHz
$\Phi_m$	Phase Margin		70	Deg
$G_m$	Gain Margin		15	dB
$e_n$	Input-Referred Voltage Noise	$f = 1\text{kHz}$ , $V_{CM} = 1\text{V}$	45	$\frac{\text{nV}}{\sqrt{\text{Hz}}}$
$i_n$	Input-Referred Current Noise	$f = 1\text{kHz}$	0.1	$\frac{\text{pA}}{\sqrt{\text{Hz}}}$
THD	Total Harmonic Distortion	$f = 1\text{kHz}$ , $A_V = +1$ $R_L = 600\Omega$ , $V_O = 1\text{V}_{PP}$	0.069	%
	Amp-to-Amp Isolation	(Note 8)	140	dB

**Note 1:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

**Note 2:** Human body model, 1.5 k $\Omega$  in series with 100pF. Machine model, 200 $\Omega$  in series with 100 pF.

**Note 3:** Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C. Output currents in excess of 45mA over long term may adversely affect reliability.

**Note 4:** The maximum power dissipation is a function of  $T_{J(max)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(max)} - T_A) / \theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

**Note 5:** Typical Values represent the most likely parametric norm.

**Note 6:** All limits are guaranteed by testing or statistical analysis.

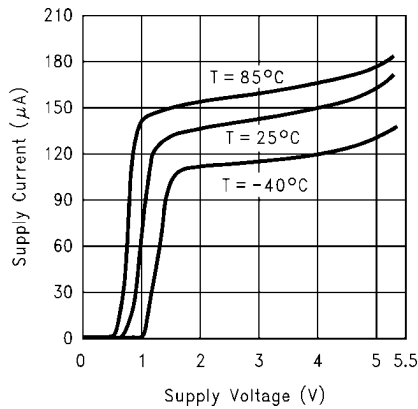
**Note 7:**  $V^+ = 5\text{V}$ . Connected as voltage follower with 5V step input. Number specified is the slower of the positive and negative slew rates.

**Note 8:** Input referred,  $V^+ = 5\text{V}$  and  $R_L = 100\text{k}\Omega$  connected to 2.5V. Each amp excited in turn with 1kHz to produce  $V_O = 3V_{PP}$ .

# Typical Performance Characteristics

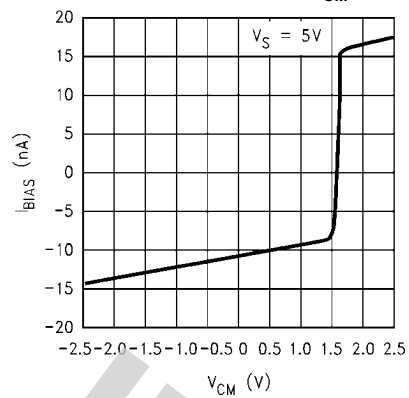
Unless otherwise specified,  $V_S = +5V$ , single supply,  $T_A = 25^\circ C$ .

**Supply Current vs. Supply Voltage (LMV921)**



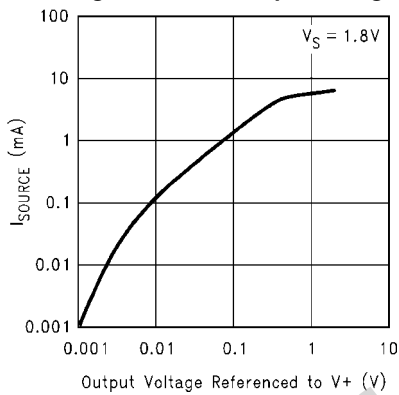
100979a1

**Input Bias Current vs.  $V_{CM}$**



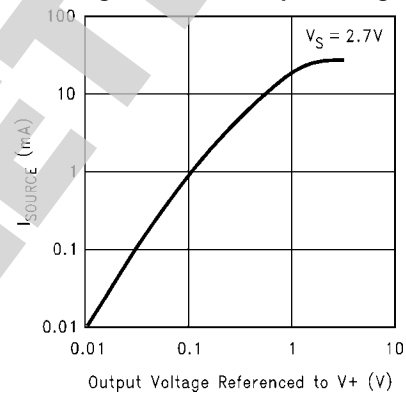
100979d5

**Sourcing Current vs. Output Voltage**



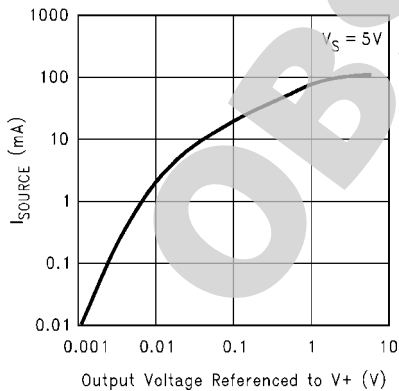
100979b3

**Sourcing Current vs. Output Voltage**



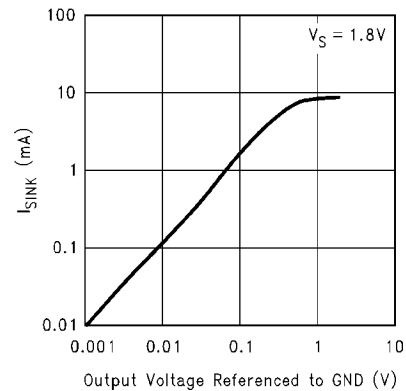
100979b8

**Sourcing Current vs. Output Voltage**



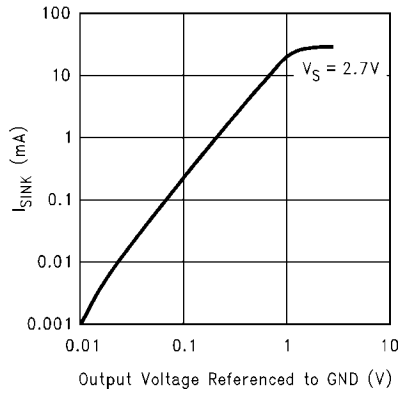
100979b2

**Sinking Current vs. Output Voltage**

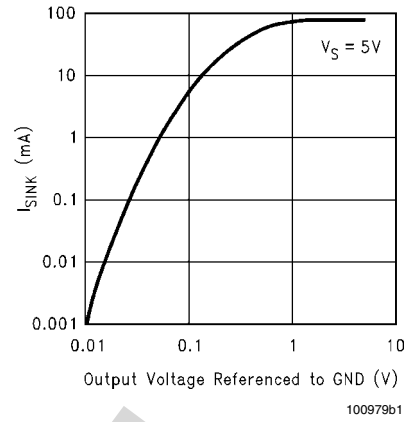


100979b4

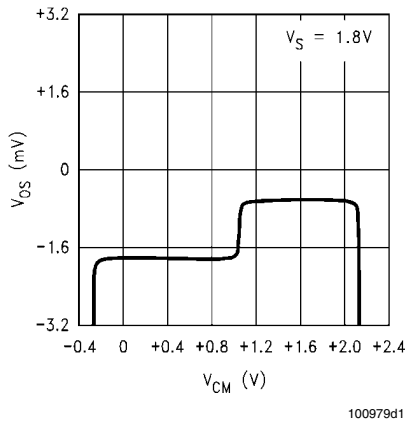
**Sinking Current vs. Output Voltage**



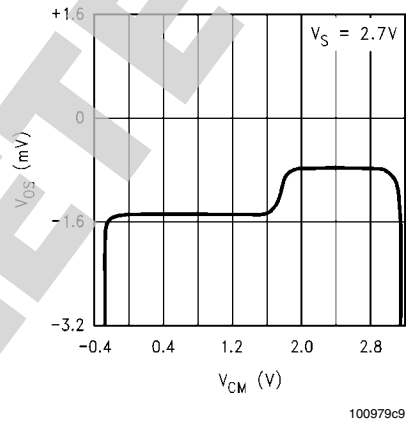
**Sinking Current vs. Output Voltage**



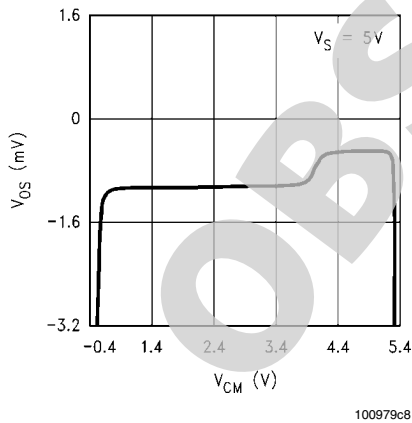
**Offset Voltage vs. Common Mode Voltage**



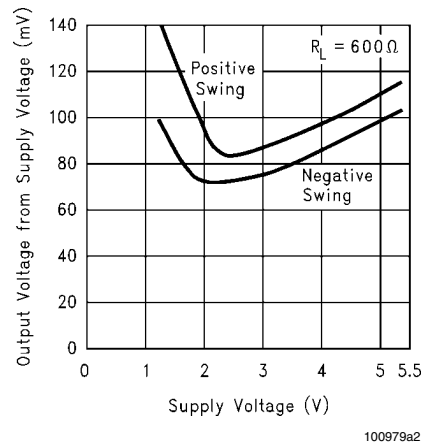
**Offset Voltage vs. Common Mode Voltage**



