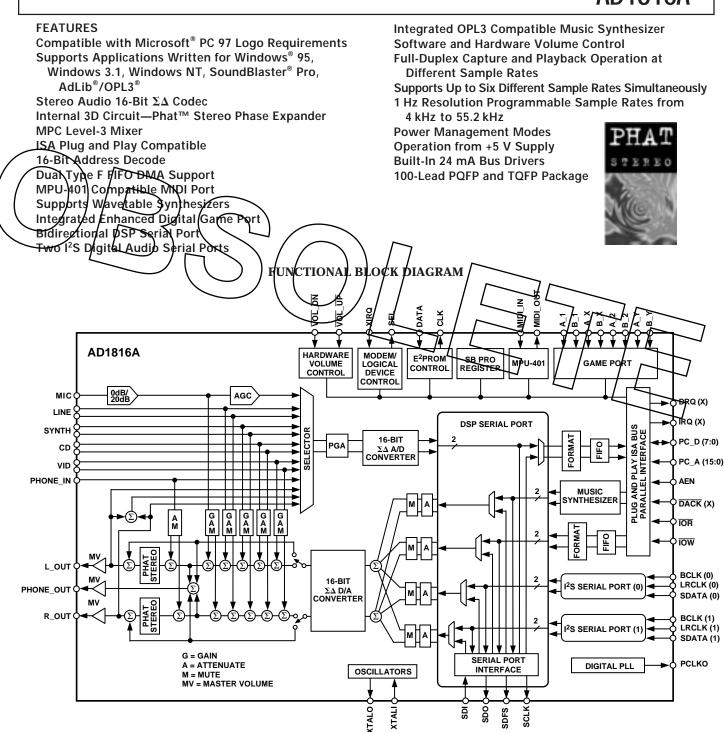


# SoundPort® Controller

## AD1816A



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#### PRODUCT OVERVIEW

The AD1816A SoundPort Controller is a single chip Plug and Play multimedia audio subsystem for concurrently processing multiple digital streams of 16-bit stereo audio in personal computers. The AD1816A maintains full legacy compatibility with applications written for SoundBlaster Pro and AdLib, while servicing Microsoft PC 97 application requirements. The AD1816A includes an internal OPL3 compatible music synthesizer, Phat

Stereo circuitry for phase expanding the analog stereo output, an MPU-401 UART, joystick interface with a built-in timer, a DSP serial port and two  $I^2S$  serial ports. The AD1816A on-chip Plug and Play routine provides configuration services for all integrated logical devices. Using an external  $E^2PROM$  allows the AD1816A to decode up to two additional external user-defined logical devices such as modem and CD-ROM.

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# **SPECIFICATIONS**

STANDARD TEST CONDITION OTHERWISE NOTED	NS UNLESS		DAC Test Conditions 0 dB Attenuation
Temperature	25	°C	Input Full Scale
Digital Supply (V <sub>DD</sub> )	5.0	V	16-Bit Linear Mode
Analog Supply (V <sub>CC</sub> )	5.0	V	100 kΩ Output Load
Sample Rate (F <sub>S</sub> )	48	kHz	Mute Off
Input Signal Frequency	1008	Hz	Measured at Line Output
Audio Output Passband	20 Hz to 2	20 kHz	ADC Test Conditions
$V_{\mathrm{IH}}$	5.0	V	0 dB Gain
$V_{\mathrm{IL}}$	0	V	Input –4 dB Relative to Full Scale
			Line Input Selected
			16-Bit Ĺinear Mode

ANALOG INPUT				
Parameter	Min	Тур	Max	Units
Full-Scale Input Voltage (RMS Valuet Assume Sine Wave Input) PHONE_IN, LINE, SYNTH, CD, VID  MIC with +20 dB Gain (MGE = 1)  MIC with 0 dB Gain (MGE = 0)  Input Impedance* Input Capacitance*		1 2.83 0.1 0.283 1 2.83 17 15	7///	V rms V p-p V rms V p-p V rms V p-p kΩ
PROGRAMMABLE GAIN AMPLIFIER—ADC				
Parameter	Min	Тур	Max	Units
Step Size (0 dB to 22.5 dB)				
(All Steps Tested)		1.5		dB
PGA Gain Range Span		22.5		dB

### CD, LINE, MICROPHONE, SYNTHESIZER, AND VIDEO INPUT ANALOG GAIN/ATTENUATORS/MUTE AT LINE OUTPUT

Parameter	Min	Тур	Max	Units
CD, LINE, MIC, SYNTH, VID				
Step Size: (All Steps Tested)				
+12 dB to -34.5 dB		1.5		dB
Input Gain/Attenuation Range		46.5		dB
PHONE_IN				
Step Size 0 dB to -45 dB: (All Steps Tested)		3.0		dB
Input Gain/Attenuation Range		45		dB

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#### DIGITAL DECIMATION AND INTERPOLATION FILTERS\*

Parameter	Min	Тур	Max	Units
Audio Passband	0		$0.4 \times F_S$	Hz
Audio Passband Ripple			$\pm 0.09$	dB
Audio Transition Band	$0.4 \times F_S$		$0.6 \times F_S$	Hz
Audio Stopband	$0.6 \times F_S$		∞	Hz
Audio Stopband Rejection	82			dB
Audio Group Delay			$12/F_S$	sec
Group Delay Variation Over Passband			0.0	μs

#### ANALOG-TO-DIGITAL CONVERTERS

Parameter	Min	Тур	Max	Units
Resolution		16		Bits
Signal to-Noise Ratio (SNR) (A-Weighted, Referenced to Full Scale)		82	80	dB
Total Harmonic Distortion (THD) (Referenced to Full Scale)		0.011	0.015	%
Audio Dynamic Range (-60 dB Input THD+N Referenced to		-79	-76.5	dB
Full-Scale, A-Weightett	79	82		dB
Audio THD+N (Referenced to Full-Scale)			0.019	%
		76	-74.5	dB
Signal-to-Intermodulation Distortion* (CCIF Method)	/ /	82		dB
ADC Crosstalk*	/ /			
Line Inputs (Input L, Ground R, Read R; Input R, Ground L Read L	)/	/95 ~	80	→ _dB
Line to MIC (Input LINE, Ground and Select MIC, Read ADC)	/   / / /	-95	- <b>\$</b> 0 <b>/</b> ∼	$\sqrt{\mathbb{D}}$
Line to SYNTH		-95	<del>/</del> 80 /	/ d/B
Line to CD		$L_{-95}$	/-80/	/ <b>(</b> B
Line to VID	' -	-95	/-8 <b>ø</b>	/ $dB$
Gain Error (Full-Scale Span Relative to Nominal Input Voltage)			/ ± <b>1/</b> 0	/ /%
Interchannel Gain Mismatch (Difference of Gain Errors)			<u>-</u>	/
ADC Offset Error	-22		+15	-mV

#### DIGITAL-TO-ANALOG CONVERTERS

Parameter	Min	Тур	Max	Units
Resolution		16		Bits
Signal-to-Noise Ratio (SNR) (A-Weighted)		83	79	dB
Total Harmonic Distortion (THD)		0.006	0.009	%
		-85	-80.5	dB
Audio Dynamic Range (-60 dB Input THD+N Referenced to				
Full Scale, A-Weighted)	79	82		dB
Audio THD+N (Referenced to Full Scale)		0.013	0.017	%
,		-78	-75.5	dB
Signal-to-Intermodulation Distortion* (CCIF Method)		95		dB
Gain Error (Full-Scale Span Relative to Nominal Input Voltage)			$\pm 10$	%
Interchannel Gain Mismatch (Difference of Gain Errors)			$\pm 0.5$	dB
DAC Crosstalk* (Input L, Zero R, Measure R_OUT;				
Input R, Zero L, Measure L_OUT)			-80	dB
Total Out-of-Band Energy (Measured from $0.6 \times F_S$ to 100 kHz				
at L_OUT and R_OUT)*			-45	dB
Audible Out-of-Band Energy (Measured from $0.6 \times F_S$ to 20 kHz				
at L_OUT and R_OUT)*			-75	dB

#### MASTER VOLUME ATTENUATORS (L\_OUT AND R\_OUT, PHONE\_OUT)

Parameter	Min	Тур	Max	Units
Master Volume Step Size (0 dB to -46.5 dB) Master Volume Output Attenuation Range Span Mute Attenuation of 0 dB Fundamental*		1.5 46.5	-80	dB dB dB
Nutre Attenuation of our Fundamental			-00	ub

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dB LSB

#### **DIGITAL MIX ATTENUATORS\***

Parameter	Min	Тур	Max	Units
Step Size: I <sup>2</sup> S (0), I <sup>2</sup> S (1), Music, ISA		1.505		dB
Digital Mix Attenuation Range Span		94.8		dB

#### ANALOG OUTPUT

Parameter	Min	Typ	Max	Units
Full-Scale Output Voltage (at L_OUT, R_OUT, PHONE_OUT)		2.8		V p-p
Output Impedance*			570	Ω
External Load Impedance*	10			kΩ
Output Capacitance*		15		pF
External Load Capacitance			100	pF
REFX	2.10	2.25	2.40	V
Veex Current Drive*		100		μΑ
N <sub>REFX</sub> Output Impedance*		6.5		kΩ
Master Volume Mute Click (Muted Analog Mixers), Muted				
Output Minus Unmuted Output at 0 dB		$\pm 5$		mV

#### SYSTEM SPICIFICATIONS

Parameter			) (	( )		/ /	Min	Тур	Max
System Frequency Response R Differential Nonlinearity Phase Linearity Deviation	ipple (	Line In	o Line	e Out)	//				1.0 †1

#### STATIC DIGITAL SPECIFICATIONS

Parameter	Min	p / Max /	Units
High Level Input Voltage (V <sub>IH</sub> )	2		**
XTALI	2.4	_	
Low Level Input Voltage (V <sub>IL</sub> )		0.8	V
High Level Output Voltage ( $V_{OH}$ ), $I_{OH} = 8 \text{ mA}^{\dagger}$	2.4		V
Low Level Output Voltage $(V_{OL})$ , $I_{OL} = 8 \text{ mA}$		0.4	V
Input Leakage Current	-10	+10	μΑ
Output Leakage Current	-10	+10	μA

#### **POWER SUPPLY**

Parameter	Min	Тур	Max	Units
Power Supply Range—Analog	4.75		5.25	V
Power Supply Range—Digital	4.75		5.25	V
Power Supply Current			221	mA
Power Dissipation			1105	mW
Analog Supply Current			51	mA
Digital Supply Current			170	mA
Analog Power Supply Current—Power-Down			2	mA
Digital Power Supply Current—Power-Down			24	mA
Analog Power Supply Current—RESET			0.2	mA
Digital Power Supply Current—RESET			10	mA
Power Supply Rejection (100 mV p-p Signal on Both Analog and Digital				
Supply Pins, Measured at ADC and Line Outputs)		40		dB

#### **CLOCK SPECIFICATIONS\***

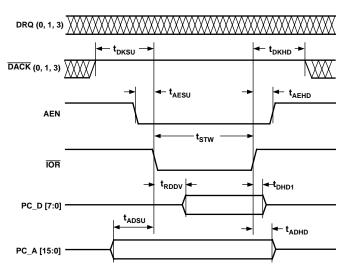
Parameter	Min	Тур	Max	Units
Input Clock Frequency Recommended Clock Duty Cycle Power-Up Initialization Time	25	33 50	75 500	MHz % ms

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TIMING PARAMETERS (Guaranteed Over Operating Temperature Range)

Parameter	Symbol	Min	Тур	Max	Units
IOW/IOR Strobe Width	$t_{STW}$	100			ns
IOW/IOR Rising to IOW/IOR Falling	$t_{ m BWDN}$	80			ns
Write Data Setup to IOW Rising	t <sub>WDSU</sub>	10			ns
IOW Falling to Valid Read Data	$t_{ m RDDV}$			40	ns
AEN Setup to <del>IOW/IOR</del> Falling	t <sub>AESU</sub>	10			ns
AEN Hold from IOW/IOR Rising	$t_{AEHD}$	0			ns
Adr Setup to IOW/IOR Falling	t <sub>ADSU</sub>	10			ns
Adr Hold from IOW/IOR Rising	$t_{ m ADHD}$	0			ns
DACK Rising to IOW/IOR Falling	$t_{ m DKSU}$	20			ns
Data Hold from IOR Rising	t <sub>DHD1</sub>			2	ns
Data Hold from IOW Rising	t <sub>DHD2</sub>	15			ns
DRQ Hold from IOW/IOR Falling	$t_{DRHD}$			25	ns
DACK Hold from IOW/IOR Rising	t <sub>DKHD</sub>	10			ns
Data [SDI] Input Setup Time to SCLK*	$t_{\rm S}$	15			ns
Data [SDI] Input Hold Time from SCLK	$t_{H}$	10			ns
Frame Sync SDFS/HI Pulse Width*	$t_{FSW}$		80		ns
Clock (SCLK) to Frame Sync [SOFS]					
Propagation Delay*	$t_{PD}$			15	ns
Clock [SCLK] to Output Data [SDØ] Valid*	\		_	15	ns
RESET Pulse Width	t t <sub>RPWL</sub>	/ / 100	/		ns
BCLK HI Pulse Width	/ \ t\ <sub>DBH</sub> / /	/ / 25 /		1 /	ns
BCLK LO Pulse Width	$t_{DBL}$	/ 25 /			ns
BCLK Period	TOBR	\[ \( \) 50 \[ \)	$\sim$		/ ns
LRCLK Setup	$t_{\rm DLS}$	5-1		/ /	/ n/s]
SDATA Setup	$t_{ m DDS}$	5		/ /	/ Ms_
SDATA Hold	$t_{\mathrm{DDH}}$	5	$\sim$ 7	/ /	/ ns
NOTES			7	LI	
*Guaranteed, not tested.				7	
†All ISA pins MIDI_OUT IOL = 24 mA. Refer to pin descript	ion for individual output driv	e levels.			7
Specifications subject to change without notice.					7

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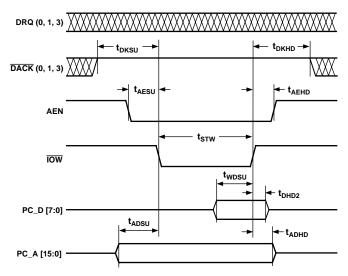


Figure 2. PIO Write Cycle

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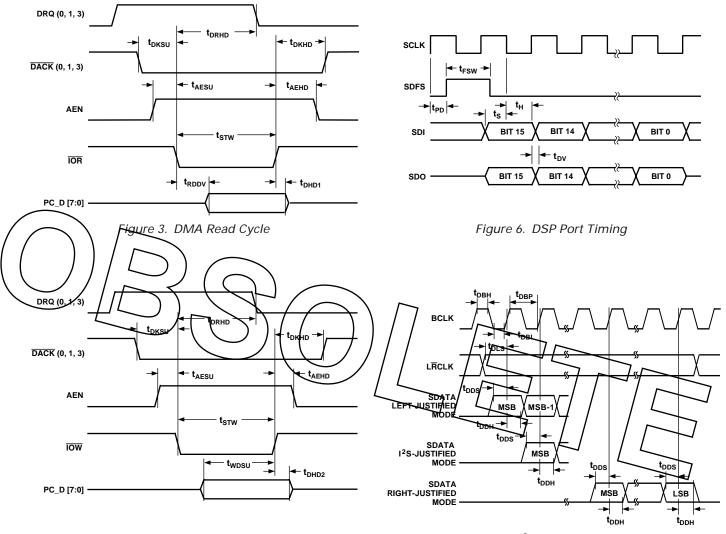
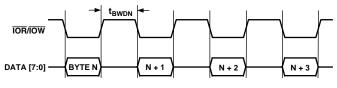


Figure 4. DMA Write Cycle

Figure 7. I<sup>2</sup>S Serial Port Timing





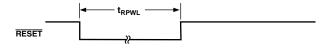


Figure 8. Reset Pulse Width

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#### ABSOLUTE MAXIMUM RATINGS\*

Parameter	Min	Max	Units
Power Supplies			
Digital $(V_{DD})$	-0.3	6.0	V
Analog (V <sub>CC</sub> )	-0.3	6.0	V
Input Current (Except Supply Pins)		$\pm 10.0$	mA
Analog Input Voltage (Signal Pins)	-0.3	$V_{CC} + 0.3$	V
Digital Input Voltage (Signal Pins)	-0.3	$V_{\rm DD} + 0.3$	V
Ambient Temperature (Operating)	0	+70	°C
Storage Temperature	-65	+150	°C

\*Stresses greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ENVIRONMENTAL CONDITIONS**

**Ambient Temperature Rating:** 

$$\begin{split} T_{AMB} &= T_{CASE} - (PD \times \theta_{CA}) \\ T_{CASE} &= Case \ Temperature \ in \ ^{\circ}C \end{split}$$

PD = Power Dissipation in W

 $\theta_{CA}$  = Thermal Resistance (Case-to-Ambient)

 $\theta_{JA}$  = Thermal Resistance (Junction-to-Ambient)

 $\theta_{JC} = Thermal \; Resistance \; (Junction-to-Case)$ 

Package	$\theta_{ m JA}$	$\theta_{ m JC}$	$\theta_{CA}$	
PQFP	35.1°C/W	7°C/W	28°C/W	
TQFP	35.3°C/W	8°C/W	27.3°C/W	

#### **ORDERING GUIDE**

Model	Temperature	Package	Package
	Range	Description	Option*
AD1816AJS	0°C to +70°C	100-Lead PQFP	S-100
AD1816AJST	0°C to +70°C	100-Lead TQFP	ST-100

= Plastic quad Flatpack; ST = Thin Quad Flatpack. JST package option availability subject to 10,000 PC minimum order quantity.

#### CAUTION.

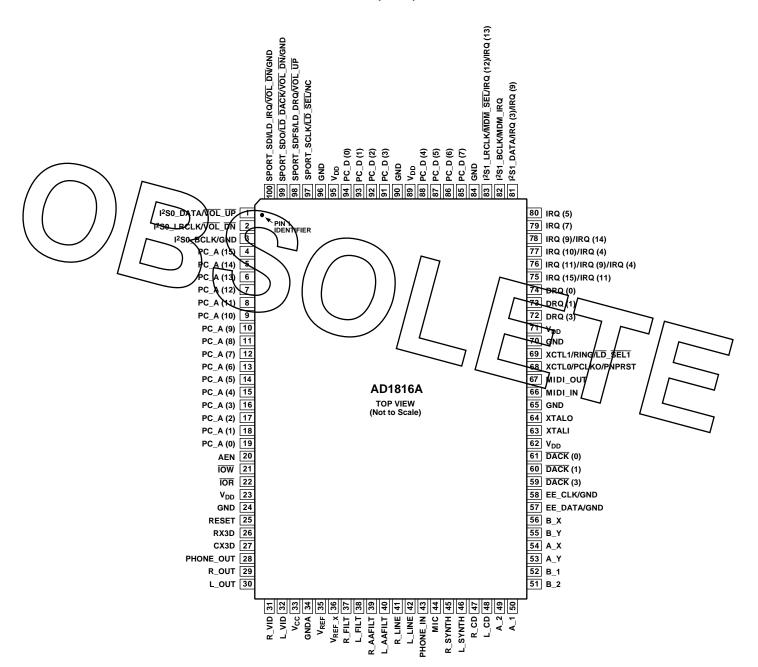
ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readil accumulate on the human body and test equipment and can discharge without detection Although the AD1816A features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

The AD1816A latchup immunity has been demonstrated at  $\geq +100$  mA/-80 mA on all pins when tested to Industry Standard/JEDEC methods.



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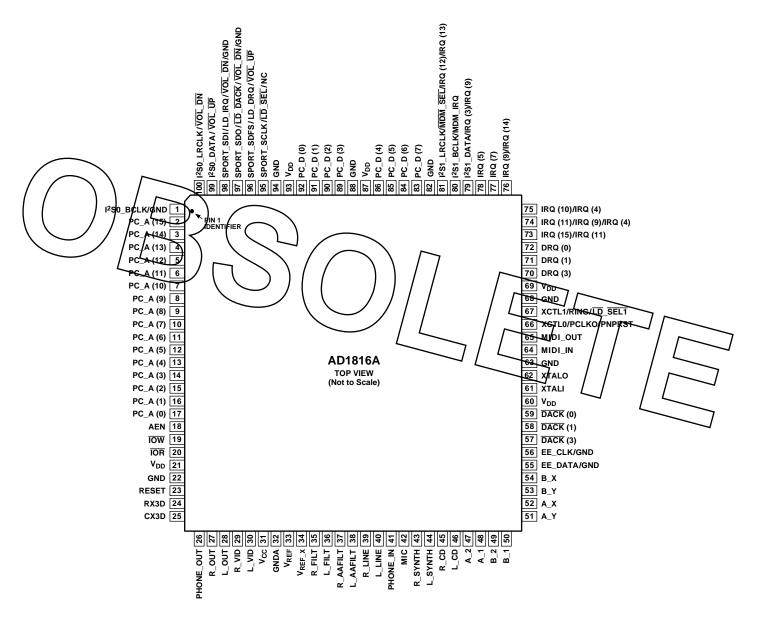
### PIN CONFIGURATION 100-Lead PQFP (S-100)



NC = NO CONNECT

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### PIN CONFIGURATION 100-Lead TQFP (ST-100)



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NC = NO CONNECT

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### PIN FUNCTION DESCRIPTIONS

Analog Signals (All Inputs must be AC-Coupled)

Pin Name	PQFP	TQFP	I/O	Description			
MIC	44	42	I	Microphone Input. The MIC input may be either line-level or $-20$ dB from line-level (the difference being made up through a software controlled 20 dB gain block). The mono Moninput may be sent to the left and right channel of the ADC for conversion, or gained/attenuated from $+12$ dB to $-34.5$ dB in 1.5 dB steps and then summed with left and right line OUT before the Master Volume stage.			
L_LINE	42	40	I	Left Line-Level Input. The left line-level input may be sent to the left channel of the ADC; gained/attenuated from $+12$ dB to $-34.5$ dB in $1.5$ dB steps and then summed with left line OUT (L_OUT).			
R_LINE	41	39	I	Right Line-Level Input. The right line-level input may be sent to the right channel of the ADC; gained/attenuated from $+12$ dB to $-34.5$ dB in 1.5 dB steps and then summed we right line OUT (R_OUT).			
L_SYNTH	46	44	I	Left Synthesizer Input. The left MIDI upgrade line-level input may be sent to the left channel of the ADC; gained/attenuated from +12 dB to -34.5 dB in 1.5 dB steps and then summed with left line OUT (L_OUT).			
R_SYNTH	45	48	I	Right Synthesizer Input. The right MIDI upgrade line-level input may be sent to the right channel of the ADC; gained/attenuated from +12 dB to -34.5 dB in 1.5 dB steps and then summed with right line OUT (RLOUT).			
L_CD	48	46		Left CD Line-Level Input. The left CD line level input may be sent to the left channel of the ADC; gained/attenuated from +12 dB to -34.5 dB in 1.5 dB steps and then summed with left line OUT (LOUT).			
R_CD	47	45	I	Right CD Line-Level Input. The right CD line-level input may be sent to the right channel of the ADC; gained/attenuated from +12 dB to -34.5 dB in 1.5 dB steps and then summed with right line OUT (R_OUT).			
L_VID	32	30	I	Left Video Input. The left audio track for a video line level input may be sent to the left channel of the ADC; gained/attenuated from +12 dB to -34.5 dB in 1.5 dB steps and then summed with left line OUT (L_OUT).			
R_VID	31	29	I	Right Video Input. The right audio track for a video line-level input may be sent to the right channel of the ADC; gained/attenuated from $+12$ dB to $-34.5$ dB in $1.5$ dB steps and then summed with right line OUT (R_OUT).			
L_OUT	30	28	О	Left Output. Left channel line-level post-mixed output. The final stage passes through the Master Volume block and may be attenuated 0 dB to $-45$ dB in 1.5 dB steps.			
R_OUT	29	27	О	Right Output. Right channel line-level post-mixed output. The final stage passes through the Master Volume block and may be attenuated 0 dB to -45 dB in 1.5 dB steps.			
PHONE_IN	43	41	I	Phone Input. Line-level input from a DAA/modem chipset.			
PHONE_OUT	28	26	0	Phone Output. Line-level output from a DAA/modem chipset.			
RX3D	26	24	О	Phat Stereo Phase Expander filter network, resistor pin.			
CX3D	27	25	I	Phat Stereo Phase Expander filter network, capacitor pin.			

