







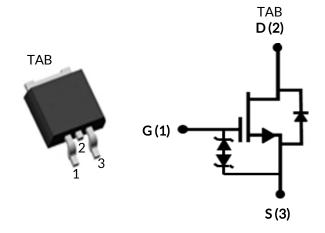








JF3C065040B3



Part Number	Package	Marking
UF3C065040B3	D ² PAK-3L	UF3C065040B3







650V-42mQ SiC FFT

Rev. C, May 2023

Description

This SiC FET device is based on a unique 'cascode' circuit configuration, in which a normally-on SiC JFET is co-packaged with a Si MOSFET to produce a normally-off SiC FET device. The device's standard gate-drive characteristics allows for a true "drop-in replacement" to Si IGBTs, Si FETs, SiC MOSFETs or Si superjunction devices. Available in the D²PAK-3L package, this device exhibits ultralow gate charge and exceptional reverse recovery characteristics, making it ideal for switching inductive loads when used with recommended RC-snubbers, and any application requiring standard gate drive.

Features

- \bullet Typical on-resistance $R_{DS(on),typ}$ of $42m\Omega$
- Maximum operating temperature of 175°C
- Excellent reverse recovery
- Low gate charge
- Low intrinsic capacitance
- ESD protected, HBM class 2
- Very low switching losses (required RC-snubber loss negligible under typical operating conditions)

Typical applications

- EV charging
- PV inverters
- Switch mode power supplies
- Power factor correction modules
- Motor drives
- Induction heating













Maximum Ratings

Parameter	Symbol	Test Conditions	Value	Units
Drain-source voltage	V_{DS}		650	V
Gate-source voltage	V _{GS}	DC	-25 to +25	V
Continuous drain current ¹	1	T _C = 25°C	41	Α
	I _D	T _C = 100°C	30	Α
Pulsed drain current ²	I _{DM}	T _C = 25°C	125	Α
Single pulsed avalanche energy ³	E _{AS}	L=15mH, I _{AS} =3.19A	76	mJ
Power dissipation	P _{tot}	T _C = 25°C	176	W
Maximum junction temperature	$T_{J,max}$		175	°C
Operating and storage temperature	T_J, T_{STG}		-55 to 175	°C
Reflow soldering temperature	T_{solder}	reflow MSL 1	245	°C

- 1. Limited by $T_{J,max}$
- 2. Pulse width t_p limited by $T_{J,max}$
- 3. Starting $T_J = 25^{\circ}C$

Thermal Characteristics

Darameter	Symbol	Test Conditions		Units		
Parameter			Min	Тур	Max	Units
Thermal resistance, junction-to-case	$R_{ heta$ JC			0.65	0.85	°C/W













Electrical Characteristics (T_J = +25°C unless otherwise specified)

Typical Performance - Static

Parameter	Symbol	Test Conditions		Units		
Parameter			Min	Тур	Max	Offics
Drain-source breakdown voltage	BV_{DS}	V_{GS} =0V, I_D =1mA	650			V
Tabal dada balan a samuah		V _{DS} =650V, V _{GS} =0V, T _J =25°C		0.7	150	- μΑ
Total drain leakage current	I _{DSS}	V _{DS} =650V, V _{GS} =0V, T _J =175°C		10		
Total gate leakage current	I _{GSS}	V _{DS} =0V, T _J =25°C, V _{GS} =-20V / +20V		6	±20	μА
Drain-source on-resistance	R _{DS(on)}	V_{GS} =12V, I_D =30A, T_J =25°C		42	52	
		V_{GS} =12V, I_{D} =30A, T_{J} =125°C		59		mΩ
		V_{GS} =12V, I_{D} =30A, T_{J} =175°C		78		
Gate threshold voltage	$V_{G(th)}$	V_{DS} =5V, I_{D} =10mA	4	5	6	V
Gate resistance	R_{G}	f=1MHz, open drain		4.5		Ω

Typical Performance - Reverse Diode

Parameter	Symbol	Test Conditions	Value			11.20
			Min	Тур	Max	Units
Diode continuous forward current ¹	I _S	T _C =25°C			41	Α
Diode pulse current ²	$I_{S,pulse}$	T _C =25°C			125	Α
Forward voltage	V _{FSD}	V _{GS} =0V, I _S =20A, T _J =25°C		1.5	1.75	V
		V _{GS} =0V, I _S =20A, T _J =175°C		1.8		V
Reverse recovery charge	Q_{rr}	I_J =175°C V_R =400V, I_S =30A, V_{GS} =-5V, R_{G_EXT} =22 Ω di/dt =1600A/ μ s, T_J =25°C		138		nC
Reverse recovery time	t _{rr}			26		ns
Reverse recovery charge	Q _{rr}	V_R =400V, I_S =30A, V_{GS} =-5V, R_{G_LEXT} =22 Ω		137		nC
Reverse recovery time	t _{rr}	di/dt=1600A/μs, T _J =150°C		26		ns













