



AC-DC Front End Power Supplies

The TET1500 is a 1500 W AC to DC power-factor-corrected (PFC) power supply that converts standard AC or HVDC power into a main output of 12 VDC for powering intermediate bus architectures (IBA) in high performance and reliability servers, routers, and network switches.

Key Features & Benefits

- High Efficiency, typ. 96% efficiency at half load
- Universal input voltage range: 90-305 VAC
- High voltage DC input: 180-400 VDC
- AC input with power factor correction
- Always-On standby output:
 - Programmable 3.3 V / 5 V (16.5 W)
- Hot-plug capable
- Parallel operation with active analog current sharing
- Digital controls for improved performance
- High density design: 35 W/in³
- Small form factor (W x H x L): 54.5 x 40.0 x 321.5 mm
 2.14 x 1.57 x 12.66 in
- I2C communication interface for control, programming and monitoring with Power Management Bus protocol
- Over temperature, output over voltage and over current protection
- 256 Bytes of EEPROM for user information
- 2 Status LEDs: OK and FAIL with fault signalling

Applications

- High Performance Servers
- Routers
- Switches



ORDERING INFORMATION

MODELS WITH PROGRAMMABLE 3.3 V / 5 V STANDBY OUTPUT

TET	1500	-	12	-	054	N	AH
Product Family	Power Level	Dash	V1 Output	Dash	Width	Airflow	Input ³
TET Front-Ends	1500 W		12 V		54 mm	N: Normal ¹ R: Reverse ²	A: C14 Socket AC: C16 Socket AH: HVDC Socket

- "N" Normal Airflow from Output connector to Input AC socket
 - Ordering PN: TET1500-12-054NA for C14 AC input connector, input range is 180 VDC ~ 350 VDC and 90 VAC ~ 264 VAC Ordering PN: TET1500-12-054NAC for C16 AC input connector, input range is 180 VDC ~ 350 VDC and 90 VAC ~ 264 VAC
- Ordering PN: TET1500-12-054NAH for both AC and HVDC (2006G1-BK) input connector, input range is 180 ~ 400 VDC and 90 ~ 305 VAC
- "R" Reverse Airflow from Input AC socket to Output connector
 - Ordering PN: TET1500-12-054RA for C14 AC input connector, input range is 180 VDC ~ 350 VDC and 90 VAC ~ 264 VAC
 - Ordering PN: TET1500-12-054RAC for C16 AC input connector, input range is 180 VDC ~ 350 VDC and 90 VAC ~ 264 VAC
- Ordering PN: TET1500-12-054RAH for both AC and HVDC (2006G1-BK) input connector, input range is 180 ~ 400 VDC and 90 ~ 305 VAC
- For difference of the AC socket and mechanical outline refer to section 14.

OVERVIEW 2.

The TET1500 Series AC/DC power supply is combination of analog and DSP control, highly efficient front-end power supply. It incorporates resonance-soft-switching technology and interleaved power trains to reduce component stresses, providing increased system reliability and very high efficiency. With a wide input operational voltage range and minimal derating of output power with input voltage and temperature, the TET1500 power supply maximizes power availability in demanding server, network, and other high availability applications. The supply is fan cooled and ideally suited for integration with a matching airflow paths. The PFC stage is digitally controlled using a state-of-the-art digital signal processing algorithm to guarantee best efficiency and unity power factor over a wide operating range.

The DC/DC stage uses soft switching resonant techniques in conjunction with synchronous rectification. An active OR-ing device on the output ensures no reverse load current and renders the supply ideally suited for operation in redundant power systems. The always-on standby output, provides power to external power distribution and management controllers. It is protected with an active OR-ing device for maximum reliability. Status information is provided with front-panel LEDs.

In addition, the power supply can be controlled and the fan speed set via the I2C bus. The I2C bus allows full monitoring of the supply, including input and output voltage, current, power, and inside temperatures. The fan speed is adjusted automatically depending on the actual power demand and supply temperature and can be overridden through the I2C bus.

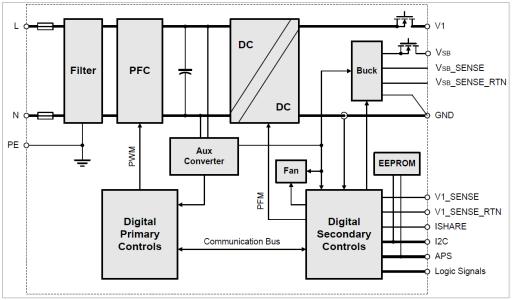


Figure 1. TET1500 Series Block Diagram



3. ABSOLUTE MAXIMUM RATINGS

Stresses in excess of the absolute maximum ratings may cause performance degradation, adversely affect long-term reliability, and cause permanent damage to the supply.

PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
Vi maxc	Maximum Innut	Continuous			305	VAC
VIIIIAXC	Maximum Input	Continuous			400	VDC

4. INPUT SPECIFICATIONS

General Condition: $T_A = 0...50$ °C unless otherwise specified.

V_{inom} Nominal Input Voltage10023027 V_i Input Voltage RangesNormal operating (V_{imin} to V_{imax})9030 V_i Vin = 100 VAC, 1000W18040 V_i Vin = 200 VAC, 1500 W30	0 VDC 5 VAC
V_i Input Voltage Ranges Normal operating (V_{imin} to V_{imax}) 180 Vin = 100 VAC, 1000W Vin = 100 VAC, 1000W	5 VAC
V_i Input Voltage Ranges Normal operating (V_{imin} to V_{imax}) 180 40 Vin = 100 VAC, 1000W 11 max Max Input Current	
Vin = 100 VAC, 1000W // Max Input Current	0 1/00
li max Max Input Current	0 VDC
Win = 200 VAC 1500 W	2 Arms
VIII = 200 VAO, 1300 VV	Arms
I_{ip} Inrush Current Limitation $V_{i min}$ to $V_{i max}$, $T_{PTC} = 25^{\circ}\text{C}$ (Figure 4)) A _p
<i>F_i</i> Input Frequency 40 50/60 7) Hz
Vi = 230 VAC/277 VAC, 50 Hz and 60 Hz	
10%, Load 0.90	W/VA
PF Power Factor 20%, Load 0.96	W/VA
50%, Load 0.98	W/VA
100%, Load 0.98	W/VA
Vi = 230 VAC/277 VAC, 50 Hz and 60 Hz	
10%, Load 1) %
<i>iTHD</i> Input iTHD 20%, Load 1) %
50%, Load	%
100%, Load	%
V _{i on} Turn-on Input Voltage ¹ Ramping up) VAC
V _{i on} Turn-on Input Voltage ¹ Ramping up 172 17	7 VDC
V _{i off} Turn-off Input Voltage Ramping down	VAC
167 17	2 VDC
V _{IN} = 230 VAC/277VAC, 10% load, T _A = 25°C 90	
Efficiency without Fan at AC VIN = 230 VAC/277VAC, 20% load, TA = 25°C 95	%
input $V_{IN} = 230 \text{ VAC}/277\text{VAC}, 50\% \text{ load}, T_A = 25^{\circ}\text{C}$ 96	%
V _{IN} = 230 VAC/277VAC, 100% load, T _A = 25°C 93	
Trout_hold 12V1 Hold-up Time After last AC zero point to $V_1 \ge 11.4$ V, V _{SB} within regulation, $V_2 = 230$ VAC, $P_{X \text{ nom}}$ 10	ms
T _{Vsb_hold} Vsb Hold-up Time V _{SB} , full load 100	ms

Table 1. Input Specification

¹ The Front-End is provided with a minimum hysteresis of 5V during turn-on and turn-off within the ranges



Asia-Pacific

Europe, Middle East

North America

+86 755 298 85888

+353 61 49 8941

4.1 INPUT FUSE

The PSU has dual pole fusing for the AC input. Time-lag 16A input fuse (5 x 20 mm) in series with the L line and N line inside the power supply protect against severe defects. The fuses are not accessible from the outside and are therefore not serviceable parts.

4.2 INRUSH CURRENT

The AC-DC power supply exhibits an X-capacitance of only 2.1 μF, resulting in a low and short peak current when the supply is connected to the mains. The internal bulk capacitor will be charged through a PTC which will limit the inrush current.

NOTE: Do not repeat plug-in/out operations within a short time, or else the internal in-rush current limiting device (PTC) may not sufficiently cool down and excessive inrush current or component failure(s) may result.

4.3 INPUT UNDER-VOLTAGE

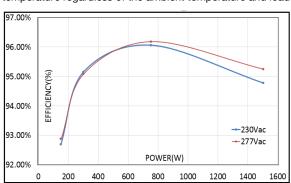
If the sinusoidal input voltage stays below the input under voltage lockout threshold Vi on, the supply will be inhibited. Once the input voltage returns within the normal operating range, the supply will return to normal operation again.

4.4 POWER FACTOR CORRECTION

Power factor correction (PFC) is achieved by controlling the input current waveform synchronously with the input voltage. A digital controller is implemented giving outstanding PFC results over a wide input voltage and load ranges. The input current will follow the shape of the input voltage.

4.5 EFFICIENCY

High efficiency (see *Figure 2*) is achieved by using state-of-the-art silicon power devices in conjunction with soft-transition topologies minimizing switching losses and a full digital control scheme. Synchronous rectifiers on the output reduce the losses in the high current output path. The speed of the fan is digitally controlled to keep all components at an optimal operating temperature regardless of the ambient temperature and load conditions.



