



SQ48T03150

Eighth-Brick DC-DC Converter

The **SQ48T03150** through-hole DC/DC converters offer a size that is only 60% of industry-standard quarter bricks, while preserving the same pinout and functionality. This is accomplished through the use of patent pending circuit and packaging techniques to achieve ultra-high efficiency, excellent thermal performance and a very low body profile.

In telecommunications applications the **SQ48** 3.5 A converters provide thermal performance that exceeds competing quarter bricks with a higher nominal rating. Low body profile and the preclusion of heatsinks minimize airflow shadowing, thus enhancing cooling for downstream devices. The use of 100% surface-mount technologies for assembly, coupled with Bel's advanced electric and thermal circuitry and packaging, results in a product with extremely high quality and reliability.

With a standard quarter brick pinout and trim equations, the **SQ48** converters are perfect drop-in replacements for existing quarter brick designs, and can result in significant board space and cost savings in new design.



Key Features & Benefits

- 43-75 Vdc Input, 3.5 A, 15 Vdc Output
- High efficiency: 88.5% @ 3.5 A, 87.5% @ 1.75 A
- Starts-up into pre-biased output
- No minimum load required
- No heatsink required
- Available in through-hole and SM packages
- Lowest profile in industry: 0.28" [7.2 mm]
- Extremely small footprint: 0.896" x 2.30" (2.06 in²)
- Industry-standard quarter brick pinout
- Meets Basic Insulation Requirements of EN60950
- Meets input voltage transient of 100 V for 100 ms
- On-board input differential LC-filter
- Fixed frequency operation
- Fully protected
- Remote output sense
- Output voltage trim range: +10%/-20%
- Trim resistor via industry-standard equations
- High reliability: MTBF 3.4 million hours, calculated per Telcordia TR-332, Method I Case 1
- Positive or negative logic ON/OFF option
- Meets conducted emissions requirements of FCC Class B and EN55022 Class B with external filter
- All materials meet UL94, V-0 flammability rating
- Approved to the latest edition and amendment of ITE Safety standards UL/CSA 60950-1



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1. ELECTRICAL SPECIFICATIONS

Conditions: $T_A = 25^\circ\text{C}$, Airflow = 300 LFM (1.5 m/s), $V_{in} = 48\text{ Vdc}$, unless otherwise specified.

PARAMETER	NOTES	MIN	TYP	MAX	UNITS
ABSOLUTE MAXIMUM RATINGS					
Input Voltage	Continuous	0		80	Vdc
Operating Ambient Temperature		-40		85	$^\circ\text{C}$
Storage Temperature		-55		125	$^\circ\text{C}$
INPUT CHARACTERISTICS					
Operating Input Voltage Range		43	48	75	Vdc
Input Under Voltage Lockout	Non-latching				
Turn-on Threshold		39	40	42	Vdc
Turn-off Threshold		38	39	40	Vdc
Input Voltage Transient	100 ms			100	Vdc
OUTPUT CHARACTERISTICS					
External Load Capacitance	Plus full load (resistive)			1,000	μF
Output Current Range		0		3.5	A _{dc}
Current Limit Inception	Non-latching	4	4.3	4.6	A _{dc}
Peak Short-Circuit Current	Non-latching, Short=10m Ω -.	4	6	8	A
RMS Short-Circuit Current	Non-latching			2	Arms
ISOLATION CHARACTERISTICS					
I/O Isolation		2000			Vdc
Isolation Capacitance			230		ρF
Isolation Resistance		10			M Ω
FEATURE CHARACTERISTICS					
Switching Frequency			415		kHz
Output Voltage Trim Range ¹	Use trim equations on Page 7	-20		+10	%
Remote Sense Compensation ¹	Percent of $V_{OUT(NOM)}$			+10	%
Output Over-Voltage Protection	Non-latching	117	122	127	%
Over-Temperature Shutdown (PCB)	Non-latching		125		$^\circ\text{C}$
Auto-Restart Period	Applies to all protection features		100		ms
Turn-On Time			4		ms
ON/OFF Control (Positive Logic)	Converter Off	-20		0.8	Vdc
	Converter On	2.4		20	Vdc
ON/OFF Control (Negative Logic)	Converter Off	2.4		20	Vdc
	Converter On	-20		0.8	Vdc

¹ V_{out} can be increased up to 10% via the sense leads or up to 10% via the trim function, however total output voltage trim from all sources should not exceed 10% of $V_{OUT(NOM)}$, in order to insure specified operation of over-voltage protection circuitry. See further discussion at end of **Output Voltage Adjust /TRIM** section.

INPUT CHARACTERISTICS					
Maximum Input Current	3.5 Adc, 15 Vdc Out @ 43 Vdc In		1.4		Adc
Input Stand-by Current	Vin = 48 V, converter disabled		3		mAdc
Input No Load Current (0 load on the output)	Vin = 48 V, converter enabled		50		mAdc
Input Reflected-Ripple Current	See Figure 24 - 25MHz bandwidth		10		mA _{PK-PK}
Input Voltage Ripple Rejection	120Hz		TBD		dB
OUTPUT CHARACTERISTICS					
Output Voltage Set Point (no load)	-40°C to 85°C	14.850	15.000	15.150	Vdc
Output Regulation					
Over Line			±4	±10	mV
Over Load			±4	±10	mV
Output Voltage Range	Over line, load and temperature	14.775		15.225	Vdc
Output Ripple and Noise - 25MHz bandwidth	Full load + 10 µF tantalum + 1 µF ceramic		100	150	mV _{PK-PK}
DYNAMIC RESPONSE					
Load Change 25% of Iout Max, di/dt = 0.1 A/µS	Co = 1 µF ceramic (Fig.19)		200		mV
	di/dt = 5 A/µS Co = 47 µF tant. + 1 µF ceramic (Fig.20)		200		mV
Setting Time to 1%			400		µs
EFFICIENCY					
100% Load			88.5		%
50% Load			87.5		%

