

# Low Voltage Fault Protection and Detection, 12 $\Omega$ R<sub>ON</sub>, Dual SPST Switch

#### **FEATURES**

- Overvoltage fault protection up to ±60 V on S1 and S2 pins
- ▶ Power-off protection up to ±60 V on S1 and S2 pins
- Overvoltage detection on S1 and S2 pins
- Known state output without digital inputs present
- Low on resistance: 12 Ω typical
- Ultraflat, on resistance: 0.005 Ω typical
- ▶ Low fault detection threshold voltage: 0.1 V typical
- ▶ 3 kV HBM ESD rating
- ▶ Latch-up immune under any circumstance
- ▶ ±1.8 V to ±2.5 V dual supply operation
- ▶ 1.8 V to 5.5 V single-supply operation
- ▶ 10-Lead, 3 mm × 2 mm LFCSP

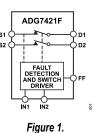
### **APPLICATIONS**

- Analog input/output modules
- Process control/distributed control systems
- Data acquisition
- Instrumentation
- Avionics
- Automatic test equipment
- Communication systems
- Relay replacement

# **COMPANION PRODUCTS**

- ADCs: AD7124-4, AD7124-8
- ► Linear Regulator: ADP162
- Additional companion products on the ADG7421F product page

### FUNCTIONAL BLOCK DIAGRAM



### **GENERAL DESCRIPTION**

The ADG7421F is a low voltage, dual single-pole/single-throw (SPST), low on-resistance switch that features overvoltage protection, power-off protection, and overvoltage detection on the source pins.

When no power supplies are present, the switch remains in the off condition, and the switch inputs are high impedance. When powered, if the analog input signal levels on the Sx pins exceed  $V_{DD}$  or  $V_{SS}$  by a threshold voltage,  $V_T$ , the switch automatically turns off and the digital FF (fault flag) pin drops to a logic low to indicate a fault.

Input signal levels up to +60 V or -60 V relative to ground are blocked, in both the powered and unpowered condition. The switches turn on with a Logic 1 input and conducts equally well in both directions with an analog signal range of  $V_{SS}$  + 0.1 V to  $V_{DD}$  – 0.55 V for a 5 V single supply. The digital input is compatible with 1.8 V logic inputs over the full operating supply range.

The ADG7421F is ideal for providing overvoltage protection for small signals such as resistance temperature device (RTD) inputs (see Figure 56) and thermocouple inputs (see Figure 55). The ability to protect against high voltages up to  $\pm 60$  V coupled with a low voltage supply can enable complete low voltage input stages for industrial applications.

### **PRODUCT HIGHLIGHTS**

- Pin S1 and Pin S2 are protected against voltages greater than the supply rails, up to ±60 V in both powered and unpowered states.
- 2. Overvoltage detection with digital output indicates operating state of switches.
- 3. Trench isolation guards against latch-up.
- The ADG7421F can operate from a dual supply range of ±1.8 V to ±2.5 V or a single-supply range of 1.8 V to 5.5 V.

Rev. 0

DOCUMENT FEEDBACK

**TECHNICAL SUPPORT** 

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# **REVISION HISTORY**

7/2021—Revision 0: Initial Version

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### Table 1. Operating Supply Voltages

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Parameter	Min	Тур	Max	Unit
SUPPLY VOLTAGE				
Dual	±1.8		±2.5	V
Single	1.8		5.5	V

# **5 V SINGLE SUPPLY**

 $V_{DD}$  = 5 V  $\pm$  10%,  $V_{SS}$  = 0 V  $\pm$  10%, and GND = 0 V, unless otherwise noted.

Parameter	+25°C	−40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
ANALOG SWITCH					V <sub>DD</sub> = 4.5 V, V <sub>SS</sub> = 0 V
Analog Signal Range	V <sub>SS</sub> + 0.1 to V <sub>DD</sub> - 0.55			V	
On Resistance, R <sub>ON</sub>	12			Ω typ	Source voltage (V <sub>S</sub> ) = 0.1 V to 3.95 V, source current (I <sub>S</sub> ) = 10 mA
	15	19	23	Ωmax	
On-Resistance Flatness, R <sub>FLAT (ON)</sub>	0.005			Ω typ	$V_{\rm S}$ = 0.1 V to 3.95 V, $I_{\rm S}$ = 10 mA
	0.03	0.05	0.1	Ωmax	
On Resistance Matching, R <sub>MATCH(ON)</sub>	0.01			Ω typ	$V_{\rm S}$ = 0.1 V to 3.95 V, $I_{\rm S}$ = 10 mA
	0.2	0.5	0.7	Ωmax	
LEAKAGE CURRENTS					V <sub>DD</sub> = 5.5 V, V <sub>SS</sub> = 0 V
Source Off Leakage, I <sub>S</sub> (Off)	±0.05			nA typ	$V_{\rm S}$ = 4.5 V to 1 V, drain voltage (V <sub>D</sub> ) = 1 V to 4.5 V
	±0.2	±1.0	±14	nA max	
			±3.5	nA max	-40°C to +105°C
Drain Off Leakage, I <sub>D</sub> (Off)	±0.05			nA typ	$V_{\rm S} = 4.5$ V to 1 V, $V_{\rm D} = 1$ V to 4.5 V
((),	±0.2	±1.0	±14	nA max	
	-0.2		±3.5	nA max	-40°C to +105°C
Channel On Leakage, I <sub>D</sub> (On), I <sub>S</sub> (On)	±0.05		20.0	nA typ	$V_{\rm S} = V_{\rm D} = 1$ V to 4.5 V
	±0.3	±1.5	±15	nA max	
	_0.0		±4.0	nA max	-40°C to +105°C
FAULT			1.0	ni/ ( max	
Threshold Voltage, V <sub>T</sub>	0.1			V typ	
Source Leakage Current, I <sub>S</sub>	0.1			1.05	
With Overvoltage			±120	µA typ	V <sub>DD</sub> = 5.5 V, V <sub>SS</sub> = 0 V, GND = 0 V, V <sub>S</sub> = ±60 V
Power Supplies Grounded or Floating			±100	µA typ	$V_{DD} = 0$ V or floating, $V_{SS} = 0$ V or floating, GND =
			100	prop	0  V, INx = 0 V or floating, V <sub>S</sub> = ±60 V
Drain Leakage Current, I <sub>D</sub>					
With Overvoltage	±0.1			nA typ	V <sub>DD</sub> = 5.5 V, V <sub>SS</sub> = 0 V, GND = 0 V, V <sub>S</sub> = ±60 V
	±0.3	±2.5	±30	nA max	
Power Supplies Grounded	±0.1			nA typ	$V_{DD}$ = 0 V, $V_{SS}$ = 0 V, GND = 0 V, $V_S$ = ±60 V, INx = 0 V
	±0.3	±2.5	±30	nA max	
Power Supplies Floating			±11	µA typ	$V_{DD}$ = floating, $V_{SS}$ = floating, GND = 0 V, $V_{S}$ = ±60 V, INx = 0 V
DIGITAL INPUTS/OUTPUTS				_	
Input Voltage High, V <sub>INH</sub>			1.3	V min	
Input Voltage Low, V <sub>INL</sub>			0.8	V max	
Input Low or High Current, I <sub>INL</sub> or I <sub>INH</sub>	0.7			µA typ	Input voltage (V <sub>IN</sub> ) = 0 V or V <sub>DD</sub>
			1	µA max	
Digital Input Capacitance, C <sub>IN</sub>	5			pF typ	