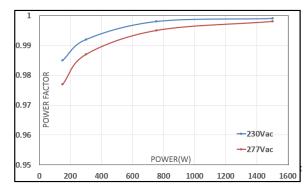


Figure 2. Efficiency vs. Load current (ratio metric loading)

Figure 3. Power factor vs. Load current

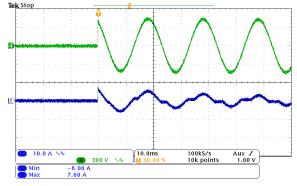


Figure 4. Inrush current, Vin = 305 VAC, 90°, CH1: Iin (10A/div), CH4: Vin (300V/div)



4.6 AC LINE TRANSIENT SPECIFICATION

AC line transient conditions shall be defined as "sag" and "surge" conditions. "Sag" conditions are also commonly referred to as "brownout", these conditions will be defined as the AC line voltage dropping below nominal voltage conditions. "Surge" will be defined to refer to conditions when the AC line voltage rises above nominal voltage, the power supply shall meet the requirements under the following AC line sag and surge conditions. The peak current shall not exceed 3 times inrush current during ride through time. There shall be 3.5s of normal AC input power between each dropout.

AC Line Sag (3.5 sec interval between each sagging)

Duration	Sag	Operating AC Voltage	Load condition	Performance Criteria*
500ms	30%	200 VAC	1* P _{1 nom}	Class A
200ms	60%	200 VAC	1* P _{1 nom}	Class B
20ms	100%	200 VAC	0.5* <i>P</i> _{1 nom}	Class A
10ms	100%	200 VAC	1* P _{1 nom}	Class A

Table 2. AC Line Sag Transient Performance

Note: 1. Performance criteria Class A is "Normal performance within the specification limits".

2. Performance criteria Class B is "Temporary degradation or loss of function or with is self-recoverable".

5. OUTPUT SPECIFICATIONS

General Condition: Ta = 0... 50°C unless otherwise specified.

Main Output Voltage V₁ nom Nominal Output Setpoint Accuracy 0.5 · h nom, Tamb = 25 °C 12 VDC V₁ set Output Setpoint Accuracy 100 M in mon, Tamb = 25 °C -0.5 +0.5 % In mon dV₁ tot Total Regulation V min to V max, 0 to 100% h nom, Tamb to Ta max -2 +2 % In mon P₁ nom Nominal Output Power 180 VAC > V _n ≥ 180 VAC, V = 12 VDC 1500 W P₁ nom Nominal Output Current 180 VAC > V _n ≥ 180 VAC, V = 12 VDC 1000 W P₁ nom Nominal Output Current 264 VAC > V _n ≥ 180 VAC, V = 12 VDC 125 ADC P₁ nom Nominal Output Current 180 VAC > V _n ≥ 180 VAC, V = 12 VDC 125 ADC P₁ nom Nominal Output Current 180 VAC > V _n ≥ 180 VAC, V = 12 VDC 125 ADC P₁ nom Nominal Output Current 180 VAC > V _n ≥ 180 VAC, V = 12 VDC 125 ADC P₁ nom 180 VAC > V _n ≥ 180 VAC, V ≥ 12 VDC 125 ADC ADC P₁ nom 180 VAC > V _n ≥ 180 VAC, V ≥ 12 VDC 125 ADC ADC	PARAMETI	ER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
V1 set Output Setpoint Accuracy 0.5 ⋅ h nom, Tamb = 25 °C -0.5 +0.5 % Vi nom dV1 tot Total Regulation V min to V max, 0 to 100% h nom, Tamin to Ta max -2 +2 % Vi nom P1 nom Nominal Output Power 305 VAC > V _{In} ≥ 180 VAC, V _I = 12 VDC 1500 W V0 VDC > V _{In} ≥ 90 VAC, V _{In} ≥ 90 VAC, V _{In} = 12 VDC 1000 W V1 nom Nominal Output Current 264 VAC > V _{In} ≥ 90 VAC, V _{In} = 12 VDC 125 ADC V1 nom Nominal Output Current 400 VDC > V _{In} ≥ 180 VAC, V _{In} = 12 VDC 125 ADC V1 nom Nominal Output Current 400 VDC > V _{In} ≥ 180 VAC, V _{In} = 12 VDC 125 ADC V1 nom V Nominal Output Current 400 VDC > V _{In} ≥ 180 VAC, V _{In} = 12 VDC 125 ADC V1 nom V Nominal Output Current V _{In} ≥ 90 VAC, V _{In} = 12 VDC 83.3 ADC V1 nom V _{In} ≥ 0 VAC, V _{In} = 12 VDC 83.3 ADC V1 nom V _{In} ≥ 0 VAC, V _{In} = 12 VDC 83.3 ADC V1 nom V _{In} ≥ 0 VAC, V _{In} = 12 VDC 83.3 ADC	Main Outpu	t V₁					
V _{I set} Output Setpoint Accuracy -0.5 +0.5 % $V_{I nom}$ dV _{I tot} Total Regulation $V_{I min}$ to $V_{I max}$, 0 to 100% $V_{I nom}$, $V_{I min}$ to $V_{I max}$. -2 +2 % $V_{I nom}$ P _{I nom} Nominal Output Power 305 VAC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 1500 W Isource V_{I nom} Nominal Output Power 180 VAC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 1000 W In nom Nominal Output Current 264 VAC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 125 ADC In nom Nominal Output Current 264 VAC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 125 ADC In nom Nominal Output Current 264 VAC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 125 ADC In nom Nominal Output Current 400 VDC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 125 ADC In nom Nominal Output Current 400 VDC > $V_{I n} ≥ 180$ VAC, $V_{I n} = 12$ VDC 125 ADC In nom Nominal Output Current In nom, $V_{I nom}$ Nom,	V _{1 nom}	Nominal Output Voltage	05.4 T = 25°C		12		VDC
Nominal Output Power 305 VAC > $V_{fh} ≥ 180$ VAC, $V_{fh} = 12$ VDC 1500 W 400 VDC > $V_{fh} ≥ 180$ VDC, $V_{fh} = 12$ VDC 1500 W 180VAC > $V_{fh} ≥ 90$ VAC, $V_{fh} = 12$ VDC 1000 W $V_{fh} > 0$ Nominal Output Current 264 VAC > $V_{fh} ≥ 180$ VAC, $V_{fh} = 12$ VDC 125 ADC 400 VDC > $V_{fh} ≥ 180$ VDC, $V_{fh} = 12$ VDC 125 ADC 400 VDC > $V_{fh} ≥ 180$ VDC, $V_{fh} = 12$ VDC 83.3 ADC V1 pp Output Ripple Voltage $V_{fh} ≥ 90$ VAC, $V_{fh} = 12$ VDC 83.3 ADC V1 pp Output Ripple Voltage $V_{fh} ≥ 00$ MHz BW (See Section 5.1) 120 mVpp dV1 Load Load Regulation $V_{fh} ≥ 00$ Mh nom, 20 MHz BW (See Section 5.1) 100 mV dV1 Line Line Regulation $V_{fh} ≥ 00$ Mh nom 100 mV dV1 share Current Sharing Deviation Deviation from $V_{fh} ≥ 00$ Mn nom, $V_{fh} ≥ 00$ M	V _{1 set}	Output Setpoint Accuracy	0.3 /1 nom, /amb – 23 O	-0.5		+0.5	% 1/1 nom
P1 nom Nominal Output Power 400 VDC > $V_{ln} \ge 180$ VDC, $V_{ln} = 12$ VDC 1500 W I 80VAC > $V_{ln} \ge 90$ VAC, $V_{ln} \ge 12$ VDC 1000 W I nom Nominal Output Current 264 VAC > $V_{ln} \ge 180$ VAC, $V_{ln} = 12$ VDC 125 ADC I nom Nominal Output Current 400 VDC > $V_{ln} \ge 180$ VAC, $V_{ln} = 12$ VDC 125 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC, $V_{ln} = 12$ VDC 83.3 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC, $V_{ln} = 12$ VDC 83.3 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC, $V_{ln} = 12$ VDC 83.3 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC, $V_{ln} \ge 90$ VAC, $V_{ln} = 12$ VDC 83.3 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC, $V_{ln} \ge 90$ VAC, $V_{ln} = 12$ VDC 83.3 ADC I nom Output Ripple Voltage $V_{ln} \ge 90$ VAC,	dV _{1 tot}	Total Regulation	\textit{V}_{1min} to $\textit{V}_{1max},0$ to 100% $\textit{I}_{1nom},\textit{T}_{amin}to\textit{T}_{amax}$	-2		+2	% 1/1 nom
In nomNominal Output Current $264 \text{ VAC} > V_{ln} ≥ 180 \text{ VAC}, V_{l} = 12 \text{ VDC} \\ 400 \text{ VDC} > V_{ln} ≥ 180 \text{ VDC}, V_{l} = 12 \text{ VDC} \\ 180 \text{ VAC} > V_{ln} ≥ 90 \text{ VAC}, V_{l} = 12 \text{ VDC} \\ 180 \text{ VAC} > V_{ln} ≥ 90 \text{ VAC}, V_{l} = 12 \text{ VDC} \\ 180 \text{ VAC} > V_{ln} ≥ 90 \text{ VAC}, V_{l} = 12 \text{ VDC} \\ 125 \text{ ADC} \\ 180 \text{ VAC} > V_{ln} ≥ 90 \text{ VAC}, V_{l} = 12 \text{ VDC} \\ 125 \text{ ADC} \\ 125 $	P _{1 nom}	Nominal Output Power	400 VDC > V _{in} ≥180 VDC, V ₁ = 12 VDC		1500		W
$I_{1 nom}$ Nominal Output Current $400 \text{ VDC} > V_{In} ≥ 180 \text{ VDC}, V_1 = 12 \text{ VDC}$ 125 ADC I_{180} VAC > $I_{In} ≥ 90 \text{ VAC}, V_1 = 12 \text{ VDC}$ 83.3ADC I_{180} VAC > $I_{In} ≥ 90 \text{ VAC}, V_1 = 12 \text{ VDC}$ 83.3ADC I_{180} VI ppOutput Ripple Voltage I_{180} N nom, I_{180} nom, I_{180} Output Ripple Voltage I_{180} MHz BW (See Section 5.1)120mVpp I_{180} Aux Coulombre Vi pp I_{180} Aux Coulombre Vi pp I_{180} Aux Coulombre Vi pp100mV I_{180} Aux Coulombre Vi pp I_{180} Aux Co							* *
180 VAC > $V_{In} \ge 90$ VAC, $V_{In} = 12$ VDC83.3ADC V_{IDP} Output Ripple Voltage V_{Inom} , h_{Inom} , 20 MHz BW (See Section 5.1)120mVpp dV_{ILoad} Load Regulation V_{Inom} , 0 - 100 % h_{Inom} 100mV dV_{ILine} Line Regulation V_{Inom} , V_{Inom} 50mV dV_{Inom} Current Sharing DeviationDeviation from h_{Iot}/N , $h > 10%$ -3+3A dV_{dyn} Dynamic Load Regulation $\frac{\Delta h}{d} = 50\% h_{Inom}$, $h = 5 \dots 100\% $	lt nom	Nominal Output Current	, .				
$dV_1 Load$ Load Regulation $V_1 = V_1 nom_1, 0 - 100 \% h_1 nom$ 100mV $dV_1 Line$ Line Regulation $V_2 = V_1 min V_1 max$ 50mV $dh_1 share$ Current Sharing DeviationDeviation from $h_1 tot/N$, $h > 10\%$ -3+3A dV_{dyn} Dynamic Load Regulation $\frac{\Delta h}{dh} = 50\% h_1 nom_1, h_1 = 5 100\% h_1 nom_1, h_2 = 5$	TI HOIII	Nominal Output Gunent	180 VAC > $V_{in} \ge 90$ VAC, $V_1 = 12$ VDC		83.3		ADC
dV_1 LineLine Regulation $V = V_1 \minV_1 \max$ 50mV dH_1 shareCurrent Sharing DeviationDeviation from h_1 tot / N, $h_1 > 10\%$ -3+3A dV_{dyn} Dynamic Load Regulation $\Delta h_1 = 50\% h_1 \min, h_1 = 5 \dots 100\% h_1 \min, h_1 = 5 \dots 100\% h_1 \min, h_2 = 5 \dots 100\% h_1 \min, h_1 = 5 \dots 100\% h_1 \min, h_2 = 5 \dots 100\% h_2 = 5 \dots 100\%$	V _{1 pp}	Output Ripple Voltage	\mathcal{N}_{nom} , \mathcal{N}_{nom} , 20 MHz BW (See Section 5.1)			120	mVpp
Current Sharing Deviation Deviation from h_{10t}/N , $h > 10\%$ -3 +3 A dV_{dyn} Dynamic Load Regulation $\frac{\Delta h}{dh} = 50\% h_{nom}, h = 5 \dots 100\% h_{no$	dV₁ Load	Load Regulation	Vi = Vi nom, 0 - 100 % In nom		100		mV
dV_{dyn} Dynamic Load Regulation $\Delta h = 50\% h_{\text{nom}}, h = 5 \dots 100\% h_{\text{nom}}, h = 0.6$ $+0.6$ V T_{rec} Recovery Time $\Delta h = 50\% h_{\text{nom}}, h = 5 \dots 100\% h_{\text{nom}}, h = 0.6$ $+0.6$ V $T_{Vout_on_delay}$ Start-up Time from AC 2 ms T_{Vout_rise} Rise Time V V V	dV1 Line	Line Regulation	$V_i = V_i \min V_i \max$		50		mV
Trec Recovery Time $\frac{\Delta h}{dt} = 1 \text{Al} / \text{µs}$ -0.6 $+0.6$ V $\Delta h = 50\% \ h \ \text{nom}, \ h = 5 \dots 100\% \ h \ \text{nom}, \ dh/dt = 1 \text{Al} / \text{µs}, \text{recovery within 1\% of } \text{Vi nom}$ 2 ms $\frac{V}{V_{\text{out_rise}}}$ Start-up Time from AC 3 sec $\frac{V}{V_{\text{out_rise}}}$ Rise Time $\frac{V}{V} = 10 \dots 90\% \ V_{\text{nom}}$ 10 ms	dI _{1 share}	Current Sharing Deviation	Deviation from I_1 tot / N, $I_1 > 10\%$	-3		+3	Α
T_{rec} Recovery Time $dh/dt = 1A/\mu s$, recovery within 1% of $V_{1 nom}$ 2ms $T_{Vout_on_delay}$ Start-up Time from AC3sec T_{Vout_rise} Rise Time $V_1 = 1090\%$ $V_{1 nom}$ 10ms	dV _{dyn}	Dynamic Load Regulation		-0.6		+0.6	V
T_{Vout_rise} Rise Time $V_1 = 1090\% \ V_{1 \text{ nom}}$ 10 ms	Trec	Recovery Time			2		ms
	T _{Vout_on_delay}	Start-up Time from AC				3	sec
C_{Load} Capacitive Loading $T_a = 25$ °C 30000 μF	T _{Vout_rise}	Rise Time	V ₁ = 1090% V _{1 nom}			10	ms
	CLoad	Capacitive Loading	$T_a = 25$ °C			30000	μF



