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August 1999 er ut takes half 235 MHz Tiny Low Power Voltage Feedback Amplifie

# National Semiconductor

## LM7121 235 MHz Tiny Low Power Voltage Feedback Amplifier

### **General Description**

The LM7121 is a high performance operational amplifier which addresses the increasing AC performance needs of video and imaging applications, and the size and power constraints of portable applications.

The LM7121 can operate over a wide dynamic range of supply voltages, from 5V (single supply) up to ±15V (see the Application Information section for more details). It offers an excellent speed-power product delivering 1300V/µs and 235 MHz Bandwidth (-3 dB, A<sub>V</sub> = +1). Another key feature of this operational amplifier is stability while driving unlimited capacitive loads.

Due to its Tiny SOT23-5 package, the LM7121 is ideal for designs where space and weight are the critical parameters. The benefits of the Tiny package are evident in small portable electronic devices, such as cameras, and PC video cards. Tiny amplifiers are so small that they can be placed anywhere on a board close to the signal source or near the input to an A/D converter.

### Features

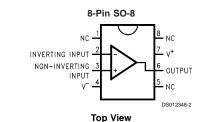
(Typical unless otherwise noted)  $V_S = \pm 15V$ 

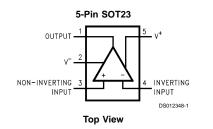
- Easy to use voltage feedback topology
- Stable with unlimited capacitive loads
- Tiny SOT23-5 package typical circuit layout takes half the space of SO-8 designs
- Unity gain frequency: 175 MHz
- Bandwidth (-3 dB,  $A_V = +1$ ,  $R_L = 100\Omega$ ): 235 MHz
- Slew rate: 1300V/µs
- Supply Voltages SO-8: 5V to ±15V SOT23-5: 5V to ±5V
- Characterized for: +5V, ±5V, ±15V
- Low supply current: 5.3 mA

### **Applications**

- Scanners, color fax, digital copiers
- PC video cards
- Cable drivers
- Digital cameras
- ADC/DAC buffers
- Set-top boxes

## **Connection Diagrams**





## **Ordering Information**

Package	Ordering Information	NSC Drawing	Package	Supplied As
		Number	Marking	
8-Pin SO-8	LM7121IM	M08A	LM7121IM	Rails
	LM7121IMX	M08A	LM7121IM	2.5k Tape and Reel
5-Pin SOT23-5	LM7121IM5	MA05A	A03A	1k Tape and Reel
	LM7121IM5X	MA05A	A03A	3k Tape and Reel

### Absolute Maximum Ratings (Note 1)

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If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

ESD Tolerance (Note 2)	2000V
Differential Input Voltage (Note 7)	±2V
Voltage at Input/Output Pin	(V+)-1.4V, (V-)+1.4V
Supply Voltage (V <sup>+</sup> -V <sup>-</sup> )	36V
Output Short Circuit to Ground	
(Note 3)	Continuous
Lead Temperature	260°C
(soldering, 10 sec)	260°C

Storage Temperature Range-65°C to +150°CJunction Temperature (Note 4)150°C

## **Operating Ratings** (Note 1)

Supply Voltage: SO-8	$4.5 \text{V} \leq \text{V}_{\text{S}} \leq 33 \text{V}$
SOT23-5	$4.5V \le V_S \le 11V$
Junction Temperature Range	$-40^{\circ}C \le T_{J} \le +85^{\circ}C$
Thermal Resistance ( $\theta_{JA}$ )	
M Package, 8-pin Surface Mount	165°C/W
SOT23-5 Package	325°C/W