### Table 2.

Parameter	+25°C	−40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
Output Voltage Low, V <sub>OL</sub>	0.4			V max	Fault flag current (I <sub>FF</sub> ) = 2 mA
DYNAMIC CHARACTERISTICS					
On Time, t <sub>ON</sub>	4.2			µs typ	Load resistance (R <sub>L</sub> ) = 300 $\Omega$ , load capacitance (C <sub>L</sub> ) = 35 pF, V <sub>S</sub> = 3 V
	5.2	5.3	5.3	µs max	
Off Time, t <sub>OFF</sub>	150			ns typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 3 V
	180	180	180	ns max	
Break-Before-Make Time Delay, t <sub>D</sub>	3.5			µs typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 3 V
			2.8	µs min	
Overvoltage Response Time, t <sub>RESPONSE</sub>					
Positive	1.2			µs typ	R <sub>L</sub> = 1 kΩ, C <sub>L</sub> = 5 pF
	1.5	1.6	1.6	µs max	
Negative	1.5			µs typ	Pull-up resistor ( $R_{PULLUP}$ ) = 1 k $\Omega$ , $C_L$ = 5 pF
	1.8	1.9	2	µs max	
Overvoltage Recovery Time, t <sub>RECOVERY</sub>	7.6			µs typ	$R_L = 1 k\Omega, C_L = 5 pF$
	9.5	10	10.3	µs max	
Interrupt Flag Response Time, t <sub>DIGRESP</sub>	500			ns typ	$R_{PULLUP} = 1 \text{ k}\Omega$ , $C_L = 12 \text{ pF}$ , pull-up voltage ( $V_{PULL}_{UP}$ ) = 5 V
	650	650	650	ns max	
Interrupt Flag Recovery Time, t <sub>DIGREC</sub>	1.2			µs typ	$R_{PULLUP}$ = 1 kΩ, C <sub>L</sub> = 12 pF, V <sub>PULL UP</sub> = 5 V
	1.5	1.5	1.5	µs max	
Charge Injection, Q <sub>INJ</sub>	-45	1.0	1.0	pC typ	$V_{\rm S}$ = 2.5 V, source resistor (R <sub>S</sub> ) = 0 $\Omega$ , C <sub>1</sub> = 1 nF
Off Isolation	-65			dB typ	$R_{I} = 50 \Omega, C_{I} = 5 pF$ , frequency (f) = 1 MHz
Channel to Channel Crosstalk	-68			dB typ	$R_{I} = 50 \Omega, C_{I} = 5 pF, f = 1 MHz$
Total Harmonic Distortion Plus Noise, THD + N	-101			dB typ	$R_{L} = 10 \text{ k}\Omega$ , $V_{S} = 3 \text{ V p-p}$ , $f = 20 \text{ Hz to } 20 \text{ kHz}$
	0.009			% typ	
Total Harmonic Distortion, THD	-132			dB typ	R <sub>I</sub> = 10 kΩ, V <sub>S</sub> = 3 V p-p, f = 1 kHz
	-125			dB typ	$R_{L} = 10 \text{ k}\Omega, V_{S} = 3 \text{ V p-p, } f = 20 \text{ kHz}$
	-112			dB typ	$R_{\rm I} = 10 \text{ k}\Omega, V_{\rm S} = 3 \text{ V p-p, } f = 100 \text{ kHz}$
−3 dB Bandwidth	580			MHz typ	
Insertion Loss	-0.92			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
Source Off Capacitance, C <sub>S</sub> (Off)	11			pF typ	$V_{\rm S} = 2.5 \text{ V}, \text{ f} = 1 \text{ MHz}$
Drain Off Capacitance, C <sub>D</sub> (Off)	11			pF typ	$V_{\rm S} = 2.5 \text{ V}, \text{ f} = 1 \text{ MHz}$
Drain On Capacitance, $O_{D}(Ch)$ Drain On Capacitance and Source On Capacitance, $C_{D}(On)$ and $C_{S}(On)$	13			pF typ	$V_{\rm S} = 2.5 \text{ V}, \text{ f} = 1 \text{ MHz}$
Drain On Capacitance and Source On Capacitance Flatness, $C_{DFLAT}$ (On) and $C_{SFLAT}$ (On)	1.6			pF typ	$V_{S}$ = 0.1 V to 3.95 V, f = 1 MHz
Capacitance Matching, C <sub>MATCH</sub> (On)	0.2			pF typ	V <sub>S</sub> = 0.1 V to 3.95 V, f = 1 MHz
POWER REQUIREMENTS					$V_{DD}$ = 5.5 V, $V_{SS}$ = 0 V, GND = 0 V, digital inputs
					V <sub>DD</sub> or GND
Normal Mode					
Positive Supply Current, I <sub>DD</sub>	850			µA typ	
	1100		1100	µA max	
GND Current, I <sub>GND</sub>	745			µA typ	
	950		950	µA max	
Negative Supply Current, I <sub>SS</sub>	105			μA typ	
	150		150	µA max	
Fault Mode				.	V <sub>S</sub> = ±60 V

Table	2.
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Parameter	+25°C	-40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
I <sub>DD</sub>	850			µA typ	
	1100		1100	µA max	
I <sub>GND</sub>	630			µA typ	
	850		850	µA max	
I <sub>SS</sub>	200			µA typ	
	270		270	µA max	

# **3 V SINGLE SUPPLY**

 $V_{DD}$  = 2.7 V to 3.6 V,  $V_{SS}$  = 0 V and GND = 0 V, unless otherwise noted.

Parameter	+25°C	−40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
ANALOG SWITCH					V <sub>DD</sub> = 2.7 V, V <sub>SS</sub> = 0 V
Analog Signal Range	V <sub>SS</sub> + 0.1 to V <sub>DD</sub> - 0.25			V	
R <sub>ON</sub>	12			Ω typ	$V_{\rm S}$ = 0.1 V to 2.45 V, $I_{\rm S}$ = 10 mA
	15	19	23	Ω max	
R <sub>FLAT (ON)</sub>	0.005			Ω typ	$V_{\rm S}$ = 0.1 V to 2.45 V, $I_{\rm S}$ = 10 mA
	0.03	0.05	0.1	Ω max	
R <sub>MATCH(ON)</sub>	0.01			Ω typ	$V_{\rm S}$ = 0.1 V to 2.45 V, $I_{\rm S}$ = 10 mA
	0.2	0.5	0.7	Ω max	
LEAKAGE CURRENTS					V <sub>DD</sub> = 3.6 V, V <sub>SS</sub> = 0 V
I <sub>S</sub> (Off)	±0.05			nA typ	$V_{\rm S}$ = 3.3 V to 1 V, $V_{\rm D}$ = 1 V to 3.3 V
	±0.2	±1.0	±14	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (Off)	±0.05			nA typ	$V_{\rm S}$ = 3.3 V to 1 V, $V_{\rm D}$ = 1 V to 3.3 V
	±0.2	±1.0	±14	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (On), I <sub>S</sub> (On)	±0.05			nA typ	$V_{\rm S} = V_{\rm D} = 1 \text{ V to } 3.3 \text{ V}$
	±0.3	±1.5	±15	nA max	
			±4.0	nA max	-40°C to +105°C
FAULT					
V <sub>T</sub>	0.1			V typ	
I <sub>S</sub>					
With Overvoltage			±120	µA typ	$V_{DD}$ = 3.6 V, $V_{SS}$ = 0 V, GND = 0 V, $V_{S}$ = ±60 V
Power Supplies Grounded or Floating			±100	µA typ	$V_{DD} = 0 V$ or floating, $V_{SS} = 0 V$ or floating, GNE = 0 V, INx = 0 V or floating, $V_S = \pm 60 V$
I <sub>D</sub>					
With Overvoltage	±0.1			nA typ	$V_{DD}$ = 3.6 V, $V_{SS}$ = 0 V, GND = 0 V, $V_{S}$ = ±60 V
	±0.3	±2.5	±30	nA max	
Power Supplies Grounded	±0.1			nA typ	$V_{DD}$ = 0 V, $V_{SS}$ = 0 V, GND = 0 V, $V_S$ = ±60 V, INx = 0 V
	±0.3	±2.5	±30	nA max	
Power Supplies Floating			±11	µA typ	$V_{DD}$ = floating, $V_{SS}$ = floating, GND = 0 V, $V_S$ = ±60 V, INx = 0 V
DIGITAL INPUTS/OUTPUTS					
V <sub>INH</sub>			1.3	V min	
V <sub>INL</sub>			0.5	V max	
I <sub>INL</sub> or I <sub>INH</sub>	0.7			µA typ	V <sub>IN</sub> = 0 V or V <sub>DD</sub>

### Table 3.

Parameter	+25°C	−40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
			1	µA max	
C <sub>IN</sub>	5			pF typ	
V <sub>OL</sub>	0.4			V max	I <sub>FF</sub> = 2 mA
DYNAMIC CHARACTERISTICS					
t <sub>ON</sub>	4			µs typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 1.5 V
	4.8	5	5	µs max	
t <sub>OFF</sub>	170			ns typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 1.5 V
	200	200	210	ns max	
t <sub>D</sub>	3.4			µs typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 1.5 V
			2.7	µs min	
t <sub>response</sub>					
Positive	1.4			µs typ	$R_L = 1 k\Omega$ , $C_L = 5 pF$
	1.7	1.8	1.9	µs max	
Negative	1.7			µs typ	$R_{PULLUP} = 1 k\Omega, C_L = 5 pF$
	2.1	2.3	2.4	µs max	
t <sub>RECOVERY</sub>	7.8			µs typ	$R_L = 1 k\Omega$ , $C_L = 5 pF$
	9.7	10.2	10.5	µs max	
t <sub>DIGRESP</sub>	450			ns typ	$R_{PULLUP} = 1 \text{ k}\Omega, C_L = 12 \text{ pF}, V_{PULL UP} = 3 \text{ V}$
2.0.120	550	550	550	ns max	
t <sub>DIGREC</sub>	1.1			µs typ	$R_{PULLUP} = 1 \text{ k}\Omega, C_L = 12 \text{ pF}, V_{PULL UP} = 3 \text{ V}$
DIONEO	1.4	1.4	1.4	µs max	
Q <sub>INJ</sub>	-40			pC typ	$V_{S} = 1.5 V, R_{S} = 0 \Omega, C_{L} = 1 nF$
Off Isolation	-64			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
Channel to Channel Crosstalk	-68			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
THD + N	-97			dB	$R_{\rm L}$ = 10 kΩ, $V_{\rm S}$ = 1.5 V p-p, f = 20 Hz to 20 kHz
	0.0013			%	
THD	-130			dB	R <sub>I</sub> = 10 kΩ, V <sub>S</sub> = 1.5 V p-p, f = 1 kHz
	-126			dB	$R_{\rm L} = 10 \text{ k}\Omega$ , $V_{\rm S} = 1.5 \text{ V}$ p-p, f = 20 kHz
	-117			dB	R <sub>L</sub> = 10 kΩ, V <sub>S</sub> = 1.5 V p-p, f = 100 kHz
−3 dB Bandwidth	570			MHz typ	$R_L = 50 \Omega, C_L = 5 pF$
Insertion Loss	-0.92			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
C <sub>S</sub> (Off)	12			pF typ	$V_{\rm S} = 1.5 \text{ V}, \text{ f} = 1 \text{ MHz}$
C <sub>D</sub> (Off)	12.5			pF typ	V <sub>S</sub> = 1.5 V, f = 1 MHz
CD (On), CS (On)	13.5			pF typ	V <sub>S</sub> = 1.5 V, f = 1 MHz
C <sub>DFLAT</sub> (On), C <sub>SFLAT</sub> (On)	1.3			pF typ	$V_{S} = 0.1$ V to 2.45 V, f = 1 MHz
C <sub>MATCH</sub> (On)	0.2			pF typ	$V_{S} = 0.1$ V to 2.45 V, f = 1 MHz
POWER REQUIREMENTS					$V_{DD}$ = 3.6 V, $V_{SS}$ = 0 V, GND = 0 V, digital inputs = GND or $V_{DD}$
Normal Mode					
I <sub>DD</sub>	630			µA typ	
	850		850	µA max	
	530			μA typ	
GND	750		750	µA max	
I <sub>SS</sub>	90			µA typ	
-55	140		140	µA max	
Fault Mode					V <sub>S</sub> = ±60 V
I <sub>DD</sub>	630			µA typ	
- טט	850		850	µA max	
	420			µA typ	