3.3/5 V _{SB} Sta	andby Output						
VSB nom	Nominal Output Voltage	0.5 · k _{B nom} . T _{amb} = 25°C	VSB_SEL = 1		3.3		VDC
VSB nom	Normal Output Voltage	0.5 '/SB nom, /amb = 25 C	$VSB_SEL = 0$		5.0		VDC
V _{SB set}	Output Setpoint Accuracy	$0.5 \cdot I_{SB \text{ nom}}, T_{amb} = 25^{\circ}C$	VSB_SEL = 0/1	-0.5		+0.5	% V _{1nom}
dV _{SB tot}	Total Regulation	V_{imin} to V_{imax} , 0 to 100% I_{SBn}	om, $T_{a \text{ min}}$ to $T_{a \text{ max}}$	-5		+5	% V _{SBnom}
P _{SB nom}	Naminal Output Power	$V_{SB} = 3.3 \text{ VDC},$			16.5		W
►SB nom	Nominal Output Power	$V_{SB} = 5.0 \text{ VDC},$			16.5		VV
1	Nominal Output Current	$V_{SB} = 3.3 \text{ VDC},$			5		ADC
ISB nom		$V_{SB} = 5.0 \text{ VDC},$			3.3		ADC
1/	Output Binnle Voltage	VsB nom, IsB nom, 20 MHz	VSB_SEL = 1			50	m\/nn
V _{SB pp}	Output Ripple Voltage	BW (See Section 5.1)	VSB_SEL = 0			80	mVpp
,	Command Lineitadian	VSB_SEL = 1,		5.25		6.2	ADC
ISB max	Current Limitation	$VSB_SEL = 0,$		3.45		4.5	ADC
dV _{SBdyn}	Dynamic Load Regulation	$\Delta k_{\rm BB} = 50\% k_{\rm SB nom}, k_{\rm SB} = 5 \dots$	100% /sB nom,	-3		+3	% VSBnom
T _{rec}	Recovery Time	$dI_0/dt = 0.5 \text{ A/}\mu\text{s}$, recovery w	$dI_0/dt = 0.5 \text{ A/}\mu\text{s}$, recovery within 1% of I_1/μ nom			250	us
T _{Vout_on_delay}	Start-up Time from AC	V _{SB} = 90% V _{SB nom}				2	sec
T _{Vsb_rise}	Rise Time	V _{SB} = 1090% V _{SB nom}		10		100	ms
<i>C</i> Load	Capacitive Loading	$T_{\text{amb}} = 25^{\circ}\text{C}$				2000	μF

Table 3. Output Specifications

5.1 OUTPUT VOLTAGE RIPPLE

The test set-up shall be following Figure 5

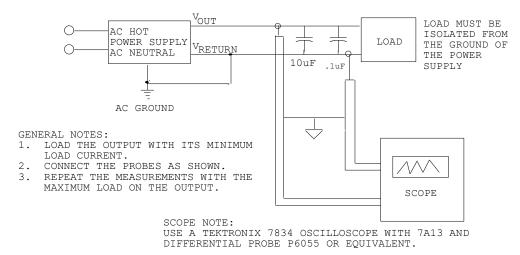


Figure 5. Output ripple test setup

Note: When performing this test, the probe clips and capacitors should be located close to the load.