## **±15V DC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ , V<sup>+</sup> = +15V, V<sup>-</sup> = -15V, V<sub>CM</sub> = V<sub>O</sub> = 0V and R<sub>L</sub> > 1 M $\Omega$ . Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit (Note 6)	Units
V <sub>os</sub>	Input Offset Voltage		0.9	8	mV
				15	max
I <sub>B</sub>	Input Bias Current		5.2	9.5	μA
				12	max
l <sub>os</sub>	Input Offset Current		0.04	4.3	μA
				7	max
R <sub>IN</sub>	Input Resistance	Common Mode	10		MΩ
		Differential Mode	3.4		MΩ
CIN	Input Capacitance	Common Mode	2.3		pF
CMRR	Common Mode	$-10V \le V_{CM} \le 10V$	93	73	dB
	Rejection Ratio			70	min
+PSRR	Positive Power Supply	$10V \le V^+ \le 15V$	86	70	dB
	Rejection Ratio			68	min
-PSRR	Negative Power Supply	$-15V \le V^- \le -10V$	81	68	dB
	Rejection Ratio			65	min
V <sub>CM</sub>	Input Common-Mode	CMRR ≥ 70 dB	13	11	V
	Voltage Range	Range			min
			-13	-11	V
					max
A <sub>V</sub>	Large Signal	$R_L = 2 k\Omega, V_O = 20 V_{PP}$	72	65	dB
	Voltage Gain			57	min
Vo	Output Swing	$R_L = 2 k\Omega$	13.4	11.1	V
				10.8	min
			-13.4	-11.2	V
				-11.0	max
		$R_{L} = 150\Omega$	10.2	7.75	V
				7.0	min
			-7.0	-5.0	V
				-4.8	max
I <sub>sc</sub>	Output Short Circuit	Sourcing	71	54	mA
	Current			44	min
		Sinking	52	39	mA
		_		34	min

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit	Units
				(Note 6)	
ls	Supply Current		5.3	6.6	mA
				7.5	max
Unless other	C Electrical Charact rwise specified, all limits guarantee pply at the temperature extremes. Parameter	ed for T, = 25°C, V⁺ = 15V, V⁻ = -	Тур	= 0V and R <sub>L</sub> > 1	MΩ. Bold Units
			(Note 5)	Limit	
				(Note 6)	
SR	Slew Rate	$A_V = +2, R_L = 1 k\Omega,$	1300		V/µs
	(Note 8)	$V_{O} = 20 V_{PP}$			
GBW	Unity Gain-Bandwidth	$R_L = 1 k\Omega$	175		MHz
φ <sub>m</sub>	Phase Margin		63		Deg
f (–3 dB)	Bandwidth	$R_{L} = 100\Omega, A_{V} = +1$	235		MHz
	(Notes 9, 10)	$R_{L} = 100\Omega, A_{V} = +2$	50		
t <sub>s</sub>	Settling Time	10 $V_{PP}$ Step, to 0.1%, R <sub>L</sub> = 500 $\Omega$	74		ns
t <sub>r</sub> , t <sub>f</sub>	Rise and Fall Time (Note 10)	$A_V = +2, R_L = 100\Omega,$ $V_O = 0.4 V_{PP}$	5.3		ns
A <sub>D</sub>	Differential Gain	$A_{V} = +2, R_{I} = 150\Omega$	0.3		%
ф <sub>D</sub>	Differential Phase	$A_V = +2, R_L = 150\Omega$	0.65		Deg
e <sub>n</sub>	Input-Referred	f = 10 kHz	17		nV
	Voltage Noise				$\frac{1}{\sqrt{Hz}}$
i <sub>n</sub>	Input-Referred	f = 10 kHz	1.9		pА
	Current Noise				√Hz
T.H.D.	Total Harmonic Distortion	2 V <sub>PP</sub> Output, $R_L = 150\Omega$ ,	0.065		%

## **±5V DC Electrical Characteristics**

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Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = -5V$ ,  $V_{CM} = V_O = 0V$  and  $R_L > 1 M\Omega$ . Bold-face limits apply at the temperature extremes.

 $A_V = +2$ , f = 5 MHz

 $2 V_{PP}$  Output, R<sub>L</sub> = 150 $\Omega$ ,

0.52

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit	Units
				(Note 6)	
Vos	Input Offset Voltage		1.6	8	mV
				15	max
I <sub>B</sub>	Input Bias Current		5.5	9.5	μA
				12	max
l <sub>os</sub>	Input Offset Current		0.07	4.3	μA
				7.0	max
R <sub>IN</sub>	Input Resistance	Common Mode	6.8		MΩ
		Differential Mode	3.4		MΩ