**Offset Voltage vs. Common Mode Voltage**

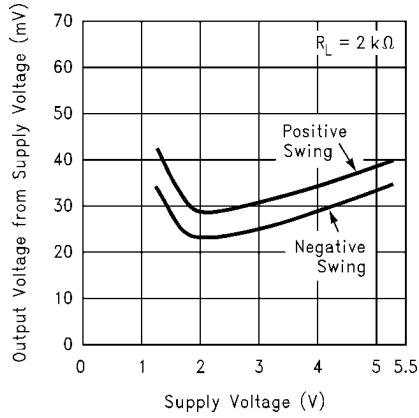


**Output Voltage Swing vs. Supply Voltage**



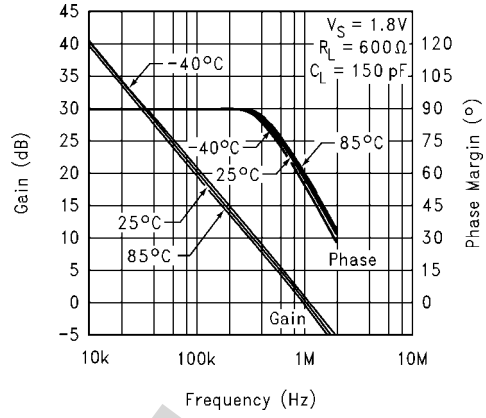


Output Voltage Swing vs. Supply Voltage



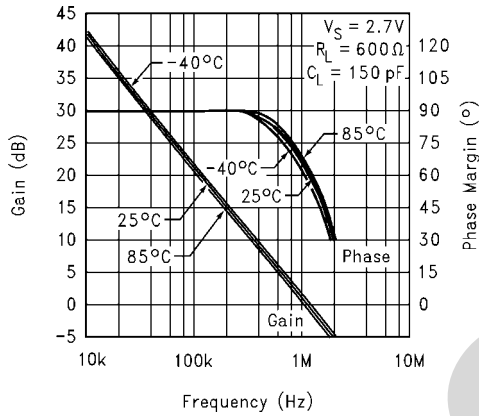
100979a3

Gain and Phase Margin vs. Frequency



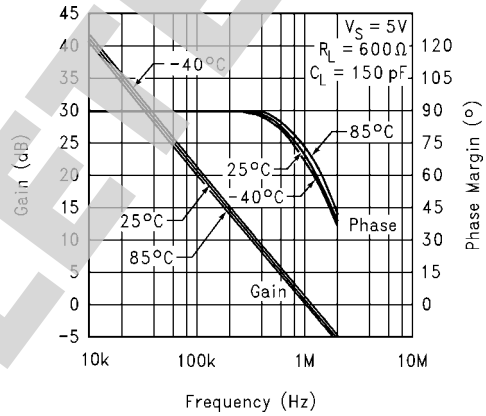
100979a6

Gain and Phase Margin vs. Frequency



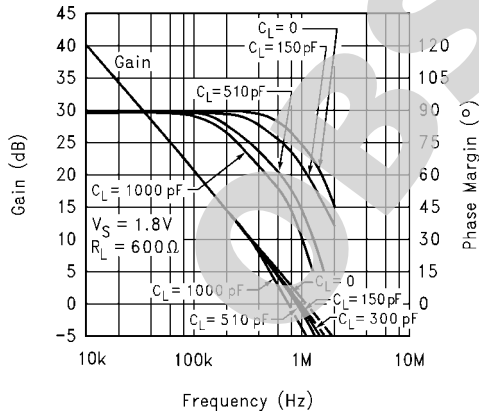
100979a5

Gain and Phase Margin vs. Frequency



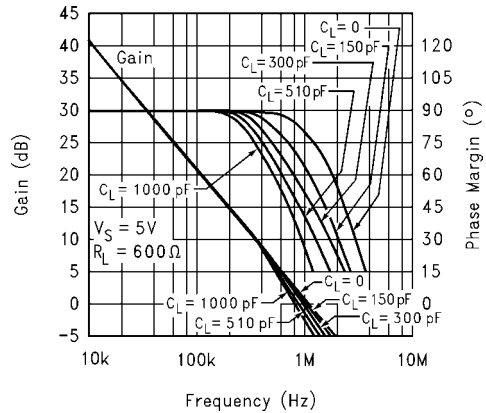
100979a4

Gain and Phase Margin vs. Frequency



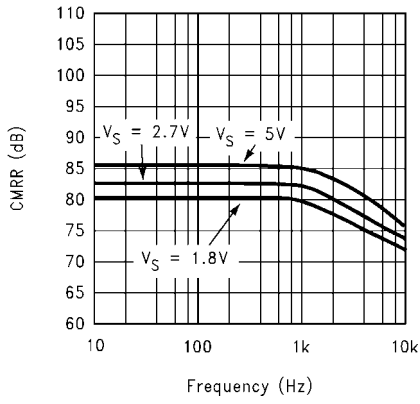
100979a8

Gain and Phase Margin vs. Frequency



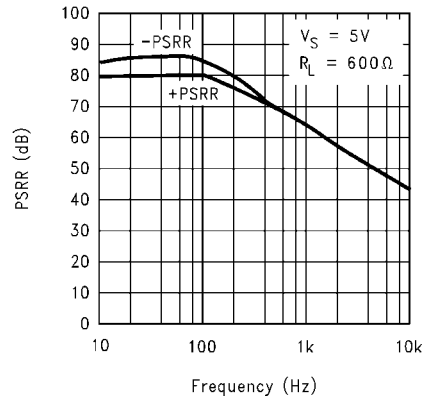
100979a7

CMRR vs. Frequency



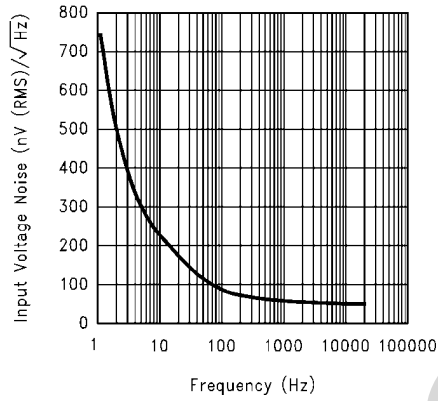
100979c7

PSRR vs. Frequency



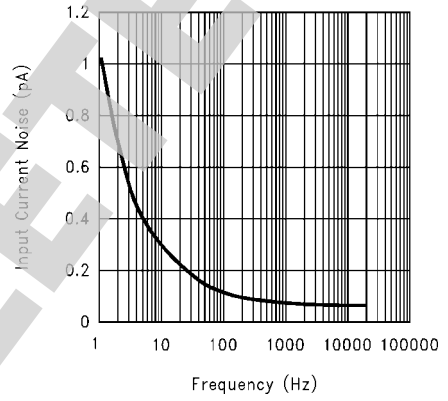
100979c6

Input Voltage Noise vs. Frequency



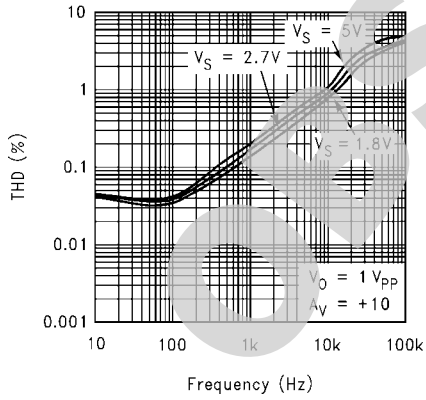
100979f4

Input Current Noise vs. Frequency



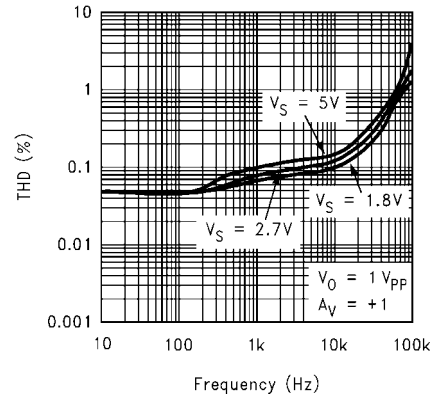
100979f5

THD vs. Frequency



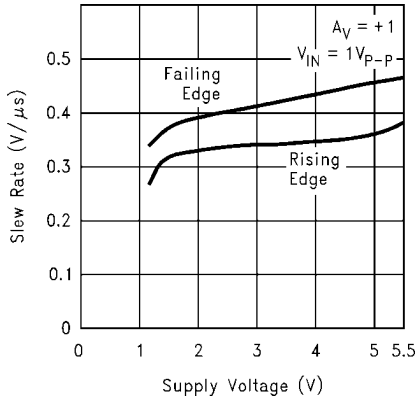
100979d4

THD vs. Frequency

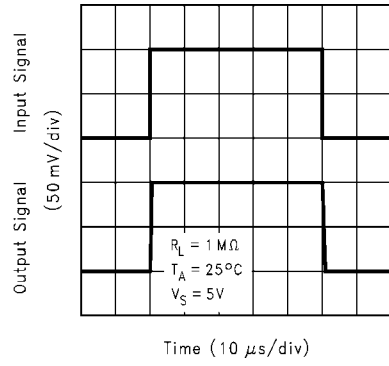


100979d3

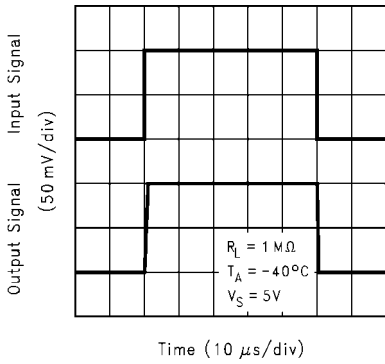
**Slew Rate vs. Supply Voltage**



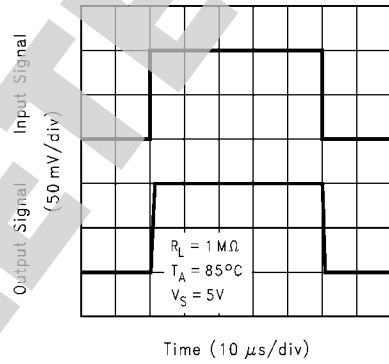
**Small Signal Non-Inverting Response**



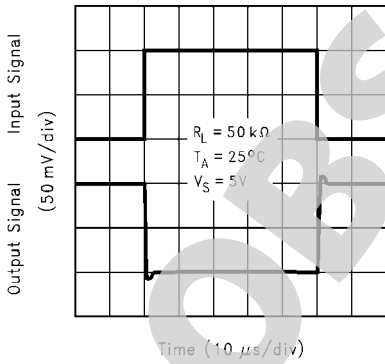
**Small Signal Non-Inverting Response**



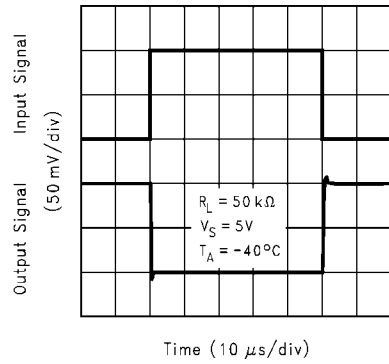
**Small Signal Non-Inverting Response**



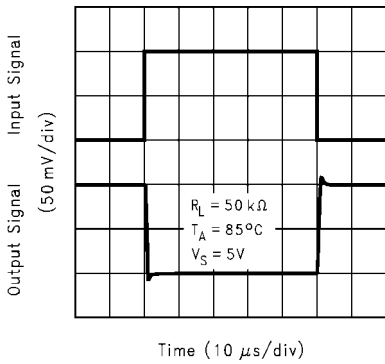
**Small Signal Inverting Response**



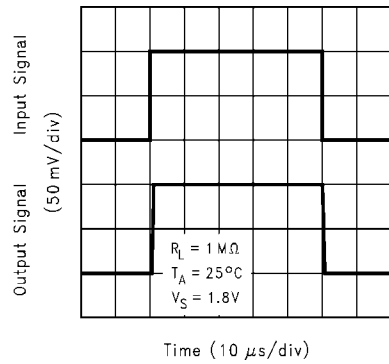
**Small Signal Inverting Response**



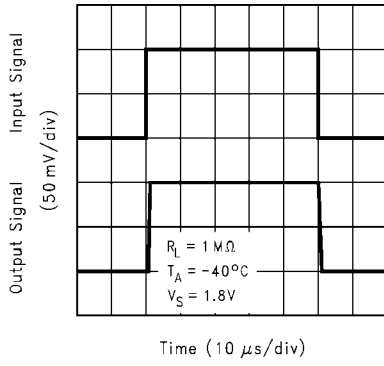
**Small Signal Inverting Response**



**Small Signal Non-Inverting Response**

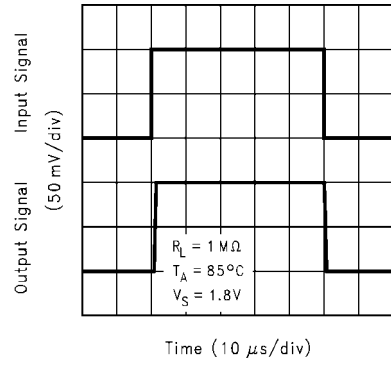


**Small Signal Non-Inverting Response**



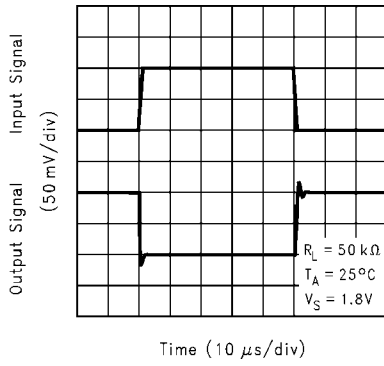
100979e7

**Small Signal Non-Inverting Response**



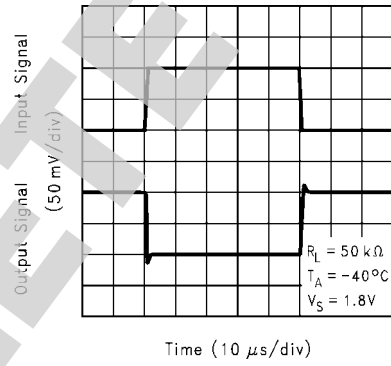
100979e5

**Small Signal Inverting Response**



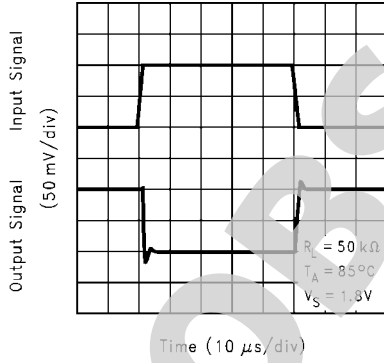
100979g3

**Small Signal Inverting Response**



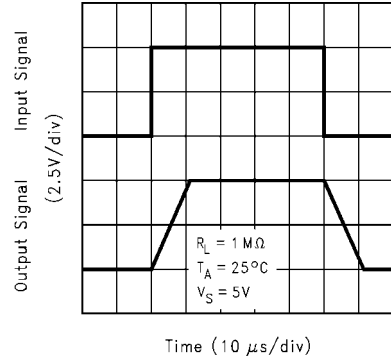
100979g2

**Small Signal Inverting Response**



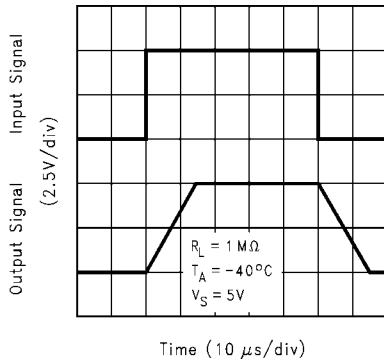
100979g1

**\*Large Signal Non-Inverting Response**



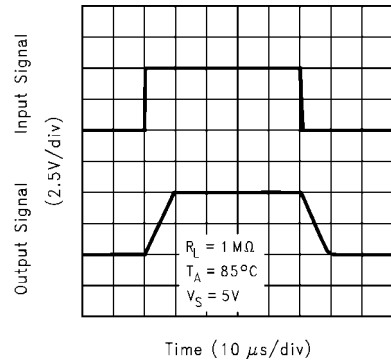
100979f0

**\*Large Signal Non-Inverting Response**



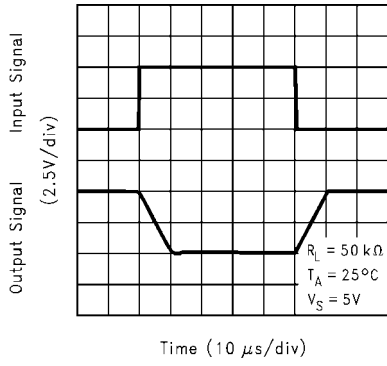
100979e9

**\*Large Signal Non-Inverting Response**



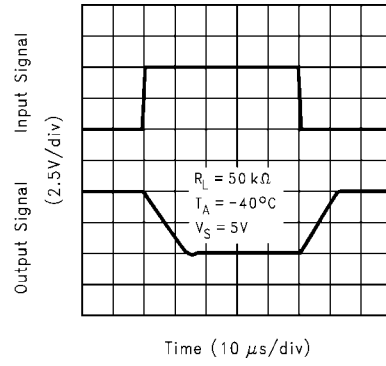
100979g0

**\*Large Signal Inverting Response**



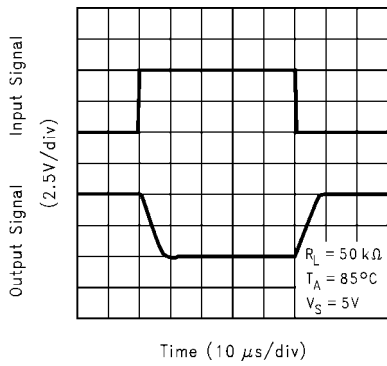
100979f9

**\*Large Signal Inverting Response**



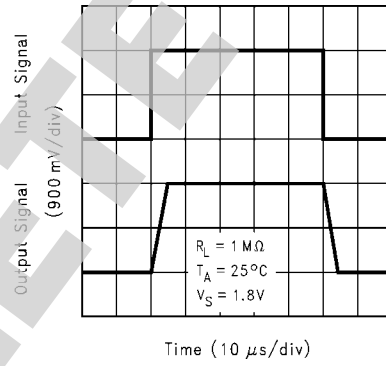
100979f8

