UMW[®]®®®&

UMW AP1501

GENERAL DESCRIPTION

The AP1501 series of regulators are monolithic integrated circuits that provide all the active functions for a step-down (buck) switching regulator, capable of driving a 3A load with excellent line and load regulation. These devices are available in fixed output voltages of 3.3V, 5V, 12V, and an adjustable output version.

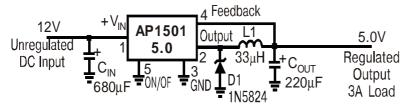
Requiring a minimum number of external components, these regulators are simple to use and include internal frequency compensation, and a fixed-frequency oscillator.

The AP1501 series operates at a switching frequency of 150 kHz thus allowing smaller sized filter components than what would be needed with lower frequency switching regulators. Available in a standard 5-lead TO-220 package with several different lead bend options, and a 5-lead TO-263 surface mount package.

A standard series of inductors are available from several different manufacturers optimized for use with the AP1501 series. This feature greatly simplifies the design of switch-mode power supplies.

Other features include a guaranteed $\pm 4\%$ tolerance on output voltage under specified input voltage and output load conditions, and $\pm 15\%$ on the oscillator frequency. External shutdown is included, featuring typically 80 µA standby current. Self protection features include a two stage frequency reducing current limit for the output switch and an over temperature shutdown for complete protection under fault conditions.

TYPICAL APPLICATION (Fixed Output Voltage Versions)



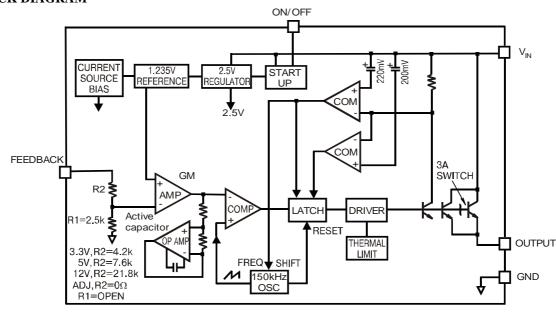
BLOCK DIAGRAM

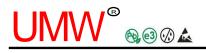


- 3.3V, 5V, 12V, and adjustable output versions
- Adjustable version output voltage range, 1.2V to 37V
- $\pm 4\%$ max over line and load conditions
- Available in TO-220 and TO-263 packages
- Guaranteed 3A output load current
- Input voltage range up to 40V
- Requires only 4 external components
- Excellent line and load regulation specifications
- 150 kHz fixed frequency internal oscillator
- TTL shutdown capability
- Low power standby mode, I_Q typically 80 μA
- High efficiency
- Uses readily available standard inductors
- Thermal shutdown and current limit protection

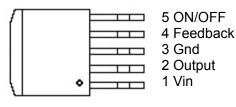
APPLICATIONS

- Simple high-efficiency step-down (buck) regulator
- On-card switching regulators
- Positive to negative converter





Pin Assignments



TO263-5L

Pin Descriptions

Name	Description
Vin	Input supply voltage
Output	Switching output
Gnd	Ground
Feedback	Output voltage feedback
ON/OFF	ON/OFF shutdown Active is "Low" or floating

PIN FUNCTIONS

 $+V_{IN}$ - This is the positive input supply for the IC switching regulator. A suitable input bypass capacitor must be present at this pin to minimize voltage transients and to supply the switching currents needed by the regulator.

Ground - Circuit ground.

Output - Internal switch. The voltage at this pin switches between (+V_{IN} - V_{SAT}) and approximately -0.5V, with a duty cycle of approximately V_{OUT} /V_{IN}. To minimize coupling to sensitive circuitry, the PC board copper area connected to this pin should be kept to a minimum.

Feedback—Senses the regulated output voltage to complete the feedback loop.

ON/OFF - Allows the switching regulator circuit to be shut down using logic level signals thus dropping the total input supply current to approximately $80 \ \mu$ A. Pulling this pin below a threshold voltage of approximately 1.3V turns the regulator on, and pulling this pin above 1.3V (up to a maximum of 25V) shuts the regulator down. If this shutdown feature is not needed, the ON /OFF pin can be wired to the ground pin or it can be left open, in either case the regulator will be in the ON condition.

ABSOLUTE MAXIMUM RATINGS (Note 1)

		. ,
Maximum Supply Voltage	45V	
ON /OFF Pin Input Voltage	$-0.3 \le V \le +25V$	
Feedback Pin Voltage	-0.3 ≤ V	≤+25V
Output Voltage to Ground		
(Steady State)		-1V
Power Dissipation	Internally limited	
Storage Temperature Range	-65° C to $+150^{\circ}$ C	
ESD Susceptibility		
Human Body Model (Note 2)	2 kV
Lead Temperature		
S Package		
Vapor Phase (60 sec.)		$+215^{\circ}C$
Infrared (10 sec.)	+245°C	
T Package (Soldering, 10 sec	2.)	$+260^{\circ}C$
Maximum Junction Tempera	ature	$+150^{0}C$

OPERATING CONDITIONS

Temperature Range	$-40^{\circ}C \le T_{J} \le +125^{\circ}C$
Supply Voltage	4.5V to 40V

AP1501-3.3

ELECTRICAL CHARACTERISTICS

Specifications with standard type face are for $T_J = 25^{0}$ C, and those with **boldface type** apply over **full Operating Temperature Range**