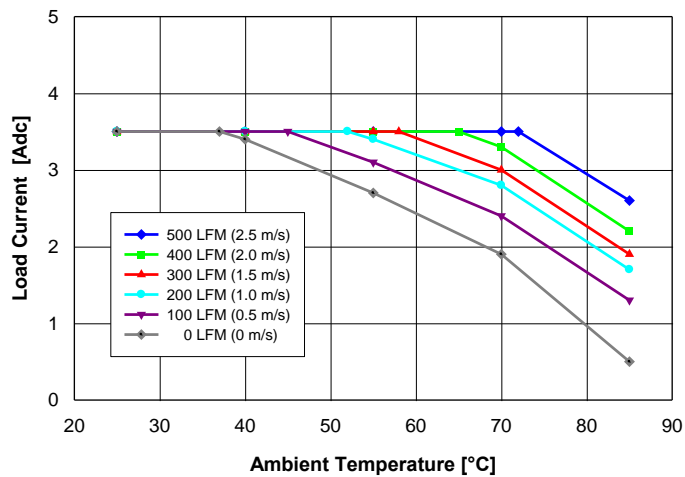


Figure 1. Available load current vs. ambient air temperature and airflow rates for SQ48T03150 converter with E height pins mounted vertically with air flowing from pin 3 to pin 1, MOSFET temperature ≤ 120°C, Vin = 48 V.



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2. OPERATION

2.1 INPUT AND OUTPUT IMPEDANCE

These power converters have been designed to be stable with no external capacitors when used in low inductance input and output circuits.

However, in many applications, the inductance associated with the distribution from the power source to the input of the converter can affect the stability of the converter. The addition of a 33 μF electrolytic capacitor with an ESR $< 1 \Omega$ across the input helps ensure stability of the converter. In many applications, the user has to use decoupling capacitance at the load. The power converter will exhibit stable operation with external load capacitance up to 1,000 μF .

2.2 ON/OFF (Pin 2)

The ON/OFF pin is used to turn the power converter on or off remotely via a system signal. There are two remote control options available, positive logic and negative logic & both are referenced to $V_{in(-)}$. Typical connections are shown in Fig. 2.

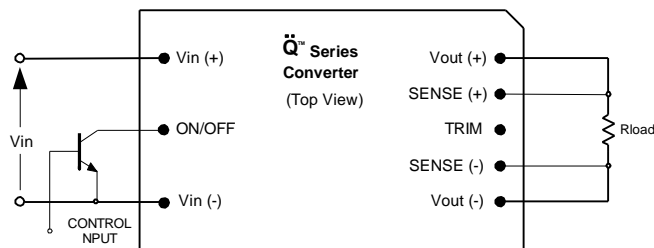


Figure 2. Circuit configuration for ON/OFF function.

The positive logic version turns on when the ON/OFF pin is at logic high and turns off when at logic low. The converter is on when the ON/OFF pin is left open.

The negative logic version turns on when the pin is at logic low and turns off when the pin is at logic high. The ON/OFF pin can be hard wired directly to $V_{in(-)}$ to enable automatic power up of the converter without the need of an external control signal.

ON/OFF pin is internally pulled-up to 5 V through a resistor. A mechanical switch, open collector transistor, or FET can be used to drive the input of the ON/OFF pin. The device must be capable of sinking up to 0.2 mA at a low level voltage of ≤ 0.8 V. An external voltage source of ± 20 V max. may be connected directly to the ON/OFF input, in which case it should be capable of sourcing 1 mA max. See the Start-up Information section for system timing waveforms associated with use of the ON/OFF pin.

2.3 REMOTE SENSE (PINS 5 AND 7)

The remote sense feature of the converter compensates for voltage drops occurring between the output pins of the converter and the load. The SENSE(-) (Pin 5) and SENSE(+) (Pin 7) pins should be connected at the load or at the point where regulation is required (see Fig. 3).

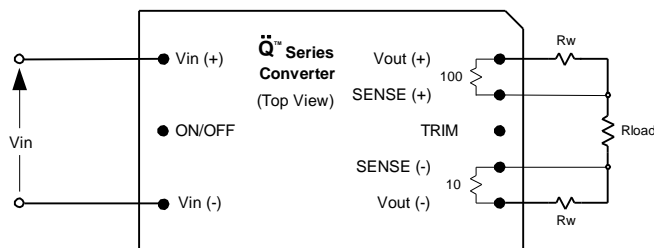


Figure 3. Remote sense circuit configuration.

If remote sensing is not required, the SENSE(-) pin must be connected to the Vout(-) pin (Pin 4), and the SENSE(+) pin must be connected to the Vout(+) pin (Pin 8) to ensure the converter will regulate at the specified output voltage. If these connections are not made, the converter will deliver an output voltage that is slightly higher than the specified value.

Because the sense leads carry minimal current, large traces on the end-user board are not required. However, sense traces should be located close to a ground plane to minimize system noise and insure optimum performance. When wiring discretely, twisted pair wires should be used to connect the sense lines to the load to reduce susceptibility to noise.

The converter's output over-voltage protection (OVP) senses the voltage across Vout(+) and Vout(-), and not across the sense lines, so the resistance (and resulting voltage drop) between the output pins of the converter and the load should be minimized to prevent unwanted triggering of the OVP.

When utilizing the remote sense feature, care must be taken not to exceed the maximum allowable output power capability of the converter, equal to the product of the nominal output voltage and the allowable output current for the given conditions. When using remote sense, the output voltage at the converter can be increased by as much as 10% above the nominal rating in order to maintain the required voltage across the load. Therefore, the designer must, if necessary, decrease the maximum current (originally obtained from the derating curves) by the same percentage to ensure the converter's actual output power remains at or below the maximum allowable output power.

2.4 OUTPUT VOLTAGE ADJUST / TRIM (PIN 6)

The converter's output voltage can be adjusted up 10% or down 20% relative to the rated output voltage by the addition of an externally connected resistor.