**\*Large Signal Inverting Response**



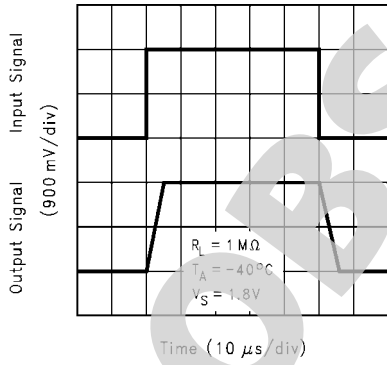
100979f7

**\*Large Signal Non-Inverting Response**



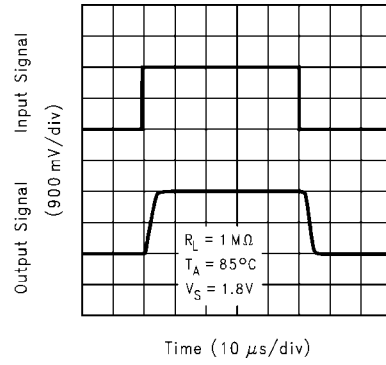
100979f1

**\*Large Signal Non-Inverting Response**



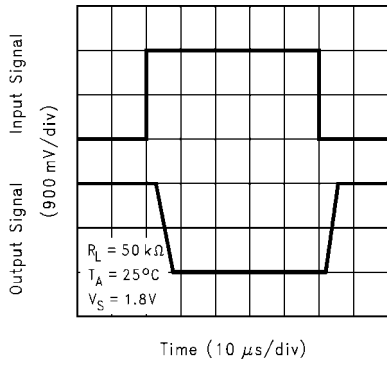
100979f2

**\*Large Signal Inverting Response**



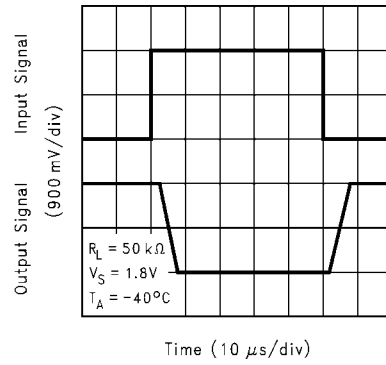
100979f6

**\*Large Signal Inverting Response**



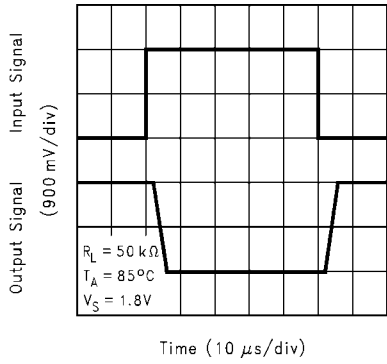
100979d6

**\*Large Signal Inverting Response**



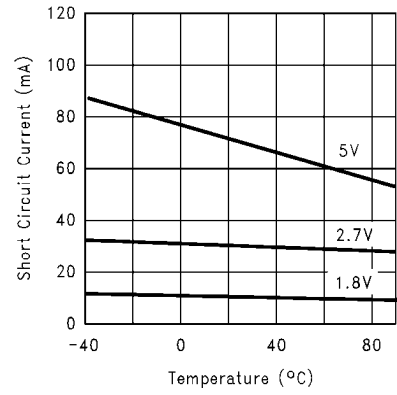
100979e1

**\*Large Signal Inverting Response**



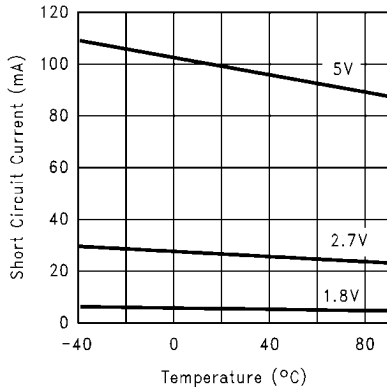
100979d7

**Short Circuit Current vs. Temperature (sinking)**



100979b5

**Short Circuit Current vs. Temperature (sourcing)**



100979b6

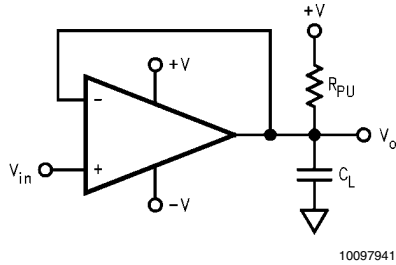
\*For large signal pulse response in the unity gain follower configuration, the input is 5mV below the positive rail and 5mV above the negative rail at 25°C and 85°C. At -40°C, input is 10mV below the positive rail and 10mV above the negative rail.

OBSOLETE

# Application Note

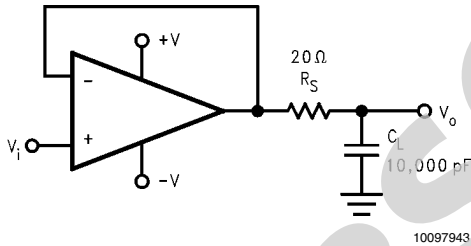
## 1.0 Unity Gain Pulse Response Considerations

The unity-gain follower is the most sensitive configuration to capacitive loading. The LMV921/LMV922/LMV924 family can directly drive 1nF in a unity-gain with minimal ringing. Direct capacitive loading reduces the phase margin of the amplifier. The combination of the amplifier's output impedance and the capacitive load induces phase lag. This results in either an underdamped pulse response or oscillation. The pulse response can be improved by adding a pull up resistor as shown in *Figure 1*



**FIGURE 1. Using a Pull-Up Resistor at the Output for Stabilizing Capacitive Loads**

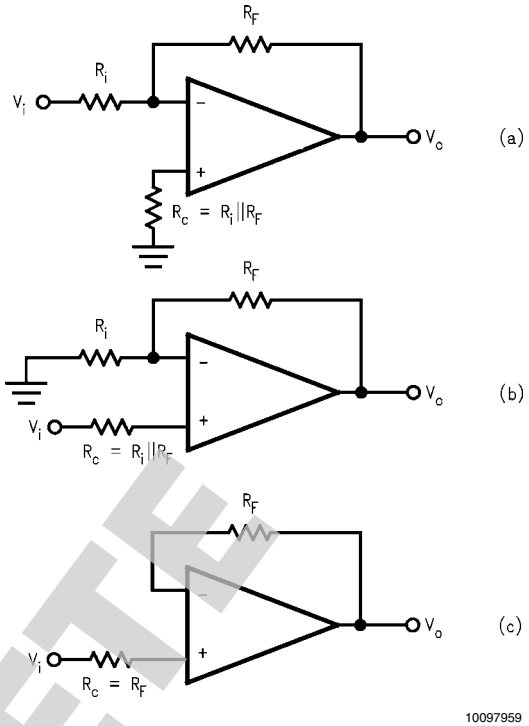
Higher capacitances can be driven by decreasing the value of the pull-up resistor, but its value shouldn't be reduced beyond the sinking capability of the part. An alternate approach is to use an isolation resistor as illustrated in *Figure 2*.



**FIGURE 2. Using an Isolation Resistor to Drive Heavy Capacitive Loads**

## 2.0 Input Bias Current Consideration

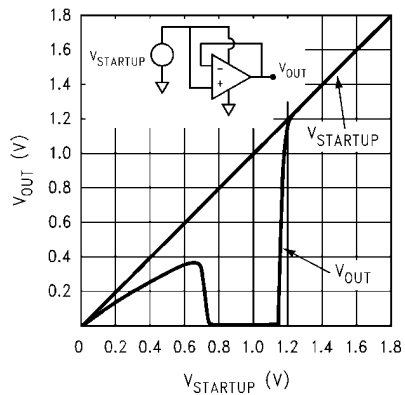
The LMV921/LMV922/LMV924 family has a bipolar input stage. The typical input bias current ( $I_B$ ) is 12nA. The input bias current can develop a significant offset voltage. This offset is primarily due to  $I_B$  flowing through the negative feedback resistor,  $R_F$ . For example, if  $I_B$  is 50nA (max room) and  $R_F$  is 100k $\Omega$ , then an offset voltage of 5mV will develop ( $V_{OS} = I_B \times R_F$ ). Using a compensation resistor ( $R_C$ ), as shown in *Figure 3*, cancels this affect. But the input offset current ( $I_{OS}$ ) will still contribute to an offset voltage in the same manner.