			AP1	Units	
Symbol	Parameter	Conditions	Typ (Note 3)	Limit (Note 4)	(Limits)
SYSTEM PA	ARAMETERS (Note 5)Tes	st Circuit Figure 1			
V _{OUT}	Output Voltage	$4.7V5 \le V_{IN} \le 40V, 0.2A \le I_{I,OAD} \le 3A$	3.3		V
				3.168/ 3.135	V(min)
				3.432/ 3.465	V(max)
η	Efficiency	$V_{IN}=12V, I_{LOAD}=3A$	73		%

AP1501-5.0

ELECTRICAL CHARACTERISTICS

Specifications with standard type face are for $T_J = 25^{0}$ C, and those with **boldface type** apply over **full Operating Temperature Range**

		AP1	Units	
Parameter	Conditions	Тур	Limit	(Limits)
		(Note 3)	(Note 4)	(Linits)
ARAMETERS (Note 5)Tes	t Circuit Figure 1			
Output Voltage	$7V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$	5.0		V
			4.800/ 4.750	V(min)
			5.200/ 5.250	V(max)
Efficiency	V _{IN} =12V, I _{LOAD} =3A	80		%
	ARAMETERS (Note 5)Tes Output Voltage	ARAMETERS (Note 5)Test Circuit Figure 1 Output Voltage $7V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$	ParameterConditionsTyp (Note 3)ARAMETERS (Note 5)Test Circuit Figure 1Output Voltage $7V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$ 5.0	ARAMETERS (Note 5)Test Circuit Figure 1 $(Note 3)$ $(Note 4)$ Output Voltage $7V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$ 5.0 $4.800/4.750$ $5.200/5.250$

AP1501-12

ELECTRICAL CHARACTERISTICS

Specifications with standard type face are for $T_J = 25^{0}$ C, and those with **boldface type** apply over **full Operating Temperature Range**

			AP1	TI		
Symbol	Parameter	Conditions	Typ (Note 3)	Limit (Note 4)	Units (Limits)	
SYSTEM PA	SYSTEM PARAMETERS (Note 5)Test Circuit Figure 1					
V _{OUT}	Output Voltage	$15V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$	12.0		V	
				11.52/ 11.40	V(min)	
				12.48/ 12.60	V(max)	
η	Efficiency	V _{IN} =12V, I _{LOAD} =3A	90		%	

AP1501-ADJ

ELECTRICAL CHARACTERISTICS

Specifications with standard type face are for $T_J = 25^{0}$ C, and those with **boldface type** apply over **full Operating Temperature Range**

			AP15	Units	
Symbol	Parameter	Conditions	Тур	Limit	(Limits)
			(Note 3)	(Note 4)	(Linnts)
SYSTEM PA	ARAMETERS (Note 5)Tes	st Circuit Figure 1			
V _{OUT}	Output Voltage	$4.5V \le V_{IN} \le 40V, 0.2A \le I_{LOAD} \le 3A$	1.230		V
		V _{OUT} programmed for 3V. Circuit of		1.193/ 1.180	V(min)
		Figure 1.		1.267/ 1.280	V(max)
η	Efficiency	$V_{IN}=12V, V_{OUT}=3V, I_{LOAD}=3A$	73		%

ALL OUTPUT VOLTAGE VERSIONS ELECTRICAL CHARACTERISTICS

Specifications with standard type face are for $T_J = 25^{0}$ C, and those with **boldface type** apply over **full Operating Temperature Rang**e. Unless otherwise specified, $V_{IN} = 12$ V for the 3.3V, 5V, and Adjustable version and $V_{IN} = 24$ V for the 12V version. $I_{LOAD} = 500$ mA

Symbol Parameter		AP1	AP1501-XX		
	Conditions	Тур	Limit	– Units (Limits)	
			(Note 3)	(Note 4)	(Linnts)
DEVICE P	ARAMETERS				-
Ib	Feedback Bias Current	Adjustable Version Only, V _{FB} =1.3V	10		nA
-				50/ 100	nA (max)
f_0	Oscillator Frequency	(Note 6)	150		kHz
				127/ 110	kHz (min)
				173/ 173	kHz (max)
V _{SAT}	Saturation Voltage	$I_{OUT}=3A$ (Notes 7, 8)	1.16		V
				1.4/ 1.5	V (max)
DC	Max Duty Cycle (ON)	(Note 8)	100		%
	Min Duty Cycle (OFF)	(Note 9)	0		
I _{CL}	Current Limit	Peak Current (Notes 7, 8)	4.5		А
				3.6/ 3.4	A (min)
				6.9/ 7.5	A (max)
I_L	Output Leakage Current	Output=0V (Notes 7, 9)		50	μA (max)
		Output=-0.9V (Note 10)	10		mA
				30	mA (max)
I_Q	Quiescent Current	(Note 9)	5		mA
				10	mA (max)
I _{STBY}	Standby Quiescent	ON/OFF pin=5V (OFF) (Note 10)	80		μΑ
	Current			200/250	$\mu A (max)$
θ_{JC}	Thermal Resistance	TO-220 or TO-263 Package, Junction to Case	2		⁰ C/W
θ_{JA}		TO-220 Package, Junction to Ambient (Note 11)	50		⁰ C/W
θ_{JA}		TO-263 Package, Junction to Ambient (Note 12)	50		⁰ C/W
θ_{JA}		TO-263 Package, Junction to Ambient (Note 13)	30		⁰ C/W
θ_{JA}		TO-263 Package, Junction to Ambient (Note 14)	20		⁰ C/W
ON/OFF C	ONTROL Test Circuit Figure	re 1			
	ON/OFF Pin Logic Input		1.3		V
V_{IH}	Threshold Voltage	Low (Regulator ON)		0.6	V (max)
V _{IL}		High (Regulator OFF)		2.0	V (min)
$I_{\rm H}$	ON/OFF Pin Input Current	V _{LOGIC} =2.5V (Regulator OFF)	5		μΑ
				15	μA (max)
I_L		V _{LOGIC} =0.5V (Regulator ON)	0.02		μΑ
				5	μA (max)