#### Table 3.

Parameter	+25°C	-40°C to +85°C	−40°C to +125°C	Unit	Test Conditions/Comments
	600		600	µA max	
I <sub>SS</sub>	180			µA typ	
	250		250	µA max	

# **1.8 V SINGLE SUPPLY**

 $V_{DD}$  = 1.8 V ± 5%,  $V_{SS}$  = 0 V and GND = 0 V, unless otherwise noted.

Parameter	+25°C	-40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
ANALOG SWITCH					V <sub>DD</sub> = 1.71 V, V <sub>SS</sub> = 0 V
Analog Signal Range	V <sub>SS</sub> + 0.1 to V <sub>DD</sub> - 0.1			V	
R <sub>ON</sub>	12			Ω typ	$V_{\rm S}$ = 0.1 V to 1.61 V, $I_{\rm S}$ = 10 mA
	15	19	23	Ω max	
R <sub>FLAT (ON)</sub>	0.005			Ω typ	$V_{\rm S}$ = 0.1 V to 1.61 V, $I_{\rm S}$ = 10 mA
	0.03	0.05	0.1	Ω max	
R <sub>MATCH (ON)</sub>	0.01				$V_{S}$ = 0.1 V to 1.61 V, $I_{S}$ = 10 mA
	0.2	0.5	0.7		
LEAKAGE CURRENTS					V <sub>DD</sub> = 1.95 V, V <sub>SS</sub> = 0 V
I <sub>S</sub> (Off)	±0.05			nA typ	$V_{S}$ = 0.6 V to 1.65 V, $V_{D}$ = 1.65 V to 0.6 V
	±0.2	±1.0	±14.0	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (Off)	±0.05			nA typ	$V_{S}$ = 0.6 V to 1.65 V, $V_{D}$ = 1.65 V to 0.6 V
	±0.2	±1.0	±14.0	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (On), I <sub>S</sub> (On)	±0.05			nA typ	$V_{S}$ = 0.6 V to 1.65 V, $V_{D}$ = 1.65 V to 0.6 V
	±0.3	±1.5	±15.0	nA max	
			±4.0	nA max	-40°C to +105°C
FAULT					
V <sub>T</sub>	0.1			V typ	
I <sub>S</sub>					
With Overvoltage			±120	µA typ	$V_{DD}$ = 1.95 V, $V_{SS}$ = 0 V, GND = 0 V, $V_{S}$ = ±60 V
Power Supplies Grounded or Floating			±100	µA typ	$V_{DD}$ = 0 V or floating, $V_{SS}$ = 0 V or floating, GND = 0 V, INx = 0 V or floating, $V_S$ = ±60 V
I <sub>D</sub>					
With Overvoltage	±0.1			nA typ	$V_{DD}$ = 1.95 V, $V_{SS}$ = 0 V, GND = 0 V, $V_{S}$ = ±60 V
	±0.3	±2.5	±30	nA max	
Power Supplies Grounded	±0.1			nA typ	V <sub>DD</sub> = 0 V, V <sub>SS</sub> = 0 V, GND = 0 V, V <sub>S</sub> = ±60 V, INx = 0 V
	±0.3	±2.5	±30	nA max	
Power Supplies Floating			±11	µA typ	$V_{DD}$ = floating, $V_{SS}$ = floating, GND = 0 V, $V_S$ = ±60 V, INx = 0 V
DIGITAL INPUTS/OUTPUTS					
V <sub>INH</sub>			1.07	V min	
V <sub>INL</sub>			0.5	V max	
I <sub>INL</sub> or I <sub>INH</sub>	0.7			µA typ	V <sub>IN</sub> = 0 V or 5 V
			1	µA max	
C <sub>IN</sub>	5			pF typ	
V <sub>OL</sub>	0.4			V max	I <sub>FF</sub> = 2 mA

#### Table 4.

Parameter	+25°C	−40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
DYNAMIC CHARACTERISTICS					
t <sub>ON</sub>	4.7			µs typ	$R_L$ = 300 Ω, $C_L$ = 35 pF, $V_S$ = 1.5 V
	6.3	7.3	7.3	µs max	
t <sub>OFF</sub>	250			ns typ	R <sub>L</sub> = 300 Ω, C <sub>L</sub> = 35 pF, V <sub>S</sub> = 1.5 V
	360	380	400	ns max	
t <sub>D</sub>	3.9			ns typ	$R_1 = 300 \Omega, C_1 = 35 pF, V_S = 1.5 V$
5			2.9	ns min	
t <sub>RESPONSE</sub>					
Positive	1.7			µs typ	$R_{L} = 1 k\Omega, C_{L} = 5 pF$
	2.1	2.2	2.4	µs max	
Negative	2			µs typ	$R_{PULLUP} = 1 k\Omega, C_L = 5 pF$
	2.5	2.6	2.8	µs max	
t <sub>RECOVERY</sub>	9	2.0	2.0	µs typ	$R_L = 1 k\Omega$ , $C_L = 5 pF$
RECOVERT	14	18.4	18.4	µs max	
t <sub>DIGRESP</sub>	400		10.1	ns typ	R <sub>PULLUP</sub> = 1 kΩ, C <sub>L</sub> = 12 pF, V <sub>PULL UP</sub> = 1.8 V
UIGRESP	450	500	500	ns max	
talaasa	450	000	000	µs typ	R <sub>PULLUP</sub> = 1 kΩ, C <sub>L</sub> = 12 pF, V <sub>PULL UP</sub> = 1.8 V
<sup>t</sup> DIGREC	1.2	1.0	1.3		$\frac{1}{12} \text{ pr, vpull_up - 1.6 v}$
0	-25	1.3	1.3	µs max	$V_{\rm S} = 0.9 \text{ V}, \text{ R}_{\rm S} = 0 \Omega, \text{ C}_{\rm I} = 1 \text{ nF}$
Q <sub>INJ</sub>				pC typ	
Off Isolation	-60			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
Channel to Channel Crosstalk	-68			dB typ	$R_{L} = 50 \Omega$ , $C_{L} = 5 pF$ , f = 1 MHz
THD + N	-97			dB typ	$R_L$ = 10 kΩ, $V_S$ = 1.5 V p-p, f = 20 Hz to 20 kHz
	0.0012			% typ	
THD	-130			dB typ	$R_L$ = 10 kΩ, $V_S$ = 1.5 V p-p, f = 1 kHz
	-121			dB typ	$R_L$ = 10 kΩ, $V_S$ = 1.5 V p-p, f = 20 kHz
	-110			dB typ	R <sub>L</sub> = 10 kΩ, V <sub>S</sub> = 1.5 V p-p, f = 100 kHz
-3 dB Bandwidth	550			MHz	$R_L = 50 \Omega, C_L = 5 pF$
				typ	
Insertion Loss	-0.92			dB typ	$R_L$ = 50 Ω, $C_L$ = 5 pF, f = 1 MHz
C <sub>S</sub> (Off)	13			pF typ	V <sub>S</sub> = 1 V, f = 1 MHz
C <sub>D</sub> (Off)	14			pF typ	V <sub>S</sub> = 1 V, f = 1 MHz
C <sub>D</sub> (On), C <sub>S</sub> (On)	14			pF typ	V <sub>S</sub> = 1 V, f = 1 MHz
C <sub>SFLAT</sub> (On), C <sub>DFLAT</sub> (On)	0.8			pF typ	V <sub>S</sub> = 0.1 V to 1.61 V, f = 1 MHz
C <sub>MATCH</sub> (On)	0.2			pF typ	V <sub>S</sub> = 0.1 V to 1.61 V, f = 1 MHz
OWER REQUIREMENTS					$V_{DD}$ = 1.95 V, $V_{SS}$ = 0 V, GND = 0 V, digital inputs = GND or $V_{DD}$
Normal Mode					
I <sub>DD</sub>	480			µA typ	
	700		700	µA max	
I <sub>GND</sub>	400			µA typ	
	600		600	µA max	
I <sub>SS</sub>	80			µA typ	
66	125		125	µA max	
Fault Mode				For the A	V <sub>S</sub> = ±60 V
I <sub>DD</sub>	520			µA typ	
עטי	800		800	µA typ	
lava	320			µA max µA typ	
I <sub>GND</sub>	550		550	µA typ µA max	
I			550	· ·	
I <sub>SS</sub>	175			µA typ	

#### Table 4.

Parameter	+25°C	−40°C to +85°C	−40°C to +125°C	Unit	Test Conditions/Comments
	245		245	µA max	

### ±2.5 V DUAL SUPPLY

 $V_{DD}$  = 2.5 V  $\pm$  10%,  $V_{SS}$  = –2.5 V  $\pm$  10%, and GND = 0 V, unless otherwise noted.