The output ripple voltage on standby output (VSB) is influenced by the main output V_1 . Evaluating standby output (VSB) output ripple must be done when maximum load is applied to V_1 .



6. PROTECTION SPECIFICATIONS

PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
F	Input Fuse (L)	Not user accessible, Time-lag(F)		16		Α
V ₁ ov	OV Threshold 1/1		13.3		14.5	VDC
V SB OV	OV Threshold VsB		110		120	% VsB
l∕ 1 lim	Over Current Limitation 1/4	$V_i \ge 180 \text{ VAC/HVDC}, T_a < 50^{\circ}\text{C}$ $V_i > 90 \text{ VAC}, T_a < 50^{\circ}\text{C}$	128 86		160 107	Α
IvsB lim	Over Current Limitation V _{SB}	T_a < 50°C for 5 V _{SB} T_a < 50°C for 3.3 V _{SB}	3.45 5.25		4.5 6.2	A A
T _{SD}	Over Temperature On Heat Sinks	Automatic shut-down		115	120	°C

Table 4. Protection Specifications

6.1 OVERVOLTAGE PROTECTION

The TET front-ends provide a fixed threshold overvoltage (OV) protection for both the main and the standby output. Once an 12V main output OV condition has been triggered, the main output will shut down and latch the fault condition. Standby output (VSB) will turn off the whole power supply and latch if standby output (VSB) over-voltage protection has been triggered. The latch for main output and standby output (VSB) can be unlocked by disconnecting the supply from the AC mains or by toggling the PSON_L input.

6.2 VSB UNDERVOLTAGE DETECTION

Both main and standby outputs (VSB) are monitored.

3.3 / 5 V_{SB}

LED and PWOK_H pin signal if the output voltage exceeds $\pm 5\%$ of its nominal voltage. Output under voltage protection is provided on the standby output (VsB) and main output V_1 . When V_{SB} falls below 75% of its nominal voltage, the main output V_1 is inhibited.

6.3 CURRENT LIMITATION

6.3.1 MAIN OUTPUT

When main output runs in current limitation mode its output will turn OFF below 2 V but will retry to recover every 1 s interval. If current limitation mode is still present after the unit retry, output will continuously perform this routine until current is below the current limitation point. The supply will go through soft start every time it retry from current limitation mode.

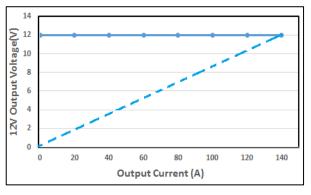


Figure 6.1 Current Limitation on V_1 ($V_i = 230 \text{ VAC}$)



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The output power derating of V1 refers to Figure 6.2 The main output will decrease if the ambient (inlet) temperature increases beyond 50°C.

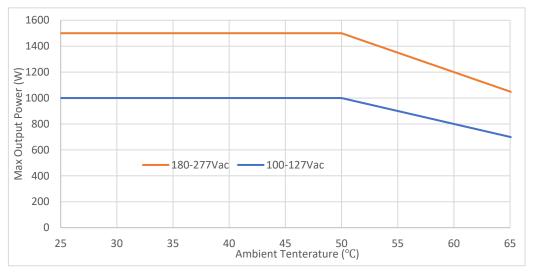


Figure 6.2 Power derating curve with Vin and Ambient Temperature

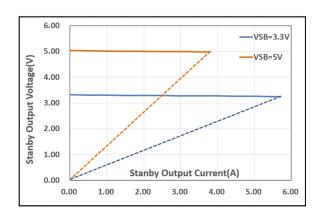
Note:

- 1. The application of power supply should also refer to installation instructions document.
- 2. The power supply has no limitation on its output current/power in the respect of meeting the operating conditions shown by the derating limits shown above. It is the responsibility of the end user to ensure operating conditions are maintained within their safety agency certification limits to assure safe and reliable operation.

6.3.2 STANDBY OUTPUT

3.3 / 5 V_{SB}

A fault of over current protection on standby output (VSB) will cause the V1 output to turn off and latch. The latch can be unlocked by disconnecting the supply from the AC mains or by toggling the PSON_L input. The current limitation of the standby output (VSB) is independent of the AC input voltage.



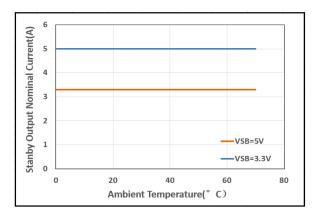


Figure 7. Current Limitation and Temperature Derating on 3.3 / 5 V_{SB}



7. MONITORING

PARAMETER	DESCRIPTION / CONDITION			MIN	NOM	MAX	UNIT
V₁ mon	Input RMS Voltage	$V_{i \min} \leq V_i \leq V_{i \min}$	ax	-2		+2	%
/ mon	Input RMS Current	<i>I</i> ⊳ 1.5A _{RMS}		-5		+5	%
P _{i mon}	True Input Power	<i>I</i> ⊳ 1.5A _{RMS}		-5		+5	%
V _{1 mon}	V ₁ Voltage	I ₁ > 20% I _{1 nom}		-2		+2	%
⅓ mon	V ₁ Current	I ₁ > 20% I _{1 nom}		-5		+5	%
P _{o nom}	Total Output Power	I ₁ > 20% I _{1 nom}		-5		+5	%
VSB mon	Standby Voltage		$3.3/5V_{SB}Models$	-0.2		+0.2	V
ÆB mon	Standby Current	$I_{SB} \le I_{SB \text{ nom}}$	3.3 / 5 V _{SB} Models	-0.5		+0.5	Α

Table 5. Monitoring parameters

8. SIGNAL & CONTROL SPECIFICATIONS

8.1 ELECTRICAL CHARACTERISTICS

PARAMETER	DESCRIPTION / CONDITION		MIN	NOM	MAX	UNIT
PSKILL_H / PSON	_L / VSB_SEL / HOTSTANDBYEN_H Input	ts				
V _{IL}	Input Low Level Voltage		-0.2		8.0	V
Ин	Input High Level Voltage		2.4		3.5	V
/ L, H	Maximum Input Sink or Source Current		0		1	mA
$R_{ m puPSKILL_H}$	Internal Pull Up Resistor on PSKILL_H			10		kΩ
RpuPSON_L	Internal Pull Up Resistor on PSON_L			10		kΩ
$R_{ m puVSB_SEL}$	Internal Pull Up Resistor on VSB_SEL			10		kΩ
RpuHOTSTANDBYEN_H	Internal Pull Up Resistor on HOTSTANDB	YEN_H		10		kΩ
<i>R</i> Low	Resistance Pin to SGND for Low Level		0		1	kΩ
<i>R</i> HIGH	Resistance Pin to SGND for High Level		50			kΩ
PWOK_H Output						
V _{OL}	Output Low Level Voltage	$I_{\text{sink}} < 4 \text{ mA}$	0		0.4	V
V 6н	Output High Level Voltage	<i>I</i> _{source} < 50 uA	2.6		3.5	V
$R_{ m puPWOK_H}$	Internal Pull Up Resistor on PWOK_H			10		kΩ
ACOK_H Output						
V ∕0L	Output Low Level Voltage	$J_{sink} < 2 \text{ mA}$	0		0.4	V
И он	Output High Level Voltage	$I_{\rm source} < 50~\mu A$	2.6		3.5	V
$R_{ m puACOK_H}$	Internal Pull Up Resistor on ACOK_H			10		kΩ
SMB_ALERT_L Ou	ıtput					
V _{ext}	Maximum External Pull Up Voltage				12	V
ИоL	Output Low Level Voltage	/source < 4 mA	0		0.4	V
Юн	Maximum High Level Leakage Current				10	μΑ
$R_{ m puSMB_ALERT_L}$	Internal Pull Up Resistor on SMB_ALERT_L			None		kΩ

Table 6. Signal & Control Specifications



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8.2 INTERFACING WITH SIGNALS

All signal pins have protection diodes implemented to protect internal circuits. When the power supply is not powered, the protection devices start clamping at signal pin voltages exceeding ±0.5 V. Therefore, all input signals should be driven only by an open collector/drain to prevent back feeding inputs when the power supply is switched off. If interconnecting of signal pins of several power supplies is required, then this should be done by decoupling with small signal schottky diodes as shown in examples in (*Figure 8*) except for SMB_ALERT_L, ISHARE and I²C pins. SMB_ALERT_L pins can be interconnected without decoupling diodes, since these pins have no internal pull up resistor and use a 15 V zener diode as protection device against positive voltage on pins. ISHARE pins must be interconnected without any additional components. This in-/output is disconnected from internal circuits when the power supply is switched off.

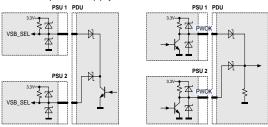


Figure 8. Interconnection of Signal Pins

8.3 FRONT LEDS

There will be 2 separate LED indicators, one blue and one amber to indicate the power supply status. There will be a (slow) blinking blue POWER LED (OK) to indicate that AC is applied to the PSU and the Standby Voltage is available. This same LED shall go steady to indicate that all the Power Outputs are available. This same LED or separate one will blink (slow) or be solid ON amber to indicate that the power supply has failed or reached a warning status and therefore a replacement of the unit is/maybe necessary. The LED are visible on the power supply's exterior face. The LED location meets ESD Requirements.