**FIGURE 3. Canceling the Voltage Offset Effect of Input Bias Current**

## 3.0 Operating Supply Voltage

The LMV921/LMV922/LMV924 family is guaranteed to operate from 1.8V to 5.0V. They will begin to function at power voltages as low as 1.2V at room temperature when unloaded. Start up voltage increases to 1.5V when the amplifier is fully loaded (600 $\Omega$  to mid-supply). Below 1.2V the output voltage is not guaranteed to follow the input. *Figure 4* below shows the output voltage vs. supply voltage with the LMV921/LMV922/LMV924 configured as a voltage follower at room temperature.



**FIGURE 4.**

#### 4.0 Input and Output Stage

The rail-to-rail input stage of this family provides more flexibility for the designer. The LMV921/LMV922/LMV924 use a complimentary PNP and NPN input stage in which the PNP stage senses common mode voltage near  $V^-$  and the NPN stage senses common mode voltage near  $V^+$ . The transition from the PNP stage to NPN stage occurs 1V below  $V^+$ . Since both input stages have their own offset voltage, the offset of the amplifier becomes a function of the input common mode voltage and has a crossover point at 1V below  $V^+$  as shown in the  $V_{OS}$  vs.  $V_{CM}$  curves.

This  $V_{OS}$  crossover point can create problems for both DC and AC coupled signals if proper care is not taken. For large input signals that include the  $V_{OS}$  crossover point in their dynamic range, this will cause distortion in the output signal. One way to avoid such distortion is to keep the signal away from the crossover. For example, in a unity gain buffer configuration and with  $V_S = 5V$ , a 5V peak-to-peak signal will contain input-crossover distortion while a 3V peak-to-peak signal centered at 1.5V will not contain input-crossover distortion as it avoids the crossover point. Another way to avoid large signal distortion is to use a gain of  $-1$  circuit which avoids any voltage excursions at the input terminals of the amplifier. In that circuit, the common mode DC voltage can be set at a level away from the  $V_{OS}$  cross-over point.

For small signals, this transition in  $V_{OS}$  shows up as a  $V_{CM}$  dependent spurious signal in series with the input signal and can effectively degrade small signal parameters such as gain and common mode rejection ratio. To resolve this problem, the small signal should be placed such that it avoids the  $V_{OS}$  crossover point.

In addition to the rail-to-rail performance, the output stage can provide enough output current to drive 600 $\Omega$  loads. Because of the high current capability, care should be taken not to exceed the 150°C maximum junction temperature specification.