Table 5.				I	
Parameter	+25°C	-40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
ANALOG SWITCH					V <sub>DD</sub> = 2.25 V, V <sub>SS</sub> = -2.25 V
Analog Signal Range	V <sub>SS</sub> + 0.1 to V <sub>DD</sub> - 0.35			V	
R <sub>ON</sub>	12			Ω typ	$V_{\rm S}$ = -2.15 V to +1.9 V, $I_{\rm S}$ = 10 mA
	15	19	23	max	
R <sub>FLAT (ON)</sub>	0.005			Ω typ	$V_{\rm S}$ = -2.15 V to +1.9 V, $I_{\rm S}$ = 10 mA
	0.03	0.05	0.1	Ω max	
R <sub>MATCH (ON)</sub>	0.01			Ω typ	$V_{\rm S}$ = -2.15 V to +1.9 V, $I_{\rm S}$ = 10 mA
	0.2	0.5	0.7	Ωmax	
LEAKAGE CURRENTS					V <sub>DD</sub> = 2.75 V, V <sub>SS</sub> = -2.75 V
I <sub>S</sub> (Off)	±0.05			nA typ	$V_{\rm S}$ = +2.25 V to -2.25 V, $V_{\rm D}$ = -2.25 V to +2.25 V
	±0.2	±1.0	±14	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (Off)	±0.05			nA typ	$V_{\rm S}$ = +2.25 V to -2.25 V, $V_{\rm D}$ = -2.25 V to +2.25 V
	±0.2	±1.0	±14	nA max	
			±3.5	nA max	-40°C to +105°C
I <sub>D</sub> (On), I <sub>S</sub> (On)	±0.05			nA typ	$V_{\rm S} = V_{\rm D} = -2.25$ V, or $V_{\rm S} = V_{\rm D} = 2.25$ V
	±0.3	±1.5	±15	nA max	
			±4.0	nA max	-40°C to +105°C
FAULT (ON Sx PINS)					
V <sub>T</sub>	0.1			V typ	
I <sub>S</sub>					
With Overvoltage			±120	µA typ	$V_{DD}$ = 2.75 V, $V_{SS}$ = -2.75 V, GND = 0 V, $V_{S}$ = ±60 V
Power Supplies Grounded or Floating			±100	μA typ	$V_{DD}$ = 0 V or floating, $V_{SS}$ = 0 V or floating, GND = 0 V, INx = 0 V or floating, $V_S$ = ±60 V
I <sub>D</sub>					
With Overvoltage	±0.1			nA typ	$V_{DD}$ = 2.75 V, $V_{SS}$ = -2.75 V, GND = 0 V, $V_{S}$ = ±60 V
	±0.3	±2.5	±30	nA max	
Power Supplies Grounded	±0.1			nA typ	$V_{\text{DD}}$ = 0 V, $V_{\text{SS}}$ = 0 V, GND = 0 V, $V_{\text{S}}$ = ±60 V, INx = 0 V
	±0.3	±2.5	±30	nA max	
Power Supplies Floating			±11	μA typ	$V_{DD}$ = floating, $V_{SS}$ = floating, GND = 0 V, $V_S$ = ±60 V, INx = 0 V
DIGITAL INPUTS/OUTPUTS					
V <sub>INH</sub>			1.07	V min	
V <sub>INL</sub>			0.5	V max	
I <sub>INL</sub> or I <sub>INH</sub>	0.7			µA typ	V <sub>IN</sub> = 0 V or 5 V
			1	µA max	
C <sub>IN</sub>	5			pF typ	
V <sub>OL</sub>	0.4			V max	I <sub>FF</sub> = 2 mA
DYNAMIC CHARACTERISTICS					
t <sub>ON</sub>	5.5			µs typ	$R_L$ = 300 $\Omega$ , $C_L$ = 35 pF, $V_S$ = 1.5 V

Parameter	+25°C	-40°C to +85°C	-40°C to +125°C	Unit	Test Conditions/Comments
	6.7	6.8	6.8	µs max	
t <sub>OFF</sub>	130			ns typ	R <sub>L</sub> = 300 Ω, C <sub>L</sub> = 35 pF, V <sub>S</sub> = 1.5 V
	170	170	170	ns max	
t <sub>D</sub>	4.9			µs typ	R <sub>L</sub> = 300 Ω, C <sub>L</sub> = 35 pF, V <sub>S</sub> = 1.5 V
2			4	µs min	
t <sub>RESPONSE</sub>					
Positive	1.3			µs typ	$R_{I} = 1 k\Omega, C_{I} = 5 pF$
	1.5	1.6	1.7	µs max	
Negative	2			µs typ	$R_{I} = 1 k\Omega, C_{I} = 5 pF$
·	2.4	2.6	2.7	µs max	
tRECOVERY	7.4			µs typ	$R_L = 1 k\Omega$ , $C_L = 5 pF$
	9.5	9.9	10.1	µs max	
t <sub>DIGRESP</sub>	500			ns typ	$R_{PULLUP}$ = 1 kΩ, C <sub>L</sub> = 10 pF, V <sub>PULL UP</sub> = 2.5 V
	550	550	600	ns max	
t <sub>DIGREC</sub>	800			ns typ	R <sub>PULLUP</sub> = 1 kΩ, C <sub>L</sub> = 10 pF, V <sub>PULL UP</sub> = 2.5 V
5101120	1.0	1.0	1.0	µs max	
Q <sub>INJ</sub>	-50			pC typ	$V_{S} = 0 V, R_{S} = 0 \Omega, C_{L} = 1 nF$
Off Isolation	-70			dB typ	$R_{I} = 50 \Omega, C_{I} = 5 pF, f = 1 MHz$
THD + N	-101			dB typ	$R_L = 10 \text{ k}\Omega$ , $V_S = 3 \text{ V} \text{ p-p}$ , f = 20 Hz to 20 kHz
	0.009			% typ	
THD	-133			dB typ	R <sub>I</sub> = 10 kΩ, V <sub>S</sub> = 3 V p-p, f =1 kHz
	-131			dB typ	$R_{\rm L} = 10 \text{ k}\Omega$ , $V_{\rm S} = 3 \text{ V}$ p-p, f = 20 kHz
	-114			dB typ	R <sub>L</sub> = 10 kΩ, V <sub>S</sub> = 3 V p-p, f = 100 kHz
−3 dB Bandwidth	590			MHz typ	$R_{\rm I} = 50 \ \Omega, C_{\rm I} = 5 \ pF$
Insertion Loss	-0.92			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
Channel to Channel Crosstalk	-68			dB typ	$R_{L} = 50 \Omega, C_{L} = 5 pF, f = 1 MHz$
C <sub>S</sub> (Off)	11			pF typ	$V_{S} = 0 V, f = 1 MHz$
C <sub>D</sub> (Off)	11			pF typ	V <sub>S</sub> = 0 V, f = 1 MHz
CD (On), C <sub>S</sub> (On)	13			pF typ	V <sub>S</sub> = 0 V, f = 1 MHz
C <sub>SFLAT</sub> (On), C <sub>DFLAT</sub> (On)	1.2			pF typ	V <sub>S</sub> = -2.15 V to +1.9 V, f = 1 MHz
C <sub>MATCH</sub> (On)	0.2			pF typ	V <sub>S</sub> = -2.15 V to +1.9 V, f = 1 MHz
OWER REQUIREMENTS					$V_{DD}$ = 2.75 V, $V_{SS}$ = -2.75 V, GND = 0 V, digital inputs = GND or $V_{DD}$
Normal Mode					
I <sub>DD</sub>	580			µA typ	
	850		850	A max	
I <sub>GND</sub>	480			μA typ	
	700		700	μA max	
I <sub>SS</sub>	100			μA typ	
	150		150	µA max	
Fault Mode					V <sub>S</sub> = ±60 V
I <sub>DD</sub>	630			μA typ	
	850		850	A max	
I <sub>GND</sub>	405			µA typ	
	600		600	µA max	
I <sub>SS</sub>	210			µA typ	
	290		290	µA max	

# CONTINUOUS CURRENT PER CHANNEL, SX OR DX, TWO CHANNELS ON

Table 6.							
Parameter	25°C	85°C	125°C	Unit	Test Conditions/Comments		
CONTINUOUS CURRENT, Sx OR Dx							
$\theta_{JA} = 170^{\circ}C/W$	80	56	38	mA max	V <sub>S</sub> = analog signal range <sup>1</sup>		

<sup>1</sup> The analog signal range for each supply range is given in Table 2 to Table 5.