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics.

Note 2: The human body model is a 100 pF capacitor discharged through a 1.5k resistor into each pin.

Note 3: Typical numbers are at 25^oC and represent the most likely norm.

Note 4: All limits guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% production tested. All limits at temperature extremes are guaranteed via correlation using standard Statistical Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

Note 5: External components such as the catch diode, inductor, input and output capacitors, and voltage programming resistors can affect switching regulator system performance. When the AP1501 is used as shown in the Figure 1 test circuit, system performance will be as shown in system parameters section of Electrical Characteristics.

Note 6: The switching frequency is reduced when the second stage current limit is activated. The amount of reduction is determined by the severity of current over-load.

Note 7: No diode, inductor or capacitor connected to output pin.

Note 8: Feedback pin removed from output and connected to 0V to force the output transistor switch ON.

Note 9: Feedback pin removed from output and connected to 12V for the 3.3V, 5V, and the ADJ. version, and 15V for the 12V version, to force the output transistor switch OFF.

Note 10: $V_{IN} = 40V$.

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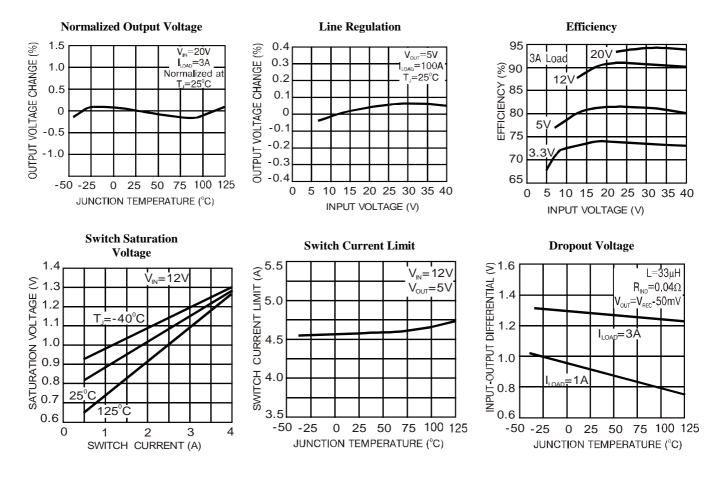
Note 11: Junction to ambient thermal resistance (no external heat sink) for the TO-220 package mounted vertically, with the leads soldered to a printed circuit board with (1 oz.) copper area of approximately 1 in^2

Note 12: Junction to ambient thermal resistance with the TO-263 package tab soldered to a single printed circuit board with 0.5 in² of (1 oz.) copper area.

Note 13: Junction to ambient thermal resistance with the TO-263 package tab soldered to a single sided printed circuit board with 2.5 in^2 of (1 oz.) copper area.

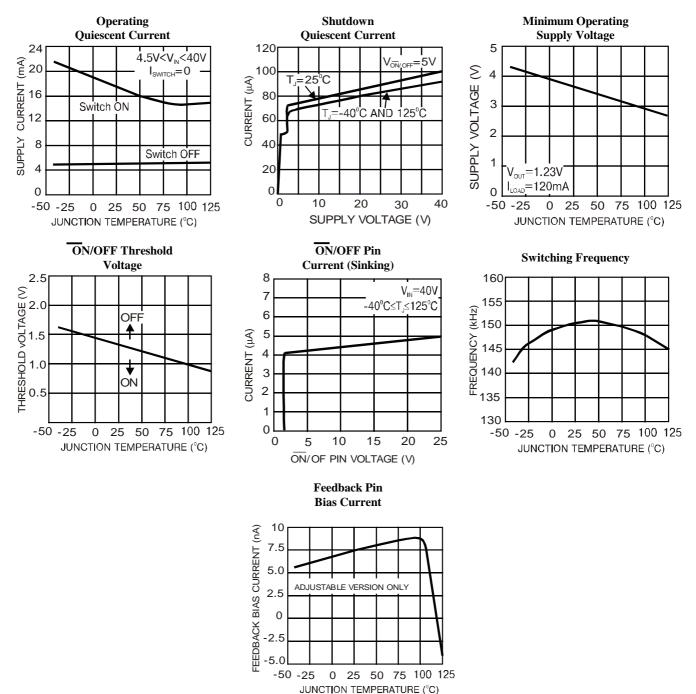
Note 14: Junction to ambient thermal resistance with the TO-263 package tab soldered to a double sided printed circuit board with 3 in^2 of (1 oz.) copper area on the AP1501S side of the board, and approximately 16 in ² of copper on the other side of the p-c board.