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## Parallel Interface (All Outputs are 24 mA Drivers)

Pin Name	PQFP	TQFP	I/O	Description
PC_D[7:0]	85-88, 91-94	83-86, 89-92	I/O	Bidirectional ISA Bus PC Data, 24 mA drive. Connects the AD1816A to the low byte data on the bus.
IRQ (x)*	75–81, 83	73–79, 81	О	Host Interrupt Request, 24 mA drive. IRQ (3)/IRQ (9), IRQ (5), IRQ (7), IRQ (9)/IRQ (14), IRQ (10)/IRQ (4), IRQ (11)/IRQ (9)/IRQ (4), IRQ (12)/IRQ (13), IRQ (15)/IRQ (11). Active HI signals indicating a pending interrupt.
DRQ (x)	72–74	70-72	О	DMA Request, 24 mA drive. DRQ (0), DRQ (1), DRQ (3). Active HI signals indicating a request for DMA bus operation.
PC_A[15:0]	4-19	2-17	I	ISA Bus PC Address. Connects the AD1816A to the ISA bus address lines.
AEN	20	18	I	Address Enable. Low signal indicates a PIO transfer.
DACK (x)	59-61	57–59	I	DMA Acknowledge. DACK (0), DACK (1), DACK (3). Active LO signal indicating that a DMA operation can begin.
ĪØR /	\ 2 <sub>4</sub> /	20	I	I/O Read. Active LO signal indicates a read operation.
tow	2 / /	19	I	I/O Write. Active HI signal indicates a write operation.
RESET /	2½ / <u> </u>	(23)	[]	Reset. Active HI.
Game Port				
Pin Name	PQFP	TQIP 🖳	I/D (	Description / /
A_1	50	48	I	Game Port A. Button #1.
A_2	49	47	I	Game Port A, Button #2.
A_X	54	52	I	Game Port A, X-Axis
A_Y	53	51	I	Game Port A, Y-Axis.
B_1	52	50	Ι	Game Port B, Button #1.
B_2	51	49	I	Game Port B, Button #2.
B_X	56	54	Ι	Game Port B, X-Axis.
B_Y	55	53	I	Game Port B, Y-Axis.

## MIDI Interface Signal (24 mA Drivers)

Pin Name	PQFP	TQFP	I/O	Description
MIDI_IN	66	64	I	RXD MIDI Input. This pin is typically connected to Pin 15 of the game port connector.
MIDI_OUT	67	65	О	TXD MIDI Output. This pin is typically connected to Pin 12 of the game port connector.

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### Muxed Serial Ports (8 mA Drivers)

Pin Name	PQFP	TQFP	I/O	Description
I <sup>2</sup> S(0)_BCLK*	3	1	I	I <sup>2</sup> S (0) Bit Clock.
I <sup>2</sup> S(0)_LRCLK*	2	100	I	I <sup>2</sup> S (0) Left/Right Clock.
$I^2S(0)_DATA^*$	1	99	I	I <sup>2</sup> S (0) Serial Data Input.
I <sup>2</sup> S(1)_BCLK*	82	80	I	I <sup>2</sup> S (1) Bit Clock.
I <sup>2</sup> S(1)_LRCLK*	83	81	I	I <sup>2</sup> S (1) Left/Right Clock.
$I^2S(1)_DATA^*$	81	79	I	I <sup>2</sup> S (1) Serial Data Input.
SPORT_SDI*	100	98	I	Serial Port Digital Serial Input.
SPORT_SCLK*	97	95	О	Serial Port Serial Clock.
SPORT_SDFS*	98	96	О	Serial Port Serial Data Frame Synchronization.
SPORT_SDO*	99	97	О	Serial Port Serial Data Output.

Miscellaneous Analog Pins

/Miscenaneous F	ynaipg Euis			
Pin Name	PQFP	TQFP	I/O	Description
V <sub>REF_X</sub>	\$6) <sub>[</sub>	7		Voltage Reference. Nominal 2.25 volt reference available for dc-coupling and level-shifting. $V_{REF\_X}$ should not be used to sink or source signal current. $V_{REF\_X}$ should be bypassed with 10 $\mu$ F and 0.1 $\mu$ F parallel capacitors.
$V_{\mathrm{REF}}$	35	93)		Voltage Reference Filter. Voltage reference filter point for external bypassing only. $N_{\rm RHF}$ should be bypassed with 10 HF and 0.1 HF parallel capacitors.
L_FILT	38	36	1	Left Channel Filter. Requires a 1.0 pF to analog ground for proper operation.
R_FILT	37	35	I	Right Channel Filter. Requires a 1.0 µF to analog ground for proper operation.
L_AAFILT	40	38	I	Left Channel Antialias Filter. This pin requires a 560 pt NPO capacitor to analog ground for proper operation.
R_AAFILT	39	37	I	Right Channel Antialias Filter. This pin requires a 560 pF NPO capacitor to analog ground for proper operation.

## **Crystal Pin**

Pin Name	PQFP	TQFP	I/O	Description
XTALO	64	62	О	33 MHz Crystal Output. If no Crystal is present leave XTALO unconnected.
XTALI	63	61	I	33 MHz Clock. When using a crystal as a clock source, the crystal should be connected between the XTALI and XTALO pins. Clock input may be driven into XTALI in place of a crystal. When using an external clock, $V_{\rm IH}$ must be 2.4 V rather than the $V_{\rm IH}$ of 2.0 V specified for all other digital inputs.

#### **External Logical Devices**

Pin Name	PQFP	TQFP	I/O	Description
LD_IRQ*	100	98	I	Logical Device IRQ.
LD_DACK*	99	97	О	Logical Device DACK.
$LD_DRQ^*$	98	96	I	Logical Device DRQ.
<u>LD_SEL</u> ∗	97	95	О	Logical Device Select.
MDM_SEL*	83	81	О	Modem Chip Set Select.
$\overline{\text{MDM}}_{\text{IRQ}}^*$	82	82	I	Modem Chip Set IRQ.
LD_SEL1*	69	67	О	Logical Device (1) Select.
PNPRST*	68	66	О	Plug and Play Reset.

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#### **Hardware Volume Pins**

Pin Name	PQFP	TQFP	I/O	Description
VOL_DN*	2, 99, 100	97, 98, 100	I	Master Volume Down. Modifies output level on pins L_OUT and R_OUT. When asserted LO, decreases Master Volume by 1.5 dB/sec. Must be asserted at least 25 ms to be recognized. When asserted simultaneously with VOL_UP, output is muted. Output level modification reflected in indirect register [41].
VOL_UP*	1, 98	96, 99	I	Master Volume Up. Modifies output level on pins L_OUT and R_OUT. When asserted LO, increases Master Volume by 1.5 dB/sec. Must be asserted at least 25 ms to be recognized. When asserted simultaneously with VOL_UP, output is muted. Output level modification reflected in indirect register [41].

#### **Control Pins**

Pin Name	PQFP	TQFP	I/O	Description
XCTLO*	68	66	0	External Control 0. The state of this pin (TTL HI or LO) is reflected in codec indexed register. This pin is an open drain driver.
PCI_KO*	)68)   <u></u>			Programmable Clock Output. This pin can be programmed to generate an output clock equal to $F_S$ , $8 \times F_S$ , $16 \times F_S$ , $32 \times F_S$ , $64 \times F_S$ , $128 \times F_S$ or $256 \times F_S$ .  MPLG decoders typically require a master clock of $256 \times F_S$ for audio
XCTL1*	69	67	0	synchronization.  External Control 1. The state of this pin (TTL HI or LO) is reflected in codec indexed register. Open drain, 8 mA active 0.5 mA pull-up resistor.
RING*	69	67	T	Ring Indicator. Used to accept the ring indicator flag from the DAA
Power Supp	lies		1	

## Dower Supplies

Power Supp	lies			
Pin Name	PQFP	TQFP	I/O	Description
$\overline{V_{CC}}$	33	31	I	Analog Supply Voltage (+5 V).
GNDA	34	32	I	Analog Ground.
$V_{DD}$	23, 62, 71, 89, 95	21, 60, 69, 87, 93	I	Digital Supply Voltage (+5 V).
GND	3*, 24, 65, 70, 84, 90, 96, 99*, 100*	1*, 22, 63, 68, 82, 88, 94, 97*, 98*	I	Digital Ground.

## **Optional EEPROM Pins**

Pin Name	PQFP	TQFP	I/O	Description
EE_CLK	58	56	О	EEPROM Clock. Open drain output, requires external pull-up.
EE_DATA	57	55	I/O	EEPROM Data. Open drain I/O, requires external pull-up.

<sup>\*</sup>The position of this pin location/function is dependent on the EEPROM data.

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#### **HOST INTERFACE**

The AD1816A contains all necessary ISA bus interface logic on chip. This logic includes address decoding for all onboard resources, control and signal interpretation, DMA selection and control logic, IRQ selection and control logic, and all interface configuration logic.

The AD1816A supports a Type "F" DMA request/grant architecture for transferring data with the ISA bus through the 8-bit interface. The AD1816A also supports DACK preemption. Programmed I/O (PIO) mode is also supported for control register accesses and for applications lacking DMA control. The AD1816A includes dual DMA count registers for full-duplex operation enabling simultaneous capture and playback on separate DMA channels.

#### Codec Functional Description

The AD1816A's full-duplex stereo codec supports business audio and multimedia applications. The codec includes stereo audio converters, complete on-chip filtering, MRC Level-2 and Level-3 compliant analog mixing programmable gain and attenuation, variable sample rate converters, extensive digital mixing and EIFOs buffering the Illug and Play ISA bus interface.

Analog Inputs

The codec contains a stereo pair of \_\_\_\_\_\_ analog to digital converters (ADC). Inputs to the ADC can be selected from the following analog signals: mono (PHONE\_IN), mono microphone (MIC), stereo line (LINE), external stereo synthesizer (SYNTH), stereo CD ROM (CD), stereo audio from a video source (VID) and post-mixed stereo or mono line output (OUT).

Analog Mixing

PHONE\_IN, MIC, LINE, SYNTH, CD and VID can be mixed in the analog domain with the stereo line OUT from the  $\Sigma\Delta$  digital-to-analog converters (DAC). Each channel of the stereo analog inputs can be independently gained or attenuated from +12 dB to -34.5 dB in 1.5 dB steps, except for PHONE\_IN, which has a range of 0 dB to -45 dB steps. The summing path for the mono inputs (MIC, and PHONE\_IN to line OUT) duplicates mono channel data on both the left and right line OUT, which can also be gained or attenuated from +12 dB to -34.5 dB in 1.5 dB steps for MIC, and +0 dB to -45.0 dB in 3 dB steps for PHONE\_IN. The left and right mono summing signals are always identical being gained or attenuated equally.

#### Analog-to-Digital Datapath

The selector sends left and right channel information to the programmable gain amplifier (PGA). The PGA following the selector allows independent gain for each channel entering the ADC from 0 dB to 22.5 dB in 1.5 dB steps.

For supporting time correlated I/O echo cancellation, the ADC is capable of sampling microphone data on the left channel and the mono summation of left and right OUT on the right channel.

The codec can operate in either a global stereo mode or a global mono mode with left channel inputs appearing at both channels of the 16-bit  $\Sigma\Delta$  converters. Data can be sampled at the programmed sampling frequency (from 4 kHz to 55.2 kHz with 1 Hz resolution).

#### Digital Mixing and Sample Rates

The audio ADC sample rate and the audio DAC sample rates are completely independent. The AD1816A includes a variable sample rate converter that lets the codec instantaneously change and process sample rates from 4 kHz to 55.2 kHz with a resolution of 1 Hz. The in-band integrated noise and distortion artifacts introduced by rate conversions are below –90 dB.

Up to four channels of digital data can be summed together and presented to the stereo DAC for conversion. Each digital channel pair can contain information encoded at a different sample rate. For example, 8 kHz .wav data received from the ISA interface, 48 kHz MPEG audio data received from I<sup>2</sup>S(0), digital 44.1 kHz CD data received from I<sup>2</sup>S(1) and internally generated 22.05 kHz music data may be summed together and converted by the DACs.

#### Digital-to-Analog Datapath

The internally generated music synthesizer data, PCM data received from the ISA interface, data received from the I $^2$ S(0) port and data received from the I $^2$ S(1) port, and the DSP serial port passes through an attenuation mute stage. The attenuator allows independent control over each digital channel, which can be attenuated from 0 dB to -94.5 dB in 1.5 dB steps before being summed together and passed to the DAC, or the channel may be muted entirely.

#### Analog Outputs and Phat Stereo

The analog output of the DAC can be summed with any of the analog input signals. The summed analog signal enters the Master Volume stage where each channel L\_OUT, R\_OUT and PHONE\_OUT may be attenuated from 0 dB to -46.5 dB in 1/.5 dB steps or muted.

Analog Outputs and Phat Stereo

The AD1816A includes ADI's proprietary Phat Steree 3D phase enhancement technology, which creates an increased sense of spiciousness using two speakers. Our unique patented feedback technology enables superior control over the width and depth of the acoustic signals arriving at the human ear. The AD1816A employs an electrical model of the speaker-to-ear path allowing precise control over a signal's phase at the ear. The Phat Stereo circuitry expands apparent sound images beyond the angle of the speakers by exploiting phase information in the audio signal and creating a more immersive listening experience.

#### Digital Data Types

The codec can process 16-bit twos complement PCM linear digital data, 8-bit unsigned magnitude PCM linear data and 8-bit  $\mu$ -law or A-law companded digital data as specified in the control registers. The AD1816A also supports ADPCM encoded in the Creative SoundBlaster ADPCM formats.

#### **Host-Based Echo Cancellation Support**

The AD1816A supports time correlated I/O data format by presenting MIC data on the left channel of the ADC and the mono summation of left and right OUT on the right channel. The ADC sample rates are independent of the DAC sample rate allowing the AD1816A to support ADC time correlated I/O data at 8 kHz and DAC data at any other sample rate in the range of 4 kHz to 55.2 kHz simultaneously.

#### **Telephony Support**

The AD1816A contains a PHONE\_IN input and a PHONE\_OUT output. These pins are supplied so the AD1816A may be connected to a modem chip set, a telephone handset or down-line phone.

#### **WSS and SoundBlaster Compatibility**

Windows Sound System software audio compatibility is built into the AD1816A.

SoundBlaster emulation is provided through the SoundBlaster register set and the internal music synthesizer. SoundBlaster Proversion 3.02 functions are supported, including record and Creative SoundBlaster ADPCM.

Virtually all applications developed for SoundBlaster, Windows Sound System, AdLib and MIDI MPU-401 platforms run on the AD1816A SoundPort Controller. Follow the same development process for the controller as you would for these other devices.

As the AD1816A contains SoundBlaster (compatible) and Windows Sound System logical devices. You may find the following related development kits useful when developing AD1816A applications.

Developer Kit for SoundBlaster Series, 2nd ed. © 1993, Creative Labs, Inc., 1901 McCarthy Blvd., Milpitas, CA 95035 Microsoft Windows Sound System Driver Development Kit (CD), Version 2.0, © 1993, Microsoft Corp., One Microsoft Way, Redmond, WA 98052

The following reference texts can serve as additional sources of information on developing applications that run on the AD1816A.

S. De Furia & A. Scacciaferro, The MIDI Implementation Book, (\$\psi\$ 1986, Third Earth, Pompton Lake)

C. Petzold, Programming Windows: the Microsoft guide to writing applications for Windows 3.1, 3rd. d., (\$19<del>92,</del> Microsoft

Press, Redmond K. Pohlmann, Principles of Digital Audio, (©

Indianapolis)
A. Stolz, *The SoundBlaster Book*, (© 1993, Abacaux, Gr

A. Stolz, *The SoundBlaster Book*, (© 1993, Abacaux, G) Rapids)

J. Strawn, Digital Audio Engineering, An Anthology, (© 1985, Kaufmann, Los Altos)

Yamamoto, *MIDI Guidebook*, 4th. ed., (© 1987, 1989, Roland Corp.)

#### **Multimedia PC Capabilities**

The AD1816A is MPC-2 and MPC-3 compliant. This compliance is achieved through the AD1816A's flexible mixer and the embedded chip resources.

#### Music Synthesis

The AD1816A includes an embedded music synthesizer that emulates industry standard OPL3 FM synthesizer chips and delivers 20 voice polyphony. The internal synthesizer generates digital music data at 22.05 kHz and is summed into the DACs digital data stream prior to conversion. To sum synthesizer data with the ADC output, the ADC must be programmed for a 22.05 kHz sample rate.



The synthesizer is a hardware implementation of Eusynth-1+ code that was developed by Euphonics, a research and development company that specializes in audio processing and electronic music synthesis.

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#### Wavetable MIDI Inputs

The AD1816A has a dedicated analog input for receiving an analog wavetable synthesizer output. Alternatively, a wavetable synthesizer's I²S formatted digital output can be directly connected to one of the AD1816A's I²S serial ports. Digital wavetable data from the AD1816A's I²S port may be summed with other digital data streams being handled by the AD1816A and then sent to the 16-bit  $\Sigma\Delta$  DAC.

#### **MIDI**

The primary interface for communicating MIDI data to and from the host PC is the compatible MPU-401 interface that operates only in UART mode. The MPU-401 interface has two built-in FIFOs: a 64-byte receive FIFO and a 16-byte transmit FIFO.

#### Game Port

An IBM-compatible game port interface is provided on chip. The game port supports up to two joysticks via a 15-pin D-sub connector. Joystick registers supporting the Microsoft Direct Input standard are included as part of the codec register map. The AD1816A may be programmed to automatically sample the game port and save the value in the Joystick Position Data Register. When enabled, this feature saves up to 10% CPU MIPS by off-loading the host from constantly polling the joystick port.

#### **Volume Control**

The registers that control the Master Volume output stage are accessible through the ISA Bus. Master Volume output can also be controlled through a 2-pin hardware interface. One pin is used to increase the gain, the other pin attenuates the output and both pins together entirely mute the output. Once muted, any further activity on these pins will unmute the AD1816A's output.

Plug and Play Configuration

The AD1816A is fully Plug and Play configurable. For mother-board applications, the built-in Plug and Play protocol can be disabled with a software key providing a back door for the BIOS to configure the AD1816A's logical devices. For information on the Plug and Play mode configuration process, see the Plug and Play ISA Specification Version 1.0a (May 3, 1994) All the AD1816A's logical devices comply with Plug and Play resource definitions described in the specification

The AD1816A may alternatively be configured using an optional Plug and Play Resource ROM. When the EEPROM is present, some additional AD1816A muxed-pin features become available. For example, pins that control an external modem logical device are muxed with the DSP serial port. Some of these pin option combinations are mutually exclusive (see Appendix A for more information).

#### REFERENCES

The AD1816A also complies with the following related specifications; they can be used as an additional reference to AD1816A operations beyond the material in this data sheet.

Plug and Play ISA Specification, Version 1.0a, © 1993, 1994, Intel Corp. & Microsoft Corp., One Microsoft Way, Redmond, WA 98052

*Multimedia PC Level 2 Specification,* © 1993, Multimedia PC Marketing Council, 1730 M St. NW, Suite 707, Washington, DC 20036

MIDI 1.0 Detailed Specification & Standard MIDI Files 1.0, © 1994, MIDI Manufacturers Association, PO Box 3173 La Habra, CA 90632-3173

Recommendation G.711-Pulse Code Modulation (PCM) Of Voice Frequencies (μ-Law & A-Law Companding), The International Telegraph and Telephone Consultative Committee IX Plenary Assembly Blue Book, Volume III - Fascicle III.4, General Aspects Of Digital Transmission Systems; Terminal Equipment's, Recommendations G.700 - G.795, (Geneva, 1988), ISBN 92-61-03341-5

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#### SERIAL INTERFACES

#### I2S Serial Ports

The two I<sup>2</sup>S serial ports on the AD1816A accept serial data in the following formats: Right-Justified, I<sup>2</sup>S-Justified and Left-Justified.