The TRIM pin should be left open if trimming is not being used. To minimize noise pickup, a 0.1 μF capacitor is connected internally between the TRIM and SENSE(-) pins.

To increase the output voltage, refer to Fig. 4. A trim resistor, R_{T-INCR}, should be connected between the TRIM (Pin 6) and SENSE(+) (Pin 7), with a value of:

$$R_{T-INCR} = \frac{5.11(100 + \Delta)V_{O-NOM} - 626}{1.225\Delta} - 10.22 \quad [k\Omega]$$

where,

R_{T-INCR} = Required value of trim-up resistor kΩ

V_{O-NOM} = Nominal value of output voltage [V]

$$\Delta = \left| \frac{(V_{O-REQ} - V_{O-NOM})}{V_{O-NOM}} \right| \times 100 \quad [\%]$$

V_{O-REQ} = Desired (trimmed) output voltage [V].

When trimming up, care must be taken not to exceed the converter's maximum allowable output power. See previous the section for a complete discussion of this requirement.

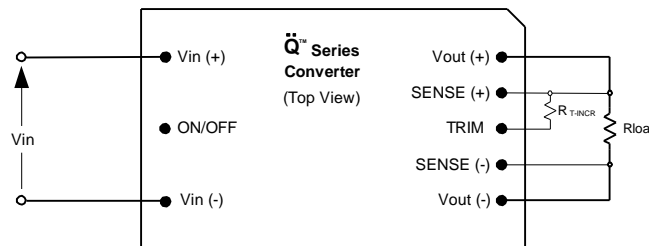


Figure 4. Configuration for increasing output voltage.

To decrease the output voltage (Fig. 5), a trim resistor, R_{T-DECR} , should be connected between the TRIM (Pin 6) and SENSE(-) (Pin 5), with a value of:

$$R_{T-DECR} = \frac{511}{\Delta} - 10.22 \quad [\text{k}\Omega]$$

where,

R_{T-DECR} = Required value of trim-down resistor [kΩ]

and Δ is as defined above.

NOTE: The above equations for calculation of trim resistor values match those typically used in conventional industry-standard quarter bricks.

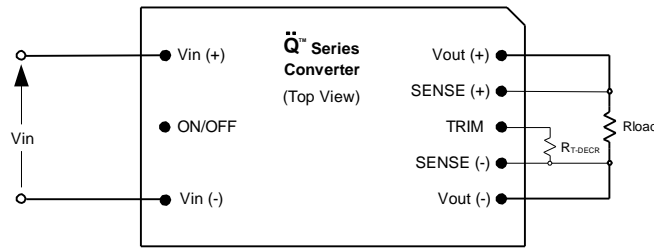


Figure 5. Configuration for decreasing output voltage.

Trimming/sensing beyond 110% of the rated output voltage is not an acceptable design practice, as this condition could cause unwanted triggering of the output over-voltage protection (OVP) circuit. The designer should ensure that the difference between the voltages across the converter's output pins and its sense pins does not exceed 1.50 V, or:

$$[V_{OUT(+)} - V_{OUT(-)}] - [V_{SENSE(+)} - V_{SENSE(-)}] \leq 1.50 \text{ [V]}$$

This equation is applicable for any condition of output sensing and/or output trim.

3. PROTECTION FEATURES

3.1 INPUT UNDERVOLTAGE LOCKOUT

Input under-voltage lockout is standard with this converter. The converter will shut down when the input voltage drops below a pre-determined voltage.

The input voltage must be at least 42 V for the converter to turn on. Once the converter has been turned on, it will shut off when the input voltage drops below 38 V. This feature is beneficial in preventing deep discharging of batteries used in telecom applications.

3.2 OUTPUT OVERCURRENT PROTECTION (OCP)

The converter is protected against over-current or short circuit conditions. Upon sensing an over-current condition, the converter will switch to constant current operation and thereby begin to reduce output voltage. When the output voltage drops below 7.5 Vdc, the converter will shut down (Fig. 23).

Once the converter has shut down, it will attempt to restart nominally every 100 ms with a 2% duty cycle (Fig 24). The attempted restart will continue indefinitely until the overload or short circuit conditions are removed or the output voltage rises above 7.5 Vdc.

3.3 OUTPUT OVERVOLTAGE PROTECTION (OVP)

The converter will shut down if the output voltage across Vout(+) (Pin 8) and Vout(-) (Pin 4) exceeds the threshold of the OVP circuitry. The OVP circuitry contains its own reference, independent of the output voltage regulation loop. Once the converter has shut down, it will attempt to restart every 100 ms until the OVP condition is removed.

3.4 OVERTEMPERATURE PROTECTION (OTP)

The converter will shut down under an over-temperature condition to protect itself from overheating caused by operation outside the thermal derating curves, or operation in abnormal conditions such as system fan failure. The over-temperature protection circuit turns the converter off when the temperature at a sensed location reaches 125°C (typical). Once the converter has shut down, it will restart when the temperature at the sensed location falls below 115°C.

3.5 SAFETY REQUIREMENTS

The converters meet North American and International safety regulatory requirements per UL60950 and EN60950. Basic Insulation is provided between input and output. To comply with safety agencies requirements, an input line fuse must be used external to the converter. A 3 A fuse is recommended for use with this product.

3.6 ELECTROMAGNETIC COMPATIBILITY (EMC)

EMC requirements must be met at the end-product system level, as no specific standards dedicated to EMC characteristics of board mounted component dc-dc converters exist. However, Bel tests its converters to several system level standards, primary of which is the more stringent EN55022, *Information technology equipment - Radio disturbance characteristics - Limits and methods of measurement*.

With the addition of a simple external filter (see application notes), all versions of the SQ48T03 converters pass the requirements of Class B conducted emissions per EN55022 and FCC, and meet at a minimum, Class A radiated emissions per EN 55022 and Class B per FCC Title 47CFR, Part 15-J. Please contact Bel Applications Engineering for details of this testing.