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit (Note 6)	Units
CIN	Input Capacitance	Common Mode	2.3	(,	pF
CMRR	Common Mode Rejection Ratio	$-2V \le V_{CM} \le 2V$	75	65 <b>60</b>	dB
+PSRR	Positive Power Supply Rejection Ratio	$3V \le V^+ \le 5V$	89	65 <b>60</b>	dB min
-PSRR	Negative Power Supply Rejection Ratio	$-5V \le V^- \le -3V$	78	65 <b>60</b>	dB min
V <sub>CM</sub> Input Common Mode Voltage Range	1 ·	CMRR ≥ 60 dB	3	2.5	V min
			-3	-2.5	V max
4 <sub>V</sub>	Large Signal Voltage Gain	$R_L = 2 k\Omega, V_O = 3 V_{PP}$	66	60 <b>58</b>	dB min
Vo	Output Swing	$R_L = 2 k\Omega$	3.62	3.0 <b>2.75</b>	V min
			-3.62	-3.0 <b>-2.70</b>	V max
		R <sub>L</sub> = 150Ω	3.1	2.5 <b>2.3</b>	V min
			-2.8	-2.15 <b>-2.00</b>	V max
	Output Short Circuit Current	Sourcing	53	38 <b>33</b>	mA min
		Sinking	29	21 <b>19</b>	mA min
Is	Supply Current		5.1	6.4 <b>7.2</b>	mA

## **±5V AC Electrical Characteristics**

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ , V<sup>+</sup> = 5V, V<sup>-</sup> = -5V, V<sub>CM</sub> = V<sub>O</sub> = 0V and R<sub>L</sub> > 1 M $\Omega$ . Bold-face limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit (Note 6)	Units
SR	Slew Rate	$A_V = +2, R_L = 1 k\Omega,$	520		V/µs
	(Note 8)	$V_{O} = 6 V_{PP}$			
GBW	Unity Gain-Bandwidth	$R_L = 1 k\Omega$	105		MHz
φ <sub>m</sub>	Phase Margin	$R_{L} = 1 \ k\Omega$	74		Deg
f (-3 dB)	Bandwidth	$R_{L} = 100\Omega, A_{V} = +1$	160		MHz
	(Notes 9, 10)	$R_{L} = 100\Omega, A_{V} = +2$	50		
t <sub>s</sub>	Settling Time	5 V <sub>PP</sub> Step, to 0.1%, R <sub>L</sub> = $500\Omega$	65		ns
t <sub>r</sub> , t <sub>f</sub>	Rise and Fall Time (Note 10)	$A_V = +2, R_L = 100\Omega,$ $V_O = 0.4 V_{PP}$	5.8		ns
Ap	Differential Gain	$A_{V} = +2, R_1 = 150\Omega$	0.3		%

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## ±5V AC Electrical Characteristics (Continued)

Unless otherwise specified, all limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = -5V$ ,  $V_{CM} = V_O = 0V$  and  $R_L > 1 \text{ M}\Omega$ . Bold-face limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit (Note 6)	Units
φ <sub>D</sub>	Differential Phase	$A_{V} = +2, R_{L} = 150\Omega$	0.65		Deg
e <sub>n</sub>	Input-Referred Voltage Noise	f = 10 kHz	17		nV √Hz
i <sub>n</sub>	Input-Referred Current Noise	f = 10 kHz	2		pA √Hz
T.H.D.	Total Harmonic Distortion	$2 V_{PP} \text{ Output, } R_L = 150\Omega,$ $A_V = +2, \text{ f} = 1 \text{ MHz}$	0.1		%
		2 V <sub>PP</sub> Output, R <sub>L</sub> = 150 $\Omega$ , A <sub>V</sub> = +2, f = 5 MHz	0.6		

## +5V DC Electrical Characteristics

Unless otherwise specified, all limits guaranteed for  $T_J$  = 25°C, V<sup>+</sup> = +5V, V<sup>-</sup> = 0V, V<sub>CM</sub> = V<sub>O</sub> = V<sup>+</sup>/2 and R<sub>L</sub> > 1 M $\Omega$ . Bold-face limits apply at the temperature extremes.

Symbol	Parameter	Conditions	<b>Typ</b> (Note 5)	LM7121I Limit (Note 6)	Units
Vos	Input Offset Voltage		2.4		mV
I <sub>B</sub>	Input Bias Current		4		μA
l <sub>os</sub>	Input Offset Current		0.04		μA
R <sub>IN</sub>	Input Resistance	Common Mode	2.6		MΩ
		Differential Mode	3.4		MΩ
CIN	Input Capacitance	Common Mode	2.3		pF
CMRR	Common Mode Rejection Ratio	$2V \le V_{CM} \le 3V$	65		dB
+PSRR	Positive Power Supply Rejection Ratio	$4.6V \le V^+ \le 5V$	85		dB
-PSRR	Negative Power Supply Rejection Ratio	$0V \le V^- \le 0.4V$	61		dB
V <sub>CM</sub>	Input Common-Mode Voltage Range	CMRR ≥ 45 dB	3.5		V min
			1.5		V max
A <sub>V</sub>	Large Signal Voltage Gain	$R_L = 2 k\Omega$ to V <sup>+</sup> /2	64		dB
Vo	Output Swing	$R_L = 2 k\Omega$ to V <sup>+</sup> /2, High	3.7		V
		$R_L = 2 k\Omega$ to V <sup>+</sup> /2, Low	1.3		1
		$R_{L} = 150\Omega$ to V <sup>+</sup> /2, High	3.48		
		$R_L = 150\Omega$ to V <sup>+</sup> /2, Low	1.59		1
I <sub>sc</sub>	Output Short Circuit	Sourcing	33		mA
	Current	Sinking	20		mA
ls	Supply Current		4.8		mA