TYPICAL PERFORMANCE CHARACTERISTICS (Circuit of Figure 1)



TYPICAL PERFORMANCE CHARACTERISTICS (Circuit of Figure 1) (Continued)

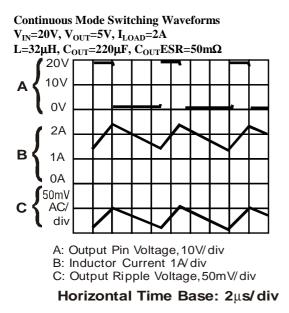
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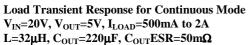


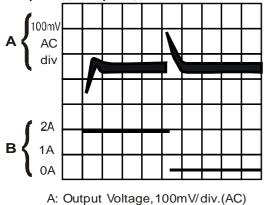
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TYPICAL PERFORMANCE CHARACTERISTICS

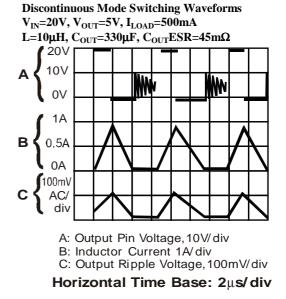




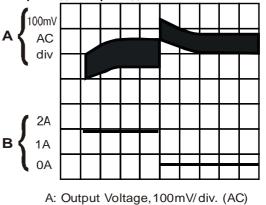


B: 500mA to 2A Load Pulse

Horizontal Time Base: 100µs/div



Load Transient Response for Discontinuous Mode $V_{IN}{=}20V,\,V_{OUT}{=}5V,\,I_{LOAD}{=}500mA$ to 2A L=10µH, $C_{OUT}{=}330\mu F,\,C_{OUT}{=}SR{=}45m\Omega$



B: 500mA to 2A Load Pulse

Horizontal Time Base: 200µs/div

TEST CIRCUIT AND LAYOUT GUIDELINES

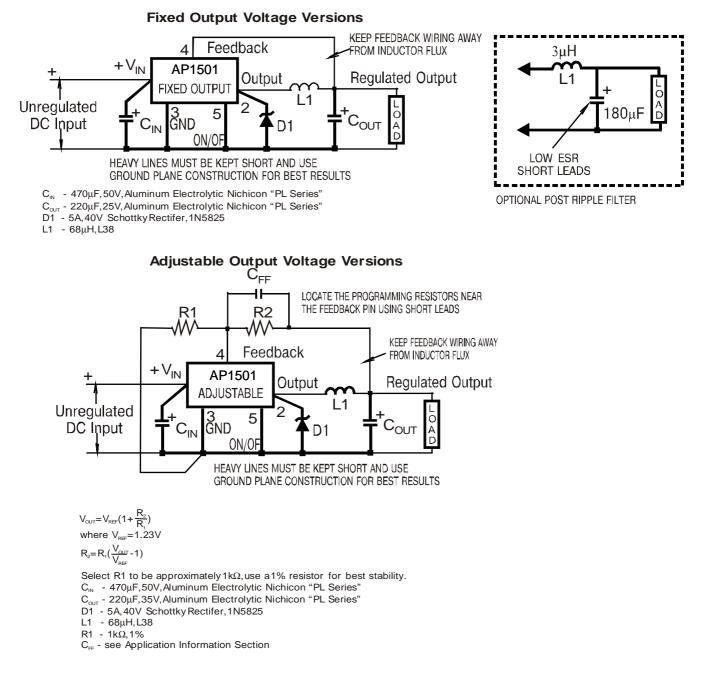


Figure 1. Standard Test Circuits and Layout Guides

As in any switching regulator, layout is very important. Rapidly switching currents associated with wiring inductance can generate voltage transients which can cause problems. For minimal inductance and ground loops, the wires indicated by **heavy lines should be wide printed circuit traces and**

should be kept as short as possible. For best results, external components should be located as close to the switcher IC as possible using ground plane construction or single point grounding.

If **open core inductors are used**, special care must be taken as to the location and positioning of this type of inductor. Allowing the inductor flux to intersect sensitive feedback, IC groundpath and C_{OUT} wiring can cause problems.

When using the adjustable version, special care must be taken as to the location of the feedback resistors and the associated wiring. Physically locate both resistors near the IC, and route the wiring away from the inductor, especially an open core type of inductor.