4. CHARACTERIZATION

4.1 GENERAL INFORMATION

The converter has been characterized for many operational aspects, to include thermal derating (maximum load current as a function of ambient temperature and airflow) for vertical and horizontal mounting, efficiency, start-up and shutdown parameters, output ripple and noise, transient response to load step-change, overload and short circuit. The following pages contain specific plots or waveforms associated with the converter. Additional comments for specific data are provided below.

4.2 TEST CONDITIONS

All data presented were taken with the converter soldered to a test board, specifically a 0.060" thick printed wiring board (PWB) with four layers. The top and bottom layers were not metalized. The two inner layers, comprising two-ounce copper, were used to provide traces for connectivity to the converter.

The lack of metalization on the outer layers as well as the limited thermal connection ensured that heat transfer from the converter to the PWB was minimized. This provides a worst-case but consistent scenario for thermal derating purposes.

All measurements requiring airflow were made in Bel's vertical and horizontal wind tunnel facilities using Infrared (IR) thermography and thermocouples for thermometry.

Ensuring components on the converter do not exceed their ratings is important to maintaining high reliability. If one anticipates operating the converter at or close to the maximum loads specified in the derating curves, it is prudent to check actual operating temperatures in the application. Thermographic imaging is preferable; if this capability is not available, then thermocouples may be used. Bel recommends the use of AWG #40 gauge thermocouples to ensure measurement accuracy. Careful routing of the thermocouple leads will further minimize measurement error. Refer to *Figure 25* for optimum measuring thermocouple location.

4.3 THERMAL DERATING

Load current vs. ambient temperature and airflow rates are given in Figs. 9-12 for through-hole version. Ambient temperature was varied between 25°C and 85°C, with airflow rates from 0 to 500 LFM (0 to 2.5 m/s), and vertical and horizontal converter mounting.

For each set of conditions, the maximum load current was defined as the lowest of:

- (i) The output current at which either any FET junction temperature did not exceed a maximum specified temperature (120°C) as indicated by the thermographic image, or
- (ii) The output current recorded prior to initiation of the over temperature protection circuit, or
- (iii) The nominal rating of the converter (3.5 A).

4.4 EFFICIENCY

Efficiency vs. load current plots are shown in Figs. 11 and 12 for ambient temperature of 25°C, airflow rate of 300 LFM (1.5 m/s), both vertical and horizontal orientations, and input voltages of 43 V, 48 V and 72 V. Also, plots of efficiency vs. load current, as a function of ambient temperature with $V_{in} = 48$ V, airflow rate of 200 LFM (1 m/s) are shown for both a vertically and horizontally mounted converter in Figs. 13 and 14, respectively.

4.5 STARTUP

Output voltage waveforms, during the turn-on transient using the ON/OFF pin for full rated load currents (resistive load) are shown without and with 1,000 μ F load capacitance in Figs. 15 and 16, respectively.

4.6 RIPPLE AND NOISE

Figure 21 shows the output voltage ripple waveform, measured at full rated load current with a 10 μ F tantalum and 1 μ F ceramic capacitor across the output. Note that all output voltage waveforms are measured across a 1 μ F ceramic capacitor. The input reflected ripple current waveforms are obtained using the test setup shown in *Figure 21*. The corresponding waveforms are shown in *Figure 20* and *Figure 22*.

4.7 STARTUP INFORMATION (USING NEGATIVE ON/OFF)

Scenario #1: Initial Start-up From Bulk Supply
 ON/OFF function enabled, converter started via application of V_{IN} . See Figure 6.

Time	Comments
t_0	ON/OFF pin is ON; system front end power is toggled on, V_{IN} to converter begins to rise.
t_1	V_{IN} crosses Under-Voltage Lockout protection circuit threshold; converter enabled.
t_2	Converter begins to respond to turn-on command (converter turn-on delay).
t_3	Converter V_{OUT} reaches 100% of nominal value.

For this example, the total converter start-up time ($t_3 - t_1$) is typically 4 ms.

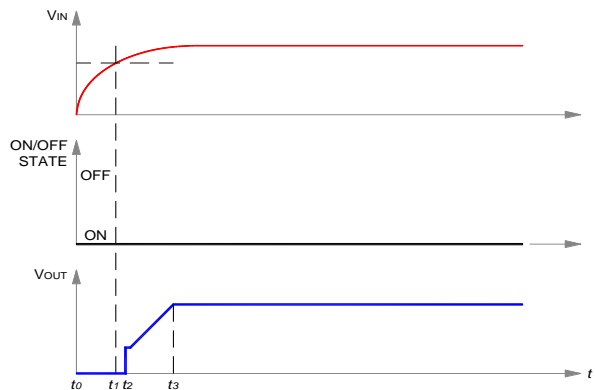


Figure 6. Startup scenario #1.

Scenario #2: Initial Start-up Using ON/OFF Pin

With V_{IN} previously powered, converter started via ON/OFF pin. See Figure 7.

Time	Comments
t_0	V_{INPUT} at nominal value.
t_1	Arbitrary time when ON/OFF pin is enabled (converter enabled).
t_2	End of converter turn-on delay.
t_3	Converter V_{OUT} reaches 100% of nominal value.

For this example, the total converter start-up time ($t_3 - t_1$) is typically 4 ms.

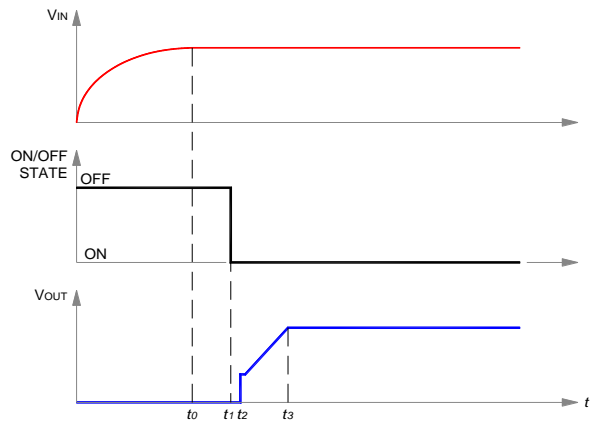


Figure 7. Startup scenario #2.