Typical Performance - Dynamic

Parameter	Symbol Test Conditions		Value			Unito
	-	Test Conditions -	Min	Тур	Max	- Units
Input capacitance	C_{iss}	V _{DS} =100V, V _{GS} =0V		1500		
Output capacitance	C_{oss}	$V_{DS}=100V, V_{GS}=0V$ = f=100kHz		200		pF
Reverse transfer capacitance	C_{rss}			2.2		
Effective output capacitance, energy related	$C_{oss(er)}$	V _{DS} =0V to 400V, V _{GS} =0V		146		pF
Effective output capacitance, time related	$C_{oss(tr)}$	V _{DS} =0V to 400V, V _{GS} =0V		325		pF
C _{OSS} stored energy	E _{oss}	V _{DS} =400V, V _{GS} =0V		11.7		μJ
Total gate charge	Q_{G}	\/ -400\/ -204		51		
Gate-drain charge	Q_{GD}	V_{DS} =400V, I_{D} =30A, V_{GS} = -5V to15V		11		nC
Gate-source charge	Q_GS	V _{GS} - 3V t013V		19		
Turn-on delay time	$t_{d(on)}$			34		ns
Rise time	t_r	V _{DS} =400V, I _D =30A, Gate		15		
Turn-off delay time	t _{d(off)}	Driver =-5V to +15V, Turn-on $R_{G,EXT}$ =1.8 Ω ,		57		
Fall time	t _f			12		
Turn-on energy including R _S energy ⁴	E _{ON}	Turn-off $R_{G,EXT}$ =22 Ω Inductive Load, FWD: same device with V_{GS} = -5V and R_{G} = 22 Ω , RC snubber: R_{S} =5 Ω and C_{S} =150pF, T_{J} =25°C		327		
Turn-off energy including R _S energy ⁴	E _{OFF}			65		
Total switching energy including R _s energy ⁴	E _{TOTAL}			392		μJ
Snubber R _S energy during turn-on	E _{RS_ON}			1.5		
Snubber R _S energy during turn-off	E _{RS_OFF}			3		
Turn-on delay time	t _{d(on)}			33		
Rise time	t _r	V _{DS} =400V, I _D =30A, Gate		15]
Turn-off delay time	t _{d(off)}	Driver =-5V to +15V, Turn-on $R_{G,EXT}$ =1.8 Ω , Turn-off $R_{G,EXT}$ =22 Ω Inductive Load, FWD: same device with V_{GS} = -5V and R_{G} = 22 Ω , RC snubber: R_{S} =5 Ω and C_{S} =150pF, T_{J} =150°C		58		ns
Fall time	t _f			13		
Turn-on energy including R _S energy ⁴	E _{ON}			314		
Turn-off energy including R _S energy ⁴	E _{OFF}			66		
Total switching energy including R _S energy ⁴	E _{TOTAL}			380		μЈ
Snubber R _S energy during turn-on	E _{RS_ON}			1.5		
Snubber R _S energy during turn-off	E _{RS_OFF}			2.9		

^{4.} The switching performance are evaluated with a RC snubber circuit as shown in Figure 24.





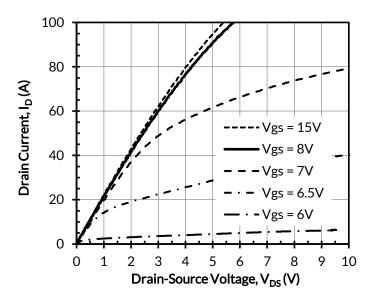








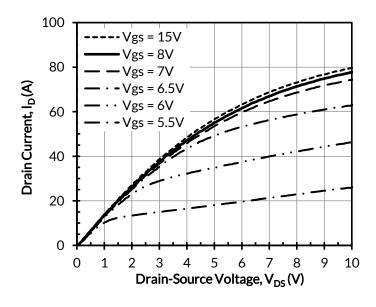
Typical Performance Diagrams



100 80 Drain Current, I_D (A) 60 - Vgs = 15V 40 Vgs = 8V Vgs = 7V Vgs = 6.5V 20 Vgs = 6V 0 0 1 2 3 5 10 Drain-Source Voltage, $V_{DS}(V)$