AP1501 SERIES BUCK REGULATOR DESIGN PROCEDURE (FIXED OUTPUT) PROCEDURE (Fixed Output Voltage Version) EXAMPLE (Fixed Output Voltage Version) Given: Given: V_{OUT} = Regulated Output Voltage (3.3V, 5V or 12V) $V_{OUT} = 5V$ V_{IN} (max) = Maximum DC Input Voltage V_{IN} (max) = 12V I_{LOAD} (max) = Maximum Load Current I_{LOAD} (max) = 3A 1. Inductor Selection (L1) 1. Inductor Selection (L1) A. Select the correct inductor value selection guide from Figures A. Use the inductor selection guide for the 5V version shown in Figure 5. Figure 4, Figure 5, or Figure 6. (Output voltages of 3.3V, 5V, or 12V respectively.) For all other voltages, see the design procedure for the adjustable version. **B.** From the inductor value selection guide, identify the inductance **B.** From the inductor value selection guide shown in Figure 5, the region intersected by the Maximum Input Voltage line and the inductance region intersected by the 12V horizontal line and the 3A Maximum Load Current line. Each region is identified by an vertical line is 33μ H, and the inductor code is L40. inductance value and an inductor code (LXX). C. Select an appropriate inductor from the four manufacturer's part C. The inductance value required is $33 \,\mu$ H. From the table in Figure 8, go to the numbers listed in Figure 8. L40 line and choose an inductor part number from any of the four manufacturers shown. (In most in-stance, both through hole and surface mount inductors are available.) 2. Output Capacitor Selection (C_{OUT}) 2. Output Capacitor Selection (C_{OUT}) A. In the majority of applications, low ESR (Equivalent Series A. See section on output capacitors in application information section. Resistance) electrolytic capacitors between 82 µF and 820 µF and low ESR solid tantalum capacitors between 10 µF and 470 µF provide the best results. This capacitor should be located close to the IC using short capacitor leads and short copper traces. Do not use capacitors larger than 820 µF. **B.** To simplify the capacitor selection procedure, refer to the quick B. From the quick design component selection table shown in Figure 2, locate the 5V output voltage section. In the load current column, choose the load current line design component selection table shown in Figure 2. This table that is closest to the current needed in your application, for this example, use the 3A contains different input voltages, output voltages, and load line. In the maximum input voltage column, select the line that covers the input currents, and lists various inductors and output capacitors that will voltage needed in your application, in this example, use the 15V line. Continuing provide the best design solutions. on this line are recommended inductors and capacitors that will provide the best overall performance. The capacitor list contains both through hole electrolytic and surface mount tantalum capacitors from four different capacitor manufacturers. It is recommended that both the manufacturers and the manufacturer's series that are listed in the table be used. In this example aluminum electrolytic capacitors from several different manufacturers are available with the range of ESR numbers needed. 330 µF 35V Panasonic HFQ Series 330 µF 35V Nichicon PL Series C. For a 5V output, a capacitor voltage rating at least 7.5V or more is needed. But C. The capacitor voltage rating for electrolytic capacitors should be even a low ESR, switching grade, 220µF 10V aluminum electrolytic capacitor at least 1.5 times greater than the output voltage, and often much would exhibit approximately 225 mW of ESR (see the curve in Figure 14 for the higher voltage ratings are needed to satisfy the low ESR ESR vs voltage rating). This amount of ESR would result in relatively high output requirements for low output ripple voltage. ripple voltage. To reduce the ripple to 1% of the output voltage, or less, a capacitor with a higher value or with a higher voltage rating (lower ESR) should be selected. A 16V or 25V capacitor will reduce the ripple volt-age by approximately half. 3. Catch Diode Selection (D1) 3. Catch Diode Selection (D1) A. The catch diode current rating must be at least 1.3 times greater than the A. Refer to the table shown in Figure 11. In this example, a 5A, 20V, 1N5823 maximum load current. Also, if the power supply design must withstand a Schottky diode will provide the best performance, and will not be overstressed even continuous output short, the diode should have a current rating equal to the for a shorted output. maximum current limit of the AP1501. The most stressful condition for this diode is an overload or shorted output condition. **B.** The reverse voltage rating of the diode should be at least 1.25 times the maximum input voltage. C. This diode must be fast (short reverse recovery time) and must be located close to the AP1501 using short leads and short printed circuit traces. Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best performance and efficiency, and should be the first choice, especially in low output voltage applications. Ultra-fast recovery, or High-Efficiency rectifiers also provide good results. Ultra-fast recovery diodes typically have reverse recovery times of 50 ns or less. Rectifiers such as the 1N5400 series are much too slow and should not be used.

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
4. Input Capacitor (C _{IN})	4. Input Capacitor (C _{IN})
A low ESR aluminum or tantalum bypass capacitor is needed between the	The important parameters for the Input capacitor are the input voltage rating and the
input pin and ground pin to prevent large volt-age transients from	RMS current rating. With a nominal
appearing at the input. This capacitor should be located close to the IC	input voltage of 12V, an aluminum electrolytic capacitor with a voltage rating
	greater than $18V (1.5 \text{ x } V_{IN})$ would be needed. The next higher capacitor voltage
should be selected to be at least 1/2 the DC load current. The capacitor	rating is 25V.
manufacturers data sheet must be checked to assure that this current rating	The RMS current rating requirement for the input capacitor in
is not exceeded. The curve shown in Figure 9 shows typical RMS current	a buck regulator is approximately 1 /2 the DC load current. In this example, with a
ratings for several different aluminum electrolytic capacitor values.	3A load, a capacitor with a RMS current rating of at least 1.5A is needed. The
For an aluminum electrolytic, the capacitor voltage rating should be	curves shown in Figure 9 can be used to select an appropriate input capacitor.
approximately 1.5 times the maximum input voltage.	From the curves, locate the 35V line and note which capacitor values have RMS
The tantalum capacitor voltage rating should be 2 times the maximum	current ratings greater than 1.5A. A 680µF/35V capacitor could be used.
input voltage and it is recommended that they be surge current tested by	For a through hole design, a 680µF/35V electrolytic capacitor (Panasonic HFQ
the manufacturer.	series or Nichicon PL series or equivalent) would be adequate. other types or other
Use caution when using ceramic capacitors for input bypassing, because it	manufacturers capacitors can be used provided the RMS ripple current ratings are
may cause severe ringing at the V_{IN} pin.	adequate.
	For surface mount designs, solid tantalum capacitors can be used, but caution must
	be exercised with regard to the capacitor surge current rating. The TPS series
	available from AVX, and the 593D series from Sprague are both surge current
	tested.