EVENT	BLUE LED STATUS	AMBER LED STATUS
12V main on and in voltage regulation band (Active mode)	Solid	OFF
12V main off (Standby mode)	1Hz Blinking	OFF
No AC input power to any of the system power supplies	OFF	OFF
No AC input power, but other PSU in the system operating	OFF	1HZ Blinking
Warning event (Output OCW/ OTW/ Fan Fail)	OFF	1HZ Blinking
Fault event (Input OVP/ Output OVP, UVP, OCP/ OTP/ Other internal fault)	OFF	Solid
FW update	OFF	2Hz Blinking

Table 7. LED Definitions

8.4 PRESENT_L

This signaling pin is recessed within the connector and will contact only once all other connector contacts are closed. This active-low pin is used to indicate to a power distribution unit controller that a supply is plugged in. The maximum current on PRESENT_L pin should not exceed 10 mA.



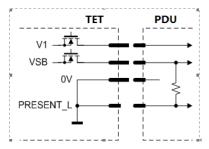


Figure 9. PRESENT_L signal pin

8.5 PSKILL_H INPUT

The PSKILL_H input is active-high and is located on a recessed pin on the connector and is used to disconnect the main output as soon as the power supply is being plugged out. This pin should be connected to SGND in the power distribution unit. The standby output will remain on regardless of the PSKILL_H input state.

8.6 TIMING REQUIREMENTS

The power supply will automatically turn-on when connected to the AC line under the condition that the PSON_L signal is pulled low and the AC line is within range. The ACOK_H signal is active-high. The timing diagram is shown in *Figure 10*. and referenced in *Table 8*.

OPERATING CONDITIO	N	MIN	MAX	UNIT
TVout_rise	Output voltage rise time for Vout		10	ms
TVsb_rise	Output voltage rise time for Vsb	10	100	ms
TVsb_on_delay	Delay from Vin being applied to Vsb being within regulation		2000	ms
TVout_on_delay	Delay from Vin being applied to Vout being within regulation		3000	ms
Tacok_on	Input voltage to ACOK_H signal delay time		2000	ms
TVout_holdup	Time of Vout output voltage stay within regulation after loss of Vin	10		ms
Tpwok_holdup	Delay from loss of Vin to de-assertion of PWOK	9		ms
Tpson_off_delay	Delay from PSON de-asserted to power supply turning off		5	ms
Tpson_on_delay	Delay from PSON active to output voltages within regulation limits	5	200	ms
Tpson_pwok	Delay from PSON de-active to PWOK being de-asserted		4	ms
Tpwok_on	Delay from output voltages within regulation limits to PWOK asserted	100	500	ms
Tpwok_off	Delay from PWOK de-asserted to output voltages dropping out of regulation limits	1		ms
Tpwok_low	Duration of PWOK being in the de-asserted state during an off/on cycle using Vin or the PSON signal	100		ms
TVsb_Vout	Delay from Vsb being in regulation to Vout being in regulation at Vin turn on	50	500	ms
TVsb_holdup	Time of Vsb output voltage stays within regulation after loss of Vin	100		ms
Tacok_de-asserted	Delay from Vin drop being 0V to de-aseertion of ACOK		5	ms
Tacok_de-asserted_Vout	ACOK de-asserted to Vout out of regulation	7		ms
Tacok_ de-asserted_Vsb	ACOK de-asserted to Vsb out of regulation	15		ms
Tpskill_off	PSKILL high to Vo1 turn OFF by PSKILL		0.1	ms

Table 8. Timing Requirements



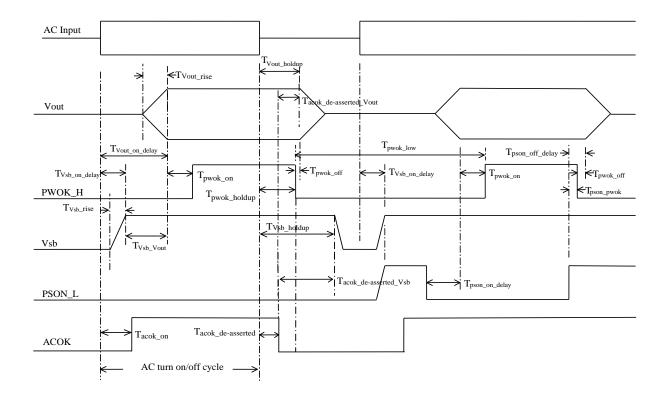


Figure 11. Turn On/Off Timing

8.7 PSON L INPUT

The PSON_L is an internally pulled-up (3.3 V) input signal to enable/disable the main output V1 of the front-end. This active-low pin is also used to clear any latched fault condition. The timing diagram is given in *Figure 10* and the parameters in the *Table 8*.

8.8 PWOK_H SIGNAL

The PWOK_H is an open drain output with an internal pull-up to 3.3 V indicating whether both V_{SB} and V_1 outputs are within regulation. This pin is active-low. The timing diagram is shown in *Figure 10* and referenced in the *Table 8*.

8.9 CURRENT SHARE

The TET front-ends have an active current share scheme implemented for V_1 . All the ISHARE current share pins need to be interconnected in order to activate the sharing function. If a supply has an internal fault or is not turned on, it will disconnect its ISHARE pin from the share bus. This will prevent dragging the output down (or up) in such cases.

The current share function uses an analog bi-directional data exchange on a recessive bus configuration to transmit and receive current share information. The controller implements a Master/Slave current share function. The power supply providing the largest current among the group is automatically the Master. The other supplies will operate as Slaves and increase their output current to a value close to the Master by slightly increasing their output voltage. The voltage increase is limited to +250 mV.

The standby output uses a passive current share method (droop output voltage characteristic).



8.10 SENSE INPUTS

Both main and standby outputs have sense lines implemented to compensate for voltage drop on load wires. The maximum allowed voltage drop is 200 mV on the positive rail and 100 mV on the power GND rail.

With open sense inputs the main output voltage will rise by 230 mV and the standby output by 50 mV. Therefore, if not used, these inputs should be connected to the power output and power GND close to the power supply connector. The sense inputs are protected against short circuit. In this case the power supply will shut down.

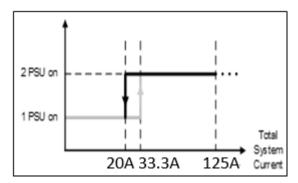
8.11 HOT-STANDBY OPERATION

The hot-standby operation is an operating mode allowing to further increase efficiency at light load conditions in a redundant power supply system. Under specific conditions one of the power supplies is allowed to disable its OR-ing device on the output. This will save the power losses associated with this power supply and at the same time the other power supply will operate in a load range having a better efficiency. In order to enable the hot standby operation, the HOTSTANDBYEN_H and the ISHARE pins need to be interconnected. A power supply will only be allowed to enter the hot-standby mode, when the HOTSTANDBYEN_H pin is high, the load current is low (see *Figure*) and the supply was allowed to enter the hot-standby mode by the system controller via the appropriate I²C command (by default disabled). The system controller needs to ensure that only one of the power supplies is allowed to enter the hot-standby mode.

If a power supply is in a fault condition, it will pull low its active-high HOTSTANDBYEN_H pin which indicates to the other power supply that it is not allowed to enter the hot-standby mode or that it needs to return to normal operation should it already have been in the hot-standby mode.

NOTE: The system controller needs to ensure that only one of the power supplies is allowed to enter the hot-standby model.

Figure shows the achievable power loss savings when using the hot-standby mode operation. A total power loss reduction of 6 W is achievable.



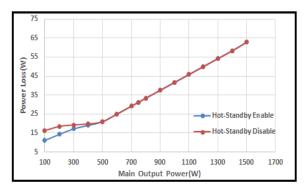


Figure 11. Hot standby enable/disable current thresholds

Figure 12. PSU power losses with/without hot-standby mode

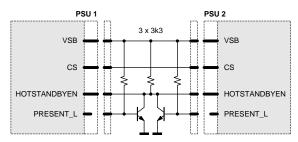


Figure 13. Recommended hot-standby configuration

In order to prevent voltage dips when the active power supply is unplugged while the other is in hot-standby mode, it is strongly recommended to add the external circuit as shown in *Figure13*. If the PRESENT_L pin status needs also to be read by the system controller, it is recommended to exchange the bipolar transistors with small signal MOS transistors or with digital transistors.



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8.12 I2C / SMBUS COMMUNICATION

The interface driver in the TET supply is referenced to the V1 Return. The TET supply is a communication Slave device only; it never initiates messages on the I2C/SMBus by itself. The communication bus voltage and timing is defined in Error! Reference s ource not found. further characterized through:

- There are no internal pull-up resistors
- The SDA/SCL IOs are 3.3/5 V tolerant
- Full SMBus clock speed of 100 kbps
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognizes any time Start/Stop bus conditions

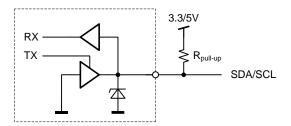


Figure 14 Physical layer of communication interface

The SMB_ALERT_L signal indicates that the power supply is experiencing a problem that the system agent should investigate. This is a logical OR of the Shutdown and Warning events.

Communication to the DSP or the EEPROM will be possible as long as the input AC voltage is provided. If no AC is present, communication to the unit is possible as long as it is connected to a life V1 output (provided e.g. by the redundant unit).