# **ABSOLUTE MAXIMUM RATINGS**

#### $T_A = 25^{\circ}C$ , unless otherwise noted.

#### Table 7.

Parameter	Value
V <sub>DD</sub> to V <sub>SS</sub>	7 V
V <sub>DD</sub> to GND	-0.3 V to +7 V
V <sub>SS</sub> to GND	-3.5 V to +0.3 V
Sx to GND	-60 V to +60 V
Sx to V <sub>DD</sub> or V <sub>SS</sub>	67 V
Sx to Dx	67 V
Dx <sup>1</sup>	$V_{SS}$ – 0.7 V to $V_{DD}$ + 0.7 V or 30 mA, whichever occurs first
Digital Inputs	GND – 0.7 V to +6 V or 30 mA, whichever occurs first
Peak Current, Sx or Dx Pin	255 mA (pulsed at 1 ms, 10% duty cycle maximum)
Continuous Current, Sx or Dx Pin	Data + 15% <sup>2</sup>
Fault Trip Frequency <sup>3</sup>	1 kHz
Digital Output	GND – 0.7 V to +6 V or 30 mA, whichever occurs first
Temperature	
Operating Range	-40°C to +125°C
Storage Range	-65°C to +150°C
Junction	150°C
Reflow Soldering Peak, Pb-Free	As per JEDEC J-STD-020

<sup>1</sup> Overvoltages at the D1 and D2 pins are clamped by internal diodes. Limit current to the maximum ratings given.

- <sup>2</sup> See Table 6.
- <sup>3</sup> Each fault trip causes a short duration peak current. Limit the fault trip frequency to the maximum ratings given.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

### THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

 $\theta_{JA}$  is the natural convection, junction to ambient thermal resistance measured in a one cubic foot sealed enclosure.

 $\theta_{JC}$  is the junction to case thermal resistance.

#### Table 8. Thermal Resistance

Package Type <sup>1</sup>	θ <sub>JA</sub>	θ <sub>JC</sub>	Unit
CP-10-16	170	58.2	°C/W

<sup>1</sup> Thermal impedance simulated values are based on a JEDEC 2S2P thermal test board with four thermal vias. See JEDEC JESD-51.

# **ELECTROSTATIC DISCHARGE (ESD) RATINGS**

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

Field induced charged device model (FICDM) per ANSI/ESDA/JE-DEC JS-002.

### **ESD Ratings for ADG7421F**

#### Table 9. ADG7421F, 10-Lead LFCSP

ESD Model	Withstand Threshold (kV)	Class
HBM <sup>1</sup>	±3	2
FICDM	±1.250	C3

<sup>1</sup> This is the HBM for the input/output port to supplies, the input/output port to input/output port, and for all other pins.

### ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

### PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



#### Figure 2. Pin Configuration

Table 10. I	Table 10. Pin Function Descriptions					
Pin No.	Mnemonic	Description				
1	S1	Overvoltage Protected Source Terminal. The S1 pin can be an input or an output.				
2	S2	Overvoltage Protected Source Terminal. The S2 pin can be an input or an output.				
3	FF	Fault Flag Digital Output. The FF pin is an open-drain output that requires an external pull-up resistor. This digital output pulls low when a fault condition occurs on either the S1 or S2 input.				
4	GND	Ground (0 V) Reference.				
5	V <sub>DD</sub>	Most Positive Power Supply Potential.				
6	V <sub>SS</sub>	Most Negative Power Supply Potential.				
7	IN2	Logic Control Input.				
8	IN1	Logic Control Input.				
9	D2	Drain Terminal. The D2 pin can be an input or an output.				
10	D1	Drain Terminal. The D1 pin can be an input or an output.				

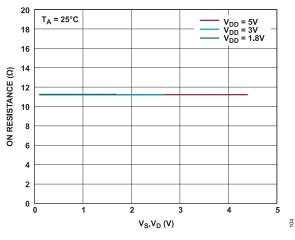


Figure 3. On Resistance as a Function of V<sub>S</sub>, V<sub>D</sub> (Single Supply)

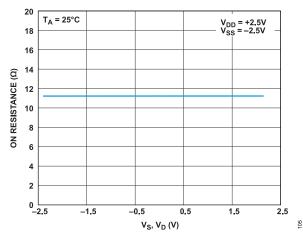


Figure 4. On Resistance as a Function of V<sub>S</sub>, V<sub>D</sub> (2.5 V Dual Supply)

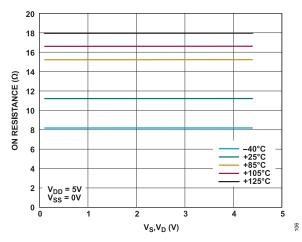


Figure 5. On Resistance as a Function of V<sub>S</sub>, V<sub>D</sub> for Different Temperatures, 5 V Single Supply

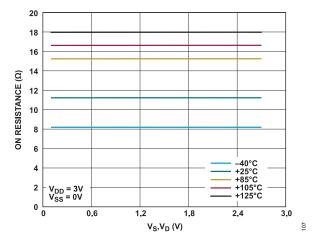


Figure 6. On Resistance as a Function of  $V_S$ ,  $V_D$  for Different Temperatures, 3 V Single Supply

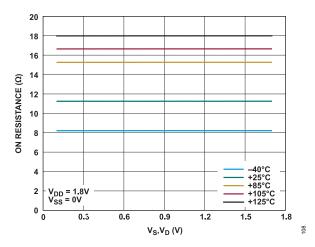


Figure 7. On Resistance as a Function of V<sub>S</sub>, V<sub>D</sub> for Different Temperatures, 1.8 V Single Supply

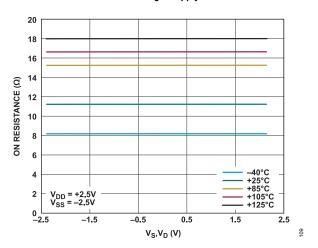


Figure 8. On Resistance as a Function of V<sub>S</sub>, V<sub>D</sub> for Different Temperatures, 2.5 V Dual Supply

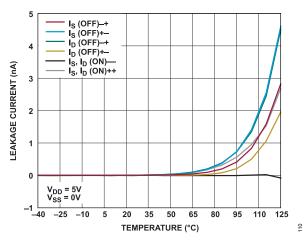


Figure 9. Leakage Current vs. Temperature, 5 V Single Supply

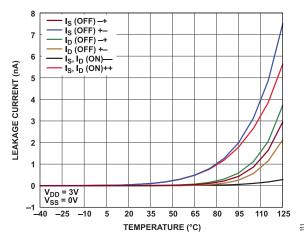


Figure 10. Leakage Current vs. Temperature, 3 V Single Supply

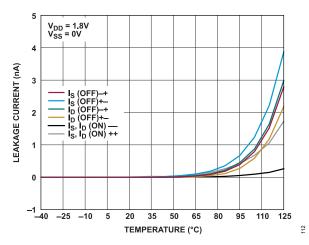


Figure 11. Leakage Current vs. Temperature, 1.8 V Single Supply

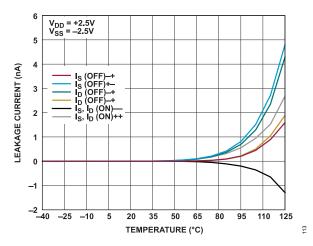


Figure 12. Leakage Current vs. Temperature, 2.5 V Dual Supply

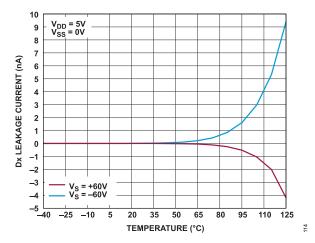


Figure 13. Dx Leakage Current vs. Temperature During Overvoltage, 5 V Single Supply

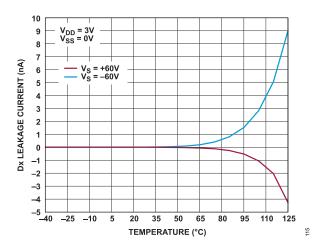


Figure 14. Dx Leakage Current vs. Temperature During Overvoltage, 3 V Single Supply

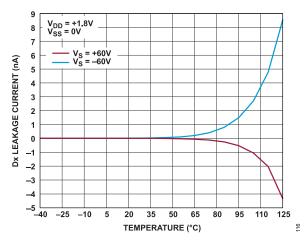


Figure 15. Dx Leakage Current vs. Temperature During Overvoltage, 1.8 V Single Supply

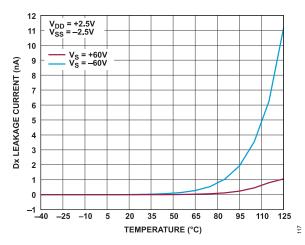


Figure 16. Dx Leakage Current vs. Temperature During Overvoltage, 2.5 V Dual Supply