Scenario #3: Turn-off and Restart Using ON/OFF Pin

With V_{IN} previously powered, converter is disabled and then enabled via ON/OFF pin. See Figure 8.

Time	Comments
t_0	V_{IN} and V_{OUT} are at nominal values; ON/OFF pin ON.
t_1	ON/OFF pin arbitrarily disabled; converter output falls to zero; turn-on inhibit delay period (100 ms typical) is initiated, and ON/OFF pin action is internally inhibited.
t_2	ON/OFF pin is externally re-enabled. If $(t_2 - t_1) \leq 100$ ms, external action of ON/OFF pin is locked out by start-up inhibit timer. If $(t_2 - t_1) > 100$ ms, ON/OFF pin action is internally enabled.
t_3	Turn-on inhibit delay period ends. If ON/OFF pin is ON, converter begins turn-on; if off, converter awaits ON/OFF pin ON signal; see Figure F.
t_4	End of converter turn-on delay.
t_5	Converter V_{OUT} reaches 100% of nominal value.

For the condition, $(t_2 - t_1) \leq 100$ ms, the total converter start-up time ($t_5 - t_2$) is typically 104 ms. For $(t_2 - t_1) > 100$ ms, start-up will be typically 4 ms after release of ON/OFF pin.

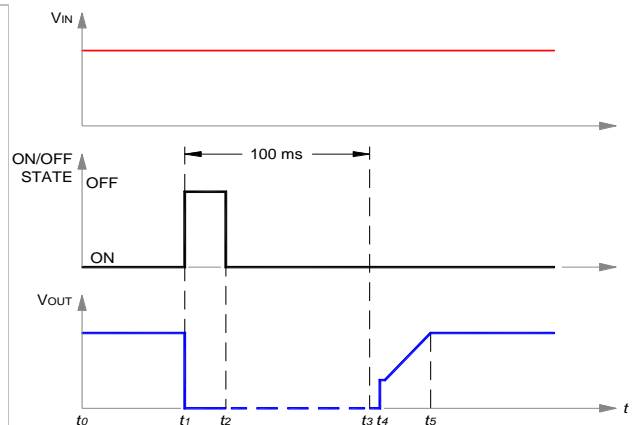


Figure 8. Startup scenario #3.



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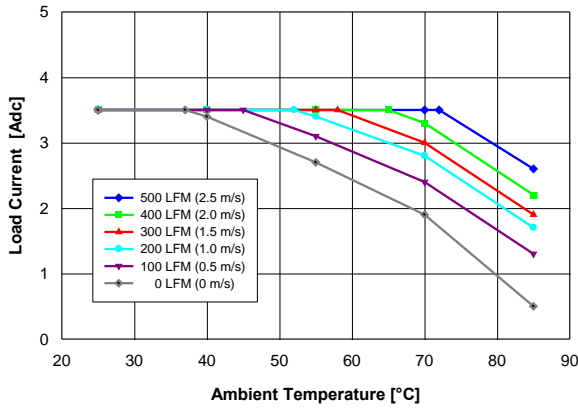


Figure 9. Available load current vs. ambient air temperature and airflow rates for SQ48T03150 converter with E height pins mounted vertically with $V_{in} = 48\text{ V}$, air flowing from pin 3 to pin 1 and maximum FET temperature $\leq 120^{\circ}\text{C}$.

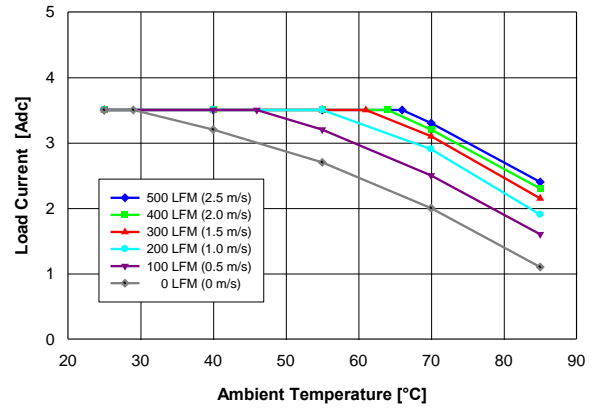


Figure 10. Available load current vs. ambient air temperature and airflow rates for SQ48T03150 converter with E height pins mounted horizontally with $V_{in} = 48\text{ V}$, air flowing from pin 3 to pin 1 and maximum FET temperature $\leq 120^{\circ}\text{C}$.

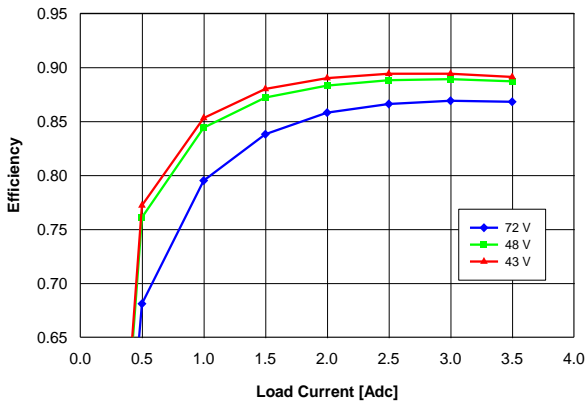


Figure 11. Efficiency vs. load current and input voltage for converter mounted vertically with air flowing from pin 3 to pin 1 at a rate of 300 LFM (1.5 m/s) and $T_a = 25^{\circ}\text{C}$.