PARAMETER	DESCRIPTION / CONDITION		MIN	IO M	UNI T
V _{iL}	Input low voltage		-0.5	1.0	V
ViH	Input high voltage		2.3	5.5	٧
V_{hys}	Input hysteresis		0.15		V
V_{oL}	Output low voltage	3 mA sink current	0	0.4	V
<i>t</i> _r	Rise time for SDA and SCL(ViLmax-0.15V to ViHmin+0.15V)	0.65V to 2.25V f _{SCL} ≤ 100 kHz	20+0.1Cb ²	1000	ns
t_{of}	Output fall time (ViHmin+0.15V to ViLmax-0.15V)	2.25V to 0.65V f _{SCL} ≤ 100 kHz	20+0.1Cb ³	300	ns
li	Input current SCL/SDA	0.1 VDD < Vi < 0.9 VDD	-10	10	μΑ
C i	Internal Capacitance for each SCL/SDA			50	pF
f _{SCL}	SCL clock frequency		0	100	kHz
R _{pu}	External pull-up resistor	f _{SCL} ≤ 100 kHz		1000 ns/ Cb	Ω
<i>thdsta</i>	Hold time (repeated) START	f _{SCL} ≤ 100 kHz	4.0		μs
tLOW	Low period of the SCL clock	f _{SCL} ≤ 100 kHz	4.7		μs
t _{HIGH}	High period of the SCL clock	$f_{SCL} \le 100 \text{ kHz}$	4.0		μS
t _{SUSTA}	Setup time for a repeated START	f _{SCL} ≤ 100 kHz	4.7		μs
t _{HDDAT}	Data hold time	f _{SCL} ≤ 100 kHz	0	3.45	μS
<i>tsudat</i>	Data setup time	f _{SCL} ≤ 100 kHz	250		ns
tsusто	Setup time for STOP condition	f _{SCL} ≤ 100 kHz	4.0		μS
<i>t_{BUF}</i>	Bus free time between STOP and START	f _{SCL} ≤ 100 kHz	5		ms

Table 9. I2C / SMBus Specification

Cb = Capacitance of bus line in pF, typically in the range of 10...400 pF



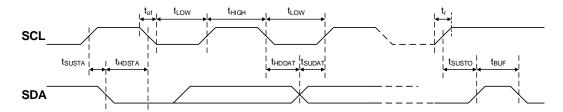


Figure 15. I2C / SMBus Timing

8.13 ADDRESS / PROTOCOL SELECTION (APS)

The APS pin provides the possibility to select the address by connecting a resistor to V1 return (0 V). A fixed addressing offset exists between the Controller and the EEPROM.

NOTE:

- If the APS pin is left open, the supply will operate with the Power Management Bus protocol at controller / EEPROM addresses 0xB6 / 0xA6.
- The APS pin is only read at start-up of the power supply. Therefore, it is not possible to change address dynamically.

Table 10. I2C / SMBus Specifications

P (0)3	Dyatasal	I2C Address 4		
R _{APS} (Ω) ³	Protocol	Controller	EEPROM	
820		0xB0	0xA0	
2700	Power	0xB2	0xA2	
5600	Management Bus	0xB4	0xA4	
8200		0xB6	0xA6	

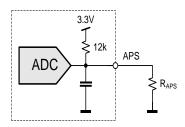


Figure 16. I2C address and protocol setting

8.14 CONTROLER AND EEPROM ACCESS

The controller and the EEPROM in the power supply share the same I2C bus physical layer (see *Figure17*). An I2C driver device assures logic level shifting (3.3/5 V) and a glitch-free clock stretching. The driver also pulls the SDA/SCL line to nearly 0 V when driven low by the DSP or the EEPROM providing maximum flexibility when additional external bus repeaters are needed. Such repeaters usually encode the low state with different voltage levels depending on the transmission direction.

The DSP will automatically set the I2C address of the EEPROM with the necessary offset when its own address is changed / set. In order to write to the EEPROM, first the write protection needs to be disabled by sending the appropriate command to the DSP. By default, the write protection is on.

The EEPROM provides 256 bytes of user memory. None of the bytes are used for the operation of the power supply.

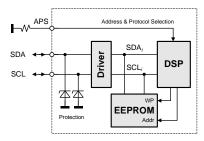


Figure 17. I2C Bus to DPS and EEPROM

⁵ The LSB of the address byte is the R/W bit



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E12 resistor values, use max 5% resistors

8.15 EEPROM PROTOCOL

The EEPROM follows the industry communication protocols used for this type of device. Even though page write / read commands are defined, it is recommended to use the single byte write / read commands.

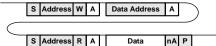
WRITE

The write command follows the SMBus 1.1 Write Byte protocol. After the device address with the write bit cleared a first byte with the data address to write to is sent followed by the data byte and the STOP condition. A new START condition on the bus should only occur after 5ms of the last STOP condition to allow the EEPROM to write the data into its memory.



READ

The read command follows the SMBus 1.1 Read Byte protocol. After the device address with the write bit cleared the data address byte is sent followed by a repeated start, the device address and the read bit set. The EEPROM will respond with the data byte at the specified location.



8.16 POWER MANAGEMENT BUS PROTOCOL

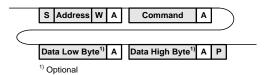
The Power Management Bus (Power Management Bus) is an open standard protocol that defines means of communicating with power conversion and other devices. For more information, please see the System Management Interface Forum web site at www.powerSIG.org.

Power Management Bus command codes are not register addresses. They describe a specific command to be executed. The TET1500 supply supports the following basic command structures:

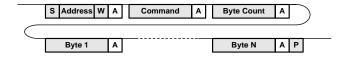
- Clock stretching limited to 1 ms
- SCL low time-out of >25 ms with recovery within 10 ms
- Recognized any time Start/Stop bus conditions

WRITE

The write protocol is the SMBus 1.1 Write Byte/Word protocol. Note that the write protocol may end after the command byte or after the first data byte (Byte command) or then after sending 2 data bytes (Word command).

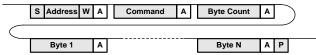


In addition, Block write commands are supported with a total maximum length of 255 bytes.



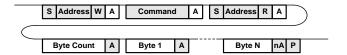
RFAD

The read protocol is the SMBus 1.1 Read Byte/Word protocol. Note that the read protocol may request a single byte or word.



In addition, Block read commands are supported with a total maximum length of 255 bytes.





8.17 POWER SUPPLY DIAGNOSTIC BLACK BOX

The power supply shall save the latest data and other pertinent data into nonvolatile memory when a critical event shuts down the power supply. This data shall be accessible via the Power Management Bus interface with an external source providing power to the

12V main output bus or standby by 5VSB output bus.

Critical Events to trigger an update to the Event Recorder includes:

- Output OVP
- Output OCP
- Input OV/UV Fault
- Fan fault
- OTP
- Other faults to cause output shutdown.

Refer to BCA.00XXX_TET1500-12-054NA Power Management Bus Communication Application Note for further information about the Power Management Bus commands to support this function.

8.18 FIRMWARE UPDATE

The power supply shall have the capability to update its firmware via the Power Management Bus interface while it is in standby mode. This

FW can be updated when in the system and in standby mode and outside the system with power applied to the 12V main output bus or standby by 5VSB output bus.

BPS standard GUI supports the firmware upgrade function.

8.19 GRAPHICAL USER INTERFACE

Bel Power Solutions provides with its "Bel Power Solutions I2C Utility" a Windows® XP/Vista/Win7 compatible graphical user interface allowing the programming and monitoring of the TET1500-12-054NA Front-End.

 $\label{thm:compower-solutions} \mbox{ and supports Power Management Bus protocols.}$

The GUI allows automatic discovery of the units connected to the communication bus and will show them in the navigation tree. In the monitoring view the power supply can be controlled and monitored.

If the GUI is used in conjunction with the SNP-OP-BOARD-01 or YTM.G1Q01.0 Evaluation Kit it is also possible to control the PSON_L pin(s) of the power supply.

Further there is a button to disable the internal fan for approximately 10 seconds. This allows the user to take input power measurements without fan consumptions to check efficiency compliance to the Climate Saver Computing Platinum specification.

The monitoring screen also allows to enable the hot-standby mode on the power supply. The mode status is monitored and by changing the load current it can be monitored when the power supply is being disabled for further energy savings. This obviously requires 2 power supplies being operated as a redundant system (as in the evaluation kit).

NOTE: The user of the GUI needs to ensure that only one of the power supplies have the hot-standby mode enabled.