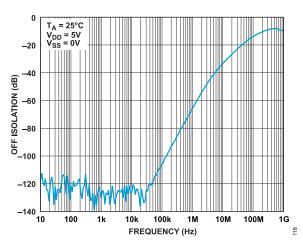


Figure 17. Off Isolation vs. Frequency, 5 V Single Supply

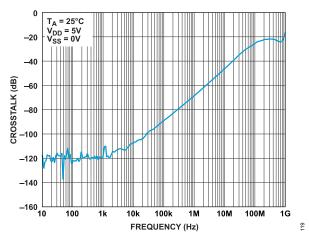


Figure 18. Crosstalk vs. Frequency, 5 V Single Supply

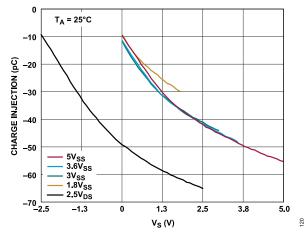


Figure 19. Charge Injection vs. V<sub>S</sub>

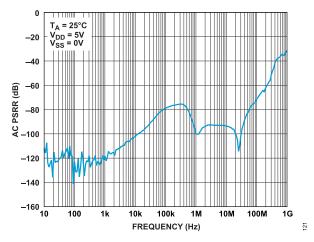
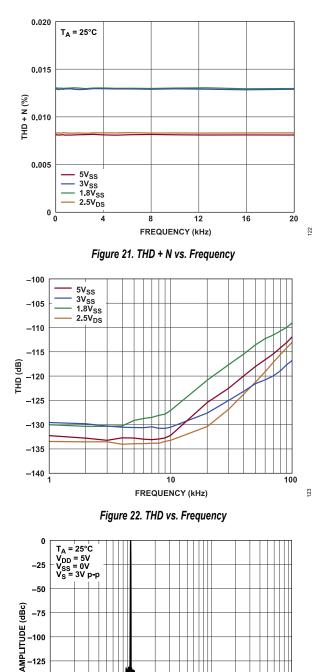
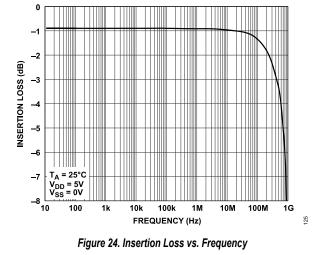


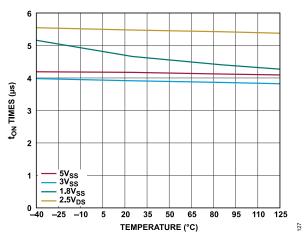
Figure 20. AC Power Supply Rejection Ratio (PSRR) vs. Frequency, 5 V Single Supply

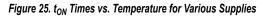


100k

124







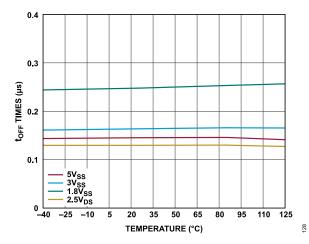


Figure 26. t<sub>OFF</sub> Times vs. Temperature for Various Supplies

-150

-175

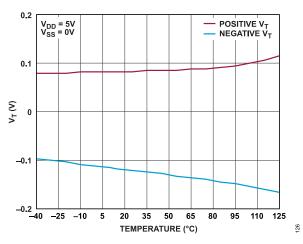
–200 └─ 100

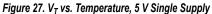
1k

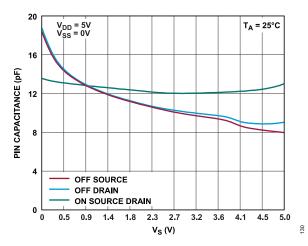
10k

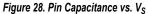
FREQUENCY (Hz)

Figure 23. THD FFT









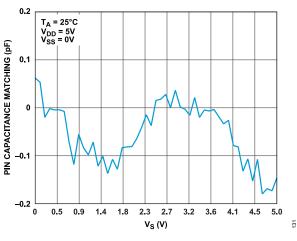


Figure 29. Pin Capacitance Matching vs. V<sub>S</sub>

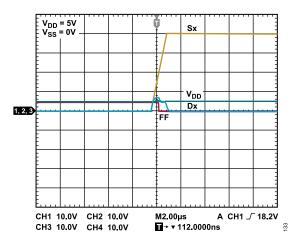


Figure 30. Drain Output Response to Positive Overvoltage

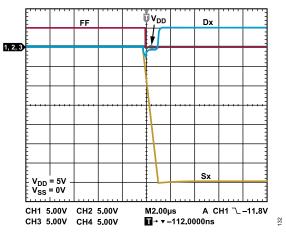


Figure 31. Drain Output Response to Negative Overvoltage

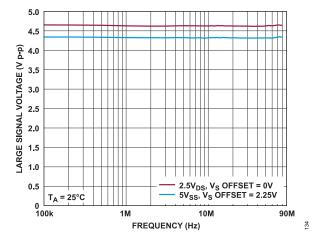


Figure 32. Large Voltage Signal Voltage vs. Frequency

# **TEST CIRCUITS**

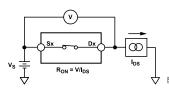


Figure 33. On Resistance (IDS Is the Drain to Source Current)

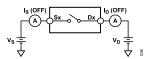


Figure 34. Off Leakage

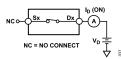


Figure 35. On Leakage

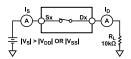


Figure 36. Switch Overvoltage Leakage

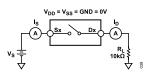


Figure 37. Switch Unpowered Leakage

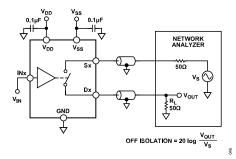


Figure 38. Off Isolation ( $V_{OUT}$  Is the Output Voltage)

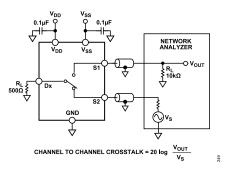


Figure 39. Channel to Channel Crosstalk

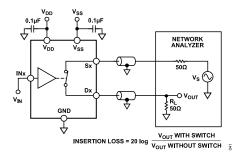


Figure 40. Bandwidth

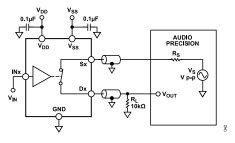


Figure 41. THD + N

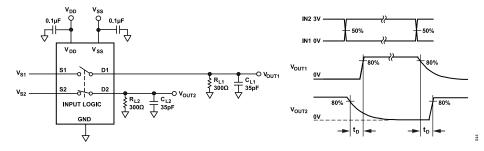


Figure 42. Break-Before-Make Time Delay, t<sub>D</sub>

# **TEST CIRCUITS**

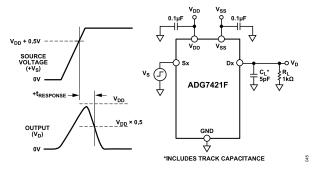


Figure 43. Overvoltage Response Time, t<sub>RESPONSE</sub>

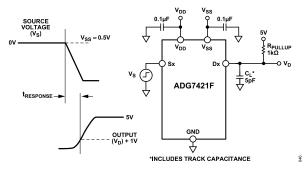
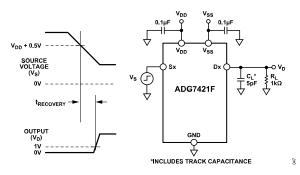


Figure 44. Negative Overvoltage Response Time, Single-Supply, t<sub>RESPONSE</sub>





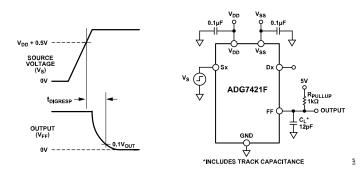


Figure 46. Interrupt Flag Response Time, t<sub>DIGRESP</sub> (V<sub>FF</sub> Is the Fault Flag Voltage)

# **TEST CIRCUITS**

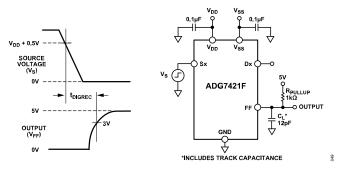


Figure 47. Interrupt Flag Recovery Time, t<sub>DIGREC</sub>

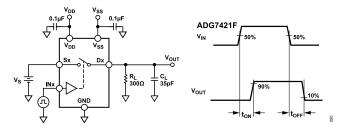


Figure 48. Switching Times, t<sub>ON</sub> and t<sub>OFF</sub>

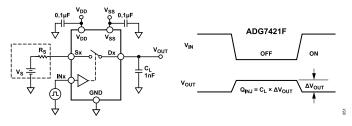


Figure 49. Charge Injection, Q<sub>INJ</sub>

# TERMINOLOGY

# I<sub>DD</sub>

 $\mathsf{I}_{\mathsf{D}\mathsf{D}}$  represents the positive supply current.

# I<sub>SS</sub>

 $I_{SS}$  represents the negative supply current.

# $V_D, V_S$

 $V_{\text{D}}$  and  $V_{\text{S}}$  represent the analog voltage on the Dx pins and the Sx pins, respectively.

# R<sub>ON</sub>

 $\mathsf{R}_{\mathsf{ON}}$  represents the ohmic resistance between the Dx pins and the Sx pins.

# R<sub>FLAT (ON)</sub>

 $R_{FLAT\ (ON)}$  is the flatness that is defined as the difference between the maximum and minimum value of on resistance measured over the specified analog signal range.

# I<sub>S</sub> (Off)

 ${\sf I}_{\sf S}$  (Off) is the source leakage current with the switch off.

# I<sub>D</sub> (Off)

 $I_D$  (Off) is the drain leakage current with the switch off.