AP1501 SERIES BUCK REGULATOR DESIGN PROCEDURE (FIXED OUTPUT) (Continued)

Conditions		Inductor		Output Capacitor				
	Conditions		Inductor		Through Hole Electrolytic		Surfac	e Mount Tantalum
Output Voltage (V)	Load Current (A)	Max Input Voltage (V)	Inductance (µH)	Inductor (#)	Panasonic HFQ Series (µF/V)	Nichicon PL Series (µF/V)	AVX TPS Series (µF/V)	Sprague 595D Series (µF/V)
3.3	3	5	22	L41	470/25	560/16	330/6.3	390/6.3
		7	22	L41	560/35	560/35	330/6.3	390/6.3
		10	22	L41	680/35	680/35	330/6.3	390/6.3
		40	33	L40	560/35	470/35	330/6.3	390/6.3
	2	6	22	L33	470/25	470/35	330/6.3	390/6.3
		10	33	L32	330/35	330/35	330/6.3	390/6.3
		40	47	L39	330/35	270/50	220/10	330/10
5	3	8	22	L41	470/25	560/16	220/10	330/10
		10	22	L41	560/25	560/25	220/10	330/10
		15	33	L40	330/35	330/35	220/10	330/10
		40	47	L39	330/35	270/35	220/10	330/10
	2	9	22	L33	470/25	560/16	220/10	330/10
		20	68	L38	180/35	180/35	100/10	270/10
		40	68	L38	180/35	180/35	100/10	270/10
12	3	15	22	L41	470/25	470/25	100/16	180/16
		18	33	L40	330/25	330/25	100/16	180/16
		30	68	L44	180/25	180/25	100/16	120/20
		40	68	L44	180/35	180/35	100/16	120/20
	2	15	33	L32	330/25	330/25	100/16	180/16
		20	68	L38	180/25	180/25	100/16	120/20
		40	150	L42	82/25	82/25	68/20	68/25

Figure 2. AP1501 Fixed Voltage Quick Design Component Selection Table AP1501 SERIES BUCK REGULATOR DESIGN PROCEDURE (ADJUSTABLE OUTPUT)

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
Given:	Given:
V _{OUT} = Regulated Output Voltage	$V_{OUT} = 20V$
V _{IN} (max) = Maximum Input Voltage	$V_{IN}(max) = 28V$
I _{LOAD} (max) = Maximum Load Current	$I_{LOAD}(max) = 3A$
F=Switching Frequency (Fixed at a nominal 150 kHz).	F=Switching Frequency (Fixed at a nominal 150 kHz).
1. Programming Output Voltage (Selecting R ₁ and R ₂ , as shown in	1. Programming Output Voltage (Selecting R ₁ and R ₂ , as shown
Figure 1)	in Figure 1)
Use the following formula to select the appropriate resistor values.	Select R_1 to be 1 k Ω , 1%. Solve for R_2 .
$V_{OUT} = V_{REF} \left(1 + \frac{R}{R} \frac{2}{R}\right) \text{ where } V_{REF} = 1.23 \text{ V}$ Select a value for R ₁ between 240Ω and 1.5kΩ. The lower resistor values minimize noise pickup in the sensitive feedback pin. (For the lowest temperature coefficient and the best stability with time, use 1% metal film resistors.)	$R_{2} = R_{1}(\frac{V_{OUT}}{V_{REF}} - 1) = 1 k(\frac{20 V}{1.23 V} - 1)$ R ₂ =1k (16.26-1)=15.26k, closest 1% value is 15.4kΩ R ₂ = 15.4 kΩ.