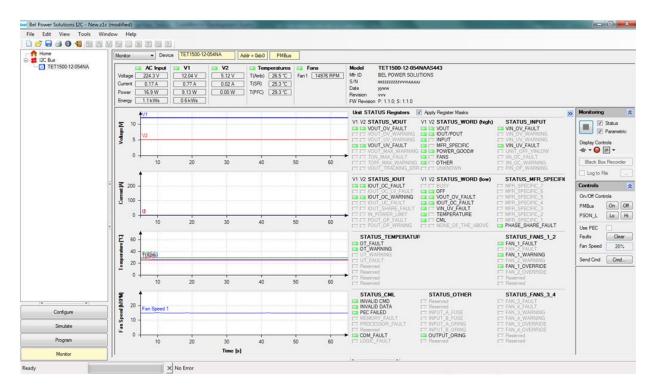


Figure 18. Monitoring dialog of the I2C Utility (for reference)

9. TEMPERATURE AND FAN CONTROL

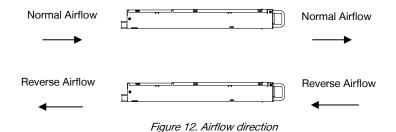
To achieve best cooling results sufficient airflow through the supply must be ensured. Do not block or obstruct the airflow at the rear of the supply by placing large objects directly at the output connector. The TET1500-12-054NA is provided with normal airflow, which means the air enters through the DC-output of the supply and leaves at the AC-inlet. The supplies have been designed for horizontal operation.

The fan inside of the supply is controlled by a microprocessor. The RPM of the fan is adjusted to ensure optimal supply cooling and is a function of output power and the inlet temperature.

For the normal airflow version additional constraints apply because of the AC-connector. In a normal airflow unit, the hot air is exiting the power supply unit at the AC-inlet.

The IEC connector on the unit is rated 120°C. If 70°C mating connector is used then end user must derated the input power to meet a maximum 70°C temperature at the front.

NOTE: It is the responsibility of the user to check the front temperature in such cases. The unit is not limiting its power automatically to meet such a temperature limitation.





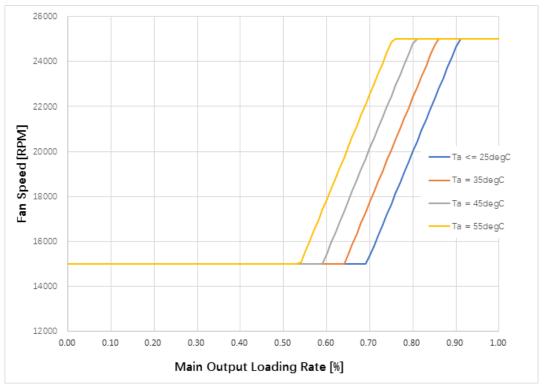


Figure 20. Fan speed VS main output load

10. ELECTROMAGNETIC COMPATIBILITY

10.1 **IMMUNITY**

NOTE: Most of the immunity requirements are derived from EN 55024:1998/A2:2003.

PARAMETER	DESCRIPTION / CONDITION	CRITERION
ESD Contact Discharge	IEC / EN 61000-4-2, ±8 kV, 25+25 discharges per test point (metallic case, LEDs, connector body)	А
ESD Air Discharge	IEC / EN 61000-4-2, ±15 kV, 25+25 discharges per test point (non-metallic user accessible surfaces)	А
Radiated Electromagnetic Field	IEC / EN 61000-4-3, 10 V/m, 1 kHz/80% Amplitude Modulation, 1 µs Pulse Modulation, 10 kHz2 GHz	А
Burst	IEC / EN 61000-4-4, level 3 AC port ±2 kV, 1 minute DC port ±1 kV, 1 minute	А
Surge	IEC / EN 61000-4-5 Line to earth: ±2 kV,2ohm Line to line: ±2 kV,2ohm	А
RF Conducted Immunity	IEC/EN 61000-4-6, Level 3, 10 Vrms, CW, 0.1 80 MHz	Α
Voltage Dips and Interruptions	IEC/EN 61000-4-11 1: Vi 230 V, 100% Load, Phase 0 °, Dip 100%, Duration 10 ms 2: Vi 230 V, 100% Load, Phase 0 °, Dip 100%, Duration 20 ms 3: Vi 230 V, 100% Load, Phase 0 °, Dip 100%, Duration >20 ms	A V _{SB} : A, V ₁ : B B

Table 11. Immunity



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10.2 EMISSION

PARAMETER	DESCRIPTION / CONDITION	CRITERION
	EN55032 / CISPR 32: 0.15 30 MHz, QP and AVG, single unit	Class A
Conducted Emission	EN55032 / CISPR 32: 0.15 30 MHz, QP and AVG, 2 units in rack system	Class A
	EN55032 / CISPR 32: 30 MHz 1 GHz, QP, single unit	Class A
Radiated Emission	EN55032 / CISPR 32: 30 MHz 1 GHz, QP, 2 units in rack system	Class A
Harmonic Emissions	IEC61000-3-2, Vin = 115 VAC / 60 Hz, & Vin = 230VAC/ 50 Hz, from 10% load to full load	Class A
AC Flicker	IEC61000-3-3, Vin = 230 VAC / 60 Hz, 100% Load	Pass

Table 12. Emission

11. SAFETY APPROVALS

Maximum electric strength testing is performed in the factory according to IEC/EN 62368-1, and UL 62368-1. Input-to-output electric strength tests should not be repeated in the field. Bel Power Solutions will not honor any warranty claims resulting from electric strength field tests.

PARA	METER	DESCRIPTION / CONDITION	MIN	NOM I	MAX	UNIT
	Agency Approvals	UL/CSA 62368-1 (USA / Canada) EN 62368-1 (Europe) IEC 62368-1 (International) CB Certificate & Report, 62368-1 (report to include all country national deviations) CE - Low Voltage Directive 2014/35/EC (Europe) GB4943.1- CNCA Certification (China) BSMI Certification - CNS15598-1 (Taiwan)				
	Isolation Strength	Input (L/N) to case (PE) Input (L/N) to output Output to case (PE)		Basic Reinforced Functional		
$d_{\mathbb{C}}$	Creepage / Clearance	Primary (L/N) to protective earth (PE) Primary to secondary		According to safety standard		mm
	Electrical Strength Test	Input to case Input to output		≤ 250 VAC) VAC < Vin ≤ 300	VAC)	VDC VDC VDC

Table 13. Safety Approvals

12. ENVIRONMENTAL SPECIFICATIONS

PARAME	TER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
T _A	Ambient Temperature	V_{lmin} to V_{lmax} , I_{lnom} , I_{SBnom} below 3000m Altitude	-5		+50	°C
<i>T</i> _{Aext}	Extended Temp. Range	Derating output power	+51		+61	°C
$\mathcal{T}_{\mathcal{S}}$	Storage Temperature	Non-operational	-40		+85	°C
	Altitude	Operational, above Sea Level, refer derating to Ta	3000			meters
N a	Audible Noise	V_{nom} , 50% I_{nom} , $T_{\text{A}} = 30^{\circ}\text{C}$		55		dBA
Llumiditu	Operating	Operating humidity noncondensing	5		90	%
Humidity	Storage	Storage humidity noncondensing	5		95	%

Table 14. Environmental Specifications



12.1 COOLING

The power supply shall include fans self-cooling. The fans shall use an acceptable anti-vibration mounting technique.

12.2 OPERATIONAL SHOCK / VIBRATION

Random Vibration - Operating

IPC 9592 Requirements for Power Conversion Devices for the Computer and Telecommunications Industries – Section 5.2.9

Random Vibration – Non-operating

IPC 9592 Requirements for Power Conversion Devices for the Computer and Telecommunications Industries - Section 5.2.10

Shock - Operating

IPC 9592 Requirements for Power Conversion Devices for the Computer and Telecommunications Industries – Section 5.2.11.

13. RELIABILITY

PARAMETER	NOTES	MIN	NOM	MAX	UNIT
Demonstrated MTBF @ 25 °C	1	300			K hours
Demonstrated MTBF @ 41 °C	1	140			K hours
Predicted MTBF (Telcordia)	2	300			K hours
Fan L10 Life@ 40 °C		70			K hours
Electrolytic Capacitor Calculated Life	3	10			Years
Field MTBF		1			M hours

Table 15. Reliability

- 1. At 25C and 41C ambient and 90% Confidence Level
- 2. At 40C and 50% part count (method 1 case 1)
- 3. To calculate electrolytic capacitor life, use capacitor supplier's equation with 40°C ambient and 80% load.
- 4. Component De-rating follow IPC9592B.

14. MECHANICAL SPECIFICATIONS

PAF	RAMETER	DESCRIPTION / CONDITION	MIN	NOM	MAX	UNIT
		Width		54.5		
	Dimensions	Height		40.0		mm
		Depth		321.5		
M	Weight			1.13		kg

Table 16. Mechanical Specifications

NOTE: A 3D step file of the power supply casing is available on request.