#### 5.0 Power-Supply Considerations

The LMV921/LMV922/LMV924 are ideally suited for use with most battery-powered systems. The LMV921/LMV922/LMV924 operate from a single +1.8V to +5.0V supply and consumes about 145 $\mu$ A of supply current per Amplifier. A high power supply rejection ratio of 78dB allows the amplifier to be powered directly off a decaying battery voltage extending battery life.

*Table 1* lists a variety of typical battery types. Batteries have different voltage ratings; operating voltage is the battery voltage under nominal load. End-of-Life voltage is defined as the voltage at which 100% of the usable power of the battery is consumed. *Table 1* also shows the typical operating time of the LMV921.

#### 6.0 Distortion

The two main contributors of distortion in LMV921/LMV922/LMV924 family is:

1. Output crossover distortion occurs as the output transitions from sourcing current to sinking current.
2. Input crossover distortion occurs as the input switches from NPN to PNP transistor at the input stage.

To decrease crossover distortion:

1. Increase the load resistance. This lowers the output crossover distortion but has no effect on the input crossover distortion.
2. Operate from a single supply with the output always sourcing current.
3. Limit the input voltage swing for large signals between ground and one volt below the positive supply.
4. Operate in inverting configuration to eliminate common mode induced distortion.
5. Avoid small input signal around the input crossover region. The discontinuity in the offset voltage will effect the gain, CMRR and PSRR.

**TABLE 1. LMV921 Characteristics with Typical Battery Systems.**

Battery Type	Operating Voltage (V)	End-of-Life Voltage (V)	Capacity AA Size (mA - h)	LMV921 Operating time (Hours)
Alkaline	1.5	0.9	1000	6802
Lithium	2.7	2.0	1000	6802
Ni - Cad	1.2	0.9	375	2551
NMH	1.2	1.0	500	3401



## Typical Applications

### 1.0 Half-wave Rectifier with Rail-To-Ground Output Swing

Since the LMV921 input common mode range includes both positive and negative supply rails and the output can also swing to either supply, achieving half-wave rectifier functions in either direction is an easy task. All that is needed are two external resistors; there is no need for diodes or matched resistors. The half wave rectifier can have either positive or negative going outputs, depending on the way the circuit is arranged.

In *Figure 5* the circuit is referenced to ground, while in *Figure 6* the circuit is biased to the positive supply. These configurations implement the half wave rectifier since the LMV921 can not respond to one-half of the incoming waveform. It can not respond to one-half of the incoming because the amplifier can not swing the output beyond either rail therefore the output disengages during this half cycle. During the other half cycle, however, the amplifier achieves a half wave that can have a peak equal to the total supply voltage.  $R_1$  should be large enough not to load the LMV921.

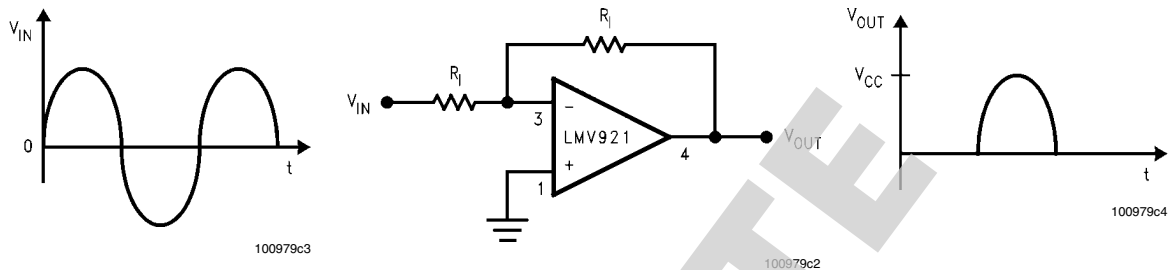


FIGURE 5. Half-Wave Rectifier with Rail-To-Ground Output Swing Referenced to Ground

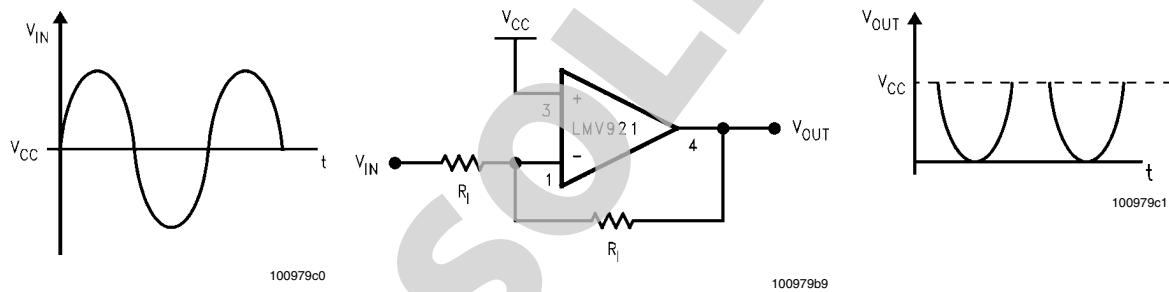


FIGURE 6. Half-Wave Rectifier with Negative-Going Output Referenced to  $V_{CC}$

### 2.0 Instrumentation Amplifier with Rail-To-Rail Input and Output

Using three of the LMV924 Amplifiers, an instrumentation amplifier with rail-to-rail inputs and outputs can be made.

Some manufacturers use a precision voltage divider array of 5 resistors to divide the common mode voltage to get a rail-to-rail input range. The problem with this method is that it also divides the signal, so in order to get unity gain, the amplifier must be run at high loop gains. This raises the noise and drift by the internal gain factor and lowers the input impedance. Any mismatch in these precision resistors reduces the CMRR as well. Using the LMV924 eliminates all of these problems. In this example, amplifiers A and B act as buffers to the differential stage. These buffers assure that the input

impedance is very high and require no precision matched resistors in the input stage. They also assure that the difference amp is driven from a voltage source. This is necessary to maintain the CMRR set by the matching  $R_1$ - $R_2$  with  $R_3$ - $R_4$ .

The gain is set by the ratio of  $R_2/R_1$  and  $R_3$  should equal  $R_1$  and  $R_4$  equal  $R_2$ .

With both rail-to-rail input and output ranges, the input and output are only limited by the supply voltages. Remember that even with rail-to-rail outputs, the output can not swing past the supplies so the combined common mode voltages plus the signal should not be greater that the supplies or limiting will occur. For additional applications, see National Semiconductor application notes AN-29, AN-31, AN-71, and AN-127.

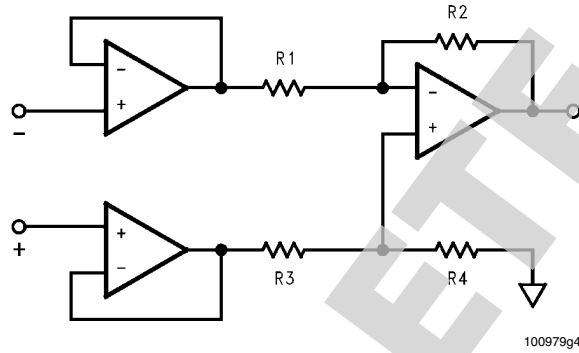
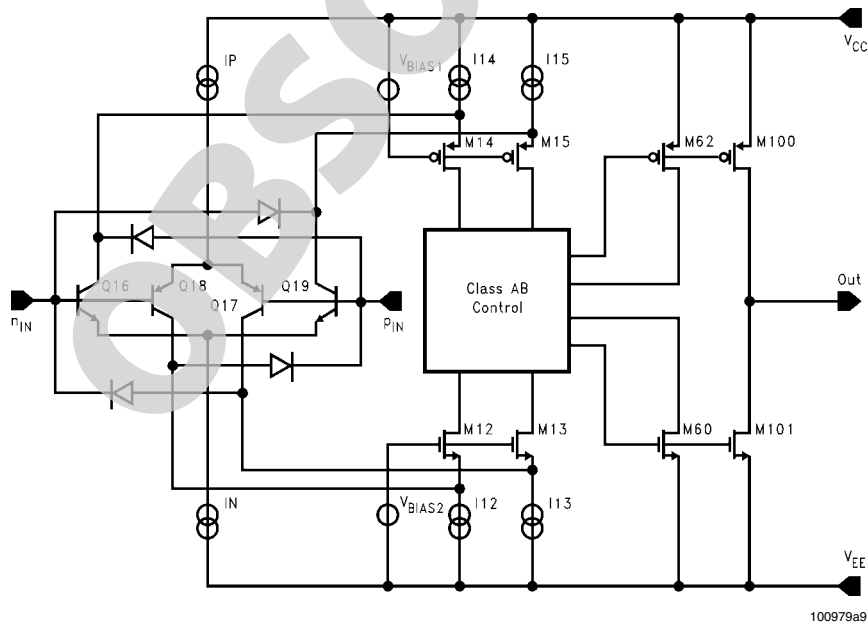


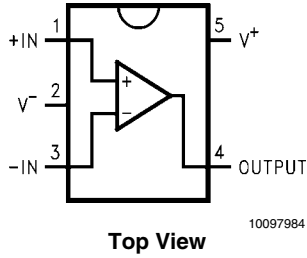
FIGURE 7. Rail-to-rail instrumentation amplifier