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
$R_{2} = R_{1}\left(\frac{V_{OUT}}{V_{REF}} - 1\right)$	
V _{REF} 2. Inductor Selection (L1) A. Calculate the inductor Volt • microsecond constant E•T (V•µs), from the following formula:	 2. Inductor Selection (L1) A. Calculate the inductor Volt • microsecond constant (E•T),
$E \bullet T = (V_{IN} - V_{OUT} - V_{SAT}) \bullet \frac{V_{OUT} + V_{D}}{V_{IN} - V_{SAT} + V_{D}} \bullet \frac{1000}{150 \text{kHz}} (V \bullet \mu s)$	$E \bullet T = (28 - 20 - 1.16) \bullet \frac{20 + 0.5}{28 - 1.16 + 0.5} \bullet \frac{1000}{150} (V \bullet \mu s)$ $E \bullet T = (6.84) \bullet \frac{20.5}{27.34} \bullet 6.67 (V \bullet \mu s) = 34.2 (V \bullet \mu s)$
where V_{SAT} = internal switch saturation voltage = 1.16V and V_D = diode forward voltage drop = 0.5V	$E \bullet T = (6.84) \bullet \frac{20.5}{27.34} \bullet 6.67(V \bullet \mu s) = 34.2(V \bullet \mu s)$
B. Use the E•T value from the previous formula and match it with the E•T number on the vertical axis of the Inductor Value Selection Guide shown in Figure 7.	B. $E \bullet T = 34.2 (V \bullet \mu s)$
C. on the horizontal axis, select the maximum load current.	C. I_{LOAD} (max) = 3A
 D. Identify the inductance region intersected by the E•T value and the Maximum Load Current value. Each region is identified by an inductance value and an inductor code (LXX). E. Select an appropriate inductor from the four manufacturer's part numbers listed in Figure 8. 	D. From the inductor value selection guide shown in Figure 7, the inductance region intersected by the 34 ($V \bullet \mu s$) horizontal line and the 3A vertical line is 47 μ H, and the inductor code is L39. E. From the table in Figure 8, locate line L39, and select an inductor part number from the list of manufacturers part numbers.
3. Output Capacitor Selection (C_{OUT}) A. In the majority of applications, low ESR electrolytic or solid tantalum capacitors between 82 μ F and 820 μ F provide the best results. This capacitor should be located close to the IC using short capacitor leads and short copper traces. Do not use capacitors larger than 820 μ F.	3. Output Capacitor Selection (C _{OUT})
B. To simplify the capacitor selection procedure, refer to the quick design table shown in Figure 3. This table contains different output voltages, and lists various output capacitors that will provide the best design solutions.	B. From the quick design table shown in Figure 3, locate the output voltage column. From that column, locate the output voltage closest to the output voltage in your application. In this example, select the 24V line. Under the output capacitor section, select a capacitor from the list of through hole electrolytic or surface mount tantalum types from four different capacitor manufacturers. It is recommended that both the manufacturers and the manufacturers series that are listed in the table be used. In this example, through hole aluminum electrolytic capacitors from several different manufacturers are available. $220 \mu\text{F/35V}$ Panasonic HFQ Series $150 \mu\text{F/35V}$ Nichicon PL Series
C. The capacitor voltage rating should be at least 1.5 times greater than the output voltage, and often much higher voltage ratings are needed to satisfy the low ESR requirements needed for low output ripple voltage.	C. For a 20V output, a capacitor rating of at least 30V or more is needed. In this example, either a 35V or 50V capacitor would work. A 35V rating was chosen, although a 50V rating could also be used if a lower output ripple voltage is needed. Other manufacturers or other types of capacitors may also be used, provided the capacitor specifications (especially the 100 kHz ESR) closely match the types listed in the table. Refer to the capacitor manufacturers data sheet for this information.
4. Feedforward Capacitor (C _{FF}) (See Figure 1) For output voltages greater than approximately 10V, an additional capacitor is required. The compensation capacitor is typically between 100 pF and 33 nF, and is wired in parallel with the output voltage setting resistor, R2. It provides additional stability for high output voltages, low input-output voltages, and/or very low ESR output capacitors, such as solid tantalum capacitors. $C_{FF} = \frac{1}{31 \times 10^3 \times R_2}$	4. Feedforward Capacitor (C_{FF}) The table shown in Figure 3 contains feed forward capacitor values for various output voltages. In this example, a 560 pF capacitor is needed.
This capacitor type can be ceramic, plastic, silver mica, etc. (Because of the unstable characteristics of ceramic capacitors made with Z5U material, they are not recommended.)	

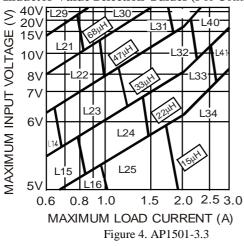
	Thro	ugh Hole Output Cap	pacitor	Surface Mount Output Capacitor			
Output Voltage (V)	Panasonic HFQ Series (µF/V)	Nichicon PL Series (µF/V)	Feedforward capacitor	AVX TPS Series (µF/V)	Sprague 595D Series (µF/V)	Feedforward Capacitor	
2	820/35	820/35	33 nF	330/6.3	470/4	33 nF	
4	560/35	470/35	10 nF	330/6.3	390/6.3	10 nF	
6	470/25	470/25	3.3 nF	220/10	330/10	3.3 nF	
9	330/25	330/25	1.5 nF	100/16	180/16	1.5 nF	
12	330/25	330/25	1 nF	100/16	180/16	1 nF	
15	220/35	220/35	680 pF	68/20	120/20	680 pF	
24	220/35	150/35	560 pF	33/25	33/25	220 pF	
28	100/50	100/50	390 pF	10/35	15/50	220 pF	

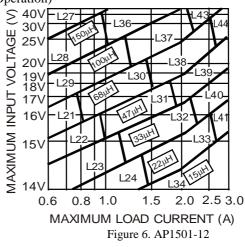
AP1501 SERIES BUCK REGULATOR DESING PROCEDURE (ADJUSTABLE OUTPUT)

Figure 3. Output Capacitor and Feedforward Capacitor Selection Table

AP1501 SERIES BUCK REGULATOR DESIGN PROCEDURE

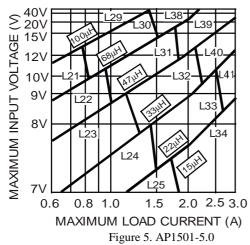
Inductor Value Selection Guides (For Continuous Mode Operation)