	oply at the temperature extremes.				
Symbol	Parameter	Conditions	Typ (Note 5)	LM7121I Limit (Note 6)	Units
SR	Slew Rate	$A_V = +2$ , $R_L = 1 \text{ k}\Omega$ to	145		V/µs
	(Note 8)	$V^{+}/2, V_{O} = 1.8 V_{PP}$			
GBW	Unity Gain-Bandwidth	$R_{L} = 1k$ , to V <sup>+</sup> /2	80		MHz
φ <sub>m</sub>	Phase Margin	$R_{L} = 1k \text{ to } V^{+}/2$	70		Deg
f (–3 dB)	Bandwidth	$R_{L} = 100\Omega$ to V <sup>+</sup> /2, $A_{V} = +1$	200		MHz
	(Notes 9, 10)	$R_{L} = 100\Omega$ to V <sup>+</sup> /2, $A_{V} = +2$	45		1
t <sub>r</sub> , t <sub>f</sub>	Rise and Fall Time	$A_{V} = +2, R_{L} = 100\Omega,$	8		ns
	(Note 10)	$V_{O} = 0.2 V_{PP}$			
T.H.D.	Total Harmonic Distortion	$0.6 V_{PP}$ Output, $R_L = 150\Omega$ ,	0.067		%
		$A_{V} = +2$ , f = 1 MHz			
		0.6 V <sub>PP</sub> Output, $R_L = 150\Omega$ ,	0.33		1
		$A_{V} = +2$ , f = 5 MHz			

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.

Note 2: Human body model, 1.5 k $\Omega$  in series with 100 pF.

Note 3: Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C.

Note 4: The maximum power dissipation is a function of  $T_{J(max)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(max)} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board.

Note 5: Typical Values represent the most likely parametric norm.

Note 6: All limits are guaranteed by testing or statistical analysis.

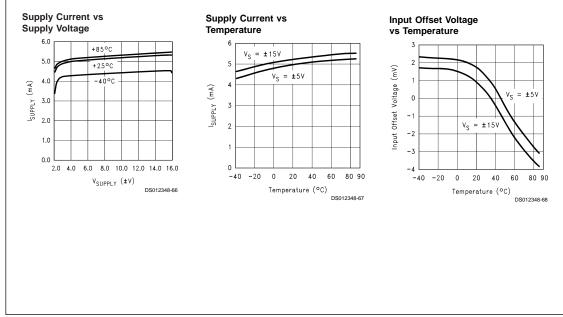
Note 7: Differential input voltage is measured at  $V_S = \pm 15V$ .

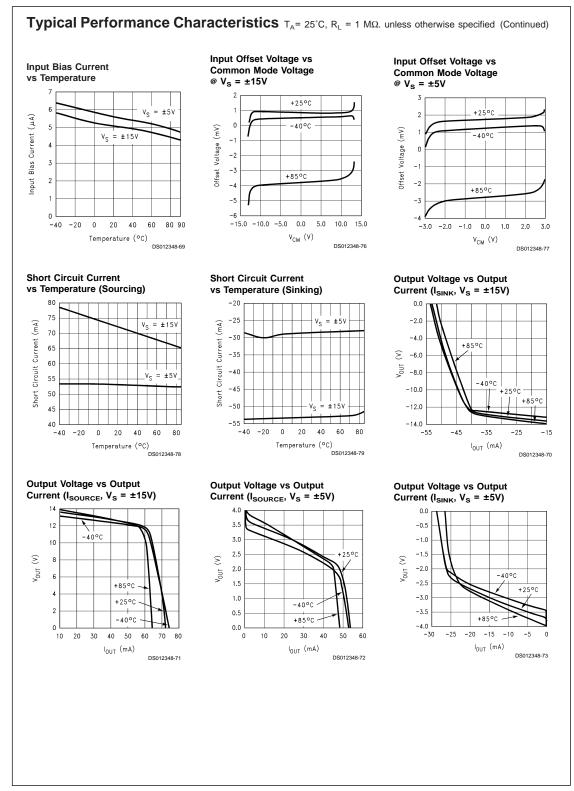
Note 8: Slew rate is the average of the rising and fallng slew rates.