Figure 9 shows the right-justified mode. LRCLK is HI for the left channel and LO for the right channel. Data is valid on the rising edge of the BCLK. The MSB is delayed 16-bit clock periods from an LRCLK transition, so that when there are 64 BCLK periods per LRCLK period, the LSB of the data will be right-justified to the next LRCLK transition.

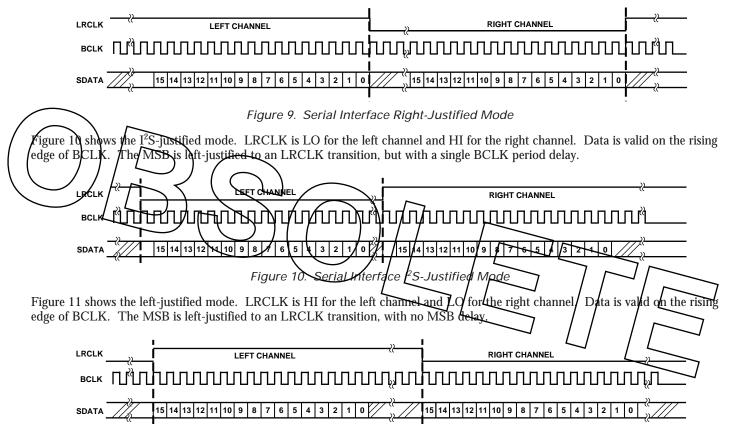


Figure 11. Serial Interface Left-Justified Mode

#### **Bidirectional DSP Serial Interface**

The AD1816A SoundPort Controller transmits and receives both data and control/status information through its DSP serial interface port (SPORT). The AD1816A is always the bus master and supplies the frame sync and the serial clock. The AD1816A has four pins assigned to the SPORT: SDI, SDO, SDFS and SCLK. The SPORT has two operating modes: monitor and intercept. The SPORT always monitors the various data streams being processed by the AD1816A. In intercept mode, any of the digital data streams can be manipulated by the DSP before reaching the final ADC or DAC stages.

The SDI and SDO pins handle the serial data input and output of the AD1816A. Communication in and out of the AD1816A requires that bits of data be transmitted after a rising edge of SCLK and sampled on the falling edge of SCLK. The SCLK frequency is always 11 MHz (or 1/3 or XTALI).

DSP Serial Port Interface time slots are mapped as shown in Table I.

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**Table I. DSP Port Time Slot Map** 

Time Slot	SDI Pin	SDO Pin
0	Control Word Input	Status Word Output
1	Control Register Data Input	Control Register Data Output
2	* SS/SB ADC Right Input (to ISA)	SS/SB ADC Right Output (from Codec)
3	* SS/SB ADC Left Input (to ISA)	SS/SB ADC Left Output (from Codec)
4	* SS/SB DAC Right Input (to Codec)	SS/SB DAC Right Output (from ISA)
5	* SS/SB DAC Left Input (to Codec)	SS/SB DAC Left Output (from ISA)
6	* FM DAC Right Input (to Codec)	FM DAC Right Output (from FM Synth Block)
7	* FM DAC Left Input (to Codec)	FM DAC Left Output (from FM Synth Block)
8	* I <sup>2</sup> S (1) DAC Right Input (to Codec)	I <sup>2</sup> S (1) DAC Right Output (from I <sup>2</sup> S Port (1))
9	* I <sup>2</sup> S (1) DAC Left Input (to Codec)	I <sup>2</sup> S (1) DAC Left Output (from I <sup>2</sup> S Port (1))
10	* I <sup>2</sup> S (0) DAC Right Input (to Codec)	I <sup>2</sup> S (0) DAC Right Output (from I <sup>2</sup> S Port (0))
<b>A</b> 1	* I <sup>2</sup> S (0) DAC Left Input (to Codec)	I <sup>2</sup> S (0) DAC Left Output (from I <sup>2</sup> S Port (0))

This data is ignored by the AD1816A unless the channel pair is in intercept mode (see below).

SB = SoundBlaster Mode

At start-up (after pin reset), there are exactly 12 time slots per frame. The frame rate will be 57,291 and 2/3 Hz (11 MHz sclk/ [16 bits × 12 slots]). Interfacing with an Analog Devices 21xx family DSP can be achieved by putting the ADSP-21xx in 24 slot per frame mode, where the first 12 and second 12 slots in the ADSP-21xx frame are identical.

The frame rate can be changed from its default by a write to the DFS(2:0) bits/in register 33. Rate choices are: Maximum (57,291 and 2/3 Hz default), SS capture rate, SS playback rate, FM rate, I<sup>2</sup>8 Port (1) rate, or I<sup>2</sup>S Port (6) rate. When the frame rate is less than 57,261 and 2/3 Hz, extra SCLK periods are added to fill up the time. The number of SCLK periods added will vary somewhat from frame to frame.

To control the sample data flow of each channel through the DSP Port, valid input, valid output and request bits are located in the control and status words. If the specified channel sample rate is equal to the frame rate these bits may be ignored since they will always be set to "1."

By default, the DSP serial port allows only codec sample data I/O to be monitored. Intercept modes must be enabled to make substitutions in sample data flow to and from the codec. There are five bits in SS register 33, which enable intercept mode for SS capture, SS playback, FM playback, I<sup>2</sup>S Port (1) playback and I<sup>2</sup>S Port (0) playback.

#### **Control Word Input (Slot 0 SDI)**

15	14	13	12	11	10	9	8
FCLR	RES	RES	SSCVI	SSPVI	FMVI	IS1VI	IS0VI
7	6	5	4	3	2	1	0
ALIVE	R/W			IA[5:0]			

IA [5:0] Indirect Register Address. Sound System Indirect Register Address defines the address of indirect registers shown

in Table VI.

R/W Read/Write request. Either a read from or a write to an SS indirect register occurs every frame. Setting this bit ini-

tiates an SS indirect register read while clearing this bit initiates an SS indirect register write.

ALIVE DSP port alive bit. When set, this bit indicates to the power-down timer that the DSP port is active. When cleared,

this bit indicates that the DSP port is inactive.

ISOVI I<sup>2</sup>S Port 0 Substitution Data Input Valid Flag. This bit is ignored if: (1) Intercept mode is not enabled for the I<sup>2</sup>S port 0 channel pair, or (2) The AD1816A did not request data from the I<sup>2</sup>S port 0 channel pair in the previous frame. Otherwise, setting this bit indicates that slots 10 and 11 contain valid right and left I<sup>2</sup>S Port 0 substitution

data. When this bit is cleared, data in slots 10 and 11 is ignored.

IS1VI I<sup>2</sup>S Port 1 Substitution Data Input Valid Flag. This bit is ignored if: (1) Intercept mode is not enabled for I<sup>2</sup>S port

1 channel pair or (2) The AD1816A did not request data from the I<sup>2</sup>S port channel pair in the previous frame. Otherwise, setting this bit indicates that Slots 8 and 9 contain valid right and left I<sup>2</sup>S Port 1 substitution data.

When this bit is cleared, data in slots 8 and 9 is ignored.

FMVI FM Synthesis Substitution Data Input Valid Flag. This bit is ignored if: (1) Intercept mode is not enabled for the

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FM synthesis channel pair or (2) The AD1816A did not request data from the FM synthesis channel pair in the previous frame (see the FMRQ Bit 9 in the status word output). Otherwise, setting this bit to 1 indicates that slots 6 and 7 contain valid right and left FM synthesis channel substitution data. When this bit is reset to 0, data in slots

6 and 7 is ignored.

SSPVI

SS/SB Playback Substitution Data Input Valid Flag. This bit is ignored if: (1) Intercept mode is not enabled for SS/SB playback or (2) The AD1816A did not request data for SS/SB playback in the previous frame (see the SSPRQ bit in the Status Word Output). Otherwise, setting this bit indicates that Slots 4 and 5 contain valid right and left SS/SB playback substitution data. If in "capture rate equal to playback rate" mode, setting this bit also indicates that valid capture substitution data is being sent to the AD1816A. If not in modem mode, right and left channel capture substitution data is accepted in Slots 2 and 3 respectively. If in modem mode, only mono capture substitution data is accepted in slots 2 and 3. When this bit is cleared, data in all slots controlled by this bit, as defined above, is ignored.

**SSCVI** 

SS/SB Capture Substitution Data Input Valid Flag. This bit is ignored if: (1) Intercept mode is not enabled for SS/SB capture or (2) The AD1816A did not request data for SS/SB capture in the previous frame (see the SSCRQ bit in the Status Word Output). Otherwise, setting this bit indicates that valid SS/SB capture substitution data is being sent to the AD1816A. If not in modem mode, or DSP port or ISA bus based, right and left channel capture data is accepted in Slots 2 and 3 respectively. If in modem mode, only mono capture substitution data is accepted in Slot 3, because Slot 2, which is mapped to the right capture channel, is being used for modem. This mono data will, however, be sent to both left and right ISA SS/SB capture channels. When this bit is cleared, data in Slots 3 and 2 is ignored.

RES FCLR

Reserved: To ensure future compatibility write "0" to all reserved bits.

DSP Port Clear Status Flag. When this bit is set, (write 1), the PNPR and PDN flag bits in the status word (Bits 15 and 14 of slots 0 SDO) are cleared. When this bit is cleared, (writing a 0), it has no effect on PNPR and PDN and preserves them in the previous states

Status Word Output (Slot 0/SDO)

15		$\int_{3}$	12	) / <sub>1</sub> /l	0	8
PDN	PNPR	RES \	SSCX/O /	/SSIPVO	FMVØ ISIVQ	TS0VQ
7	6	5	4	/ /3	$\int 2$ 1	7 ~ 1
MB1	MB0	RES	SSCRQ	/ SSPRQ	FMRQ ISIRQ	/ISORQ/

IS0RQ

I<sup>2</sup>S Port (0) Input Request Flag. This bit is set if intercept mode is enabled for I<sup>2</sup>S Port (0) and its four-word stereo input buffer is not full.

IS1RQ

I<sup>2</sup>S Port (1) Input Request Flag. This bit is set if intercept mode is enabled for I<sup>2</sup>S Port (1) and its four-word stereo input buffer is not full.

**FMRQ** 

FM Synthesis Input Request Flag. This bit is set if intercept mode is enabled for FM synthesis and its four-word stereo input buffer is not full.

SSPRQ

SS/SB Playback Input Request Flag. This bit is set if intercept mode is enabled for SS/SB playback and its four-word stereo input buffer is not full.

SSCRQ

SS/SB Capture Input Request Flag. This bit is set if intercept mode is enabled for SS/SB capture and its four-word stereo input buffer is not full.

MB0

Mailbox 0 Status Flag. This bit is set if the most recent action to SS indirect register 42 (DSP port Mail Box 1) was a write, and is cleared if the most recent action was a read. The status of this bit is also reflected in SS indirect register 33. It may be used as a handshake bit to facilitate communication between a DSP on the DSP port and a host CPU on the ISA bus.

MB1

Mailbox 1 Status Flag. This bit is set if the most recent action to SS indirect register 43 (DSP port Mail Box 1) was a write and is cleared if the most recent action was a read. The status of this bit is also reflected in SS indirect register 33. It may be used as a handshake bit to facilitate communication between a DSP on the DSP port and a host CPU on the ISA bus.

**ISOVO** 

I<sup>2</sup>S Port 0 Valid Out. This bit is set if Slots 10 and 11 contain valid right and left I<sup>2</sup>S Port 0 data.

IS1V1

I<sup>2</sup>S Port 1 Valid Out. This bit is set if Slots 8 and 9 contain valid right and left I<sup>2</sup>S Port 1 data.

FMVO

FM Synthesis Valid Out. This bit is set if Slots 6 and 7 contain valid left and right FM synthesis data.

**SSPVO** 

SS/SB Playback Valid Out. This bit is set if Slots 4 and 5 contain valid right and left SS/SB playback data.

SSCVO

SS/SB Capture Valid Out. This bit is set if valid SS/SB capture data is being transmitted. If not in a modem mode, Slots 2 and 3 will contain valid right and left SS/SB capture data. If in modem mode, only Slot 3 will contain valid left SS/SB capture data as Slot 2 and the ADC right channel are used by the modem.

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**PNPR** 

Plug and Play Reset flag. This bit is set by an AD1816A reset (RESETB pin asserted LOW) or by a Plug and Play reset command. This bit is cleared by the assertion of the FCLR bit in the control word. While this bit is set, all attempts to write an SS indirect register via the DSP port will be ignored and fail. This is to ensure that Plug and Play resets are immediately applied to the application running on the DSP, without requiring them to continuously poll the Plug and Play reset status bit. During the frame in which this bit is cleared (by asserting FCLR), an attempt to write an SS indirect register will succeed. If the FCLR bit is continuously asserted, writes to indirect registers via the DSP port will always be enabled. A Plug and Play reset command will set this PNPR bit HIGH during at least one frame.

**PDN** 

Power-Down flag. This bit is set by an AD1816A reset (RESETB pin asserted LOW), or by an AD1816A power-down. Before an AD1816A power-down sequence shuts down the DSP port, at least one frame will be sent with this bit set. This bit can be cleared by the assertion of the FCLR (DSP port status clear) bit in the control word, providing the AD1816A is no longer in power-down.

The SDFS pin is used for the serial interface frame synchronization. New frames are marked by a one SCLK duration HI pulse, driven out on SDFS, one serial clock period before the frame begins. Upon initializing, there are exactly 12 time slots per frame and 16 bits per time slot. The frame rate is 57,291 and 2/3 Hz (11 MHz SCLK /(16 bits × 12 slots)). The frame rate can also be changed from the default value by reprogramming the rate in registers. The frame rate can run at the default rate or be programmed to match the modem sample rate. ADC capture rate, DAC playback rate, music sample rate, I<sup>2</sup>S(1) sample rate or I<sup>2</sup>S(0) sample rate. When the frame rate is not equivalent to the sample rate, Valid Out, Request In and Valid In bits are used to control the sample data flow. When the frame rate is equivalent to the sample rate. Valid and Request bits can be ignored.

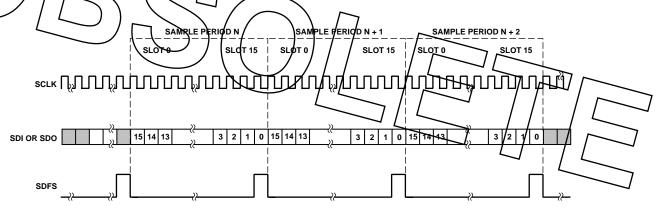


Figure 12. DSP Serial Interface (Default Frame Rate)

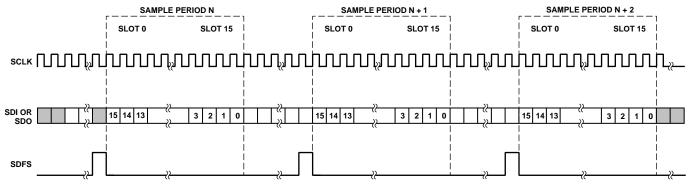
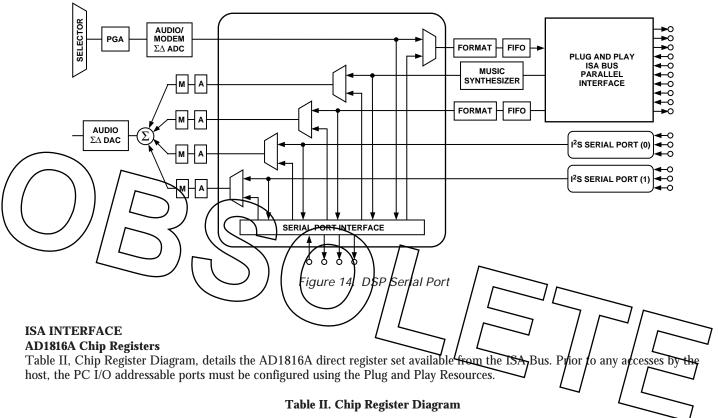


Figure 13. DSP Serial Interface (User Programmed Frame Rate)

Figure 14 illustrates the flexibility of the DSP Serial Port interface. This port can monitor or intercept any of the digital streams managed by the AD1816A. Any ADC or DAC data stream can be intercepted by the port, shipped to an external DSP or ASIC manipulated, and returned to any DAC summing path or to the ADC.



Register Type-Register Name	Register PC I/O Address
Plug and Play ADDRESS	0x279
WRITE_DATA	0xA79
READ_DATA	Relocatable in Range 0x203 – 0x3FF
TICHID_DITTI	relocatable in range 0x200 0x011
Sound System Codec	
CODEC REGISTERS	0x(SS Base+0 – SS Base+15)
	Relocatable in Range 0x100 – 0x3FF
	See Table V
SoundBlaster Pro	
Music0: Address (w), Status (r)	(SB Base) Relocatable in Range 0x100 – 0x3F0
Music0: Data (w)	(SB Base+1)
Music1: Address (w)	(SB Base+2)
Music1: Data (w)	(SB Base+3)
Mixer Address (w)	(SB Base+4)
Mixer Data (w)	(SB Base+5)
Reset (w)	(SB Base+6 or 7)
Music0: Address (w)	(SB Base+8)
Music0: Data (w)	(SB Base+9)
Input Data (r)	(SB Base+A or +B)
Status (r), Output Data (w)	(SB Base+C or +D)
Status (r)	(SB Base+E or +F)

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Register Type-Register Name	Register PC I/O Address
AdLib	
Music0: Address (w), Status (r)	(AdLib Base) Relocatable in Range 0x100 – 0x3F8
Music0: Data (w)	(AdLib Base+1)
Music1: Address (w)	(AdLib Base+2)
Music1: Data (w)	(AdLib Base+3)
MIDI MPU-401	
MIDI Data (r/w)	(MIDI Base) Relocatable in Range 0x100 – 0x3FE
MIDI Status (r), Command (w)	(MIDI Base+1)
Game Port Game Port I/O	(Game Base +0 to Game Base +7) Relocatable in Range
0x100 - 0x3F8	(Suine Pase 10 to Suine Pase 17) reciocatable in realige

AD 1816A Plug and Play Device Configuration Registers

The AD1816A may be configured according to the Intel/Microsoft Plug and Play Specification using the internal ROM. Alternatively, the PnII configuration sequence may be hypassed using the "Alternate Key Sequence" described in Appendix A.

The operating system configures the AIJ1816A Plug and Play Logical Devices after system boot. There are no "boot-devices" among the Plug and Play Logical Devices in the AD1816A. Non Plug and Play BIOS systems configure the AD1816A's Logical Devices after boot using drivers. Depending on BIOS implementations, Plug and Play BIOS systems may configure the AD1816A's Logical Devices before POST or after Boot. See the Plug and Play ISA Specification Version 1.0a for more information on configuration control. To complete this configuration, the system reads resource data from the AD1816A's on-chip resource ROM or optional EEPROM and from any other Plug and Play cards in the system, and then arbitrates the configuration of system resources with a heuristic algorithm. The algorithm maximizes the number of active devices and the acceptability of their configurations.

The system considers all Plug and Play logical device resource data at the same time and makes a conflict-free assignment of resources to the devices. If the system cannot assign a conflict-free resource to a device, the system does not configure or activate the device. All configured devices are activated.

The system's Plug and Play support selects all necessary drivers, starts them and maintains a list of system resources allocated to each logical device. As an option, system resources can be reassigned at runtime with a Plug and Play Resource Manager. The custom setup created using the manager can be saved and used automatically on subsequent system boots.

Plug and Play Device IDs (embedded in the logical device's resource data) provide the system with the information required to find and load the correct device drivers. One custom driver, the AD1816A Sound System driver from Analog Devices, is required for correct operation. In the other cases (MIDI, Game Port), the system can use generic drivers. Table III lists the AD1816A's Logical Devices and compatible Plug and Play device drivers.

Table III. Logical Devices and Compatible Plug and Play Device Drivers

<b>Logical Device Number</b>	Emulated Device	Compatible (Device ID)	Device ID
0	Sound System	—	ADS7180
1	MIDI MPU401 Compatible	PNPB006	ADS7181
2	Game/Joystick Port	PNPB02F	ADS7182

The configuration process for the logical devices on the AD1816A is described in the *Plug and Play ISA Specification Version 1.0a* (*May 5, 1994*). The specification describes how to transfer the logical devices from their start-up *Wait For Key* state to the *Config* state and how to assign I/O ranges, interrupt channels and DMA channels. See Appendix A for an example setup program and specific Plug and Play resource data.

Table IV describes in detail the I/O Port Address Descriptors, DMA Channels, Interrupts for the functions required for the AD1816A Logical Device groups.

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**Table IV. Internal Logical Device Configuration** 

LDN	PnP Function	Description
0	I/O Port Address Descriptor (0x60-0x61)	The SoundBlaster Pro address range is from 0x100 to 0x3F0. The typical address is 0x220. The range is 16 bytes long and must be aligned to a 16 byte memory boundary.
0	I/O Port Address Descriptor (0x62-0x63)	The AdLib address range is from 0x100 to 0x3F8. The typical address is 0x388. The range is 4 bytes long and must be aligned to an 8 byte memory boundary.
0	I/O Port Address Descriptor (0x64-0x65)	The Codec address range is from 0x100 to 0x3F8. The range is 16 bytes long and must be aligned to a 16 byte memory boundary.
0	Interrupt Request Level Select (0x70-0x71)	This IRQ is shared between the SB Pro device and the Codec. These devices require one of the following IRQ channels: 5, 7, 9, 11, 12 or 15. Typically, the IRQ is set to 5 or 7 for this device.
	DMA Playback Channel Select (0x74)	This 8-bit channel is shared between the SB Pro device and the Codec for playback. These devices require one of the following DMA channels: 0, 1, 3. Typically, DMA channel 1 is set.
	DMA Capture Channel Select (0x75)	This the DMA channel used for capturing Codec data. The Codec operates in single channel mode if a separate DMA channel for capture and playback is not assigned. The following DMA channels may be programmed, 0, 1, 3. DMA Channel 4 indicates single channel mode.
1	I/O Port Address Descriptor (0x60-0x61)	The MPU-401 compatible device address range is 0x100 to 0x3FE.  Typical configurations use 0x330. The range is 2 bytes long and must be aligned to a 2 byte memory boundary.
1	Interrupt Request Level Select (0x70-0x71)	The MIDI device requires one of the following IRQ channels: 5, 7, 9, 11, 12 or 15.
2	I/O Port Address Descriptor (0x60-0x61)	The Game Port address range is from 0x 100 to 0x3F8. The typical address is 0x200. The range is 8 bytes long and must be aligned to an 8 byte memory boundary.