Figure 1. Typical output characteristics at T_J = - 55°C, tp < 250 μ s

Figure 2. Typical output characteristics at $T_J = 25$ °C, $tp < 250\mu s$



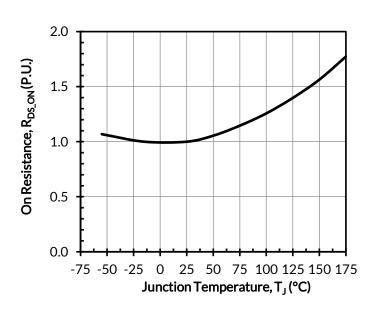


Figure 3. Typical output characteristics at T_J = 175°C, tp < 250 μ s

Figure 4. Normalized on-resistance vs. temperature at V_{GS} = 12V and I_D = 30A



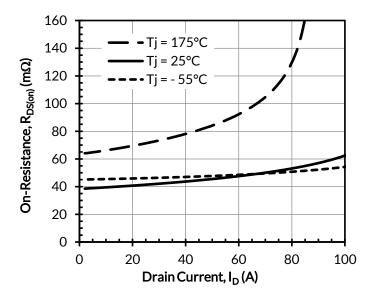












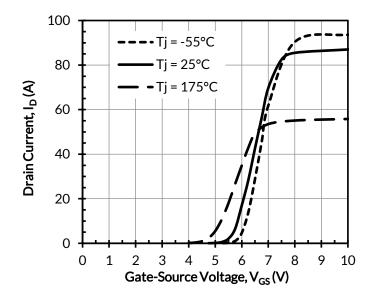
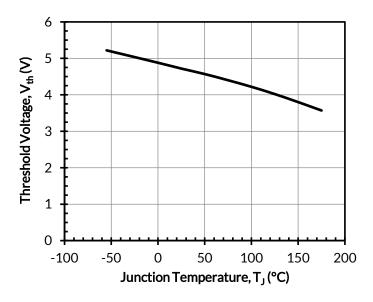


Figure 5. Typical drain-source on-resistances at V_{GS} = 12V

Figure 6. Typical transfer characteristics at V_{DS} = 5V



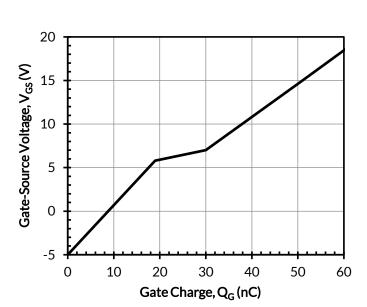


Figure 7. Threshold voltage vs. junction temperature at V_{DS} = 5V and I_{D} = 10mA

Figure 8. Typical gate charge at V_{DS} = 400V and I_{D} = 30A













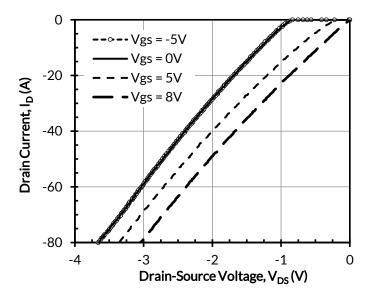


Figure 9. 3rd quadrant characteristics at $T_J = -55$ °C

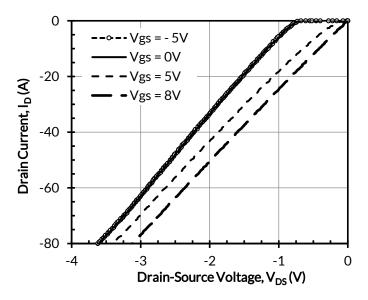


Figure 10. 3rd quadrant characteristics at $T_J = 25$ °C

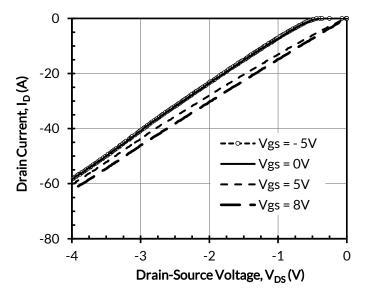


Figure 11. 3rd quadrant characteristics at $T_J = 175$ °C

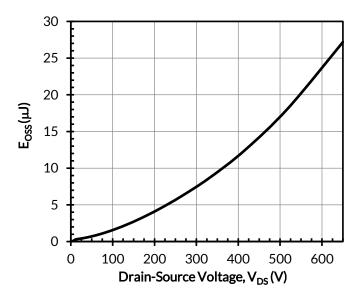


Figure 12. Typical stored energy in C_{OSS} at $V_{GS} = 0V$



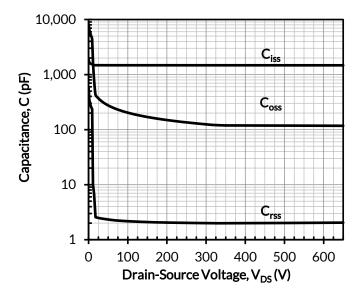












45 40 35 30 25 20 20 -75 -50 -25 0 25 50 75 100 125 150 175 Case Temperature, T_c (°C)