### Simplified Schematic

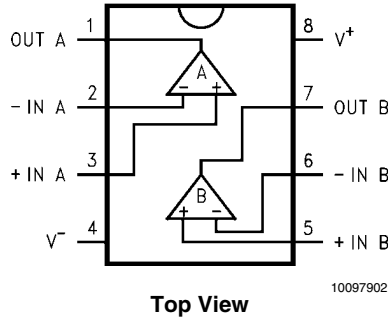


## Connection Diagrams

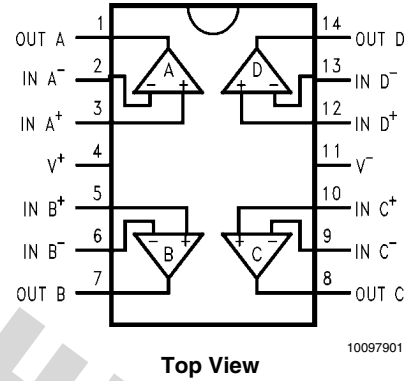
5-Pin SC70-5/SOT23-5



8-Pin MSOP/SOIC



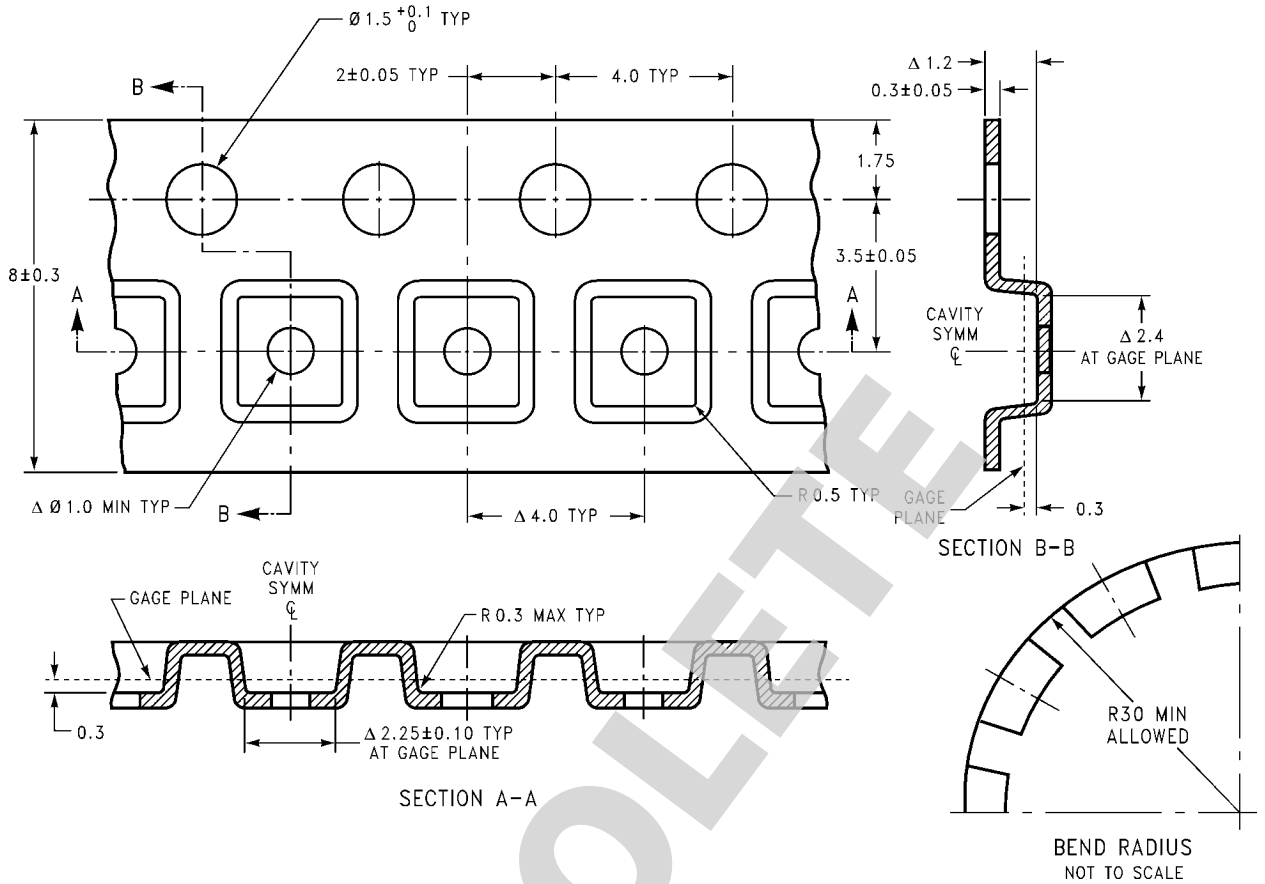
14-Pin TSSOP/SOIC



## Ordering Information

Package	Temperature Range Industrial -40°C to +85°C	Package Marking	Transport Media	NSC Drawing
5-Pin SC70-5	LMV921M7	A21	1k Units Tape and Reel	MAA05A
	LMV921M7X	A21	3k Units Tape and Reel	
5-Pin SOT-23	LMV921M5	A29A	1k Units Tape and Reel	MF05A
	LMV921M5X	A29A	3k Units Tape and Reel	
8-Pin MSOP	LMV922MM	LMV922	1k Units Tape and Reel	MUA08A
	LMV922MMX	LMV922	3.5k Units Tape and Reel	
14-Pin TSSOP	LMV924MT	LMV924	Rails	MTC14
	LMV924MTX	LMV924	2.5k Units Tape and Reel	
8-Pin SOIC	LMV922M	LMV922M	Rails	M08A
	LMV922MX	LMV922M	2.5k Units Tape and Reel	
14-Pin SOIC	LMV924M	LMV924M	Rails	M14A
	LMV924MX	LMV924M	2.5k Units Tape and Reel	

### SC70-5 Tape Dimensions



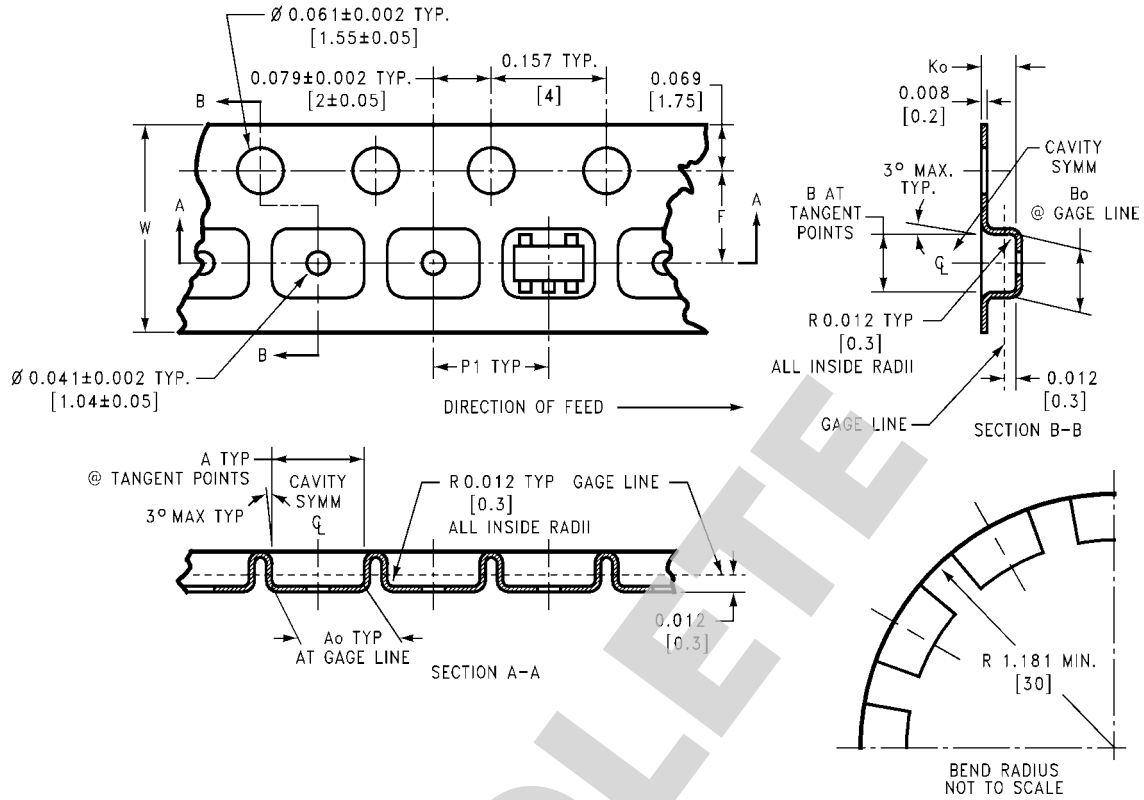
10097996

### SOT23-5 and SC70-5 Tape Format

#### Tape Format

Tape Section	# Cavities	Cavity Status	Cover Tape Status
Leader (Start End)	0 (min)	Empty	Sealed
	75 (min)	Empty	Sealed
Carrier	3000	Filled	Sealed
	250	Filled	Sealed
Trailer (Hub End)	125 (min)	Empty	Sealed
	0 (min)	Empty	Sealed