NOTE

DMA channel 4 indicates single-channel mode.

#### **Sound System Direct Registers**

The AD1816A has a set of 16 programmable Sound System Direct Registers and 36 Indirect Registers. This section describes all the AD1816A registers and gives their address, name and initialization state/reset value. Following each register table is a list (in ascending order) of the full register name, its usage and its type: (RO) Read Only, (WO) Write Only, (STKY) Sticky, (RW) Read Write and Reserved (res). Table V is a map of the AD1816A direct registers.

**Table V. Sound System Direct Registers** 

Direct								
Address	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
SSBASE + 0	CRDY	VBL		-	INADR	[5:0]	•	
SSBASE + 1	PI	CI	TI	VI	DI	RI	GI	SI
SSBASE + 2				Indirect SS Dat	a [7:0]			
SSBASE + 3				Indirect SS Dat	ta [15:8]			
SSBASE + 4	Rl	ES	PUR	COR	ORR	[1:0]	(	ORL [1:0]
SSBASE + 5	PFH	PDR	PLR	PUL	CFH	CDR	CLR	CUL
SSBASE + 6				PIO Playback/C	Capture [7:0]	-	•	
SSBASE + 7				RESERV	ED			
SSBASE + 8	TRD	DAZ	PFM	Γ [1:0]	PC/L	PST	PIO	PEN
SSBASE + 9	RES		CFM	Γ [1:0]	CC/L	CST	CIO	CEN
SSBASE + 10				RESE	RVED			
SSBASE + 11				RESE	RVED			
SSBASE + 12				JOYSTICK DA	TA [7:0]			
SSBASE + 13	JRDY	JWRP	JSEL [1	:0]		JMSF	K [3:0]	·
SSBASE + 14				JAXIS	5 [7:0]	·		
SSBASE + 15				JAXIS	5 [15:8]	•		·

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#### [Base+0] Chip Status/Indirect Address CRDY VBI INADR[5:0 RESET = [0x00]INADR [5:0] (RW) Indirect Address for Sound System (SS). These bits are used to access the Indirect Registers shown in Table VIII. All registers data must be written in pairs, low byte followed by high byte, by loading the Indirect SS Data Registers, (Base +2) and (Base +3). **VBL** Volume Button Location. When using an EEPROM to configure the PnP state of the AD1816A, this bit determines whether PQFP Pins 1 and 2 (TQFP Pins 99 and 100) are used for VOL\_UP and VOL\_DN or I2SO\_DATA and I<sup>2</sup>S0\_LRCLK respectively. I2S0\_DATA and I2S0\_LRCLK VOL\_UP and VOL\_DN CRDY (RO) AD1816A Ready. The AD1816A asserts this bit when AD1816A can accept data. AD1816A not ready AD1816A ready Base+1] Inter PΙ RI GI SI RESET = [0x00]generated Interrupt (RO)undBlaste SI No interrupt SoundBlaster GI (RW) Game Interrupt (Sticky, Write Clear No interrupt An interrupt is pending due to Digital Game/Port data ready RI (RW) Ring Interrupt (Sticky, Write "0" to Clear). No interrupt An interrupt is pending due to a Hardware Ring pin being asserted DI (RW) DSP Interrupt (Sticky, Write "0" to Clear). No interrupt An interrupt is pending due to a write to the DIT bit in indirect register [33] bit <13> (RW) Volume Interrupt (Sticky, Write "0" to Clear). VI No interrupt An interrupt is pending due to Hardware Volume Button being pressed ΤI (RW) Timer Interrupt. This bit indicates there is an interrupt pending from the timer count registers. (Sticky, Write "0" to Clear). No interrupt Interrupt is pending from the timer count register (RW) Capture Interrupt. This bit indicates that there is an interrupt pending from the capture DMA count register. CI (Sticky, Write "0" to Clear). No interrupt Interrupt is pending from the capture DMA count register (RW) Playback Interrupt. This bit indicates that there is an interrupt pending from the playback DMA count PΙ register. (Sticky, Write "0" to Clear). No interrupt Interrupt is pending from the playback DMA count register **Indirect SS Data Low Byte** [Base+2] Indirect SS Data [7:0] RESET = [0xXX][Base+3] **Indirect SS Data High Byte** Indirect SS Data [15:8] RESET = [0xXX]Indirect SS Indirect Sound System Data. Data in this register is written to the Sound System Indirect Register specified by the Data [15:0] address contained in INDAR [5:0], Sound System Direct Register [Base +0]. Data is written when the Indirect SS

Data High Byte value is loaded.

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#### [Base+4] PIO Debug

7	6	5	4	3	2	1	0	_	
RES		PUR	COR	ORR[	1:0]	ORL	[1:0]	RESET	= [0x00]

All bits in this register are sticky until any write that clears all bits to 0.

ORL/ORR (RO) [1:0]

Overrange Left/Right detect. These bits record the largest output magnitude on the ADC right and left channels and are cleared to 00 after any write to this register. The peak amplitude as recorded by these bits is "sticky," i.e., the largest output magnitude recorded by these bits will persist until these bits are explicitly cleared. They are also cleared by powering down the chip.

	ORL/ORR	Over/Under Range Detection	
	00	Less than -1 dB Underrange	
	01	Between –1 dB and 0 dB Underrange	
	10	Between 0 dB and 1 dB Overrange	
	11	Greater than 1 dB Overrange	
capture JIFO fills codec clears this bit Playback Under Ru ter the playback FII	When COR is immediately n. The codec FO enupties. It set, the playba	ets (1) this bit when capture data is not read as set, the FIFO is full and the codec discardiffer a 4 byte capture sample is read.  Sets (1) this bit when playback data is not fine codec clears (0) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) this bit immediately and ck channel has "run out" of daya and either the codec clears (1) the codec c	rds any new data generated. The t written within one sample period af- after a 4 byte playback sample is writ-
6 5 H PDR PL		$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	L RESET = [0x00]
or lower byte of the  1 Lower byte re  1 Upper byte re	channel. eady eady or any 8-	This bit indicates whether the PIO capture bit mode bit indicates whether the PIO capture da	
or the right channel 0 Right channel 1 Left channel	l ADC. I	on marches mount in 110 capears an	a material is for the fore channel 122 c
used only when direction of the original original original or the original	ect programm Do not rerea	pture Data register contains data ready for red I/O data transfers are desired (FIFO h d the information	

COR

[Base+5]

**CUL** 

**CLR** 

Left channe 1

**PIO Status** 

PFH

(RO)

(RO)

**CDR** Capture Data Rea (RO) used only when o

ADC data is fresh. Ready for next host data read

Capture FIFO Half Full. (FIFO has at least 32 bytes before full.) **CFH** (RO)

**PUL** Playback Upper/Lower Sample. This bit indicates whether the PIO playback data needed is for the upper or (RO) lower byte of the channel.

> Lower byte needed 0

Upper byte needed or any 8-bit mode

**PLR** (RO) Playback Left/Right Sample. This bit indicates whether the PIO playback data needed is or the left channel DAC or the right channel DAC.

Right channel needed

1 Left channel or mono

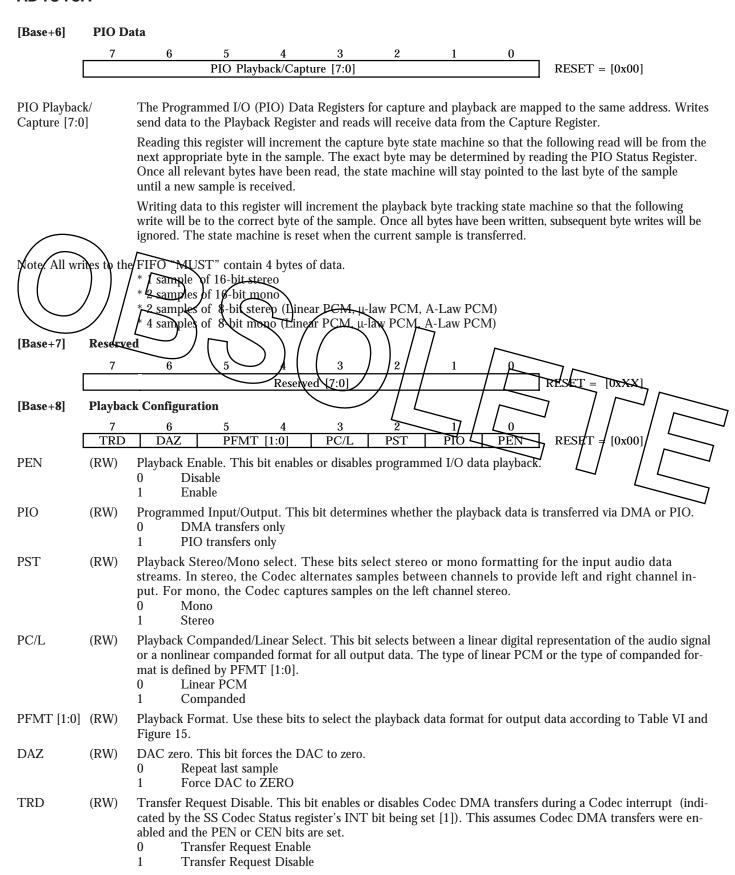
**PDR** (RO) Playback Data Ready. The PIO Playback data register is ready for more data. This bit should only be used when direct programmed I/O data transfers are desired (FIFO can take at least 4 bytes).

> 0 DAC data is still valid. Do not overwrite

DAC data is stale. Ready for next host data write value

**PFH** Playback FIFO Half Empty. FIFO can take at least 32 bytes, eight groups of 4 bytes. (RO)

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After setting format bits, sample data into the AD1816A must be ordered according to Figure 15, Table VI.

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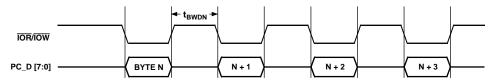


Figure 15. Codec Transfers

**Table VI. Codec Transfers** 

ST	FMT1 FMT0 C/L	Format	Byte 3 MSB LSB	Byte 2 MSB LSB	Byte 1 MSB LSB	Byte 0 MSB LSB
0	000	Mono Linear, 8-Bit Unsigned	Sample 3 8 Bits Left Channel	Sample 2 8 Bits Left Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Left Channel
ĺ	000	Stereo Linear, 8-Bit Unsigned	Sample 1 8 Bits Right Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Right Channel	Sample 0 8 Bits Left Channel
	061	Mono µ-Law, 8-Bit Companded	Sample 3 8 Bits Left Channel	Sample 2 87Bits Left Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Left Channel
1	001	Stereo μ-Law, 8-Bit Companded	Sample 1 8 Bits Right Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Right Charinel	Sample 0 8 Bits Left Channel
0	010	Mono Linear 16-Bit Little Endian	Sample 1 Upper 8 Bits Left Channel	Sample 1 Lower 8 Bits Left Channel	Sample 0/ Upper 8 Bits Deft Channel	Sample 0 Lower 8 Bits Left Channel
1	010	Stereo Linear 16-Bit Little Endian	Sample 0 Upper 8 Bits Right Channel	Sample 0 Lower 8 Bits Right Channel	Sample 0 / Upper 8 Bits Left Channel	Sample 0 Lower 8 Bits Left Channel
0	011	Mono A-Law, 8-Bit Companded	Sample 3 8 Bits Left Channel	Sample 2 8 Bits Left Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Left Channel
1	011	Stereo A-Law, 8-Bit Companded	Sample 1 8 Bits Right Channel	Sample 1 8 Bits Left Channel	Sample 0 8 Bits Right Channel	Sample 0 8 Bits Left Channel
0	100	Reserved				
1	100	Reserved				
0	101	Reserved				
1	101	Reserved				
0	110	Mono Linear, 16-Bit Big Endian	Sample 1 Lower 8 Bits Left Channel	Sample 1 Upper 8 Bits Left Channel	Sample 0 Lower 8 Bits Left Channel	Sample 0 Upper 8 Bits Left Channel
0	110	Stereo Linear, 16-Bit Big Endian	Sample 0 Lower 8 Bits Right Channel	Sample 0 Upper 8 Bits Left Channel	Sample 0 Lower 8 Bits Left Channel	Sample 0 Upper 8 Bits Left Channel
0	111	Reserved				
1	111	Reserved				

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#### [Base+9] **Capture Configuration** RES RESET = [0x00](RW) Capture Enable. This bit enables or disables data capture. CEN Disable Enable CIO Capture Programmed I/O. This bit determines whether the capture data is transferred via DMA or PIO. **DMA** PIO 1 (RW) Capture Stereo/Mono Select. This bit selects stereo or mono formatting for the input audio data streams. **CST** In stereo, the Codec alternates samples between channels to provide left and right channel input. For mono, the Codec captures samples on the left channel. Mono 1 Stereo Capture Companded/Linear Select. This bit selects between a linear digital representation of the audio siggalor a nonlinear, companded format for all output data. The type of linear PCM or the type of companded format)s defined by CFMT [1:0]. Linear PCM Companded select the format for capture data according to the following Table VI and CFMT [1:0] Figure. Reserved [Base+10] RESERVES [Base+11] Reserved RESERVED Joystick RAW DATA [Base+12] Joystick Data [7:0] RESET = [0xF0]Joystick Data (RO) Joystick Data. Joystick Data (identical to LDN 2): Writes to this register are ignored. [Base+13] **Joystick Control** JSEL [1:0] JRDY JWRP JMSK [3:0] RESET = [0xF0]JMSK [3:0] (RW) Joystick Axis Mask. JRDY bit calculated based on axes selected by JMSK only. xxx1 Enable AX xx1x Enable AY

JSEL [1:0]	(RW)	Joystick Select.	Selects one of for	ır joystick ax	is register sets	according to t	he following table:

x1xx

1xxx

00	Read AX (16 Bits) from [Base+14] & [Base+15]
01	Read AY (16 Bits) from [Base+14] & [Base+15]
10	Read BX (16 Bits) from [Base+14] & [Base+15]
11	Read BY (16 Bits) from [Base+14] & [Base+15]

Enable BX Enable BY

JWRP (RW) Joystick Wrapmode. Continuous Joystick sampling mode—sampling automatically restarted every ~16 ms.

JRDY (RO) Joystick Ready. Sampling complete, joystick data ready for reading.

Note: Sampling must be started manually if JWRP is set before any sampling cycles are run. To start sampling after setting the JWRP bit, write to the joystick port [Base+14].

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#### [Base+14] Joystick Position Data Low Byte

7 6 5 4 3 2 1 0 JAXIS [7:0] RESET = [0xFF]

JAXIS [7:0] (RO) Joystick Axis Low Byte.

Note: Axis to be read through this register is selected by the JSEL bits in the control register. A write to this register starts a sampling cycle.

#### [Base+15] Joystick Position Data High Byte

7	6	5	4	3	2	1	0	_
			JAXIS	[15:8]				RESET = [0xFF]

JAXIS [15:8] (RO) Joystick Axis High Byte.

Note: Axis to be read through this register is selected by the JSEL bits in the control register. A write to this register starts a sampling cycle

### Sound System Indirect Registers

Writing Indirect Registers

All Indirect Registers must be written in pairs: low byte followed by high byte. The Indirect Address Register [SSBASE+0] holds the address for a register pair, the Indirect Low Data Byte [SSBASE+2] is used to write low data byte and the Indirect High Data Byte [SSBASE+3] is used to write the high data byte. The low data byte is held in the temporary register until the upper byte is written.

#### **Programming Example**

"Write Sample Rate for Voice Playback at 11,000 Hz (0x2AF8)

1) Write [SSBASE+0] with 0x02 indirect register for voice/playback sample ra

2) Write [SSBASE+2] with 0xF8 ; low byte of 16-bit sample rate register

3) Write [SSBASE+3] with 0x2A ; high byte of 16-bit sample late register

#### Reading Indirect Registers

All indirect registers can be individually read. The Sound System Indirect Address Register [SSBASE+0] holds the address for a register pair, the Indirect Low Data Byte [SSBASE+2] is used to read low data byte and Indirect High Data Byte [SSBASE+3] is used to read the High data byte.

#### **Programming Example**

"Read Sample Rate for Voice Playback set to 11,000 Hz (0x2AF8)"

Write [SSBASE+0] with 0x02
 Read [SSBASE+2]
 Read [SSBASE+3]
 ; indirect register for voice playback sample rate
 ; low byte of 16-bit sample rate register set to 0xF8
 ; high byte of 16-bit sample rate register set to 0x2A

#### ISR Saves and Restores

For Interrupt Service Routines, ISRs, it is necessary to save and restore the Indirect Address and the Low Byte Temporary Data holding registers inside the ISR.

#### **Programming Example**

"Save/Restore during an ISR"

Beginning of ISR:

1) Read [SSBASE+0] ; save Indirect Address register to TMP\_IA
2) Write [SSBASE+0] with 0x00; ; indirect Register for Low Byte Temporary Data
3) Read [SSBASE+2] ; save Low Byte Temporary data to TMP\_LBT

4) ISR Code ; ISR routine

5) Write [SSBASE+2] with TMP\_LBT ; restore Low Byte Temporary data TMP\_LBT 6) Write [SSBASE+0] with TMP\_IA ; restore Indirect Address Register to TMP\_IA

7) Return from Interrupt ; return from ISR

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Table VII. Indirect Register Map and Reset/Default States

	Address	Register Name	Reset/ Default State
	00	Low Byte TMP	0xXX
	01	Interrupt Enable and External Control	0x0102
	02	Voice Playback Sample Rate	0x1F40
	03	Voice Capture Sample Rate	0x1F40
	04	Voice Attenuation	0x8080
	05	FM Attenuation	0x8080
	06	I <sup>2</sup> S(1) Attenuation	0x8080
	07	I <sup>2</sup> S(0) Attenuation	0x8080
	08	Playback Base Count	0x0000
	09	Playback Current Count	0x0000
	10	Capture Base Count	0x0000
	<del> </del>	Capture Current Count	0x0000
/ / \ \	/12	Timer Base Count	0x0000
	/13/	Timer Current Count	0x0000
	14	Master V <del>olu</del> me Attenuation	0x0000
	15	CD Gail Attenuation	0x8888
	(6)	Synth Gain/Attenuation	0x8888
L	17	Video Gam/Attenuation	0x8888
	18	Line Cain/A)tenuation	0x8888
	19	MixPHONE_IX Gain(Attenuation)	0x8888
	20	ADC Source Select and ADC PGA	0x0000
	32	Chip Configuration	0x00F0
	33	DSP Configuration	0x0000
	34	FM Sample Rate	0x5622
	35	I <sup>2</sup> S(1) Sample Rate	OXAC44
	36	I <sup>2</sup> S(0) Sample Rate	0xAC44
	37	Reserved	0x0000
	38	Programmable Clock Rate	0xAC44
	39	3D Phat Stereo Control/PHONE_OUT Gain Attenuation	0x8000
	40	Reserved Hardware Volume Button Modifier	0x0000 0xXX1B
	41 42	DSP Mailbox 0	0x0000
	42 43	DSP Mailbox 0 DSP Mailbox 1	0x0000 0x0000
	43 44	Power-Down and Timer Control	0x0000 0x0000
	44 45	Version ID	0xXXXX
	45 46	Reserved	0x0x0x 0x0000
	-10	IVOSCI YCU	UAUUUU

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**Table VIII. Sound System Indirect Registers** 

				(High	n Byte)								(Low	Byte)			
	ADDRESS	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
	00 (0x00)				RI	ES							LBTI	D [7:0]			
	01 (0x01)	PIE	CIE	TIE	VIE	DIE	RIE	JIE	SIE	TE						XC1	XC0
	02 (0x02)				VPSR	[15.8]							VPSR	? [7:0]			
	03 (0x03)				VCSR	. ,							VCSF	R [7:0]			
	04 (0x04)	LVM	RES				[5:0]			RVM	RES				A [5:0]		
	05 (0x05)	LFMM	RES				A [5:0]			RFMM	RES				/IA [5:0]		
	06 (0x06)	LS1M	RES				A [5:0]			RS1M	RES				IA [5:0]		
	07 (0x07)	LS0M	RES				A [5:0]			RS0M	RES				OA [5:0]		
	08 (0x08)				PBC								PBC				
	09 (0x09)				PCC									[7:0]			
	10 (0x0A)				CBC									[7:0]			
	11 (0x0B)				CCC	. ,							CCC	. ,			
	12_(0x0C)				TBC									[7:0]			
	13 (0x0D)				TCC								TCC	[7:0]			
/	14 (0x0E)	LMVM/	_	ES			LMVA [4:0			RMVM		ES			RMVA [4:		
/ /	15 (0x0F)	1 CDM		ES			LCDA [4:0			RCDM		ES			RCDA [4:		
	16 (0x10)	LSYM		PS )		$\overline{}$	LSYA [4:0]			RSYM		ES			RSYA [4:0		
/ /	17 (0x11)	LVDM	R		$H \subset$	$\overline{}$	LVDA [4:0			RVDM		ES			RVDA [4:		
	18 (0x/2) 19 (0x13)	LLM M¢M	M20 R	ES	$\mathcal{L}$	$\rightarrow$	LLA [4:0] MCA [4:0]	$\overline{}$		RLM PIM		ES ES			RLA [4:0		RES
\	20 (0x13)	LAGC	MZU			$\overline{}$		[3:0]	$\overline{}$	RAGC	K.	RAS [2:0]		1	PIA [3:0]	I G [3:0]	KES
	32 (0x20)	WSE	CDE	LAS [2:0]	CNP	$\vdash$		[3:0] FS	$\rightarrow$	KAGC	COE	(3:0)		1901	F1 [1:0]		0 [1:0]
	32 (0x20) 33 (0x21)	DS1	DS0	DIT	RI	<del>}</del>	AIDR	IIT	I IOT	CPI	PBI	FMI		I01	F1 [1:0]	DFS [2:0]	-
	34 (0x22)	D31	D30	DII	FSMR		ADK	1111	1 101	CFI	FDI	I IVII	-	R [7:0]		DF3 [2.0]	
	35 (0x23)				SISR		+	-/	<del>//</del>	<del> /                                    </del>	-	-	SISR		_		
	36 (0x24)				SOSR	. ,	$\overline{}$		<del>/  </del>	/	-+			2 [7:0]	$\overline{}$	<del></del>	
	37 (0x25)					ES	$\overline{}$		$-\!$	$\leftarrow$	-/-	<u></u>		ES ES	$+ \sim$	<del>/ _ `</del>	$\overline{}$
	38 (0x26)					[15:8]			-		7 /	<del>/                                    </del>	PCR		/	$+ \cap$	egthinspace =  egt
	39 (0x27)	3DDM	R	ES		. ,	[3:0]		RES	POM		ES		<del>'                                    </del>	POA [4:0		
	40 (0x28)	,			RI		11					_	7 RI	es/ /	<del> </del>	<u> </u>	$\overline{}$
	41 (0x29 )				RI					VMU	VUP	VDN	r'	1 1	BM [4.0	<del>-                                    </del>	7
	42 (0x2A)				MB0R	[15:8]							MB0F	R [7:0]	<del>- 1</del>	$\overline{}$	
	43 (0x2B)				MB1R									R [7:0]		$\overline{}$	$\overline{}$
	44 (0x2C)	CPD	RES	PIW	PIR	PAA	PDA	PDP	PTB	3D	PD3D	GPSP	RES	DM		RES	$\supset$
	45 (0x2D)			•	VER [15:8]		•		•				VER	[7:0]	•		
	46 (0x2E)				RES								R	ES			

[00] I	NDIRE	CT LO	N BYTI	E TMP									DEFAU	J <b>LT</b> =	[0xXX]
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
		RI	ES								LBTE	[7:0]			

LBTD [7:0] Low Byte Temporary Data holding latch for register pair writes; Written on any write to [SSBase + 2], Read from [SSBase + 2] when the indirect address is 0x00.