Figure 13. Typical capacitances at f = 100kHz and V_{GS} = 0V

Figure 14. DC drain current derating

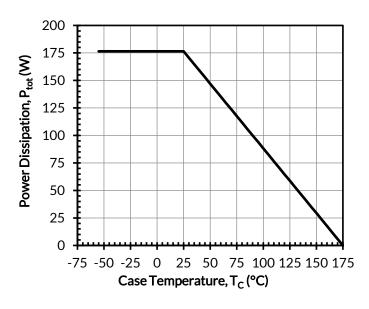


Figure 15. Total power dissipation

Figure 16. Maximum transient thermal impedance



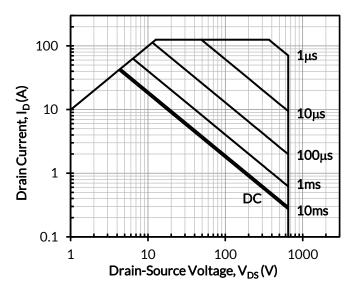








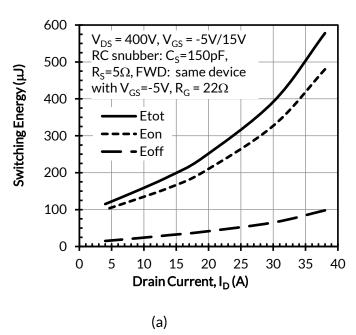




150 125 100 Qrr (nC) 75 $V_{DS} = 400V$, $I_{S} = 30A$, $di/dt = 1600A/\mu s$, 50 $V_{GS} = -5V, R_{G} = 22\Omega$ 25 0 25 0 50 75 100 125 150 175 Junction Temperature, T_J (°C)

Figure 17. Safe operation area at T_C = 25°C, D = 0, Parameter t_p

Figure 18. Reverse recovery charge Qrr vs. junction temperture



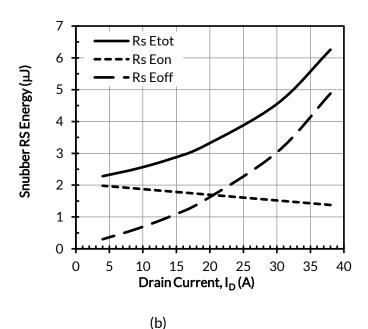


Figure 19. Clamped inductive switching energy (a) and RC snubber energy loss (b) vs. drain current at T_J = 25°C, turn-on R_{G_EXT} = 1.8 Ω , and turn-off R_{G_EXT} = 22 Ω



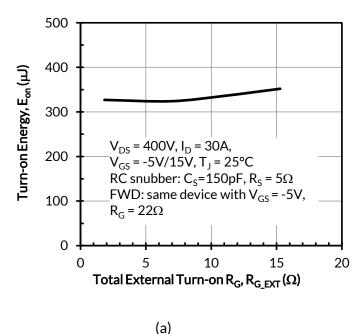












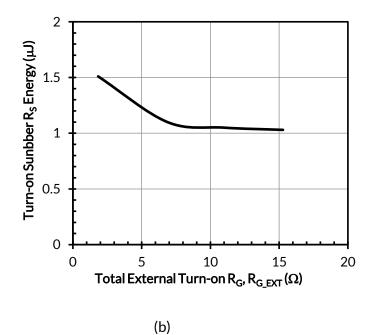
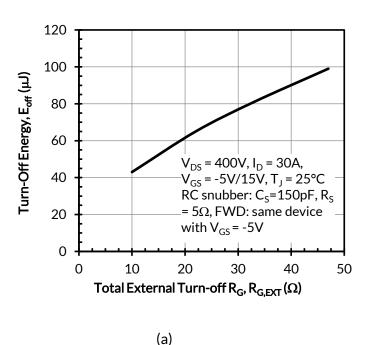


Figure 20. Clamped inductive switching turn-on energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-on gate resistor $R_{G,EXT}$



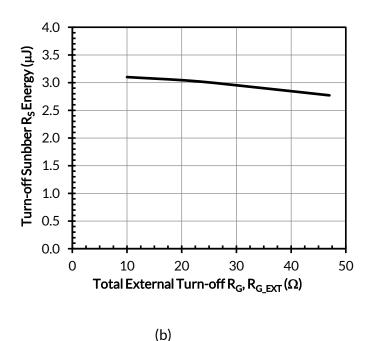


Figure 21. Clamped inductive switching turn-off energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of total external turn-off gate resistor $R_{G EXT}$