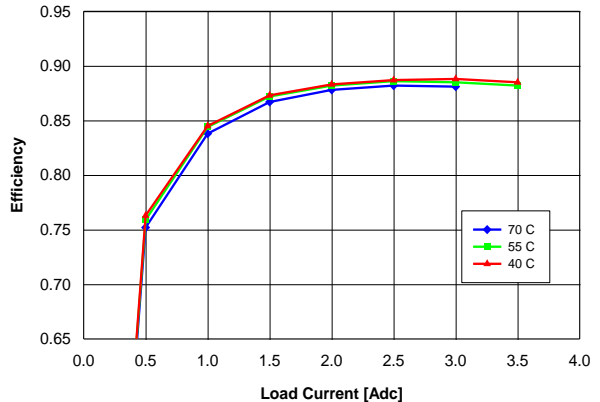


Figure 12. Efficiency vs. load current and ambient temperature for converter mounted vertically with $V_{in} = 48\text{ V}$ and air flowing from pin 3 to pin 1 at a rate of 200 LFM (1.0 m/s).

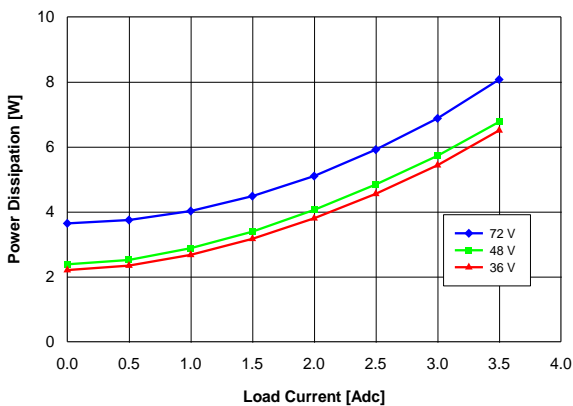


Figure 13. Power dissipation vs. load current and input voltage for converter mounted vertically with air flowing from pin 3 to pin 1 at a rate of 300 LFM (1.5 m/s) and $T_a = 25^{\circ}\text{C}$.

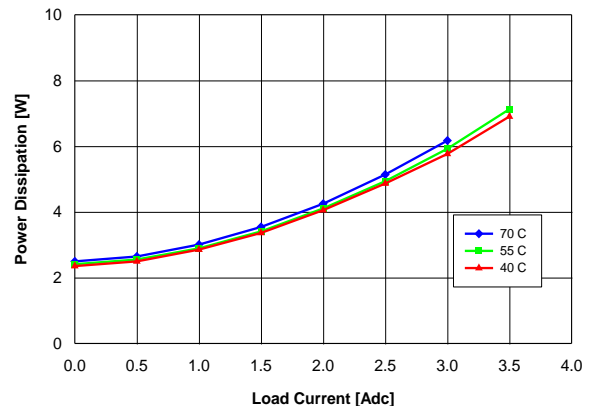


Figure 14. Power dissipation vs. load current and ambient temperature for converter mounted vertically with $V_{in} = 48\text{ V}$ & air flowing from pin 3 to pin 1 at a rate of 200 LFM (1.0 m/s).

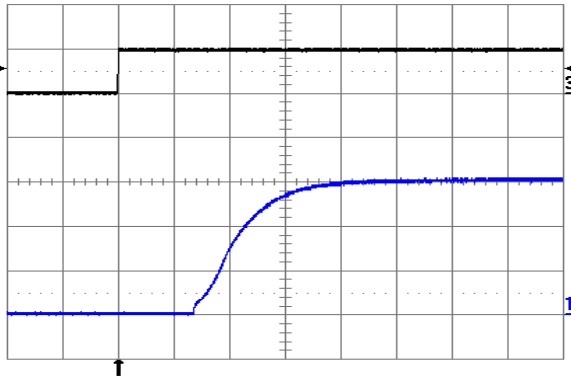


Figure 15. Turn-on transient at full rated load current (resistive) with no output capacitor at $V_{in} = 48\text{ V}$, triggered via ON/OFF pin. Top trace: ON/OFF signal (5 V/div.). Bottom trace: output voltage (5 V/div.) Time scale: 1 ms/div.

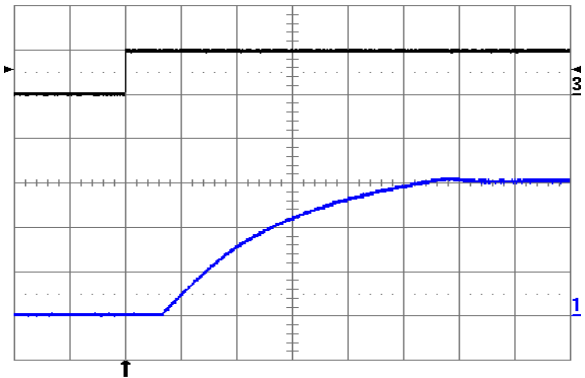


Figure 16. Turn-on transient at full rated load current (resistive) plus 1,000 μF at $V_{in} = 48\text{ V}$, triggered via ON/OFF pin. Top trace: ON/OFF signal (5 V/div.). Bottom trace: output voltage (5 V/div.). Time scale: 2 ms/div.

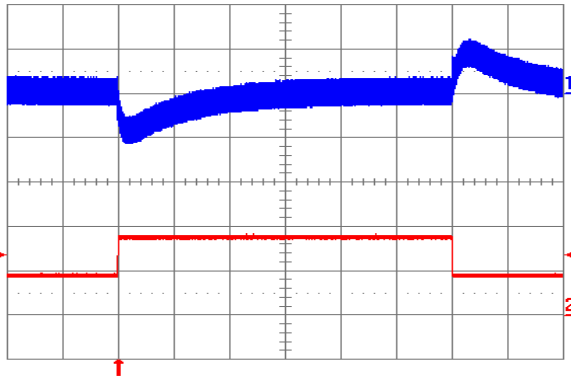


Figure 17. Output voltage response to load current step-change (0.875 A – 1.75 A – 0.875 A) at $V_{in} = 48\text{ V}$. Top trace: output voltage (200 mV/div.). Bottom trace: load current (1 A/div.). Current slew rate: 0.1 A/ μs . $C_o = 1\ \mu\text{F}$ ceramic. Time scale: 0.5 ms/div.