[01] I	NTERE	UPT E	NABLE	AND I	EXTER	NAL C	ONTRO	L					DEF	AULT = [	0x0102]
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
PIE	CIE	TIE	VIE	DIE	RIE	JIE	SIE	TE			RES			XC1	XC0
XC0	F	RW									e XCTL0 to be disa			also mux 2].	ed with
XC1	F	RW									e XCTL1 ull-up ~ 0		CTL1 m	ay also be	used for
TE	F	RW	Time	er Enabl	e Bit.										
SIE	I	RW	Sour 0 1	Sc	oundBlas	ster Inte	ole; This errupt di errupt er	sabled	t be set	to en	able Curi	ent Co	unt Tim	er.	
JIE	F	RW	Joyst 0 1		ystick Ir	terrupt	disabled enabled								

RIE	RW	Ring In	iterrupt Enal Ring Inte	ole; errupt disa	ıbled								
		1		errupt ena									
DIE	RW	DSP In 0 1		ole; errupt disa errupt ena									
VIE	RW	=		-		software	increm	ents/dec	rements	RUT	TON MOD	IFIE	R via
VIL	1000		pt routine an Volume		buttons disabled						oes not chan		
TIE	RW	Timer 1 0 1		able; iterrupt di iterrupt en									
CIE	RW	Captur	e Interrupt H										
	\ _	0		Interrupt Interrupt									
PIÆ	RW /	Playbac	ck Interrupt		епаріец								
			Płayback	Interrupt Interrupt									
[02] VO	ICE PLAY	BACK SA	MPLE RAY	£ /			_				DEFAULT	Γ =	[0x1F40]
7	$\frac{6}{5}$	-4	L3 /5	1	0	<del>\ \ \</del>	<u> </u>	5	4	3	2	1	
		VPSR	[15:8]	-) [ [	'	<del>IJ</del> ,	<u> </u>		VPSR	[7:0]			
VIDOD [45	, ol 17 1 D			/ \ \		!	/ ,			$\sim 1$		_	
VPSR [15	5:0] Voice P		mple Rate. 1 mple rate is 8		rate can	be progr	animed	from 4/kl	HZ <del>(0-55</del>	.2 KH	z in Thertz ii	ncre	nents. The
[00] Y/O	•	v	•			_		¬ /	/	_/	DEHALT		0x1F40]
	ICE CAPT							J			DEFAULT	= 4	
7			9 9	1	Λ	7	G	<b>*</b>	1	~ ?	191	1 /	0 -
7 	6 5	VCSR			0	7	6	5	VCSR			1/	
VCSR [15	5:0] Voice C nored if	VCSR   apture San CNP bit in	[15:8] nple Rate. Th n SS [32] = 0	ne sample i in which ca	rate can se VPSR	be progra	mmed fi	pture rate	Iz to 55. e. The de	7:0] 2 kHz		ple ra	nents, Ig- te is 8 kHz.
VCSR [15]	5:0] Voice C nored if	VCSR   apture San CNP bit in	[15:8]  nple Rate. That is a SS [32] = 0.	ne sample i	rate can	be progra	mmed f		Iz to 55.	7:0] 2 kHz efault	in 1 hertz in capture samp	ple ra <b>Г</b> =	rents. Ig- te is 8 kHz.
VCSR [15]	5:0] Voice C nored if DICE ATTE 6 5 RES     Right V range is Right V   Left Vo	VCSR   capture San CNP bit in NUATION 4  oice Atten 0 dB to -9 oice Attenu 0 dB to -9	[15:8]  Inple Rate. The SS [32] = 0 and SS [32	ne sample i in which ca 1 ayback ch uted, 1 = 1 yback cha	rate can se VPSR 0 annel. 7 Muted.	be progra [15:0] co 7 RVM The LSB	mmed fintrols ca  6 RES represer	5 nts –1.5 c	Iz to 55 e. The do 4	2 kHz efault 3	in 1 hertz in capture samp DEFAULT	ple ra <b>Γ</b> =   1 the	nents, Ig- te is 8 kHz.
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM	5:0] Voice C nored if DICE ATTE 6 5 RES   ] Right V range is Right V l Left Vo	VCSR   apture San CNP bit in NUATION 4  foice Atten 0 dB to -5 oice Mute ice Attenu 0 dB to -5 ice Mute.	[15:8]  Inple Rate. The SS [32] = 0 and SS [32	ne sample i in which ca 1 ayback ch uted, 1 = 1 yback cha	rate can se VPSR 0 annel. 7 Muted.	be progra [15:0] co 7 RVM The LSB	mmed fintrols ca  6 RES represer	5 nts –1.5 c	Iz to 55 e. The do 4	2 kHz efault 3	c in 1 hertz in capture samp  DEFAULT 2  RVA [5:0]  = 0 dB and the capture sample samp	ple ra $\Gamma =  $ 1 the	nents, Ig- te is 8 kHz. (0x8080]
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM	5:0] Voice C nored if DICE ATTE 6 5 RES     Right V range is Right V   Left Vo	VCSR   apture San CNP bit in NUATION 4  foice Atten 0 dB to -5 oice Mute ice Attenu 0 dB to -5 ice Mute.	[15:8]  Inple Rate. The SS [32] = 0 and SS [32	ne sample i in which ca 1 ayback ch uted, 1 = 1 yback cha	rate can se VPSR 0 annel. 7 Muted.	be progra [15:0] co 7 RVM The LSB	mmed fintrols ca  6 RES represer	5 nts –1.5 c	Iz to 55 e. The do 4	2 kHz efault 3	in 1 hertz in capture samp DEFAULT 2 RVA [5:0]	ple ra $\Gamma =  $ 1 the	nents, Ig- te is 8 kHz. (0x8080]
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM  [05] FN 7	5:0] Voice Conored if  DICE ATTE 6 5  RES  Right Vorange is Right Vorange is Left Vo	vcsr   apture San cNP bit in NUATION 4  oice Atten 0 dB to -9 oice Mute ice Attenu 0 dB to -9 oice Mute.	[15:8]  Inple Rate. The SS [32] = 0 and SS [32	ne sample in which ca  1  ayback chauted, 1 = 1 yback chauted, 1 = M	rate can se VPSR 0 annel. 7 Muted. nnel. Tl	be progra 2 [15:0] co 7 RVM The LSB	mmed fintrols ca  6  RES  represent	5 nts –1.5 d	Iz to 55.  2  4  dB, 0000	7:0] 2 kHz 2 kHz 3 000 =	a in 1 hertz in capture samp  DEFAULT 2  RVA [5:0]  = 0 dB and the  DEFAULT  O dB and the	ple ra  F =    1  the	nents, Ig- te is 8 kHz.  [0x8080] 0
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM  [05] FN 7  LFMM F  RFMA [5]	5:0] Voice Conored if  DICE ATTE 6 5  RES  Right Voice Conored if Conored if Conored in the cono	apture San CNP bit ir NUATION 4  oice Atten 0 dB to -9 oice Mute ice Attenu 0 dB to -9 ice Mute. 4  ATION 4  Music Att Music M	nple Rate. The SS [32] = 0 or SS [32	ne sample in which car in which can be a seed, 1 = M in the international in which in the international in which car in which in which car in which c	rate can se VPSR  0  annel. 7  Muted. nnel. Tl  futed.  0  nal Mus  = Mute	be progra c [15:0] co 7 RVM The LSB ne LSB re 7 RFMM ic Synthe d.	mmed fintrols can feet a second can be called	5  nts -1.5 di  s -1.5 di  he LSB r	Iz to 55e. The do	7:0] 2 kHz 2 kHz 3 000 = 3 tts -1.	a in 1 hertz in capture samp  DEFAULT  2  RVA [5:0]  = 0 dB and the  DEFAULT  2  RFMA [5:0]  5 dB, 000000	pple rate $\Gamma = 1$ the second $\Gamma = 1$	(0x8080) 0 (0x8080) 0 (0x8080)
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM  [05] FN 7  LFMM F  RFMA [5]  RFMM [5]	5:0] Voice Conored if  DICE ATTE 6 5 RES    Right Voice Conored if 1 Right Voice Conored if 2 Right Voice Conored in the conor	apture San CNP bit in NUATION 4 Toice Atten 0 dB to -9 toice Mute ice Attenu 0 dB to -9 toice Mute. ATION 4 Music Atte where is 0 dB to Music Mute of dB to -9 to dB to -9	nple Rate. The SS [32] = 0 or SS [32	ane sample in which car in which can be sample in the international in the internation which in the internation which is in the internation which can be a supplied to the internation which is in the internation wh	rate can se VPSR  0 annel. 7 Muted. nnel. Tl futed.  0 nal Muse = Mute	be progra [15:0] co  7 RVM The LSB The LSB ref RFMM The Synthesis	mmed fintrols can feet a second can be called	5  nts -1.5 di  s -1.5 di  he LSB r	Iz to 55e. The do	7:0] 2 kHz 2 kHz 3 000 = 3 tts -1.	c in 1 hertz in capture samp  DEFAULT 2  RVA [5:0]  = 0 dB and the  DEFAULT 2  RFMA [5:0]	pple rate $\Gamma = 1$ the second $\Gamma = 1$	(0x8080) 0 (0x8080) 0 (0x8080)
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  LVM  [05] FN 7  LFMM F  RFMA [5]  RFMA [5]	5:0] Voice Conored if  DICE ATTE 6 5  RES  RES  Right Vorange is Right Vorange is Left Vo M ATTENU 6 5  RES  :0] Right F the rang Right F :0] Left F N range is Left F N	vcsr   apture San cNP bit in NUATION 4 oice Atten 0 dB to -9 oice Mute ice Attenu 0 dB to -9 oice Mute. 4 ATION 4 Music Att ge is 0 dB t Music Mu Music Attenu 0 dB to -9 Music Mu	nple Rate. The SS [32] = 0 or SS [32	ane sample in which car in which can be sample in the international in the internation which in the internation which is in the internation which can be a supplied to the internation which is in the internation wh	rate can se VPSR  0 annel. 7 Muted. nnel. Tl futed.  0 nal Muse = Mute	be progra [15:0] co  7 RVM The LSB The LSB ref RFMM The Synthesis	mmed fintrols can feet a second can be called	5  nts -1.5 di  s -1.5 di  he LSB r	Iz to 55e. The do	7:0] 2 kHz 2 kHz 3 000 = 3 tts -1.	a in 1 hertz in capture samp  DEFAULT  2  RVA [5:0]  = 0 dB and the  DEFAULT  2  RFMA [5:0]  5 dB, 000000  dB, 0000000	pple rate $\mathbf{r} = 1$ the second $\mathbf{r} = 1$	nents, Ig- te is 8 kHz.  (0x8080] 0  (0x8080] 0  0 dB and dB and the
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM  [05] FN 7  LFMM [5]  RFMA [5]  RFMA [5]  RFMM [5]  LFMM [06] I <sup>2</sup> S	5:0] Voice Conored if  DICE ATTE 6 5 RES  Right Voice Conored if 1 Right Voice Conored if 2 Right Voice Conored in the conored	apture San CNP bit in NUATION 4 Toice Atten 0 dB to -9 toice Mute ice Attenu 0 dB to -9 toice Mute. 4  Music Atten Music Atte 0 dB to -9 toice Music Music Music Music Atte 0 dB to -9 Music Mus	nple Rate. The SS [32] = 0 or SS [32	ayback charted, 1 = Manuted, 1	rate can se VPSR  0  annel. 7  Muted. nnel. Tl  futed.  0  al Muse Muted Muted	be progra [15:0] co  7 RVM The LSB ne LSB re  7 RFMM ic Synthesi	mmed fintrols can be seen to be s	pture rate  5  nts -1.5 di  5  he LSB re	Iz to 55e. The de 4  HB, 0000  3, 00000  4  represents	7:0] 2 kHz 2 kHz 3 000 = 00 = 3 tts -1.5	in 1 hertz in capture samp  DEFAULT 2  RVA [5:0]  = 0 dB and the description of the composition of the composition of the capture sample of the capture sa	ple rapper	[0x8080] 0 0 dB and dB and the
VCSR [15]  [04] VC 7  LVM F  RVA [5:0]  RVM  LVA [5:0]  LVM  [05] FN 7  LFMM F  RFMA [5  RFMM  LFMA [5  LFMM  [06] I <sup>2</sup> S 7	5:0] Voice Conored if  DICE ATTE 6 5  RES  RES  Right Vorange is Right Vorange is Left Vo M ATTENU 6 5  RES  :0] Right F the rang Right F :0] Left F N range is Left F N	vcsr   apture San cNP bit in NUATION 4 oice Atten 0 dB to -9 oice Mute ice Attenu 0 dB to -9 oice Mute. 4 ATION 4 Music Att ge is 0 dB t Music Mu Music Attenu 0 dB to -9 Music Mu	nple Rate. The SS [32] = 0 or SS [32	ayback charted, 1 = Mayback ch	rate can se VPSR  0 annel. 7 Muted. nnel. Tl futed.  0 nal Muse = Mute	be progra [15:0] co  7 RVM The LSB The LSB ref RFMM The Synthesis	mmed fintrols can feet a second can be called	5  nts -1.5 di  s -1.5 di  he LSB r	Iz to 55e. The do	7:0] 2 kHz 2 kHz 3 000 = 3 tts -1.	a in 1 hertz in capture samp  DEFAULT  2  RVA [5:0]  = 0 dB and the  DEFAULT  2  RFMA [5:0]  5 dB, 000000  dB, 0000000	pple rate $\mathbf{r} = 1$ the second $\mathbf{r} = 1$	nents, Ig- te is 8 kHz.  (0x8080] 0  (0x8080] 0  0 dB and dB and the

RS1A [5:0] Right  $I^2S(1)$  Attenuation register. The LSB represents -1.5 dB, 000000 = 0 dB and the range is 0 dB to -94.5 dB.

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RS1M Right  $I^2S(1)$  Mute. 0 = Unmuted, 1 = Muted. LS1A [5:0] Left  $I^2S(1)$  Attenuation register. The LSB represents -1.5 dB, 000000 = 0 dB and the range is 0 dB to -94.5 dB. LS1M Left  $I^2S(1)$  Mute. 0 = Unmuted, 1 = Muted. [07] I<sup>2</sup>S(0) ATTENUATION DEFAULT = [0x8080]LS0M RES RS0M RES RS0A [5:0] Right I<sup>2</sup>S(0) Attenuation register. The LSB represents -1.5 dB, 000000 = 0 dB and the range is 0 dB to -94.5 dB. RS0A [5:0] Right  $I^2S(0)$  Mute. 0 = Unmuted, 1 = Muted. RS0M LS0A [5:0] Left I<sup>2</sup>S(0) Attenuation register. The LSB represents -1.5 dB, 000000 = 0 dB and the range is 0 dB to -94.5 dB. LS0M Left  $I^2S(0)$  Mute. 0 = Unmuted, 1 = Muted. [08] PLAYBACK BASE COUNT DEFAULT = [0x0000]PBC [15:8] PBC [7:0] PBC [15:0 Playback Base Count. This register is for loading the Playback DMA Count. Writing a value to this register also loads the came data into the Rlayback Current Count register. You must load this register when Playback Enable (PEN) is deasserted. When PEN is asserted, the Playback Current Count decrements once for every four bytes transferred via a DMA excle. The next transfer, after zero is reached in the Playback Current Count, will generate an interrupt/and reload the Rlayback Current Count with the value in the Playback Base Count. The Playback Base Count should always be programmed to Number Bytes divided by four, minus one ((Number Bytes/4) –1). The circular software DMA buffer must be divisible by four to ensure proper operation. DEFAULT [0x0000] [09] PLAYBACK CURRENT COUNT 5 3 [7:0] PCC [15:8] PCC PCC [15:0] Playback Current Count register. Contains the current Playback DMA Count/Reads/and/Writes must be done when PEN is deasserted. [10] CAPTURE BASE COUNT DEFAULT (0000xQ CBC [15:8] CBC [7:0] CBC [15:0] Capture Base Count. This register is for loading the Capture DMA Count. Writing a value to this register also loads the same data into the Capture Current Count register. Loading must be done when Capture Enable (CEN) is deasserted. When CEN is asserted, the Capture Current Count decrements once for every four bytes transferred

is deasserted. When CEN is asserted, the Capture Current Count decrements once for every four bytes—transferred via a DMA cycle. The next transfer, after zero is reached in the Capture Current Count, will generate an interrupt and reload the Capture Current Count with the value in the Capture Base Count. The Capture Base Count should always be programmed to Number Bytes divided by four, minus one ((Number Bytes/4) –1). The circular software DMA buffer must be divisible by four to ensure proper operation.

[11]	CAPTU	RE CU	RRENT	COUN	$\mathbf{T}$							]	DEFAU	LT = [(	)x0000]
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
			CCC	[15:8]							CCC	[7:0]			

CCC [15:0] Capture Current Count register. Contains the current Capture DMA Count. Reading and Writing must be done when CEN is deasserted.

[12]	TIMER	BASE	COUNT									]	DEFAU	LT = [0]	)x0000]
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
			TBC	[15:8]							TBC	[7:0]			

TBC [15:0] Timer Base Count. Writing a value to this register loads data into the Timer Current Count register. Loading must be done when Timer Enable (TE) is deasserted. When TE is asserted, the Timer Current Count register decrements once for every specified time period. The time period (10 µs or 100 ms) is programmed via the PTB bit in SS [44]. When TE is asserted, the Timer Current Count decrements once every time period. The next count, after zero is reached in the Timer Current Count register, will generate an interrupt and reload the Timer Current Count register with the value in the Timer Base Count register.

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LLA [4:0] LLM

Left Line Mute. 0 = Unmuted, 1 = Muted.