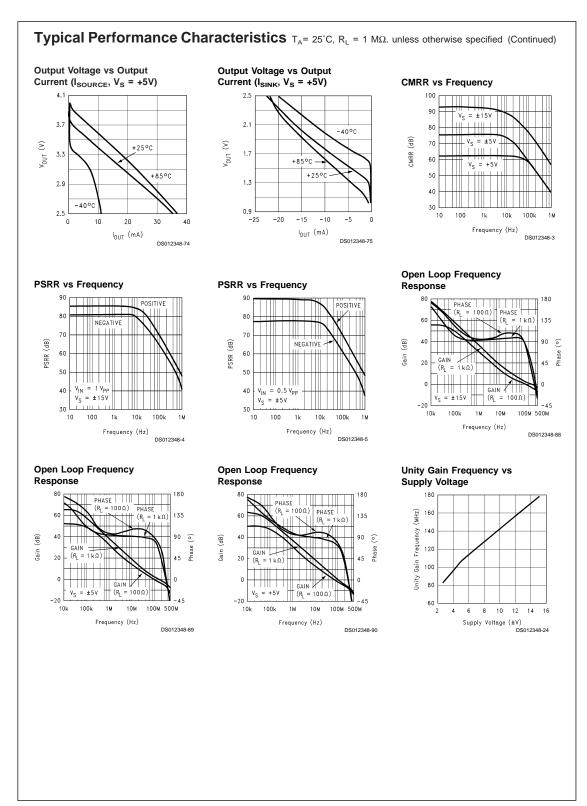
**Note 9:** Unity gain operation for  $\pm$ 5V and  $\pm$ 15V supplies is with a feedback network of 510 $\Omega$  and 3 pF in parallel (see the Application Information section). For +5V single supply operation, feedback is a direct short from the output to the inverting input.

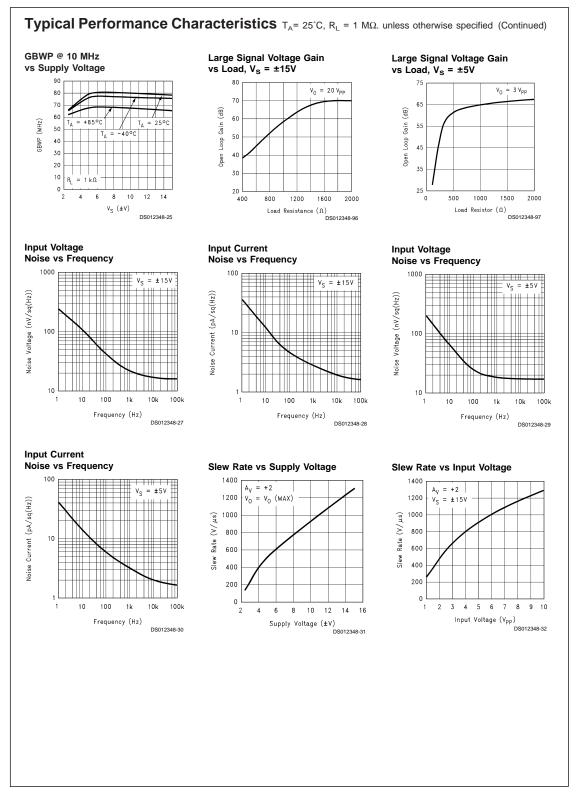
Note 10:  $A_V$  = +2 operation with 2 k $\Omega$  resistors and 2 pF capacitor from summing node to ground.