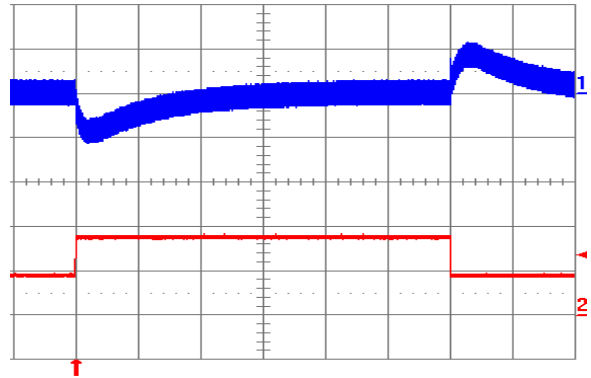


Figure 18. Output voltage response to load current step-change (0.875 A – 1.75 A – 0.875 A) at $V_{in} = 48\text{ V}$. Top trace: output voltage (200 mV/div.). Bottom trace: load current (1 A/div.). Current slew rate: 5 A/ μs . $C_o = 47\ \mu\text{F}$ tantalum + 1 μF ceramic. Time scale: 0.5 ms/div.

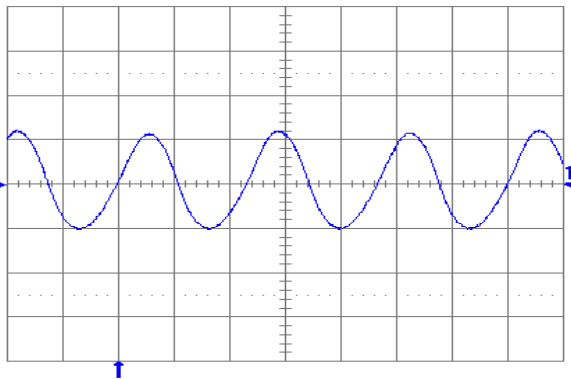


Figure 19. Output voltage ripple (50 mV/div.) at full rated load current into a resistive load with $C_o = 10\ \mu\text{F}$ tantalum + 1 μF ceramic and $V_{in} = 48\text{ V}$. Time scale: 1 μs /div.

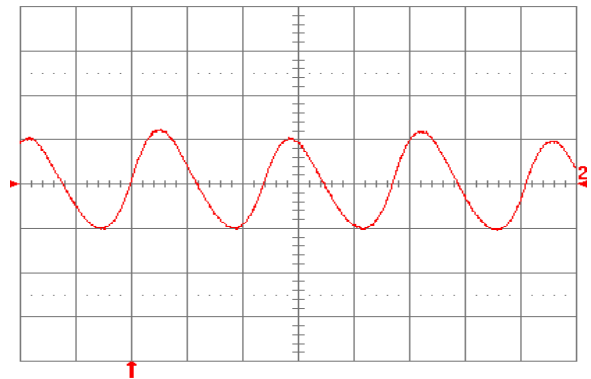


Figure 20. Input reflected ripple current, i_c (100 mA/div), measured at input terminals at full rated load current and $V_{in} = 48\text{ V}$. Refer to Fig. 21 for test setup. Time scale: 1 μs /div.

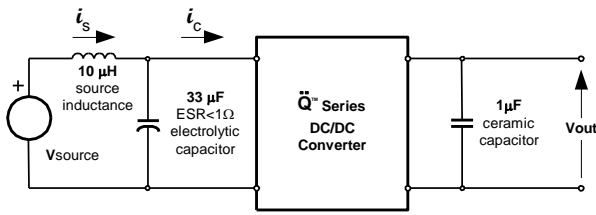


Figure 21. Test Set-up for measuring input reflected ripple currents, i_c and i_s .

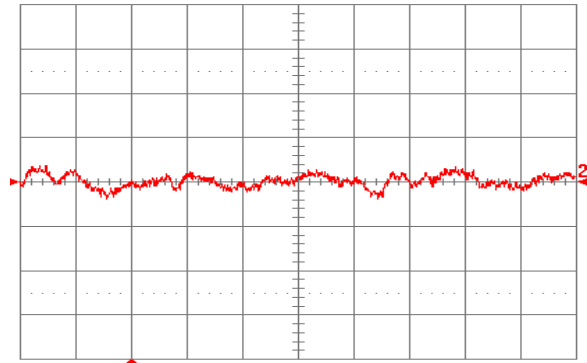


Figure 22. Input reflected ripple current, i_s (10 mA/div), measured through 10 μ H at the source at full rated load current and $V_{in} = 48$ V. Refer to Fig. 21 for test setup. Time scale: 1 μ s/div.

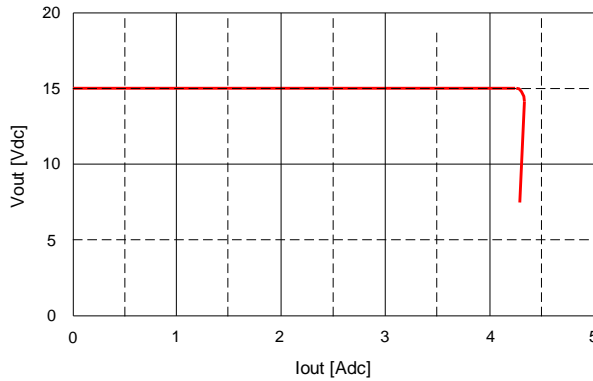


Figure 23. Output voltage vs. load current showing current limit point and converter shutdown point. Input voltage has almost no effect on current limit characteristic.