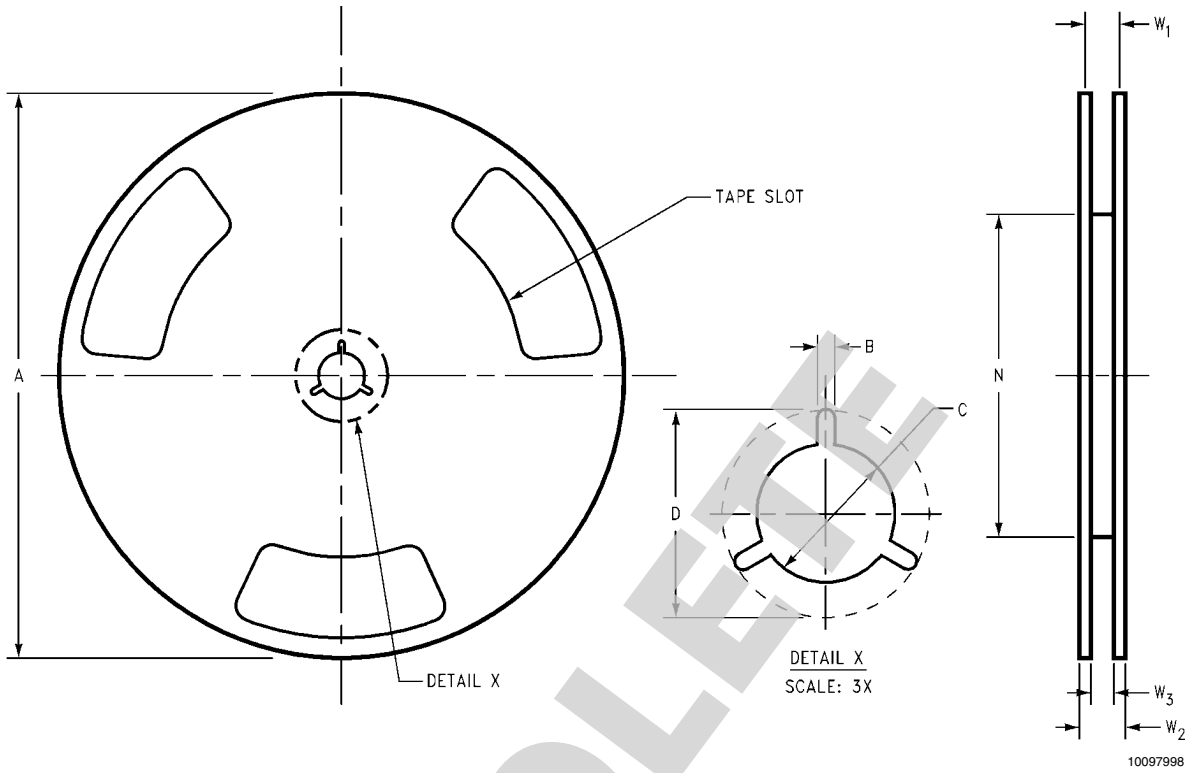
# SOT23-5 Tape Dimensions



10097997

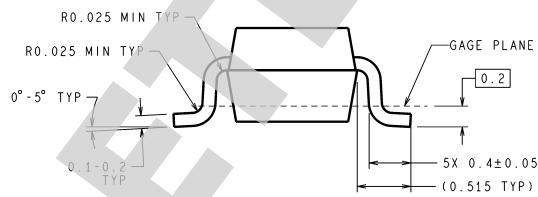
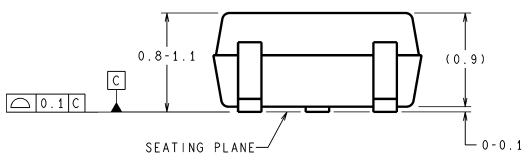
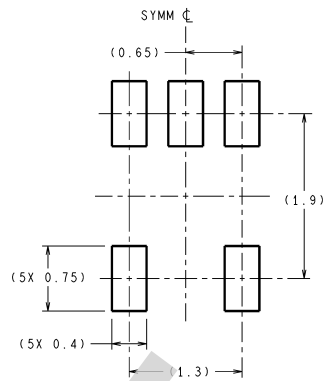
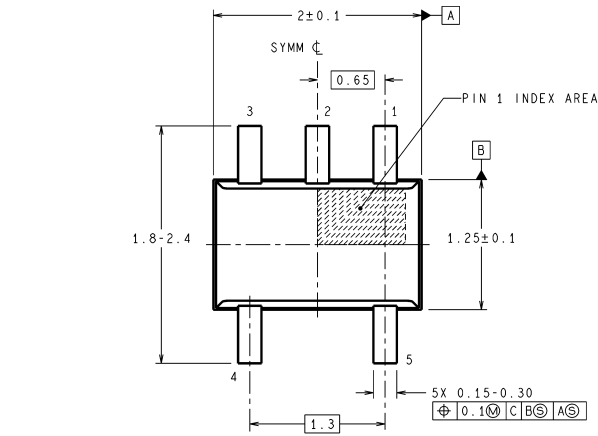
<b>8 mm</b>	<b>0.130</b> <b>(3.3)</b>	<b>0.124</b> <b>(3.15)</b>	<b>0.130</b> <b>(3.3)</b>	<b>0.126</b> <b>(3.2)</b>	<b>0.138 ± 0.002</b> <b>(3.5 ± 0.05)</b>	<b>0.055 ± 0.004</b> <b>(1.4 ± 0.11)</b>	<b>0.157</b> <b>(4)</b>	<b>0.315 ± 0.012</b> <b>(8 ± 0.3)</b>
Tape Size	DIM A	DIM Ao	DIM B	DIM Bo	DIM F	DIM Ko	DIM P1	DIM W

### SOT23-5 and SC70-5 Reel Dimensions



<b>8 mm</b>	<b>7.00</b>	<b>0.059</b>	<b>0.512</b>	<b>0.795</b>	<b>2.165</b>	<b>0.331 + 0.059/-0.000</b>	<b>0.567</b>	<b>W1 + 0.078/-0.039</b>
	<b>330.00</b>	<b>1.50</b>	<b>13.00</b>	<b>20.20</b>	<b>55.00</b>	<b>8.40 + 1.50/-0.00</b>	<b>14.40</b>	<b>W1 + 2.00/-1.00</b>
Tape Size	A	B	C	D	N	W1	W2	W3

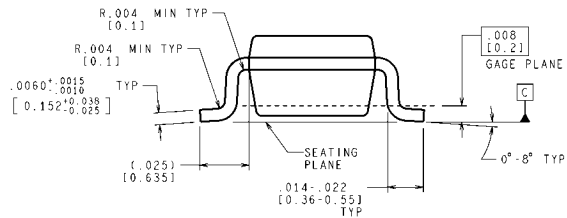
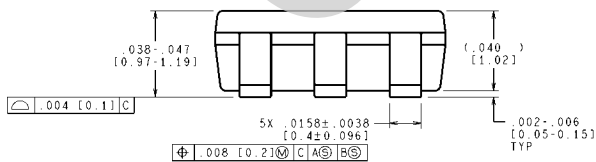
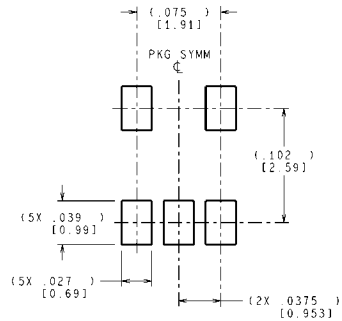
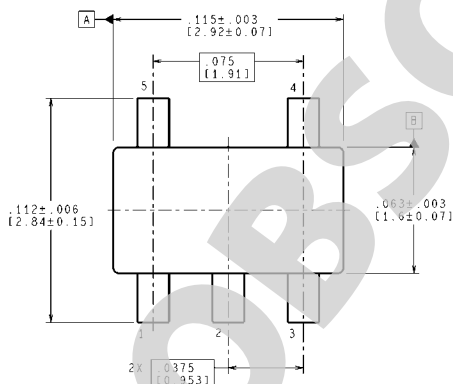
**Physical Dimensions** inches (millimeters) unless otherwise noted



DIMENSIONS ARE IN MILLIMETERS  
DIMENSIONS IN ( ) FOR REFERENCE ONLY

MAA05A (Rev D)

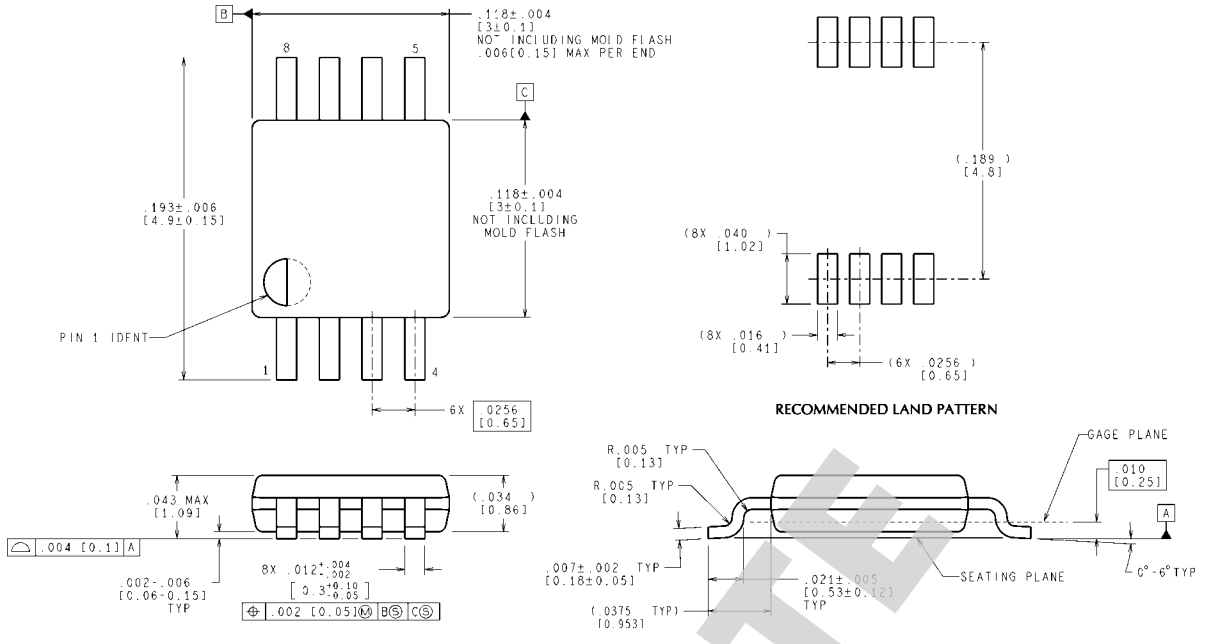
**SC70-5**  
**NS Package Number MAA05A**