22 [12 2]			3	2	1	0	7	6	5	4	3	2	1	0
		TCC [	15:8]							TCC	[7:0]			
CC [15:0]		OMA Curr easserted.	rent Co	unt registe	er. Con	tains	the curren	t time	r count.	Reading a	and V	Vriting must	be do	ne wh
[14] MAS	TER VOI	LUME A	TTEN	UATION								DEFAUL	Γ = [0	x0000
7 6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
MVM	RES		LI	MVA [4:0]			RMVM	F	RES			RMVA [4:0]		
MVA [4:0] MVM MVA [4:0]	-46.5 dl Volume Right M Left Ma -46/5 dJ	B. This reg attenuation Iaster Volun Ster Volun B. This reg	gister is n level. ume Mu ne Atte gister is	added with See Hardy Ite. 0 = U nuation. Tadded with	h the H ware Vo nmuted The LS h the H	ardwa llume d, 1 = B repi ardwa	re Volume Button Me Muted. resents –1 re Volume	Butto odifier 5 dB, Butto	n Modifi Register 00000 = n Modifi	er value to descriptio 0 dB and er value to	proon proon	e range is 0 of duce the final more details. range is 0 dl duce the final more details.	DAC B to DAC	
MVM/		ster Volun						_	24082002	acserpee				
_	GAIN/ATT	ENUATI	1 1	//	//		) )	17		_		DEFAUL	$\Gamma = [0]$	x8888
7 6		4/	$\sqrt{3}$	2) )	1 /	0	7 RCDM	9	5	/4	3	2	1	0
CDM	RES	<u> </u>	L	CDA [4:0]	+		JKC DM/	<del>    1</del>	RES	-	$\longrightarrow$	RGDA [4:0]	_	
CDA [4:0]	Right C	D Attenua	ation T	Γhο I SR r	onrocer	ote _1	5/1B (00	$d_0 = .$	12 dB/2	and the ra	ngo i	s +12/dB <b>f</b> o	-2A 5/	dR
CDA [4.0] CDM	_	D Mute. 0			-			- OWO	-12 db/s	illy the ra		3 +12/ub/o	-34.3	ub. /~
	0						4D 0000	0 -	Tallon	d the	ao io	+12/dB/to -	34.5 d	$_{\mathrm{D}} \subset$
CDA [4:0]					-		ab, 0000	10 = +	12 an an	id the <del>ran</del>	ge is 7	+12/06/10 -	34.9 a	Б. —
CDM	Len CD	) Mute. 0 :	= Unm	utea, 1 =	Mutea	•					7	L		/
[16] SYN'	TH GAIN	/ATTENU	J <b>ATIO</b>	N										
7 6	5											DEFAUL	Γ <u></u> [0	x8888
	DEG	4	3	2	1	0	7	6	5	4	3	2	Γ <u> </u> [0	x8888 0_
	RES	4			1	0	7 RSYM		5 RES	4	3		Γ /= [0 1	x8888 
SYM [4:0] SYM SYM [4:0] SYM	Right SY Right SY Left SY	YNTH Att YNTH Mu NTH Atte NTH Mut	tenuation tenuation tenuation te. 0 = 1	2 SYA [4:0] on. The L Unmuted on. The LS	SB repred, 1 = 1	resent Mute	RSYM   s -1.5 dB; d1.5 dB;	0000	$\frac{\text{RES}}{0 = +12}$	dB and tl	ne ran	2 RSYA [4:0] age is +12 dl ge is +12 dB	1 B to -34	34.5 dB
SYM [4:0] SYM SYM [4:0] SYM 17] VID	Right SY Right SY Left SY Left SY	YNTH Att YNTH Mu NTH Atte NTH Mut	tenuation tenuation tenuation te. 0 = 1	2 SYA [4:0] on. The L Unmuted 1. The LS Unmuted	SB repred, 1 = N	resent Mutee esents Muted	RSYM   RSYM   s -1.5 dB d1.5 dB, (	00000	RES $0 = +12$ $= +12$ d	dB and the	ne ran	2 RSYA [4:0]  age is +12 dl  ge is +12 dB  DEFAUL	1 B to -34	34.5 dB
SYM [4:0] SYM SYA [4:0] SYM SYA [4:0] SYM [17] VID 7 6	Right SY Right SY Left SY	YNTH Att YNTH Mu NTH Atte NTH Mut	tenuation tenuation te. 0 = 1 te. 0 = 1 te. 0 = 1	2 SYA [4:0] on. The L Unmuted on. The LS	SB repred, 1 = 1	resent Mute	RSYM   s -1.5 dB; d1.5 dB;	00000	$\frac{\text{RES}}{0 = +12}$	dB and tl	ne ran	2 RSYA [4:0] age is +12 dl ge is +12 dB	1 B to -34	34.5 dl
SYM [4:0] SYA [4:0] SYM SYA [4:0] SYM  [17] VID 7 6 VDM  CVDA [4:0] CVDM  VVDA [4:0] VVDM  [18] LINE	Right SY Right SY Left SY Left SY  GAIN/AT 5 RES  Right VI Right VI Left VII Left VII	YNTH Atte YNTH Mu NTH Atte NTH Mut  TENUAT 4  ID Attenua ID Mute. () D Attenua D Mute. 0	tenuation te. 0 = 1 tenuation te. 0 = 1 tenuation 3 LV action. T 0 = Un attion. T = Unm	2 SYA [4:0] on. The L Unmuted The LS Unmuted,  2 VDA [4:0] The LSB r mute, 1 = The LSB re nuted, 1 =	SB repred, 1 = M  1  represere Muteo	resent Mutedesents Auted. 0 nts -1.1 d.	RSYM   RSYM   s -1.5 dB, d1.5 dB, 000   5 dB, 000	$\frac{6}{0000}$ $\frac{6}{10000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$	RES $0 = +12$ $= +12$ d d d $= +12$ d d d $= +12$ d d d d $= +12$ d d d d d d d d d d d d d d d d d d d	dB and the  4  and the raind the raind the rain	and a rang	2 RSYA [4:0]  rige is +12 dB  ge is +12 dB  DEFAULT 2  RVDA [4:0]  s +12 dB to  +12 dB to  DEFAULT	$\frac{1}{1}$ B to $-34$ T = [0 $\frac{1}{1}$ $-34.5$ $-34.5$ $\Gamma$ = [0	34.5 dB 1.5 dB 0 dB.
SYM [4:0] SYM SYA [4:0] SYM [17] VID 7 6 VDM VDA [4:0] VDM VDA [4:0] VDM	Right SY Right SY Left SY Left SY  GAIN/AT 5 RES Right VI Right VI Left VII Left VII	YNTH Atte YNTH Mu NTH Atte NTH Mut  TENUAT 4  ID Attenua ID Mute. 0 D Attenua D Mute. 0	tenuation te. 0 = 1 tenuation te. 0 = 1 tenuation 3 LV action. T 0 = Un attion. T = Unm TION 3	2 SYA [4:0] on. The L Unmuted The LS Unmuted,  2 VDA [4:0] The LSB r mute, 1 =	SB repred, 1 = M  1  represer	resent Mutedesents Auted.  0  nts -1. d. ts -1.5	RSYM   RSYM   s -1.5 dB, d1.5 dB, (	$\frac{6}{10000}$ $\frac{6}{10000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$ $\frac{6}{1000}$	RES $0 = +12$ $= +12$ d  5 RES $+12$ dB a	dB and the  4  and the ra	ne range range is	2 RSYA [4:0] age is +12 dB ge is +12 dB DEFAULT 2 RVDA [4:0] s +12 dB to +12 dB to	3 to $-34$ to $-34$ $\Gamma = [0]$ $-34.5$ $-34.5$	34.5 dB 1.5 dB <b>x8888</b> 0 dB.

Left LINE Attenuation. The LSB represents -1.5 dB, 00000 = +12 dB and the range is +12 dB to -34.5 dB.

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REV. A

[19] MIC/P	<b>PHONE_I</b>	N GAIN	I/ATTEN 3	NUATI	<b>ON</b> 1	0	7	6	5	4	3	DEFAUI 2	LT = [0	0 <b>x8888</b> ]
MCM M20				CA [4:0]			PIM		RES		PIA			RES
PIA [3:0] PIM MCA [4:0] M20 MCM	PHONI Micropi Micropi	E_IN M hone Att	ute. tenuation dB Gain	ı. The I	LSB repr	resents	-1.5 dB,	00000	0 dB and t $0 = +12 dB$ $0 = +20 dB$	and th	ne range			4.5 dB.
[20] ADC SO	•			DC PC	<u>.</u> v						ī	DEFAUI	T = [0	10000x
7 6	5 5	4	3	2	1	0	7	6	5	4	3	2	1	0
LAGC	LAS [2:0	]		LAG	[3:0]		RAGC		RAS [2:0]			RAG	[3:0]	
RAG [3:0]  RAG [C LAG [3:0]  LAG	and the Right A Left AD and the	range is utomati C Gain range is	0 dB to c Gain C Control 7 0 (B to	+22.5 (ontrol ABC so +22 5 (	dB. (AGC) F urce sele dB.	Enable,	1 = Ena Gain. Fo	bled, 0 r Gain,	n, LSB repro = Disable LSB repres = Disabled	d. sents +				
RAS [2:0] 000 001 010 011 100 101 111	R_LINI R_OUT R_CD R_SYN R_VID Mono N Reserve Reserve	TH Mix ed	ut Sour	<i>)</i> (				AS [2: 00 01 10 11 10 11 10 11	L L C L S L V MIC Rese	INE OUT YNTH ID ONE_I erved	N $\int$	ource		
Note: When [32] CHIP			_	ontrol	settings f	for the .	ADC PC	A are	overridden	for all	-	DEFAUI	LT = [(	
7 6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
WSE   CDI	E RES	CNP		RI	ES			CO	F [3:0]		I <sup>2</sup> SF1	[1:0]	I <sup>2</sup> SF0	[1:0]
I <sup>2</sup> SF0 [1:0] I <sup>2</sup> SF1 [1:0]	I <sup>2</sup> S Por	(		abled	ied l	pe.								
COF [3:0]	PCLKC	0 = 256	requency × PCR/2' F > 11, th	<sup>COF</sup> whe	ere COF	r = 0:11	and PC	on PCI R is th	LKO pin is e value of t	detern he Pro	nined us ogramma	ing the fo	ollowin k Rate	g formula Register,
CNP	0 = Cap	oture equ	ual to Pla uals Playl t equal to	back. T		ıre sam	ple rate i	is deter	mined by t	the play	yback sa	mple rate	e in SS	[02].
CDE	CD Ena	able, Set	-	vhen a (	CD play			to I <sup>2</sup> S (	(0), maps S	SoundE	Blaster C	D mixer	contro	ls from
WSE	Sound S 0 = Sou 1 = Sou Note: V	System I ındBlast ınd Syste	Enable. er Mode. em Mode SoundBla	e under	Window	vs.	•	nd DAG	C channels	will be	e used so	olely for o	convert	ing

REV. A -35-

POA [4:0]

[ <b>33</b> ] <b>D</b> S	SP CO	ONFIGU 5	RATION 4	3	2	1	0	7	6	5	4	3	<b>DEF</b> 2	AULT = [0]	<b>xuuuu</b> ] 0
	DS0	DIT	RE		ADR	I1T	IOT	CPI	PBI	FMI	III	IOI	T ~	DFS [2:0	
FS [2:0		000—Ma 001—I <sup>2</sup> S 010—I <sup>2</sup> S 011—Mu	me Sync aximum F (0) Samp (1) Samp usic Synth	'rame le Rat le Rat lesizer	Rate te te Sample	Rate		me Sync	accord	ling to th	ne follow	ing sou	rce.		
		101—So 111—Re		m Ca	pture Sa	mple Ra	te	٥							
0I			ata Interc	•			-								
II MI			ata Interc ic Synthe								fusic Da	ta Enab	iled.		
BI	Ι,	. , –	Data Int			-			•			u Diiub	icu.		
CPI	)	1 °1 F	Data Int	•											
如	//	1 ~ 1	akeover J	~		`									
$_{1T}$	//	I2\$(1)T	akeqver\D	ata.0	= Disab	<del>le,</del> )1 =∕1	Enabled.								
DR			esync.Wr												
OIT			rrupt A									_			
OSO OS1			ilbox 0 St ilbox 1 St											_	
				atus. (	$0 = \frac{1}{10}$ st a	iccess in	atcates	eau, 1	last ac	cess ma	icates wi	_	$\sum$		
		MPLE R		0	0	1	$\smile$	٦/	<u></u>	_ [		<b>~</b> /	DEFA	.ULT = {0	_
7	6	5	4	3	2	1	0	7 🖳	- 6	751	/ 4	$\sim_3$	12	/ 1 /	/0 ~
MSR [1	5:0]		ASR [15:8 Sample F	3]					ogramı	ned fror	$\overline{}$	[7:0] [0 27.6	k <b>j</b> Hz i		
		F Music <b>AMPLE</b> 5	Sample F RATE 4	3] Pate re					ogramı 6	med from	n 4 kHz	27.6 E	$\sim$	1 hertz in  JLT = [0x	remen
[35] I <sup>2</sup> S	S(1) S	F Music <b>AMPLE</b> 5	Sample F	3] Pate re	gister. T	he samp	le rate c	an be pr			n 4 kHz	27.6	DEFAU	1 hertz in	remen
7 7 11SR [15	<b>S(1) S</b> 6 :0]	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Program	Sample F RATE 4 SR [15:8 ample Rat ming this	B] Cate re  3 B] e regis	gister. T 2 ster. The	he samp	le rate o	an be pr	6 ramme	5 d from 4	4 S1SF	27.6 E 3 2 [7:0] 55.2 kF	DEFAU 2 Hz in 1	1 hertz in 1 hertz in 1 hertz incr	rements.
7 1SR [15	<b>S(1) S</b> 6 :0]	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programs	Sample F RATE 4 SR [15:8 ample Rat ming this	ate re  3  e registregiste	gister. T  2  ster. The er has no	he samp	le rate o	an be pr	6 ramme   is ena	5 d from 4 bled.	4 S1SF	27.6 E 3 2 [7:0] 55.2 kF	DEFAU 2 Hz in 1	1 hertz in  JLT = 10x	rements.
[35] I <sup>2</sup> S 7 1SR [15 [36] I <sup>2</sup> S	(1) S (6) (1) S (1) S (1) S (1) S	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programs  AMPLE 5	Sample FATE  4 SR [15:8 ample Rat ming this RATE	3 3 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8	gister. T 2 ster. The	he samp	ole rate o	an be pr  7 be prog SF1 [1:0	6 ramme	5 d from 4	4 S1SF 4 kHz to	27.6 I 3 2 [7:0] 55.2 kH	DEFAU  2  Hz in 1  DEFA	1 hertz in  JLT = 10x  1 hertz incr	remenda C44]
[35] I <sup>2</sup> S 7 1SR [15 [36] I <sup>2</sup> S	(a) S (b) S (c) S	F Music <b>AMPLE</b> 5  S1  I <sup>2</sup> S(1) Sa  Programm <b>AMPLE</b> 5  S0  I <sup>2</sup> S(0) Sa	Sample F  RATE  4  SR [15:8  Imple Rate hing this  RATE  4  SR [15:8	3 Bleate re  3 Bleate registre	gister. T  2  ster. The er has no	he sample sample sample	ole rate o	an be pr 7 be prog FF1 [1:0	6 ramme l is ena 6	5 d from 4 bled.	4 S1SF 4 kHz to 4 S0SR	27.6 3 2 [7:0] 55.2 kI	DEFAU  2  Hz in 1  DEFA  2	1 hertz in  JLT = 10x  1 hertz incr	ements  (AC44)  0
[35] I <sup>2</sup> S 7 21SR [15 [36] I <sup>2</sup> S 7	(a) S (b) S (c) S	F Music  AMPLE 5 S1 I <sup>2</sup> S(1) Sa Program  AMPLE 5 S0 I <sup>2</sup> S(0) Sa Program	Sample F  RATE  4  SR [15:8  Imple Rate  ming this  RATE  4  SR [15:8  Imple Rate  Annual Rate	3 Bleate re  3 Bleate registre	gister. T  2  ster. The er has no	he sample sample sample	ole rate o	an be pr 7 be prog FF1 [1:0	6 ramme l is ena 6	5 d from 4 bled.	4 S1SF 4 kHz to 4 S0SR	27.6 3 2 [7:0] 55.2 kI	DEFAU  2  Hz in 1  DEFA  2  in 1 h	1 hertz in  1 hertz incr  1 hertz incr  ULT = [0:	ements  (AC44)  0  nents.
[35] I <sup>2</sup> S 7 21SR [15 [36] I <sup>2</sup> S 7 20SR [15	(a) S (b) S (c) S	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programm 5  AMPLE 5 S0  I <sup>2</sup> S(0) Sa Programm EVED 5	Sample F  RATE  4  SR [15:8  Imple Rate  ming this  RATE  4  SR [15:8  Imple Rate  Annual Rate	3 Bleate re  3 Bleate registre	gister. T  2  ster. The er has no	he sample sample sample	ole rate o	an be pr 7 be prog FF1 [1:0	6 ramme l is ena 6	5 d from 4 bled.	4 S1SF 4 kHz to 4 S0SR kHz to 55	27.6 3 2 [7:0] 55.2 kI	DEFAU  2  Hz in 1  DEFA  2  in 1 h	h 1 hertz in  JLT = 10x  1  hertz incr  ULT = 10x  1  ertz incren	ements  (AC44)  0  nents.
[35] I <sup>2</sup> S 7 31SR [15 [36] I <sup>2</sup> S 7 30SR [15	(a) S (b) S (c) S	F Music  AMPLE 5 S1  I²S(1) Sa Programm  AMPLE 5 S0  I²S(0) Sa Programm  EVED	Sample F  RATE  4  SR [15:8  Imple Rate of this of thi	3 Bl e registregistregister  gister	gister. T  2  ster. The er has no 2  ster. The has no e	he sample sample sample ffect un	ole rate of the rate can less I <sup>2</sup> SF	an be pr 7 be prografication of the prografi	famme disconnection of the control o	d from 4 led.	4 S1SF 4 kHz to 4 S0SR kHz to 55	3 2 [7:0] 55.2 kI 3 [7:0] 5.2 kHz	DEFAU  2  Hz in 1  DEFA  2  in 1 h	h 1 hertz in  JLT = 10x  1 hertz incr  ULT = 10x  1 ertz increm	(AC44] 0 ements 0 nents.
[35] I <sup>2</sup> S 7 31SR [15 [36] I <sup>2</sup> S 7 30SR [15	(a) S (b) S (c) S	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programm 5  AMPLE 5 S0  I <sup>2</sup> S(0) Sa Programm EVED 5	Sample F  RATE  4  SR [15:8  Imple Rate of this of thi	3 Bl e registregistregister  gister	gister. T  2  ster. The er has no 2  ster. The has no e	he sample sample sample ffect un	ole rate contrate can less I <sup>2</sup> SF	an be pr 7 be prografication of the prografi	famme disconnection of the control o	d from 4 led.	4 S1SF 4 kHz to 4 S0SR kHz to 55	3 2 [7:0] 55.2 kI 3 [7:0] 5.2 kHz	DEFAU  2  Hz in 1  DEFA  2  in 1 h	h 1 hertz in  JLT = 10x  1 hertz incr  ULT = 10x  1 ertz increm	(AC44] 0 ements 0 nents.
[35] I <sup>2</sup> S 7 51SR [15 [36] I <sup>2</sup> S 7 50SR [15 [37] RI 7	(1) S (6) (1) S (6) S (6) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1	F Music  AMPLE 5 S1  I²S(1) Sa Programi  AMPLE 5 S0  I²S(0) Sa Programi  EVED 5 RES	Sample F  RATE  4  SR [15:8  Imple Rate of this of thi	3 B] e register  gister  3 3 3 3 3 3 3 3 3 3 3	gister. T  2  ster. The er has no exter. The has no exter.	he sample sample sample ffect un	ole rate contrate can less I <sup>2</sup> SF	an be pr 7 be prografication of the prografi	famme disconnection of the control o	d from 4 led.	4 S1SF 4 kHz to 4 S0SR kHz to 55	3 2 [7:0] 55.2 kI 3 [7:0] 5.2 kHz	DEFAU  2  Hz in 1  DEFA  2  in 1 h  DEFA  2	h 1 hertz in  JLT = 10x  1 hertz incr  ULT = 10x  1 ertz increm	0 ements.  (AC44] 0 ements.  (AC44] 0 ements.
[35] I <sup>2</sup> S 7  1SR [15  [36] I <sup>2</sup> S 7  0SR [15  [37] RI 7	(1) S (6) (1) S (6) S (6) (1) (1) (1) (1) (1) (1) (1) (1) (1) (1	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programm  AMPLE 5 S0  I <sup>2</sup> S(0) Sa Programm  EVED 5 RES	Sample F  RATE  4  SR [15:8  Imple Rate of the second seco	3 Bl e register  3 Bl e register  3 Bl e register  3 Bl e regis	gister. T  2  ster. The er has no exter. The has no exter.	he sample sample sample ffect un	ole rate contrate can less I <sup>2</sup> SF	an be pr 7 be prografication of the prografi	famme disconnection of the control o	d from 4 led.	4 S1SF 4 kHz to 4 S0SR kHz to 55 4 RES	3 2 [7:0] 555.2 kH 3 [7:0] 5.2 kHz	DEFAU  2  Hz in 1  DEFA  2  in 1 h  DEFA  2	h 1 hertz in  JLT = 10x  1  hertz incr  ULT = 10x  1  ertz increm  AULT = 11	0 ements.  (AC44] 0 ements.  (AC44] 0 ements.
[35] I <sup>2</sup> S 7 51SR [15 536] I <sup>2</sup> S 7 50SR [15 7 50SR [15	(1) S (6) S (6) S (6) S (7) S (8) S (8) S (9) S	F Music  AMPLE 5 S1  I <sup>2</sup> S(1) Sa Programm  AMPLE 5 S0  I <sup>2</sup> S(0) Sa Programm  EVED 5 RES	Sample F  RATE  4  SR [15:8  Imple Rate of the second seco	3 Bl e register  3 Bl e register  3 Bl e register  3 Bl e regis	gister. T  2  ster. The er has no expense has no expense 2  RATE	he sample sample sample ffect unit	ole rate of the rate car of the rate can oless I <sup>2</sup> SF	an be programmed for the program	fammed s enable	d from 4 l from 4 l ed.	4 S1SF 4 kHz to 4 S0SR kHz to 55 4 RES	3 2 [7:0] 555.2 kH 3 [7:0] 5.2 kHz	DEFAU  2 Hz in 1  DEFA  2 in 1 h  DEFA  2	1 hertz in  JLT = 10x  1 hertz incr  ULT = 10x  1 ertz incren  AULT = 1  ULT = 1	rements.  (AC44] 0 nents.  (AC44] 0 AC44]
[35] I <sup>2</sup> S 7 51SR [15 [36] I <sup>2</sup> S 7 50SR [15 [37] RI 7 [38] PI 7	(1) S (6) (1) S (6) (1) S (6) S (6) (1) S (6) S (6) S (6) S (7) S (8) S (9) S (9) S (10) S (1	F Music  AMPLE 5 S1 I²S(1) Sa Program  AMPLE 5 S0 I²S(0) Sa Program  VED 5 RES  RAMMA 5 P Progra increm 256 × 1	Sample F RATE 4 SR [15:8 Imple Rate of the second s	ate re  3 3 8 8 8 9 9 9 9 9 9 9 9 9 9 9 9 9 9	gister. T  2  ster. The er has no experience and experience are gister. The has no experience are gister is on SS [32]	he sample sample sample iffect unit	ole rate containess I <sup>2</sup> SF  o  o  o  o  o  o  o  o  o  o  o  o  o	an be pr 7 be progra 7 be progra 7 [1:0] i 7  rate can e COF b the value	fammed s enable 6	d from 4 led.  5  from 4 led.  5  grammer S [32] a	4 S1SF 4 kHz to 4 S0SR kHz to 55 4 RES 4 PCR	3 [7:0] 55.2 kHz 3 [7:0] 5.2 kHz	DEFAU  2 Hz in 1 DEFA  2 in 1 h  DEFA  2 OEFA  2 o 50 k ultiplie	h 1 hertz in  JLT = [0x  1  hertz incr  ULT = [0x  1  ertz increm  AULT = [0x  1  Hz in 1 hertz increm  r factor. Po	0 ements.  (AC44] 0 enerts.  (AC44] 0 enerts.
[35] I <sup>2</sup> S 7 21SR [15 236] I <sup>2</sup> S 7 20SR [15 27 20SR [15:	(1) S (6) (1) S (6) (2) S (6) (3) S (6) (4) S (6) S (6) S (7) S (8) S (9) S (9) S (9) S (10)	F Music  AMPLE 5 S1 I²S(1) Sa Program  AMPLE 5 S0 I²S(0) Sa Program  EVED 5 RES  RAMMA 5 P Progra increm 256 × it Stereo	Sample F  RATE  4  SR [15:8  Imple Rate of the second seco	ate re  3 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8 8	gister. T  2  ster. The er has no expected as no expected as no expected as no expected as since a second as since as si	he sample sample offect unitation of the sample of the sam	ole rate control of the clock when the t	an be programmer of the value iion	fammed s enable 6	d from 4 led.  5  from 4 led.  5  grammee S [32] a DF.	4 S1SF 4 kHz to 4 S0SR kHz to 55 4 RES 4 PCR d from 2. re set for	3 [7:0] 5.2 kHz 3 [7:0] 5 kHz t the mu	DEFAU  2 Hz in 1 DEFA  2 in 1 h  DEFA  2  OEFA  2  O 50 k  ultiplie	h 1 hertz in  JLT = 10x  1 hertz incr  ULT = 10x  1 ertz increm  AULT = 10x  1 Hz in 1 hertz increm  AULT = 10x  AULT = 10x	rements.  AC44] 0 ements.  AC44] 0 extz CLKO 0x8000]
[35] I <sup>2</sup> S 7 51SR [15 [36] I <sup>2</sup> S 7 50SR [15 [37] RI 7 [38] PI 7	(1) S (6) (1) S (6) (2) S (6) (3) S (6) (4) S (6) S (6) S (7) S (8) S (9) S (9) S (9) S (10)	F Music  AMPLE 5 S1 I²S(1) Sa Program  AMPLE 5 S0 I²S(0) Sa Program  VED 5 RES  RAMMA 5 P Progra increm 256 × 1	Sample F RATE 4 SR [15:8 Imple Rate of the second s	ate re  3 Bl e regist 3 Bl e regist 3 Clock I 3 Clock s regis 7. See and I	gister. T  2  ster. The er has no experience and experience are gister. The has no experience are gister is on SS [32]	he sample sample sample iffect unit	ole rate containess I <sup>2</sup> SF  o  o  o  o  o  o  o  o  o  o  o  o  o	an be pr 7 be progra 7 be progra 7 [1:0] i 7  rate can e COF b the value	fammed s enable 6  6  6  6  6  6  6  6  6  6  6  6  6	d from 4 led.  5  from 4 led.  5  grammer S [32] a	4 S1SF 4 kHz to 4 S0SR kHz to 55 4 RES 4 PCR	3 [7:0] 55.2 kHz 3 [7:0] 5.2 kHz 3 [7:0] 5 kHz to the mu	DEFAU  2 Hz in 1 DEFA  2 in 1 h  DEFA  2 OEFA  2 o 50 k ultiplie	h 1 hertz in  JLT = 10x  1  hertz incr  ULT = 10x  1  ertz increm  AULT = 10x  1  Hz in 1 hertz increm  AULT = 10x  1	0 ements.  (AC44] 0 enerts.  (AC44] 0 enerts.