# I<sub>D</sub> (On), I<sub>S</sub> (On)

 $I_{\text{D}}\left(\text{On}\right)$  and  $I_{\text{S}}\left(\text{On}\right)$  represent the channel leakage currents with the switch on.

# V<sub>INL</sub>

VINL is the maximum input voltage for Logic 0.

# V<sub>INH</sub>

V<sub>INH</sub> is the minimum input voltage for Logic 1.

# $I_{\rm INL}, I_{\rm INH}$

 $\mathbf{I}_{\text{INL}}$  and  $\mathbf{I}_{\text{INH}}$  represent the low and high input currents of the digital inputs.

# C<sub>D</sub> (Off)

 $C_{\text{D}}$  (Off) represents the off switch Dx pin capacitance, which is measured with reference to ground.

# C<sub>S</sub> (Off)

 ${\rm C}_{\rm S}$  (Off) represents the off switch Sx pin capacitance, which is measured with reference to ground.

# $C_D$ (On), $C_S$ (On)

 $C_D\left(On\right)$  and  $C_S\left(On\right)$  represent on switch capacitances, which are measured with reference to ground.

C<sub>IN</sub> is the digital input capacitance.

# t<sub>ON</sub>

 $t_{\text{ON}}$  represents the delay between applying the digital control input and the output switching on (see Figure 48).

# toff

 $t_{\sf OFF}$  represents the delay between applying the digital control input and the output switching off (see Figure 48).

### t<sub>DIGRESP</sub>

 $t_{\text{DIGRESP}}$  is the time required for the FF pin to go low (0.3 V), measured with respect to voltage on the source pin exceeding the supply voltage by 0.5 V.

# t<sub>DIGREC</sub>

 $t_{\text{DIGREC}}$  is the time required for the FF pin to return high, measured with respect to voltage on the Sx pin falling below the supply voltage plus 0.5 V.

# **t**<sub>RESPONSE</sub>

 $t_{\mbox{RESPONSE}}$  represents the delay between the source voltage exceeding the supply voltage by 0.5 V and the drain voltage falling to 90% of the supply voltage.

### **t**<sub>RECOVERY</sub>

 $t_{\sf RECOVERY}$  represents the delay between an overvoltage on the Sx pin falling below the supply voltage plus 0.5 V and the drain voltage rising from 0 V to 10% of the supply voltage.

# **Off Isolation**

Off isolation is a measure of unwanted signal coupling through an off switch.

# **Charge Injection**

Charge injection is a measure of the glitch impulse transferred from the digital input to the analog output during switching.

# -3 dB Bandwidth

-3~dB bandwidth is the frequency at which the output is attenuated by 3 dB.

# **On Response**

On response is the frequency response of the on switch.

# Insertion Loss

Insertion loss is the loss due to the on resistance of the switch.

# TERMINOLOGY

### THD + N

THD + N is the ratio of the harmonic amplitude plus noise of the signal to the fundamental.

# AC Power Supply Rejection Ratio (AC PSRR)

AC PSRR is the ratio of the amplitude of signal on the output to the amplitude of the modulation. AC PSRR is a measure of the ability of

the device to avoid coupling noise and spurious signals that appear on the supply voltage pin to the output of the switch. The dc voltage on the device is modulated by a sine wave of 0.62 V p-p.

# VT

 $V_T$  is the voltage threshold at which the overvoltage protection circuitry engages (see Figure 27).

# THEORY OF OPERATION

# SWITCH ARCHITECTURE

The ADG7421F consists of two switch channels of N channel diffused metal-oxide semiconductor (DMOS) transistors. This construction provides excellent R<sub>ON</sub> performance in a small area. The ADG7421F operates as a standard switch when input signals within the analog signal range are applied. For example, the on resistance is 12  $\Omega$  typically, and the INx pins control when the switches open or close.

Additional internal circuitry enables the switches to detect overvoltage inputs by comparing the voltage on both the S1 and S2 pins with the V<sub>DD</sub> and V<sub>SS</sub> pins. A signal is considered overvoltage when the signal exceeds the supply voltages by V<sub>T</sub>. V<sub>T</sub> is typically 0.1 V but can change over temperature. See Figure 27 to see the change in V<sub>T</sub> with the operating temperature.

When an overvoltage condition is detected on either the S1 or S2 pins, the switch automatically opens regardless of the digital logic state (INx). The S1 to D1 and S2 to D2 pins become high impedance and ensure that no current flows through the switches. In Figure 30, the voltage on the Dx pin follows the voltage on the Sx pins until the main channel switch turns off completely, and the drain voltage discharges through the load. The maximum voltage on the drain is limited by the internal ESD diodes and the rate at which the output voltage discharges is dependent on the load at the Dx pins.

The maximum voltage that can be applied to any source input is +60 V or -60 V. During overvoltage conditions, the leakage current into and out of the Sx pins is 120  $\mu$ A typical and only 30 nA maximum for the Dx pins. This limit protects the switches and connected circuitry from overstresses and restricts the current drawn from the signal source.

# **ESD** Performance

The ADG7421F has an ESD rating of 3 kV for the HBM.

The Dx pins have ESD protection diodes to the rails and the voltage at these pins must not exceed the supply voltage. The Sx pins have specialized ESD protection that allows the signal voltage to reach  $\pm 60$  V regardless of the supply voltage level. See Figure 50 for the switch channel overview.

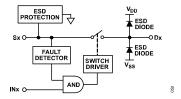


Figure 50. Switch Channel and Control Function

# **Trench Isolation**

In the ADG7421F, an insulating oxide layer (trench) is placed between the N channel DMOS (NDMOS) and the P channel DMOS

(PDMOS) transistors in the circuit. Parasitic junctions that occur between the transistors in the junction isolated switches are eliminated, and the result is a switch that is latch-up immune under all circumstances. These devices pass the JESD78D latch-up test.

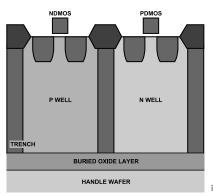


Figure 51. Trench Isolation

# **OVERVOLTAGE FAULT PROTECTION**

When the voltage at the Sx inputs exceeds  $V_{DD}$  or  $V_{SS}$  by  $V_T$ , the switches turn off or, if the device is unpowered, the switches remain off. Both switch inputs remain high impedance regardless of the digital input state or the load resistance, and the output acts as a virtual open circuit. Signal levels up to +60 V and -60 V are blocked in both the powered and unpowered condition.

# **Power-On Protection**

To activate the switches, the three following conditions must be met:

- ▶ The minimum supply operating conditions in Table 1.
- ▶ The input signal must be within the analog signal range.
- ▶ The digital logic control input, INx, is on.

When the switches are on, signal levels within the analog signal range are passed.

When the voltage on either of the Sx pins exceeds  $V_{DD}$  or  $V_{SS}$  by  $V_T$ , the switches respond by turning off. The absolute input voltage limits are –60 V and +60 V. The switches remain off until the  $V_S$  voltage at the Sx pins returns to within the analog signal range.

When powered by the 5 V single supply, the positive overvoltage response time ( $t_{RESPONSE}$ ) is typically 1.2 µs, and  $t_{RECOVERY}$  is 7.6 µs. These values vary with different supply voltage and output load conditions.

Exceeding ±60 V on either the Sx inputs may damage the ESD protection circuitry on the ADG7421F.

### **Power-Off Protection**

When no power supplies are present, the switches remain in an off state and the switch inputs are high impedance. This state

# THEORY OF OPERATION

ensures that no current flows and prevents damage to the switches or downstream circuitry. The switch outputs are open circuit.

The switches remain off regardless of whether the V<sub>DD</sub> and V<sub>SS</sub> supplies are 0 V or floating. A 0 V GND reference must always be present to ensure proper operation. Signal levels of up to  $\pm 60$  V are blocked when the switches are powered off.

### **Overvoltage Interrupt Flag**

The voltages on the Sx inputs of the ADG7421F are continuously monitored, and the open-drain output pin, FF, indicates the fault state.

The voltage on the FF pin indicates if either of the Sx input pins is experiencing a fault condition. The FF pin is an open-drain output that requires an external pull-up resistor. The output of the FF pin is high impedance when both the Sx pins are within the normal operating range. If the voltage of either Sx pin exceeds the supply voltage ( $V_{DD}$  or  $V_{SS}$ ) by  $V_T$ , the FF output provides a low impedance path to GND.

The ADG7421F overvoltage protected switches provide a robust solution for instrumentation, industrial, aerospace, and other harsh environments where overvoltage signals can be present, and the system must remain operational after an overvoltage occurs.

### POWER SUPPLY RAILS

To guarantee correct operation of the device, 0.1  $\mu F$  decoupling capacitors are required on both  $V_{DD}$  and  $V_{SS}$  to GND.

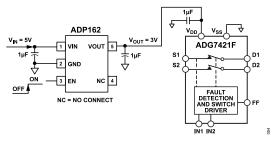
The ADG7421F can operate with bipolar supplies between ±1.8 V and ±2.5 V. Note that the V<sub>DD</sub> and V<sub>SS</sub> supplies do not have to be symmetrical, but the supply voltage range must not exceed 5.5 V. The ADG7421F can also operate with single supplies between 1.8 V and 5.5 V with V<sub>SS</sub> connected to GND.

The ADG7421F is fully specified at the +5 V, +3 V , +1.8 V, and  $\pm 2.5$  V supply ranges.