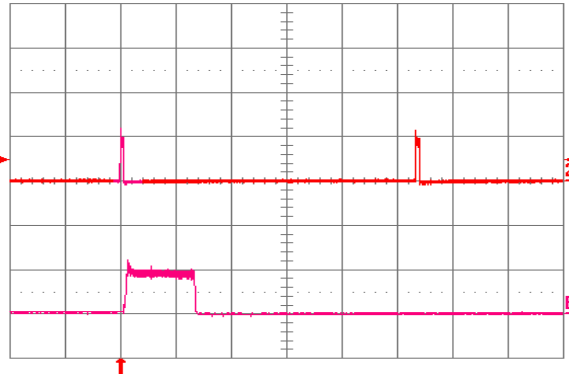


Figure 24. Load current (top trace, 5 A/div, 20 ms/div) into a 10 $m\Omega$ short circuit during restart, at $V_{in} = 48$ V. Bottom trace (5 A/div, 1 ms/div) is an expansion of the on-time portion of the top trace.

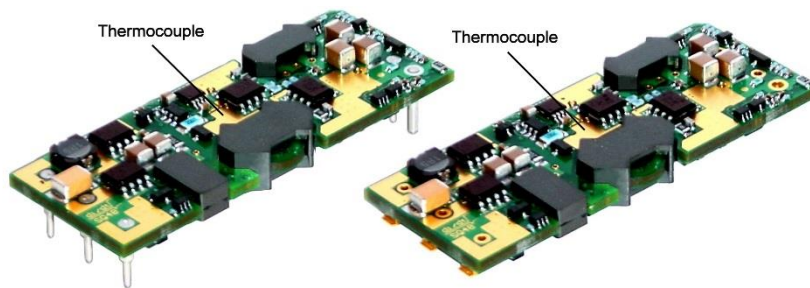
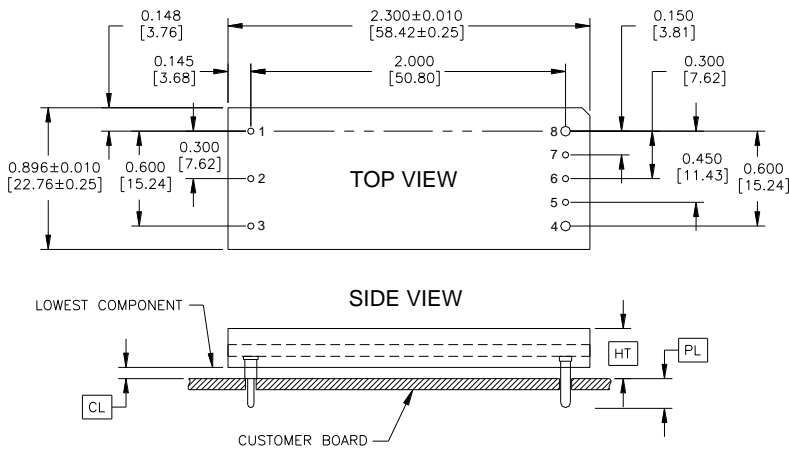


Figure 25. Location of the thermocouple for thermal testing.

5. MECHANICAL PARAMETERS



- All dimensions are in inches [mm]
- Pins 1-3 and 5-3.5 Are $\varnothing 0.040''$ [1.02] with $\varnothing 0.078''$ [1.98] shoulder
- Pins 4 and 8 are $\varnothing 0.062''$ [1.57] without shoulder
- Pin Material: Brass
- Pin Finish: Tin/Lead over Nickel
- Converter Weight: 0.53 oz [15 g]

PAD/PIN CONNECTIONS	
Pad/Pin #	Function
1	Vin (+)
2	ON/OFF
3	Vin (-)
4	Vout (-)
5	SENSE(-)
6	TRIM
7	SENSE(+)
8	Vout (+)

Pin Option	PL Pin Length
	± 0.005 [± 0.13]
A	0.188 [4.77]
B	0.145 [3.68]
C	0.110 [2.79]

Height Option	HT (Max. Height)	CL (Min. Clearance)
		+0.000 [+0.00]
	-0.038 [- 0.97]	-0.000 [- 0.00]
A	0.303 [7.69]	0.030 [0.77]
B	0.336 [8.53]	0.063 [1.60]
C	0.500 [12.70]	0.227 [5.77]
D	0.400 [10.16]	0.127 [3.23]
E	0.282 [7.16]	0.009 [0.23]

6. ORDERING INFORMATION

Product Series	Input Voltage	Mounting Scheme	Rated Load Current	Output Voltage	ON/OFF Logic	Maximum Height [HT]	Pin Length [PL]	Special Features	RoHS	
SQ	48	T	03	150	-	N	B	A	0	G
1/8 th Brick Format	43-75 V	T \Rightarrow Through-hole	3.3 A	150 \Rightarrow 15V	N \Rightarrow Negative P \Rightarrow Positive	Through hole A \Rightarrow 0.303" B \Rightarrow 0.336" C \Rightarrow 0.500" D \Rightarrow 0.400" E \Rightarrow 0.282"	Through hole A \Rightarrow 0.188" B \Rightarrow 0.145" C \Rightarrow 0.110"	0 \Rightarrow STD	G \Rightarrow RoHS compliant for all six substances	

The example above describes P/N SQ48T03150-NBA0: 43-75 V input, through-hole mounting, 3.5 A @ 15 V output, negative ON/OFF logic, a maximum height of 0.336", and a through the board pin length of 188". Please consult factory regarding availability of a specific version.

For more information on these products consult: tech.support@psbel.com

NUCLEAR AND MEDICAL APPLICATIONS - Products are not designed or intended for use as critical components in life support systems, equipment used in hazardous environments, or nuclear control systems.

TECHNICAL REVISIONS - The appearance of products, including safety agency certifications pictured on labels, may change depending on the date manufactured. Specifications are subject to change without notice.



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