PHONE-OUT Attenuation. The LSB represents -1.5 dB, 0000 = 0 dB and the range is 0 dB to -46.5 dB.

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POM PHONE-OUT Mute. 0 = Unmuted. 1 = Muted. 3DD [3:0] 3D Depth Phat Stereo Enhancement Control. The LSB represents 6 2/3% phase expansion, 0000 = 0% and the range is 0% to 100%. 3DDM 3D Depth Mute. Writing a "1" to this bit has the same affect as writing 0s to 3DD [3:0] bits, and causes the Phat 3D Stereo Enhancement to be turned off. 0 = Phat Stereo is on, 1 = Phat Stereo is off. [40] RESERVED DEFAULT = [0x0000]1 RES RES [41] HARDWARE VOLUME BUTTON MODIFIER DEFAULT = [0xXX1B]0 7 RES **VMU** VUP **VDN** BM [4:0] **Button Modifier** BM [4:0] Volume Down **VUP** Volume Up VMU olume Mute This register contains a Master Yolume attenuation offset, which can be incremented or decremented via the Hardware Volume Pins. This register is summed with the Master Volume attenuation to produce the actual Master Volume DAC attenuation. A momentary grounding of greater than 50 ms on the WOL UR pin will cause a decrement (decrease in Attenuation) in this register. Holding the pin LO for greater than 200 ms will cause an auto-decrement every 200 ms. This is also true for a momentary grounding of the VOL\_DN pin. A momentary grounding of both the VOL\_UP/and VOL\_DN causes a mute and no increment or decrement to occur. When Muted, an unmute is possible by a momentary grounding of both the VOL\_UP and VOL\_DN pins together, a momentary grounding of VOL\_UP (this also causes a volume increase), of molecularly grounding of VOL\_DN (this/also causes a volume decrease) or a write of "0" to the VI bit in SS [BASE+1]. [42] DSP MAILBOX 0 = [0x0000]**EFAUL/**T 3 7 5 0 MB0R [15:8] MB0R [7:0 MB0R [15:0] This register is used to send data and control information to and from the DSP. [43] DSP MAILBOX 1 DEFAULT = [0x0000]MB1R [15:8] MB1R [7:0] MB1R [15:0] This register is used to send data and control information to and from the DSP. [44] POWERDOWN AND TIMER CONTROL DEFAULT = [0x0000]6 3 CPD RES PIW PIR PAA PDA PDP PTB 3D PD3D GPSP RES RES DM The AD1816A supports a timeout mechanism used in conjunction with the Timer Base Count and Timer Current Count registers to generate a power-down interrupt. This interrupt allows software to power down the entire chip by setting the CPD bit. This power-down control feature lets users program a time interval from 1 ms to approximately 1.8 hours in 1 ms increments. Five power-down count reload enable bits are used to reload the Timer Current Count from the Timer Base Count when activity is seen on that particular channel. Programming Example: Generate Interrupt if No ISA Reads or Writes occur within 15 Minutes. 1) Write [SSBASE+0] with 0x0C: Write Indirect address for TIMER BASE COUNT "register 12" 2) Write [SSBASE+2] with 0x28; Write TIMER BASE COUNT with  $(15 \min \times 60 \text{ sec/min} \times 100 \text{ ms}) = 0x2328$ ; Note: PTB = 1, timer decrements every 100 ms 3) Write [SSBASE+3] with 0x23; Write High byte of TIMER BASE COUNT 4) Write [SSBASE+0] with 0x2C; Write Indirect address for POWER-DOWN and TIMER CONTROL register 5) Write [SSBASE+2] with 0x00; Write Low byte of POWER-DOWN and TIMER CONTROL register 6) Write [SSBASE+3] with 0x31; Set Enable bits for PIW and PIR

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8) Write [SSBASE+2] with 0x82; Set the TE (Timer Enable) bit

9) Write [SSBASE+3] with 0x20; Set the TIE (Timer Interrupt Enable) bit

7) Write [SSBASE+0] with 0x01; Write Indirect address for INTERRUPT CONFIG register

DM	DAC	Mute. T	his bit r	nutes the	digita	l DAC o	utput en	tering tl	he analog	g mixer.					
GPSP	Game 0 1	Slow C	oeed Sele Game Po Jame Poi	rt	ts the o	operating	g speed o	of the ga	me port.						
PD3D	Power 0 1	r-Down On Off	3D. Tur	ns off in	ternal I	Phat Ster	reo circui	itry.							
3D	ultima	ate flexib AC outp 3D Ph	oility for out. at Stereo		nd any d for D	combin AC Out									
PTB PDP	Power	r-down o	count rel	oad on I	OSP Po	rt enable	ms, 0 = ed; "1" = Alive (Bit	Reload	d count i		ort enab	oled. DS	SP Port is	s enable	ed when
PIDA	Powel activi	-down d y on (I	count rel S0, I <sup>2</sup> S1,	oad on I FM ør l	Digital A	Activity; SACK).	"1" = R	eload co	ount on l	Digital A	Activity.	Digital	Activity	is defin	ed as any
PAN	Power arralog	r-down α g/input ι	count rel	oad on A (LINE,	nalog CD, S	Activity; YNTH,	"1" - R MIC, PI	eload co	ount on . IN) or N	Analog . //ASTE	Activity. R VOLU	Analog JME un	Activity muting.	is defir	ned as any
PIR	Power logica	down o	count rel inside th	oad on I ie AD 8	SA Rea	id; "1" =	Reload	count o	on ISA re	ead. ISA	Read is	defined	d as a rea	d from	any active
PIW				oad on I ie AD18		ite; 1"\	Reload	count	on ISA v	vrite. IS	A Write	defined —	as a wri	te to an	y active
CPD For Pow	1 0	Power	-Down; -Up	oll the [	SSBAS	E+01 C1	RY hit fo	r "1" be	efore wri	J L	reading	J logic	call devic	- / e /	
	ERSIO		ono ara p	on the L	32212		VI DIC 10		01010 1111		· cuaing		EFAUL	- /	XXXXI
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	
		V	ER [15:	8]						V	'ER [7:0	]			
[46] R	RESERV	VED										]	<b>DEFAU</b>	LT = [0	0x0000]
7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
			RES								RES				

Test register. Should never be written or read under normal operation.

### SB Pro; AdLib Registers

The AD1816A contains sets of ISA Bus registers (ports) that correspond to those used by the SoundBlaster Pro audio card from Creative Labs and the AdLib audio card from AdLib Multimedia. Table IX lists the ISA Bus SoundBlaster Pro registers. Table X lists the ISA Bus AdLib registers. Because the AdLib registers are a subset of those in the SoundBlaster card, you can find complete information on using both of these registers in the *Developer Kit for SoundBlaster Series, 2nd ed. © 1993*, Creative Labs, Inc., 1901 McCarthy Blvd., Milpitas, CA 95035.

Table IX. SoundBlaster Pro ISA Bus Registers

Register Name	ISA Bus Address
Music0: Address (w), Status (r)	(SB Base) Relocatable in range 0x100 - 0x3F0
Music0: Data (w)	(SB Base+1)
Music1: Address (w)	(SB Base+2)
Music1: Data (w)	(SB Base+3)
Mixer Address (w)	(SB Base+4)
Mixer Data (w)	(SB Base+5)
Reset (w)	(SB Base+6)
Music0: Address (w)	(SB Base+8)
Music0: Data (w)	(SB Base+9)
Input Data (r)	(SB Base+A)
Status (r), Output Data (w)	(SB Base+C)
Status (r)	(SB Base+E)

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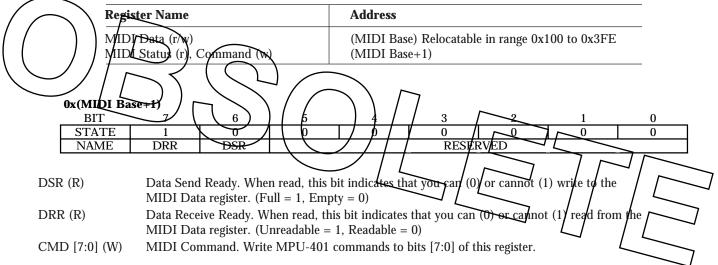
Table X. AdLib ISA Bus Registers

Register Name	ISA Bus Address
Music0: Address (w), Status (r) Music0: Data (w) Music1: Address (w) Music1: Data (w)	(AdLib Base) Relocatable in range 0x100 – 0x3F8 (AdLib Base+1) (AdLib Base+2) (AdLib Base+3)

### **MPU-401 Registers**

The AD1816A contains a set of ISA Bus registers (ports) that correspond to those used by the ISA bus MIDI audio interface cards. Table XI lists the ISA Bus MIDI registers. These registers support commands and data transfers described in *MIDI 1.0 Detailed Specification and Standard MIDI Files 1.0*, © 1994, MIDI Manufacturers Association, PO Box 3173 La Habra, CA 90632-3173.

Table XI. MPU-401 ISA Bus Registers



#### NOTES

The AD1816A supports *only* the MPU-401 0xFF (reset) and 0x3F (UART) commands. The controller powers setup for Smart mode, but must be put in pass-through mode. To start MIDI operations, send a reset command (0xFF) and then send a UART mode command (0x3F). The MPU-401 data register contains an acknowledge byte (0xFE) after each command transfer unless it is in UART mode..

All commands return an ACK byte in "smart" mode.

Status commands (0xAx) return ACK and a data byte; all other commands return ACK.

All commands except reset (0xFF) are ignored in UART mode. No ACK bytes are returned.

"Smart" mode data transfers are not supported.

#### **Game Port Registers**

The AD1816A contains a Game Port ISA Bus Register that is compatible with the IBM joystick standard.

Table XII. Game Port ISA Bus Registers

Register Name	Address
Game Port I/O	(Game Port Base+0 to Game Port Base+7) Relocatable in the range 0x100 to 0x3F8

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#### APPENDIX A

Start dependent function, suboptimal config PLUG AND PLAY INTERNAL ROM Note: All addresses are depicted in hexadecimal notation. IRQ: channel(s) 5 7 9 10 11 15 type(s) active-high, edge-triggered Vendor ID: ADS7181 DMA: channel(s) 0 1 3 Serial Number: FFFFFFF Checksum: 2F DMA: NULL PNP Version: 1.0, vendor version: 20 ASCII string: "Analog Devices AD1816A" Logical Device ID: ADS7180 not a boot device, implements PNP register(s) 31 End all dependent functions Start dependent function, best config Logical Device ID: ADS7181 IRQ: channel(s) 5 7 type(s) active-high, edge-triggered Compatible Device ID: PNPB006 DMA: channel(s) 1 Start dependent function, best config Type F, count-by byte nonbus-mastering, 8-bit only DMA: channel s) 0/1 3 IRQ: channel(s) 5 7 9 11 type(s) active-high, edge-triggered Type F, count-by-byte, nonbus-mastering, 8-bit only I/O: 16-bit/decode, range [0220,0240] mod 20, length 10 I/O: 16-bit decode, range [0388,0388] mod 08, length 04 I/O: 16-bit decode, range [0500 0560] mod 10, length 10, Start dependent function, acceptable config IRQ: channel(s) 5 7 9 10 11 15 type(s) active-high, edge-triggered Start dependent function, acceptable config IRQ: channel(s) 5 7 10 End all dependent functions. type(s) active-high, edge-triggered Logical Device ID: ADS 182 DMA: channel(s) 0 1 3 Type F, count-by-byte, nonbus-mastering, 8-bit only Compatible Device/ID: PNPB02F DMA: channel(s) 0 1 3 Start dependent function, best config Type F, count-by-byte, nonbus-mastering, 8-bit only I/O: 16-bit decode, range [0220,0240] mod 20, length 10 Start dependent function, acceptable config I/O: 16-bit decode, range [0388,0388] mod 08, length 04 I/O: 16-bit decode, range [0500,0560] mod 10, length 10 End all dependent functions Start dependent function, acceptable config End: IRQ: channel(s) 5 7 9 10 11 15 type(s) active-high, edge-triggered DMA: channel(s) 0 1 3 Type F, count-by-byte, nonbus-mastering, 8-bit only DMA: channel(s) 0 1 3 Type F, count-by-byte, nonbus-mastering, 8-bit only I/O: 16-bit decode, range [0220,02E0] mod 20, length 10 I/O: 16-bit decode, range [0388,03B8] mod 08, length 04

I/O: 16-bit decode, range [0500,0560] mod 10, length 10

Type F, count-by-byte, nonbus-mastering, 8-bit only I/O: 16-bit decode, range [0220,02E0] mod 20, length 10 I/O: 16-bit decode, range [0388,03B8] mod 08, length 04 I/O: 16-bit decode, range [0500,0560] mod 10, length 10 not a boot device, implements PNP register(s) 31 I/O: 16-bit decode, range [0300,0330] mod 30, length 02 O: 16-bit decode, range [0300,0420] mod 30, length 02 not a boot device implements PNP register(s) 31 I/O: 16-bit decode, range [0200,0200] mod 08, length 08 I/O: 16-bit decode, range [0200,0208] mod 08, length 08

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### PLUG AND PLAY KEY AND "ALTERNATE KEY" SEQUENCES

One additional feature of the AD1816A is an alternate programming method used, for example, if a BIOS wants to assume control of the AD1816A and present DEVNODES to the OS (rather than having the device participate in Plug and Play enumeration). The following technique may be used.

Instead of the normal 32 byte Plug and Play key sequence, an alternate 126 byte key is used. After the 126 byte key, the AD1816A device will transition to the Plug and Play "sleep" state. It can then be programmed as usual using the standard Plug and Play ports. After programming, the AD1816A should be sent to the Plug and Play "WFK" (wait for key) state. Once the AD1816A has seen the alternate key, it will no longer parse for the Plug and Play key (and therefore never participate in Plug and Play enumeration). It can be reprogrammed by reissuing the alternate key again.

Both the Plug and Play key and the alternate key are sequences of writes to the Plug and Play address register, 0x279. Below are the ISA data values of both keys.

This is the standard Plug and Play sequence:

6a	b5	da	ed	f6	fb	7d	be	df	6f	37	1b	0d	86	c3	61
60	58	2c	16	8b	45	a2	d1	e8	74	3a	9d	ce	e7	73	39
7 his	is the lor	nger, 126	-byte alt	ernate ke	ey. It is g	enerated	d by the	function	:						
$\int f[n+1]$	[] = (f[n])	] > <b>∤</b> 1) <b>/</b> [	(((f[n] <b>\</b>	(f[n] >>	1)) & 0	x01) <<	6) f[0] :	= 0x01							
01	<b>#</b> 0 /	/20/_	_1) )	0,8	_04	02	41	60	30	18	0c	06	43	21	50
\ \\28	/14/	/ 0a	45	<b>6</b> 2 (	ر الر	78	3c	1e	<b>4</b> f	27	13	09	44	22	51
68	34	/ 1/a	4d\	66	73	39	$\frac{5c}{}$	2e	57	2b	15	4a	65	72	79
7c	3e	<u> 5f-</u>	<b>L</b> f )	$-\frac{17}{2}$	\d0	$\binom{05}{50}$	42	$\setminus 61$	770	38	1c	0e	47	23	11
48	24	12	_49/	64	32	59	6c	36/	/ 5b	24	56	6b	35	5a	6d
76 50	7b	3d	5e	6f	-37	116	0d 3b	146	63	3/1	58	$7\frac{2c}{hc}$	16	4b	25 5.2
52 29	69 54	74	3a 55	5d— 6a	<del>5e</del> 75	\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\	<u> </u>	/ 1/1 / /e /	4e 7f	87 L	33	J [19]	07	26	53
29	54	2a	33	oa	73	NA.	-7d	$I^{\rm e} L$	/1	\( \begin{align*}     ali		0f	μ	/	
										/ /		/	/	1 /	$\sim$
									$\overline{}$	[ -	_	/	/	/	
											$\sim$ 7	/	/ /	′ ~	
											7		/	1	7
												_	L		_
															olimits

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#### **AD1816 AND AD1816A COMPATIBILITY**

The AD1816 and AD1816A are pin for pin and functionally compatible. The AD1816A may be dropped directly into an existing AD1816 design. However, the AD1816A has greater pin assignment flexibility to accommodate a wider range of applications and for controlling extra logical devices such as a modem chip set or an Enhanced IDE controller. Pin assignments are controlled by the external EEPROM. Consequently, the optional EEPROM must be reprogrammed to configure the AD1816A.

#### USING AN EEPROM WITH THE AD1816 OR AD1816A

The AD1816 and AD1816A support an optional Plug and Play resource ROM. If present, the ROM must be a two-wire serial device (e.g. Xicor X24C02) and the clock and data lines should be wired to EE CLK and EE DATA pins; pull-up resistors are required on both signals. The EEPROM's A2 and A1 pins (also A0 for 256-byte EEPROMs) must all be tied to ground. The write control pin (WC\*) must be tied to power if you wish to program the EEPROM in place; otherwise, we recommend tying it to ground to prevent accidental writes.

The EEPROM interface logic examines the state of the EE\_CLK pin shortly after RESET is deasserted and whenever the Plug and Play reset register (02h) is written with a value X such that ([X & 1]  $\neq$  0). If an EEPROM is connected, EE\_CLK is pulled high and the EEPROM logic attempts to read the first ROM byte (page 0, byte 0). If EE\_CLK is tied low, the internal ROM is used; in this case EE\_DATA is used to set the state of VOL\_EN, and should also be tied high or low. EE\_CLK is not used as an input at any other time.

The initial part of the ROM is not part of the Plug and Play resource data. It consists of a number of flags that enable optional functionality. The number of flag bytes and the purpose of each <u>bit d</u>epend on whether an AD1816 or an AD1816A is being used.

#### AD1816 FLAG BYTE The AD1816 has a single flag byte that is used as shown below: XTR SIXE XTRA\_E XTRA\_IRQ 1 0 0 MODEM EN VOL vice has an I/O range and an IRQ Program to one to enable the modem logical device. This logical de MODEM\_EN

range has the following requirements:

- Length of eight bytes
- Alignment of eight bytes
- 16-bit address decode

Program to zero to enable I<sup>2</sup>S Port 1.

XTRA\_EN

Program to one to enable the XTRA logical device. This logical device has an I/O range, an optional IRQ, and an optional DMA. The I/O range has the following requirements:

- Length of eight bytes or 16 bytes, selectable by XTRA\_SIZE
- Alignment of eight bytes or 16 bytes, matches length
- 16-bit address decode

Program to zero to enable the DSP serial port.

XTRA\_IRQ

Program to one to include an IRQ in the XTRA logical device. When enabled, the IRQ level and type are programmed through PnP registers 0x70 and 0x71. (Note: For the 1816, the IRQ type is hard coded and rising edge triggered.)

VOL EN

Program to one to enable hardware volume control.