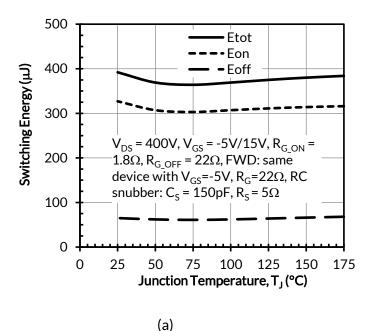












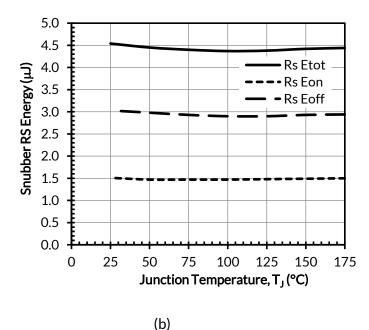
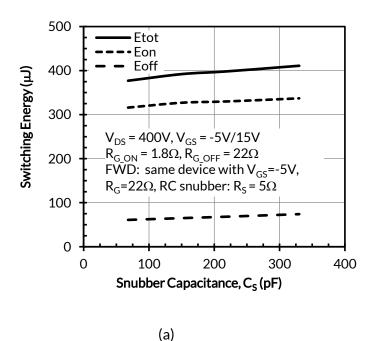


Figure 22. Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of junction temperature at $I_D = 30A$



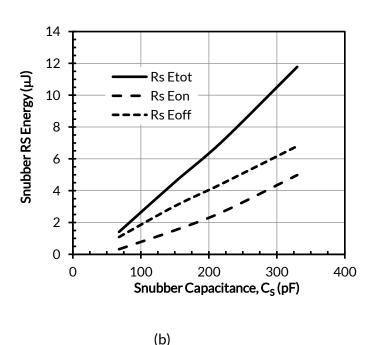


Figure 23. Clamped inductive switching energy including RC snubber energy loss (a) and RC snubber energy loss (b) as a function of snubber capacitance at I_D = 30A and T_J = 25°C













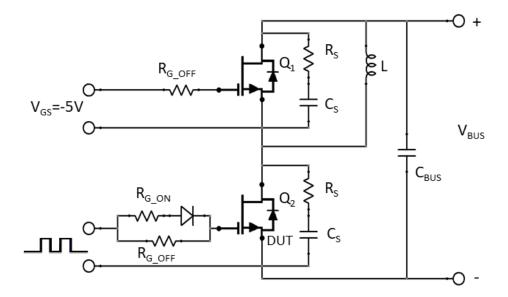


Figure 24. Clamped inductive load switching test circuit An RC snubber ($R_S = 5\Omega$ and $C_S = 150$ pF) is required to improve the turn-off waveforms.

Applications Information

SiC FETs are enhancement-mode power switches formed by a high-voltage SiC depletion-mode JFET and a low-voltage silicon MOSFET connected in series. The silicon MOSFET serves as the control unit while the SiC JFET provides high voltage blocking in the off state. This combination of devices in a single package provides compatibility with standard gate drivers and offers superior performance in terms of low on-resistance ($R_{DS(on)}$), output capacitance (C_{oss}), gate charge (C_{oss}), and reverse recovery charge (C_{oss}) leading to low conduction and switching losses. The SiC FETs also provide excellent reverse conduction capability eliminating the need for an external anti-parallel diode.

Like other high performance power switches, proper PCB layout design to minimize circuit parasitics is strongly recommended due to the high dv/dt and di/dt rates. An external gate resistor is recommended when the FET is working in the diode mode in order to achieve the optimum reverse recovery performance. For more information on SiC FET operation, see www.unitedsic.com.

A snubber circuit with a small $R_{(G)}$, or gate resistor, provides better EMI suppression with higher efficiency compared to using a high $R_{(G)}$ value. There is no extra gate delay time when using the snubber circuitry, and a small $R_{(G)}$ will better control both the turn-off $V_{(DS)}$ peak spike and ringing duration, while a high $R_{(G)}$ will damp the peak spike but result in a longer delay time. In addition, the total switching loss when using a snubber circuit is less than using high $R_{(G)}$, while greatly reducing $E_{(OFF)}$ from mid-to-full load range with only a small increase in $E_{(ON)}$. Efficiency will therefore improve with higher load current. For more information on how a snubber circuit will improve overall system performance, visit the UnitedSiC website at www.unitedsic.com













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