# POWER SUPPLY RECOMMENDATIONS

Analog Devices, Inc., has a wide range of power management products to meet the requirements of most high performance signal chains.

An example of a unipolar power solution is shown in Figure 52. The ADP162 ultralow quiescent current, 150 mA, CMOS linear regulator generates a positive supply rail for the ADG7421F amplifier and/or a precision converter in a typical signal chain.





# POWER SUPPLY SEQUENCING PROTECTION

When the device is off, the switch channels remain open and the signals from -60 V to +60 V can be applied without damaging the device. The switch channels only close when the supplies are connected, a suitable digital control signal is placed on the INx pin, and the signal is within the normal operating range. Note that placing the ADG7421F between external connectors and sensitive components offers protection in systems where a signal is presented to the S1 or S2 pin before the supply voltages are available.

# SIGNAL RANGE

The ADG7421F switches have overvoltage detection circuitry on the S1 and S2 pins that compares the voltage levels with  $V_{DD}$  and  $V_{SS}$ . To protect downstream circuitry from overvoltages, supply the ADG7421F with voltages that match the intended signal range. The ADG7421F uses an NDMOS only architecture, and to achieve a

near rail-to-rail signal range, an internal charge pump is used. The internal charge pump frequency is 20 MHz. A signal that exceeds the supply rail by  $V_T$  is then blocked. This signal block offers protection to both the device and any downstream circuitry. To avoid any false trips, the analog signal range is reduced slightly inside the supply voltages,  $V_{DD}$  and  $V_{SS}$ . Table 11 shows the signal range for the specified voltage supply ranges.

Table 11. Signa	l Range for	Specified \	Voltage Su	nnlv Ranges
Table II. Signa	i nange ior	opecilieu v	vonaye Ju	ppiy nanges

Power Supply	Signal Range
5 V Single Supply	V <sub>SS</sub> + 0.1 V to V <sub>DD</sub> – 0.55 V
3 V Single Supply	$V_{SS}$ + 0.1 V to $V_{DD}$ – 0.25 V
1.8 V Single Supply	$V_{SS}$ + 0.1 V to $V_{DD}$ – 0.1 V
2.5 V Dual Supply	$V_{SS}$ + 0.1 V to $V_{DD}$ – 0.35 V

# INTELLIGENT FAULT DETECTION

The ADG7421F digital output pin (FF) can interface with a microprocessor or control system and be used as an interrupt flag. This feature provides real-time diagnostic information on the state of the device and the system to which the device connects.

The control system can use the digital interrupt to start a variety of actions, such as the following:

- Initiating investigation into the source of the overvoltage fault
- Shutting down critical systems in response to the overvoltage
- Data recorders marking data during these events as unreliable or out of specification

For systems that are sensitive during a start-up sequence, the active low operation of the flag allows the system to ensure that the ADG7421F powers on and that all input voltages are within the normal operating range before initiating operation. The FF pin is an open drain that requires an external pull-up resistor, typically 1 k $\Omega$ , which allows signals to be combined into a single interrupt for larger modules that contain multiple devices.

# SHORT DURATION FAULT CURRENT

When a fault voltage is present on either S1 or S2, the switch channel opens within typically 1.2  $\mu$ s. During this short duration, the internal diodes on the drain side of the switch, shown in Figure 53, starts to clamp the voltage on the Dx pins until the switch is fully open. These internal diodes shunt the majority of the short duration current to the power supply. As the voltage on the Dx pin stabilizes, there may be some residual current allowed to flow from the ADG7421F into a connected device.

Figure 54 is a measurement of the current that flows out of the ADG7421F and into external diodes. The external diodes in Figure 53 are used to represent the internal ESD protection diodes that are usually present on the input of a downstream device in a signal chain (for example, an ADC or an amplifier).

Before the switch opens, the majority of the short duration fault current that flows during a fault event flows through the internal diodes to the supplies (shown in Figure 53 as  $I_1$ ). The remaining current,  $I_2$  (approximately 20 mA), flows out of the switch drain pin and through the succeeding external diodes that represent the next device in the signal chain.

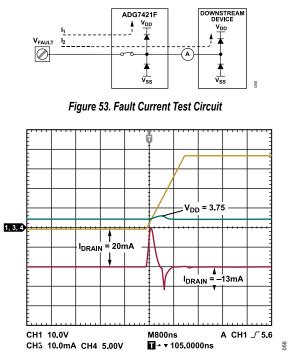


Figure 54. Fault Current Measurement

Depending on the nature of the fault and the power supply setup, the short duration current that flows through the internal diodes on the ADG7421F may cause a variation on supply voltage. If this variation is an issue, use larger decoupling capacitors (10  $\mu$ F) on the supply to stabilize V<sub>DD</sub> and V<sub>SS</sub>.

### THERMOCOUPLE INPUT PROTECTION

Figure 55 shows a typical configuration to protect thermocouple inputs to the AD7124-4. The signal from a thermocouple is small. Therefore, the thermocouple voltage must be suitably biased to stay within the specified signal range of the ADG7421F. The AD7124-4 includes an internal bias voltage generator to set the bias voltage to  $AV_{DD}/2$ , which is available on any input channel. For a dual supply solution, the thermocouple voltages can be centered around ground.

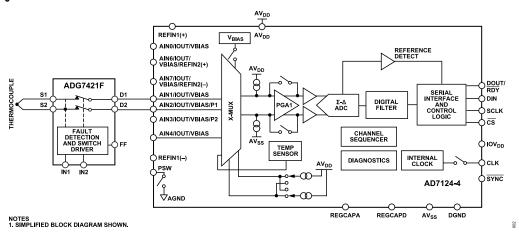


Figure 55. Thermocouple Input Configuration

### **RTD INPUT PROTECTION**

The ADG7421F can be used to protect the inputs for 2-, 3-, or 4-wire RTD configurations. Figure 56 shows one possible configuration diagram to protect a 4-wire RTD input to the AD7124-4 ADC. For 4-wire RTD configurations, two ADG7421F devices are required to protect the four system input nodes accessible by the user.

In this configuration, the voltage developed across the RTD device is sensed by the two ADC input channels, AIN1 and AIN2. These ADC input channels are connected to the external environment through two channels of the ADG7421F to provide protection against overvoltages up to ±60 V.

The excitation current required for the RTD is provided by the AD7124-4 on the AIN0/IOUT/VBIAS pin. The excitation current

flows from AIN0/IOUT/VBIAS through one ADG7421F channel, through the RTD device, through a second ADG7421F channel, and finally to GND via the reference resistor,  $R_{REF}$ , and the head-room resistor,  $R_{HEADROOM}$ . The reference resistor in this setup is used to provide a ratiometric measurement. This reference resistor generates the ADC reference voltage from the excitation current so that any variation in excitation current does not affect the accuracy of the measurement.

The headroom resistor is required because the reference input buffers on the ADC may require some headroom and the switch is not fully rail to rail.

For more information on thermocouple and RTD measurements using the AD7124-4 ADC, see the AD7124-4 data sheet.

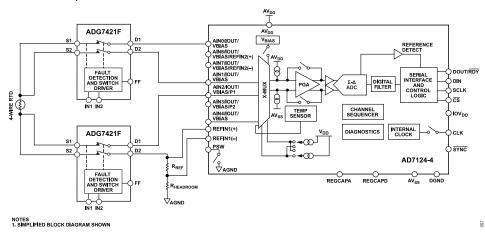


Figure 56. RTD Input Configuration

#### SWITCHES IN A KNOWN STATE

If no digital inputs are present on the switch control lines (INx), the switches remain in an off state, which prevents unwanted signals passing through the switches.

### **HIGH VOLTAGE SUPPRESSION**

To achieve protection from high voltage transients, such as IEC 61000-4-2 ESD, IEC 61000-4-4 electrical fast transient (EFT), and IEC61000-4-5 surge, implement the circuit shown in Figure 57 by using a discrete resistors and a transient voltage suppression (TVS) device.

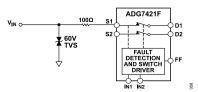


Figure 57. High Voltage Transient Protection

Table 12 details the results achieved by using the discrete protection circuit shown in Figure 57. To replicate the harshest environments, the surge test was performed by striking the Sx pins directly through a 40  $\Omega$  resistor and a 0.5  $\mu F$  capacitor coupling network. The EFT test was performed by zapping the Sx pins directly without any capacitive coupling through cables.

#### Table 12. High Voltage Transient Protection

IEC 61000-4 Transient	Protection Level (kV)
ESD (Contact)	±8
EFT	±4
Surge	±1

# **OUTLINE DIMENSIONS**

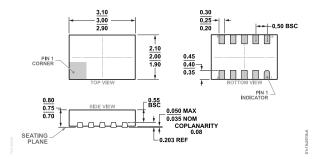


Figure 58. 10-Lead Lead Frame Chip Scale Package [LFCSP] 3 mm × 2 mm Body and 0.75 mm Package Height (CP-10-16) Dimensions shown in millimeters

### ORDERING GUIDE

Model <sup>1</sup>	Temperature Range	Package Description	Package Option
ADG7421FBCPZ-RL7	-40°C to +125°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-16
EVAL-ADG7421FEBZ		Evaluation Board	

<sup>1</sup> Z = RoHS Compliant Part.