XTRA\_SIZE/ VOL\_SEL

The function of this bit depends on XTRA\_EN. If XTRA\_EN is one, this bit selects the size of the XTRA device's I/O range. Program to one to make the XTRA logical device I/O length 16 bytes. Program to zero to set the XTRA logical device I/O length to eight bytes. The alignment specified in the resource data must be an integer multiple of the length. If XTRA EN is zero (and VOL EN is one), then this bit selects the location of the hardware volume control pins. Program to zero to replace I<sup>2</sup>SO with the volume control pins; program to one to replace the SPORT.

The three MSBs in the first byte of the AD1816 EEPROM are used to verify that the EEPROM data is valid. The bits are compared to the values shown; if a mismatch is found, then the EEPROM will be ignored. The internal ROM will be used to perform PnP enumeration, and the MODEM and XTRA logical devices will not be available. Hardware volume will be enabled on the I<sup>2</sup>SO port. The SPORT is disabled.

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#### **USING THE AD1816 WITHOUT AN EEPROM**

If the EEPROM is absent (EE\_CLK pin = GND), the flags are set as shown below:

MODEM\_EN = XTRA\_EN = XTRA\_IRQ = VOL\_SEL = 0

 $VOL_EN = EE_DATA pin$ 

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#### **AD1816A FLAG BYTES**

The AD1816A has four flag bytes that are used as shown below:

(\*) AD1816-compatible setting.

#### Byte 0

MODEM\_EN

TRA

7	6	5	4	3	2	1	0
1	0	0	XTRA_HV	I <sup>2</sup> S0_HV	SUPER_EN	XTRA_EN	MODEM_EN

Program to one to enable the modem logical device. This logical device has an I/O range and an IRQ.

The I/O range has the following requirements:

- Length of eight bytes

- Alignment of eight bytes

- 16-bit address decode

Program to zero to enable I<sup>2</sup>S Port 1 (SUPER EN and IRQ EN must also be zero).

Program to one to enable the XTRA logical device. This logical device has an I/O range, an optional IRQ, and an optional DMA. The I/O range has the following requirements:

Length of 1 to 16 bytes selectable by XTRASZ0[3:0]

Alignment of 1 to 16 bytes, matches length

16-bit address decode

Program to zero to enable the DSP serial port (XTRA\_HV A second I/O range is available (see

must also be zero

Program to one to merge the XTRX and mode in logical devices. If this bit is set lo lone, XTRA EN and IRQ\_EN SUPER\_EN must be set to one and MODEN. EN must be set to zero. The combined device has up to two I/D ranges, two

IRQs and one DMA. The two I/O ranges are both taken from the XTRA device; the modelm I/O range is disabled. The first IRQ is the XTRA device IRQ, the second is the modern IRQ. Program to zero for distinct modern and

XTRA devices. (\*)

I<sup>2</sup>S0 HV Program to one to enable hardware volume inputs on the I<sup>2</sup>S port 0 pins.

XTRA HV Program to one to enable hardware volume inputs on the DSP serial port pins. Do not enable both XTRA

and I<sup>2</sup>SO\_HV. Program to zero to enable the XTRA device DMA or the DSP serial port.

The three MSBs in the first byte of the AD1816A EEPROM are used to verify that the EEPROM data is valid. The bits are compared to the values shown; if a mismatch is found, the EEPROM will be ignored. The internal ROM will be used to perform PnP enumeration, and the MODEM and XTRA logical devices will not be available. Hardware volume will be enabled on the I<sup>2</sup>S0 port. The SPORT is disabled.

Byte 1

7	6	5	4	3	2	1	0
	RESERVED		0	0	RSTB_EN	IRQSEL3_9	IRQSEL12_13

Program to one to enable IRQ 13. IRQSEL12 13

Program to zero to enable IRQ 12.

IRQ\_EN must be one and MODEM\_EN must be zero, or this bit has no effect.

Program to one to enable IRQ 9. IRQSEL3\_9

Program to zero to enable IRQ 3. (\*)

MODEM\_EN or IRQ\_EN must be one, or this bit has no effect.

RSTB\_EN Program to one to enable an active-low RESET output on the XCTRLO pin.

Program to zero to enable XCTRL0/PCLKO. (\*)

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#### Byte 2

7	6	5	4	3	2	1	0
IRQSEL4_9_11	IRQSEL9_14	IRQSEL11_15	IRQSEL4_10		XTRAS	SZ0[3:0]	

XTRASZ0[3:0] Sets the XTRA device I/O range 0 length. The XTRASZ0 bits set the length of the first XTRA device I/O range as follows:

XTRASZ0	I/O Range Length
0000	16
1000	8
1100	4
1110	2
1111	1

All other combinations should be avoided.

ROSEL4\_10 Program to one to enable IRQ 10. (\*, if MODEM\_EN is zero) Program to zero to enable IRQ 4. (\*, if MODEM\_EN is one)

/ Program to one to enable IRQ 15. (\*) Program to zero to enable IRQ 11.

IRQSEL9\_14 Program to one to enable IRQ 19
Program to zero to enable IRQ 9

IRQSEL4\_9\_11 Program to one to enable IRQ 11. (\*)
Program to zero to enable IRQ 4 (if MODEM\_EN is zero)

Program to zero to enable IRQ 4 (if MODEM\_EN is zero)

Byte 3



XIRQINV Program to one to make LD\_IRQ active-low.

Program to zero to make LD\_IRQ active-high. (\*)

MIRQINV Program to one to make MDM\_IRQ active-low.

Program to zero to make MDM\_IRQ active-high. (\*)

IRQ\_EN Program to one to enable additional IRQ options on the ISA bus. If MODEM\_EN is zero, then two IRQs are

added; if MODEM\_EN is one, this bit is ignored. Program to zero to enable I<sup>2</sup>S port 1 (SUPER\_EN and

MODEM\_EN must also be zero). (\*)

XTRA\_CS Program to one to enable a second I/O range for the XTRA or SUPER logical devices. It is identical to

the first I/O range, except its size is controlled by XTRASZ1[3:0]. Program to zero to enable the XCTR1/

RING\_IN pin. (\*) Always considered to be zero if XTRA\_EN is zero.

XTRASZ1[3:0] Sets the XTRA device I/O range one length. The XTRASZ1 bits set the length of the second XTRA device I/O range as follows:

XTRASZ1	I/O Range Length
0000	16
1000	8
1100	4
1110	2
1111	1

All other combinations should be avoided.

#### **USING THE AD1816A WITHOUT AN EEPROM**

If the EEPROM is absent (EE\_CLK pin = GND), then the flags are set as shown below:

MODEM EN = XTRA EN = SUPER EN = XTRA HV = RSTB EN = IRQ EN = 0

 $IRQSEL9_14 = MIRQINV = XIRQINV = 0$ 

IRQSEL4 10 = IRQSEL11 15 = IRQSEL4 9 11 = 1

 $I^2S0_HV = EE_DATA pin$ 

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#### MAPPING THE AD1816 EEPROM INTO THE AD1816A EEPROM

The equations below map AD1816 flags onto AD1816A flags:

 $MODEM_EN = MODEM_EN$ 

 $XTRA_EN = XTRA_EN$ 

 $SUPER_EN = 0$ 

 $I^2S0_HV = VOL_EN * \overline{VOL_SEL}$ 

 $XTRA_HV = VOL_EN * VOL_SEL$ 

 $IRQSEL12_13 = X$  (don't care)

 $IRQSEL3_9 = 0$ 

 $RSTB_EN = 0$ 

 $XTRASZ0[3] = \overline{XTRA SIZE}$ 

XTRASZ0[2:0] = 000

 $IRQSEL4_10 = \overline{MODEM\_EN}$ 

 $IRQSEL11_15 = 1$ 

 $RQSEL9_14 = 0$ 

 $IRQSEL4_9_11$ XIRQINV = 0

MIRQINV = 0

IRQINV = 0  $IRQ_EN = 0$ 

(TRASZ1[3:0] = XXXX)(dqn't care)

PIN MUXING IN THE AD 816 AND AD 1816A

Some AD1816 and AD1816A options are mutually exclusive because there are a limited number of pins on the device to support them all. The tables below map functions to pin, and show how the flags must be set to assign functions to pins. For each pin, the first function listed is the default; that function is used if the EEPRQM is absent or invalid.

Table XIII. AD1816 Pin Muxing

PQFP	TQFP	Pin Function	I/O	Flags Required
1	99	I <sup>2</sup> S0_DATA	I	VOL_EN + (XTRA_EN* VOL_SEL)
		VOL_UP	I	VOL_EN * (XTRA_EN + VOL_SEL)
2	100	I <sup>2</sup> S0_LRCLK	I	$\overline{VOL}_{EN} + (\overline{XTRA}_{EN} * VOL_{SEL})$
		VOL_DN	I	$VOL_EN * (XTRA_EN + \overline{VOL_SEL})$
3	1	I <sup>2</sup> S0_BCLK	I	$\overline{\text{VOL}_{\text{EN}}} + (\overline{\text{XTRA}_{\text{EN}}} * \text{VOL}_{\text{SEL}})$
		GND	I	$VOL_EN * (XTRA_EN + \overline{VOL_SEL})$
77	75	IRQ(10)	O (1)	MODEM_EN
		IRQ(4)	O (1)	MODEM_EN
81	79	I <sup>2</sup> S1_DATA	I	MODEM_EN
		IRQ(3)	O (1)	MODEM_EN
82	80	I <sup>2</sup> S1_BCLK	I	MODEM_EN
		MDM_IRQ	I	MODEM_EN
83	81	I <sup>2</sup> S1_LRCLK	I	MODEM_EN
		MDM_SEL	O (2)	MODEM_EN
97	95	SPORT_SCLK	О	$\overline{\text{XTRA\_EN}} * (\overline{\text{VOL\_EN}} * \overline{\text{VOL\_SEL}})$
		LD_SEL	0	XTRA_EN
		No Connect	О	XTRA_EN * VOL_EN * VOL_SEL
98	96	SPORT_SDFS	O (2)	$\overline{\text{XTRA\_EN}} * (\overline{\text{VOL\_EN}} * \overline{\text{VOL\_SEL}})$
		LD_DRQ	I	XTRA_EN
		VOL_UP	I	XTRA_EN * (VOL_EN * VOL_SEL)
99	97	SPORT_SDO	0	$\overline{\text{XTRA\_EN}} * \overline{\text{(VOL\_EN * VOL\_SEL)}}$
		LD_DACK	0	XTRA_EN
		No Connect	О	XTRA_EN * VOL_EN * VOL_SEL
100	98	SPORT_SDI	I	XTRA_EN * (VOL_EN * VOL_SEL)
		LD_IRQ	l I	XTRA_EN * XTRA_IRQ
		VOL_DN		XTRA_EN * (VOL_EN * VOL_SEL)
		GND	1	XTRA_EN * XTRA_IRQ

<sup>(1)</sup> IRQ pins are three-stated if not assigned to a logical device.

<sup>(2)</sup> A pull-up or pull-down resistor may be required if EEPROM is used, because this pin is three-stated while EEPROM is read.

Table XIV. AD1816A Pin Muxing

PQFP	TQFP	Pin Function	I/O	Flags Required
1	99	I <sup>2</sup> S0_DATA VOL_UP	I	I <sup>2</sup> S0_HV I <sup>2</sup> S0_HV
2	100	I <sup>2</sup> S0_LRCLK VOL_DN	I I	I <sup>2</sup> S0_HV I <sup>2</sup> S0_HV
3	1	I <sup>2</sup> S0_BCLK GND	I I	I <sup>2</sup> S0_HV I <sup>2</sup> S0_HV
68	66	XCTL0/PCLKO PNPRST	0 0	RSTB_EN RSTB_EN
69	67	XCTL1/RING LD_SEL1	O (1)	XTRA_EN + XTRA_CS XTRA_EN * XTRA_CS
75	73	IRQ(15) IRQ(11)	O (2) O (2)	IRQSEL15_11 IRQSEL15_11
76	)  74	IRQ(0) IRQ(0) IRQ(4)	O (2) O (2) O (2)	IRQSEL4_9_11 IRQSEL4_9_11* MODEM_EN IRQSEL4_9_11* MODEM_EN
77	15	IRQ(10)	0 (2)	IRQSFL4/10 IRQSFL4/10
78	76	IRQ(9) IRQ(14)	0 (2)	
81	79	I <sup>2</sup> S1_DATA IRQ(3)	I O (2)	MODEM_EN * SUPER_EN * JRQ_EN (MODEM_EN) + SUPER_EN + IRQ_EN) + IRQSEL3 9
82	80	IRQ(9) I <sup>2</sup> S1_BCLK MDM_IRQ	O (2) I I	(MODEM_EN + SUPER_EN + IRQ_EN) * IRQSEL3_9 MODEM_EN MODEM_EN
83	81	I <sup>2</sup> S1_LRCLK MDM_SEL IRQ(12)	I O (4) O (2)	MODEM_EN * SUPER_EN * IRQ_EN MODEM_EN *SUPER_EN (MODEM_EN + SUPER_EN) * IRQ_EN * IRQSEL12_13
		IRQ(13)	O (2)	(MODEM_EN + SUPER_EN) * IRQ_EN * IRQSEL12_13
97	95	SPORT_SCLK LD_SEL0 No Connect	0 0 0	XTRA_EN * XTRA_HV XTRA_EN XTRA_EN * XTRA_HV
98	96	SPORT_SDFS LD_DRQ VOL_UP	O (3) I	XTRA_EN * XTRA_HV XTRA_EN * XTRA_HV XTRA_HV
99	97	SPORT_SDO LD_DACK VOL_DN GND	O (3) O (3) I I	XTRA_EN * XTRA_HV XTRA_EN * XTRA_HV (XTRA_EN + XTRA_CS) * XTRA_HV XTRA_EN * XTRA_HV * XTRA_CS
100	98	SPORT_SDI LD_IRQ VOL_DN GND	I I I I	XTRA_EN XTRA_HV XTRA_CS  XTRA_EN * XTRA_HV  XTRA_EN  XTRA_EN * XTRA_HV * XTRA_CS  XTRA_EN * XTRA_HV * XTRA_CS

The direction of some pins (input vs. output) depends on the flags. In order to prevent conflicts on pins that may be both inputs and outputs, the AD1816 and AD1816A disable the output drivers for those pins while the flags are being read from the EEPROM, and keep them disabled if the EEPROM data is invalid.

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<sup>(1)</sup> Open-drain driver with internal weak pull-up.
(2) PC\_IRQ pins are three-stated if not assigned to a logical device.
(3) A pull-up or pull-down resistor may be required if EEPROM is used, because this pin is three-stated while EEPROM is read.

<sup>(4)</sup> An internal pull-up holds this pin deasserted until the EEPROM is read.

#### PROGRAMMING EXTERNAL EEPROMS

Below are the details for programming an external EEPROM or an ADI-supplied PC Program may be used. The PnP EEPROM can be written only in the "Alternate Key State"; this prevents accidental EEPROM erasure when using standard PnP setup. The procedure for writing an EEPROM is:

- 1) Enter PnP configuration state and fully reset the part by writing 0x07 to PnP register 0x02. This step can be eliminated if the part has not been accessed since power-up, a previous full PnP reset or assertion of the ISA bus RESET signal.
- 2) Send the alternate initiation key to the PnP address port. EEPROM writes are disabled if the standard PnP key is used.
- 3) Enter isolation state and write a CSN to enter configuration state. Do not perform any isolation reads.
- 4) Poll PnP register 0x05 until it equals 0x01 and wait at least 336 microseconds (ensures that EEPROM is idle).
- 5) Write the second byte of your serial identifier to PnP register 0x20.
- 6) Read PnP register 0x04.
- 7) Wait for at least 464 microseconds, plus the EEPROM's write cycle time (up to 10 ms for a Xicor X24C02).
- 8) Repeat steps 4 through 7 for each byte in your PnP ROM, starting with the third byte of the serial identifier and ending with the final checksum byte. You must then continue to write filler bytes until 512 bytes, minus one more than the number of flag bytes, have been written. Finally, write the flag byte(s) (described above) and the first byte of the serial identifier.
- (9) Fully reset the part by writing 0x07 to PnP register 0x02.
- The AD18/16 of AD18/16A will now act according to the contents of the EEPROM.

#### NOTES

Programming will not work if more than one part uses the same alternate initiation key in the system. Parts that use this alternate initiation key are the AD1816 and AD1816A.

If a 256-byte EEPROM is used, it is not necessary to wait 10 ms after writing bytes 255 to 511 because the EEPROM will ignore them anyway.

You can skip over bytes that you don't care to write by just performing a ROM read instead of a ROM write followed by a ROM read.

#### REFERENCE DESIGNS AND DEVICE DRIVERS

Reference designs and device drivers for the AD1816A are available via the Analog Devices Home Page of the World Wide Web at http://www.analog.com. Reference designs may also be obtained by contacting your local Analog Devices Sales representative or authorized distributor.

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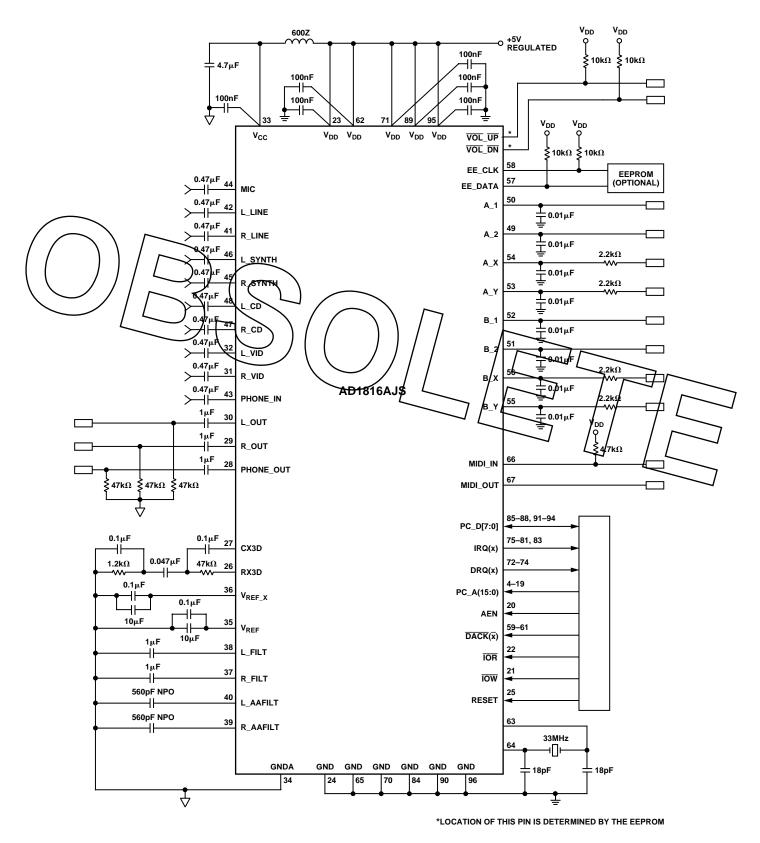


Figure 16. Recommended Application Circuit

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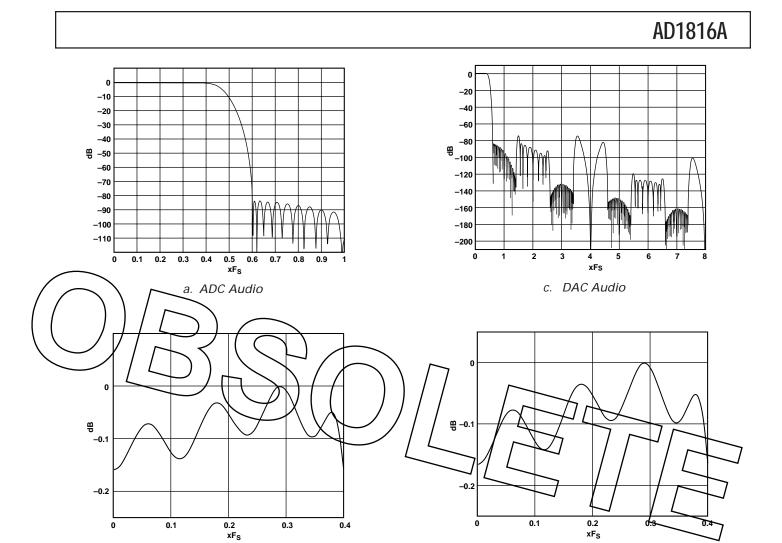


Figure 17. AD1816A Frequency Response Plots (Full-Scale Line-Level Input, 0 dB Gain). The Plots Do Not Reflect the Additional Benefits of the AD1816A Analog Filters. Out-of-Band Images Will Be Attenuated by an Additional 31.4 dB at 100 kHz.

d. DAC Audio Passband (Including Out-of-Band Spectrum)

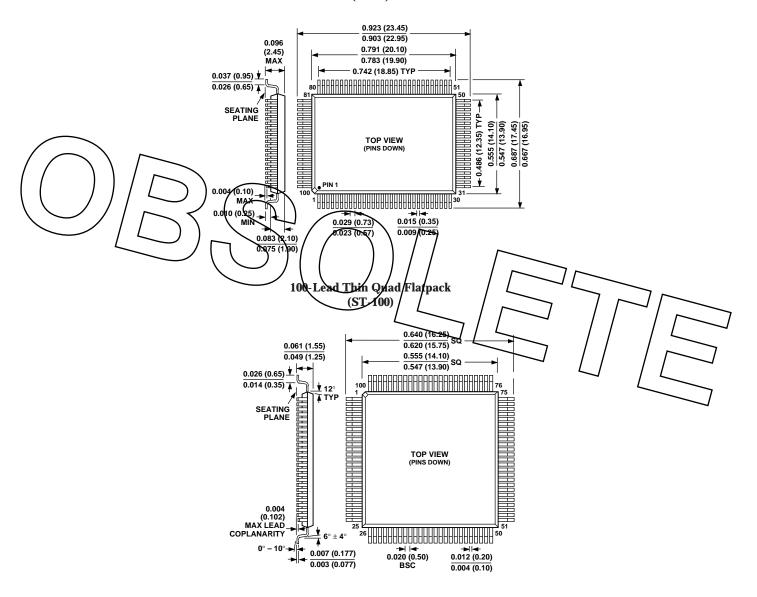
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b. ADC Audio Passband

### **OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).

# 100-Lead Plastic Quad Flatpack (S-100)



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