TET1500-12-054NA/RA: C14 type Input AC connector RongFeng SS-120-1.0B-2.8BV or equivalent

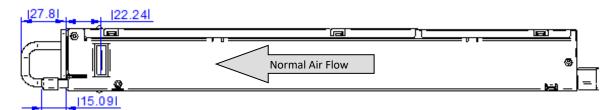


Figure 21. Side View 1

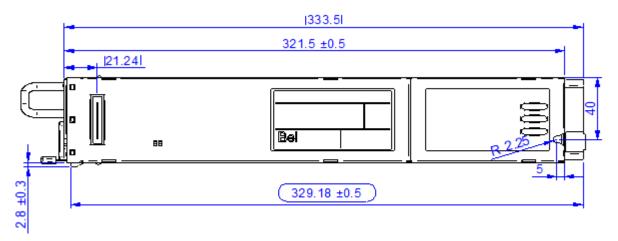


Figure 22. Top View

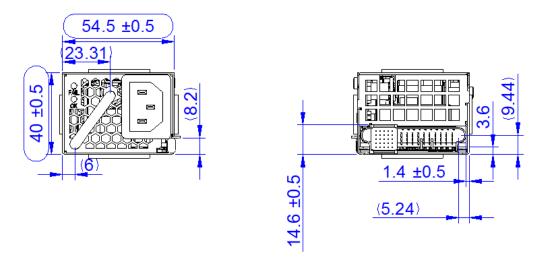


Figure 23. Front and Rear View



TET1500-12-054NAC/RAC: C16 Type Input AC connector, RongFeng SS-120B-1.0-4.0Ad or equivalent

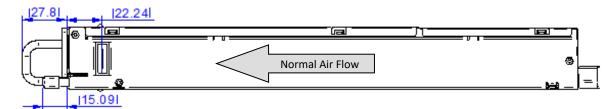


Figure 24. Side View 1

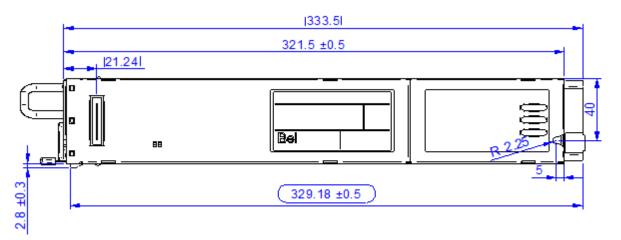


Figure 25. Top View

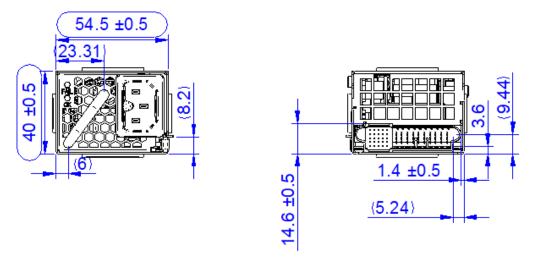


Figure 26. Front and Rear View



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TET1500-12-054NAH/RAH: ANDERSON POWER PRODUCTS 2006G1-BK or equivalent

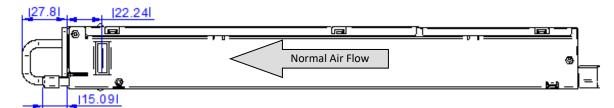


Figure 27. Side View 1

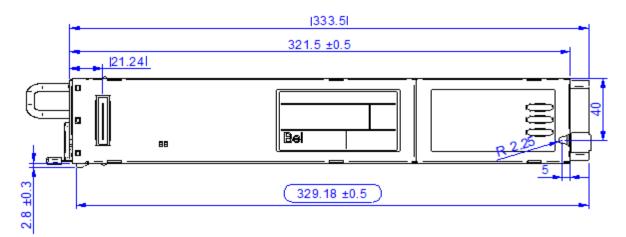


Figure 28. Top View

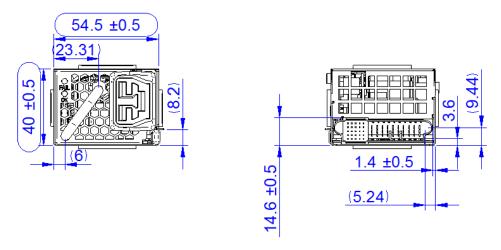


Figure 29. Front and Rear View



15. CONNECTIONS

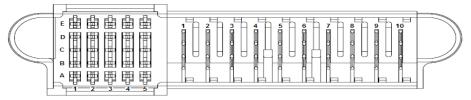
AC INPUT CONNECTOR:

TET1500-12-054NAH/RAH: Power supplier connector: ANDERSON POWER PRODUCTS 2006G1-BK

Mating connector: Anderson Saf-D-Grid Power cord 2034KZ2 or equivalent, http://www.andersonpower.com/

TET1500-12-054NA/RA: Power supplier connector: IEC320 C14 type TET1500-12-054NAC/RAC: Power supplier connector: IEC320 C16 type

DC OUTPUT CONNECTOR:



Power Supply Connector: Tyco Electronics P/N 1926736-3 or FCI connector 10133129-002LF or equivalent

(NOTE: Column 5 is recessed (short pins))

Mating Connector: Tyco Electronics P/N 2-1926739-5 or FCI 10108888-R10253SLF

PIN	NAME	DESCRIPTION
Output		
6, 7, 8, 9, 10	V1	+12 VDC main output
1, 2, 3, 4, 5	PGND	Power ground (return)
Control Pins		
A1	VSB	Standby positive output (+3.3/5 V _{SB})
B1	VSB	Standby positive output (+3.3/5 V _{SB})
C1	VSB	Standby positive output (+3.3/5 V _{SB})
D1	VSB	Standby positive output (+3.3/5 V _{SB})
E1	VSB	Standby positive output (+3.3/5 VsB)
A2	SGND	Signal ground (return)
B2	SGND	Signal ground (return)
C2	HOTSTANDBYEN_H	Hot standby enable signal: active-high
D2	VSB_SENSE_R	Standby output negative sense
E2	VSB_SENSE	Standby output positive sense
A3	APS	I ² C address and protocol selection (select by a pull down resistor)
B3	N/C	Reserved
C3	SDA	I ² C data signal line
D3	V1_SENSE_R	Main output negative sense
E3	V1_SENSE	Main output positive sense
A4	SCL	I ² C clock signal line
B4	PSON_L	Power supply on input (connect to A2/B2 to turn unit on): active-low
C4	SMB_ALERT_L	SMB Alert signal output: active-low
D4	N/C	Reserved
E4	ACOK_H	AC input OK signal: active-high
A5	PSKILL_H	Power supply kill (lagging pin): active-high
B5	ISHARE	Current share bus (lagging pin)
C5	PWOK_H	Power OK signal output (lagging pin): active-high
D5	VSB_SEL	Standby voltage selection (lagging pin)
E5	PRESENT_L	Power supply present (lagging pin): active-low

Table 17. Connector Pins Configurations



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16. ACCESSORIES

ITEM	DESCRIPTION	ORDERING PART NUMBER	SOURCE
	Bel Power Solutions I²C Utility Windows XP/Vista/7 compatible GUI to program, control and monitor TET Front-Ends (and other I ² C units)	N/A	belfuse.com/power-solutions
	Dual Connector Board Connector board to operate 2 TET units in parallel. Includes on-board USB to I ² C converter (use <i>Bel Power Solutions FC Utility</i> as desktop software).	SNP-OP-BOARD-01 or YTM.G1Q01.0	belfuse.com/power-solutions



17. REVISION HISTORY

DATE	REVISION	SECTION	ISSUE	PREPARED BY	APPROVED BY
2018/11/05	001	/	First release	Jemrry Zhang	BaoJun Zeng
2019/09/17	002	1.0	Description of adding AH in model name	Rick Luo	BaoJun Zeng
2019/09/17	002	5.0	Remove droop voltage spec for Vsb	Rick Luo	BaoJun Zeng
2019/09/17	002	7.0	Change lin & Pin monitoring condition to I;>2A from $I_1\!\!>\!\!20\%I_{1NOM}$	Rick Luo	BaoJun Zeng
2019/09/17	002	8.19	Add TBD on figure 18	Rick Luo	BaoJun Zeng
2020/03/24	003	8.19	Update figure 18	Rick Luo	BaoJun Zeng
2020/03/30	004	14&15	14: Add NA(C) and RA(C); 15: Add models and connectors	Ryan Li	Andrew Li
2020/09/22	005	1.0	Description of adding Ordering PN	Eisen Xu	BaoJun Zeng
2020/09/22	005	4.0	Change Ii max to Vin=100VAC,1000W 12Arms from Vin=100VAC,836W 10Arms; Change PF of 20% Load to 0.96 from 0.98	Eisen Xu	BaoJun Zeng
2020/09/22	005	4.6	Remove 10s AC Line Sag Transient Performance and Low Line Sag Transient Performance	Eisen Xu	BaoJun Zeng
2020/09/22	005	5.0	Change P1nom of 180 VAC > Vin \geq 90 VAC, V1 = 12 VDC to 1000W from 836W; Change I1nom of 180 VAC > Vin \geq 90 VAC, V1 = 12 VDC to 83.3A from 69.5A	Eisen Xu	BaoJun Zeng
2020/09/22	005	6.0	Change Iv1 lim of $V_1 > 90$ VAC, $T_a < 50$ °C to 86Amin,107Amax from 72Amin,90Amax	Eisen Xu	BaoJun Zeng
2020/09/22	005	6.3.1	Add Power derating curve	Eisen Xu	BaoJun Zeng
2021/07/21	Α	/	Upgrade to version A	Rick Luo	BaoJun Zeng
2021/10/26	Α	6.3.1	OCP change from latch to auto recovery	Joe Wu	BaoJun Zeng
2022/03/11	В	4.6	Change 500ms AC Line Sag Performance criteria from Class A to B	Sunny Zeng	BaoJun Zeng
2022/04/06	С	1.0	Change RF-203-D-1.0 to 2006G1-BK	Ryan Li	Andrew Li
2023/02/22	D	4.6	Change 500ms AC Line Sag Performance criteria from Class B to A	Sunny Zeng	Baojun Zeng
2023/08/08	E	15	Power Supply Connector: FCI connector 10133129- 002LF was added.		

For more information on these products consult: tech.support@psbel.com

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