CONTROLLING DIMENSION IS INCH  
VALUES IN [ ] ARE MILLIMETERS  
DIMENSIONS IN ( ) FOR REFERENCE ONLY

MF05A (Rev D)

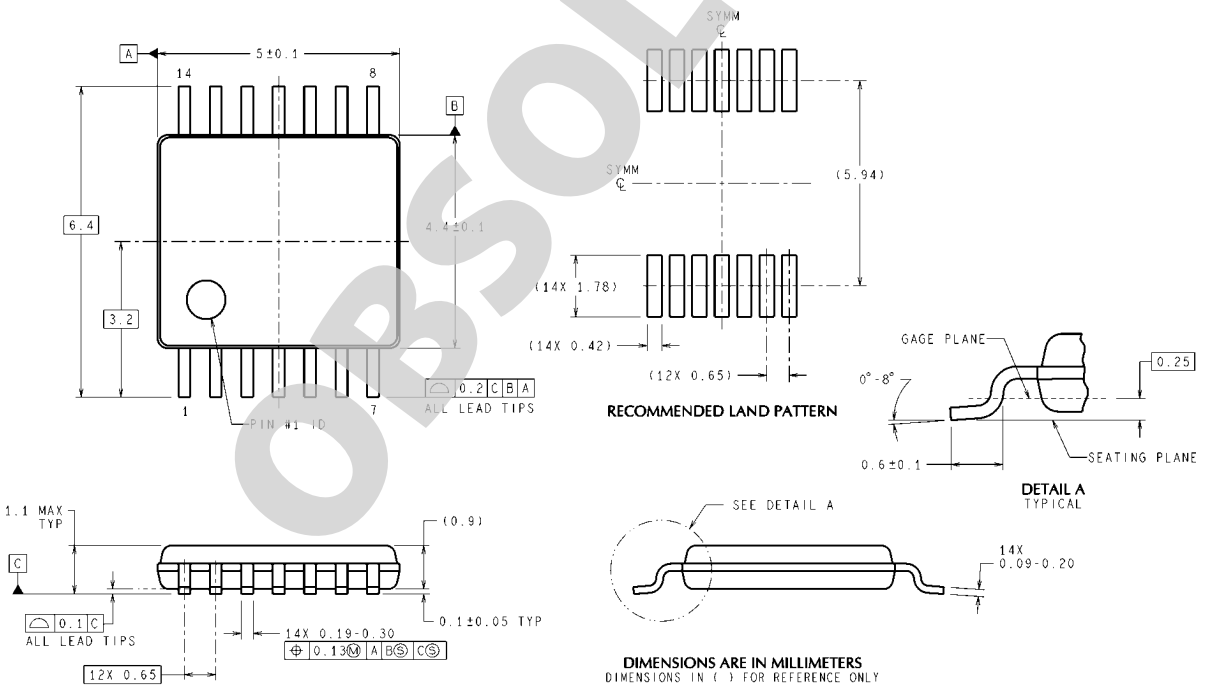
**5-Pin SOT-23**  
**NS Package Number MF05A**



CONTROLLING DIMENSION IS INCH  
VALUES IN [ ] ARE MILLIMETERS

**8-Pin MSOP**  
NS Package Number MUA08A

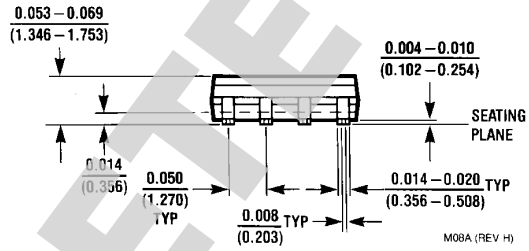
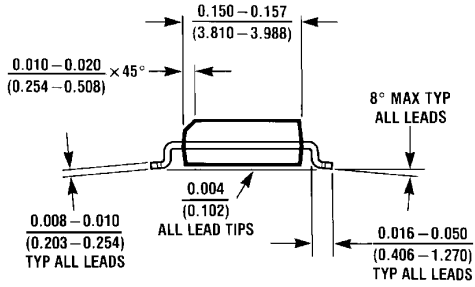
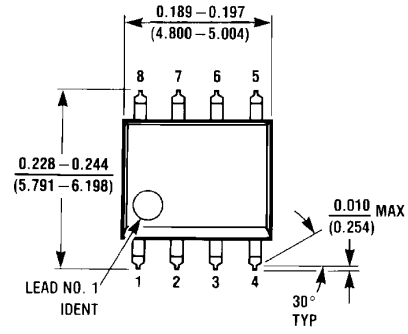
MUA08A (Rev F)



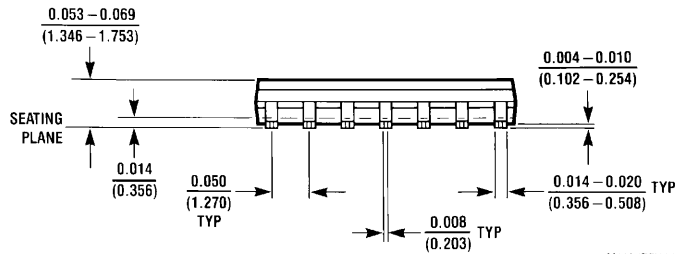
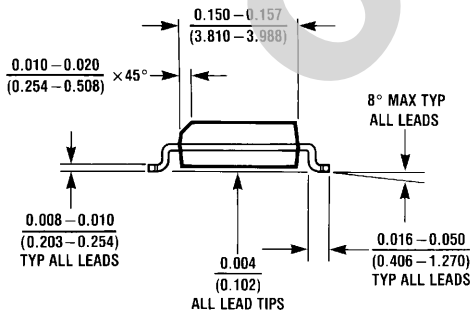
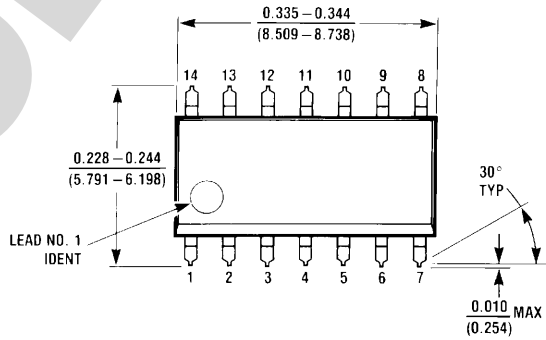
**14-Pin TSSOP**  
NS Package Number MTC14

MTC14 (Rev D)





8-Pin SOIC  
NS Package Number M08A



14-Pin SOIC  
NS Package Number MA14

## Notes

---

### TI/NATIONAL INTERIM IMPORTANT NOTICE

---

Texas Instruments has purchased National Semiconductor. As of Monday, September 26th, and until further notice, products sold or advertised under the National Semiconductor name or logo, and information, support and interactions concerning such products, remain subject to the preexisting National Semiconductor standard terms and conditions of sale, terms of use of website, and Notices (and/or terms previously agreed in writing with National Semiconductor, where applicable) and are not subject to any differing terms and notices applicable to other TI components, sales or websites. To the extent information on official TI and National websites and business social networking media, etc., pertains to both TI and National-branded products, both companies' instructions, warnings and limitations in the above-referenced terms of use apply.

Following are URLs where you can obtain information on other Texas Instruments products and application solutions:

Products		Applications	
Audio	<a href="http://www.ti.com/audio">www.ti.com/audio</a>	Communications and Telecom	<a href="http://www.ti.com/communications">www.ti.com/communications</a>
Amplifiers	<a href="http://amplifier.ti.com">amplifier.ti.com</a>	Computers and Peripherals	<a href="http://www.ti.com/computers">www.ti.com/computers</a>
Data Converters	<a href="http://dataconverter.ti.com">dataconverter.ti.com</a>	Consumer Electronics	<a href="http://www.ti.com/consumer-apps">www.ti.com/consumer-apps</a>
DLP® Products	<a href="http://www.dlp.com">www.dlp.com</a>	Energy and Lighting	<a href="http://www.ti.com/energy">www.ti.com/energy</a>
DSP	<a href="http://dsp.ti.com">dsp.ti.com</a>	Industrial	<a href="http://www.ti.com/industrial">www.ti.com/industrial</a>
Clocks and Timers	<a href="http://www.ti.com/clocks">www.ti.com/clocks</a>	Medical	<a href="http://www.ti.com/medical">www.ti.com/medical</a>
Interface	<a href="http://interface.ti.com">interface.ti.com</a>	Security	<a href="http://www.ti.com/security">www.ti.com/security</a>
Logic	<a href="http://logic.ti.com">logic.ti.com</a>	Space, Avionics and Defense	<a href="http://www.ti.com/space-avionics-defense">www.ti.com/space-avionics-defense</a>
Power Mgmt	<a href="http://power.ti.com">power.ti.com</a>	Transportation and Automotive	<a href="http://www.ti.com/automotive">www.ti.com/automotive</a>
Microcontrollers	<a href="http://microcontroller.ti.com">microcontroller.ti.com</a>	Video and Imaging	<a href="http://www.ti.com/video">www.ti.com/video</a>
RFID	<a href="http://www.ti-rfid.com">www.ti-rfid.com</a>	Wireless	<a href="http://www.ti.com/wireless-apps">www.ti.com/wireless-apps</a>
RF/IF and ZigBee® Solutions	<a href="http://www.ti.com/lprf">www.ti.com/lprf</a>	<b>TI E2E Community Home Page</b>	<a href="http://e2e.ti.com">e2e.ti.com</a>

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265

[www.ti.com](http://www.ti.com)

Copyright© 2011 Texas Instruments Incorporated