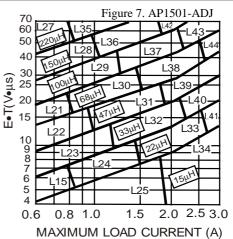






UMW AP1501





AP1501 SERIES BUCK REGULATOR DESIGN PROCEDURE (Continued)

	Inductance	Cur-	Cur- Schott		Renco		Pulse Engineering		Coilcraft
	(µH)	rent	Through	Surface	Through Hole	Surface	Through	Surface	Surface Mount
		(A)	Hole	Mount		Mount	Hole	Mount	
L15	22	0.99	67148350	67148460	RL-1284-22-43	RL1500-22	PE-53815	PE-53815-S	DO3308-223
L21	68	0.99	67144070	67144450	RL-5471-5	RL1500-68	PE-53821	PE-53821-S	DO3316-683
L22	47	1.17	67144080	67144460	RL-5471-6	-	PE-53822	PE-53822-S	DO3316-473
L23	33	1.40	67144090	67144470	RL-5471-7	-	PE-53823	PE-53823-S	DO3316-333
L24	22	1.70	67148370	67148480	RL-1283-22-43	-	PE-53824	PE-53825-S	DO3316-223
L25	15	2.10	67148380	67148490	RL-1283-15-43	-	PE-53825	PE-53824-S	DO3316-153
L26	330	0.80	67144100	67144480	RL-5471-1	-	PE-53826	PE-53826-S	DO5022P-334
L27	220	1.00	67144110	67144490	RL-5471-2	-	PE-53827	PE-53827-S	DO5022P-224
L28	150	1.20	67144120	67144500	RL-5471-3	-	PE-53828	PE-53828-S	DO5022P-154
L29	100	1.47	67144130	67144510	RL-5471-4	-	PE-53829	PE-53829-S	DO5022P-104
L30	68	1.78	67144140	67144520	RL-5471-5	-	PE-53830	PE-53830-S	DO5022P-683
L31	47	2.20	67144150	67144530	RL-5471-6	-	PE-53831	PE-53831-S	DO5022P-473
L32	33	2.50	67144160	67144540	RL-5471-7	-	PE-53932	PE-53932-S	DO5022P-333
L33	22	3.10	67148390	67148500	RL-1283-22-43	-	PE-53933	PE-53933-S	DO5022P-223
L34	15	3.40	67148400	67148790	RL-1283-15-43	-	PE-53934	PE-53934-S	DO5022P-153
L35	220	1.70	67144170	-	RL-5473-1	-	PE-53935	PE-53935-S	-
L36	150	2.10	67144180	-	RL-5473-4	-	PE-54036	PE-54036-S	-
L37	100	2.50	67144190	-	RL-5472-1	-	PE-54037	PE-54037-S	-
L38	68	3.10	67144200	-	RL-5472-2	-	PE-54038	PE-54038-S	-
L39	47	3.50	67144210	-	RL-5472-3	-	PE-54039	PE-54039-S	-
L40	33	3.50	67144220	67148290	RL-5472-4	-	PE-54040	PE-54040-S	-
L41	22	3.50	67144230	67148300	RL-5472-5	-	PE-54041	PE-54041-S	-
L42	150	2.70	67148410	-	RL-5473-4	-	PE-54042	PE-54042-S	-
L43	100	3.40	67144240	-	RL-5473-2	-	PE-54043		-
L44	68	3.40	67144250	-	RL-5473-3	-	PE-54044		-

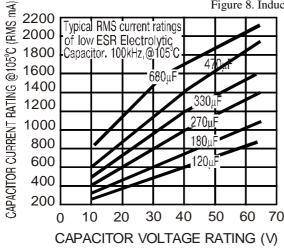


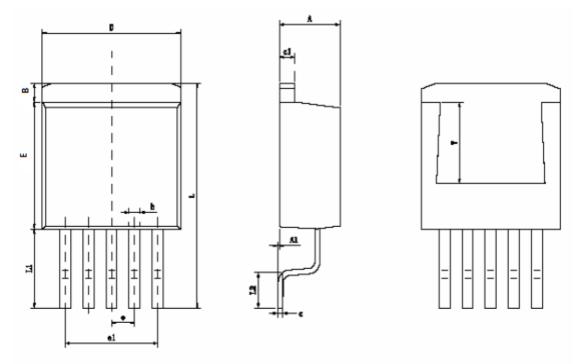
Figure 8. Inductor Manufacturers Part Numbers

Figure 9. RMS Current Ratings for Low ESR Electrolytic Capacitors (typical)

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PACKAGES DIMENSION : TO263-5L



Symphol	Dimensions	In Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
Α	4.470	4.670	0.176	0.184	
A1	0.000	0.150	0.000	0.006	
В	1.560	1.760	0.061	0.069	
b	0.710	0.910	0.028	0.036	
С	0.310	0.530	0.012	0.021	
c1	1.170	1.370	0.046	0.054	
D	9.880	10.180	0.389	0.401	
E	8.200	8.600	0.323	0.339	
е	1.70	DTYP	0.067TYP		
e1	6.700	6.900	0.264	0.272	
L	15.140	15.540	0.596	0.612	
L1	5.080	5.480	0.200	0.216	
L2	2.340	2.740	0.092	0.108	
V	5.600	REF	0.220REF		