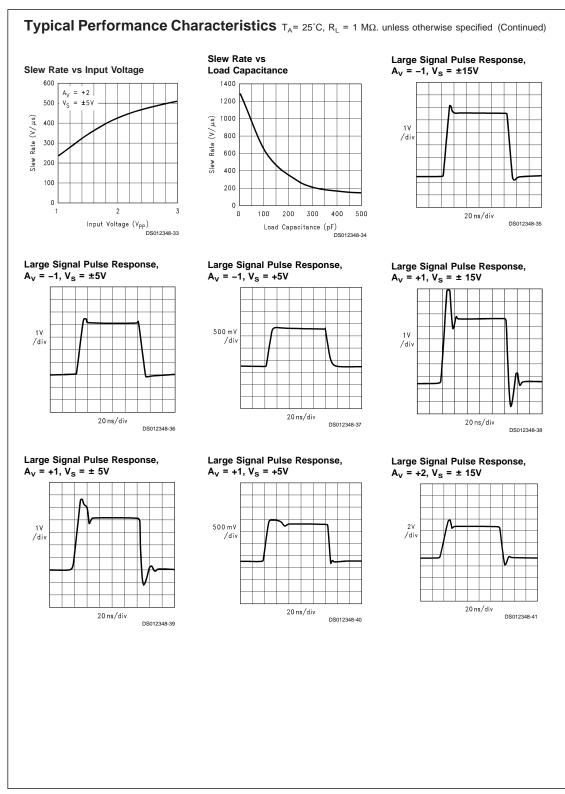
## Typical Performance Characteristics $T_A$ = 25°C, $R_L$ = 1 M $\Omega$ . unless otherwise specified

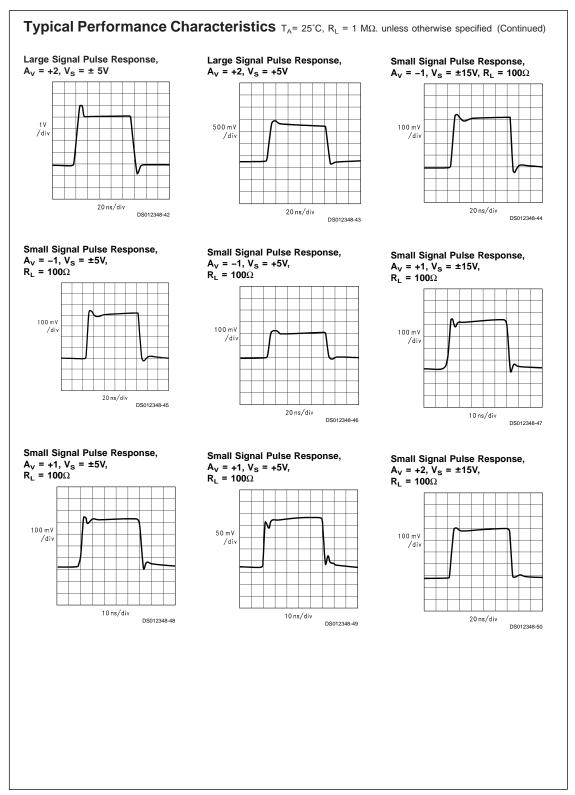


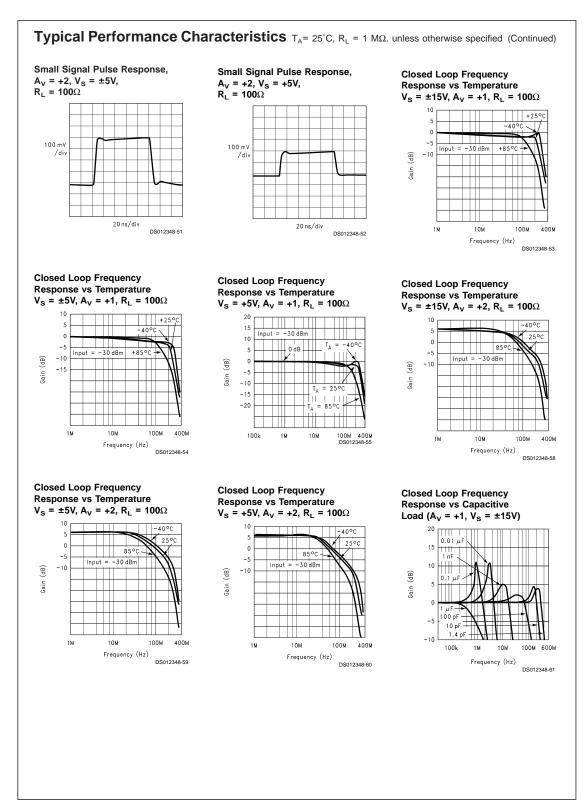


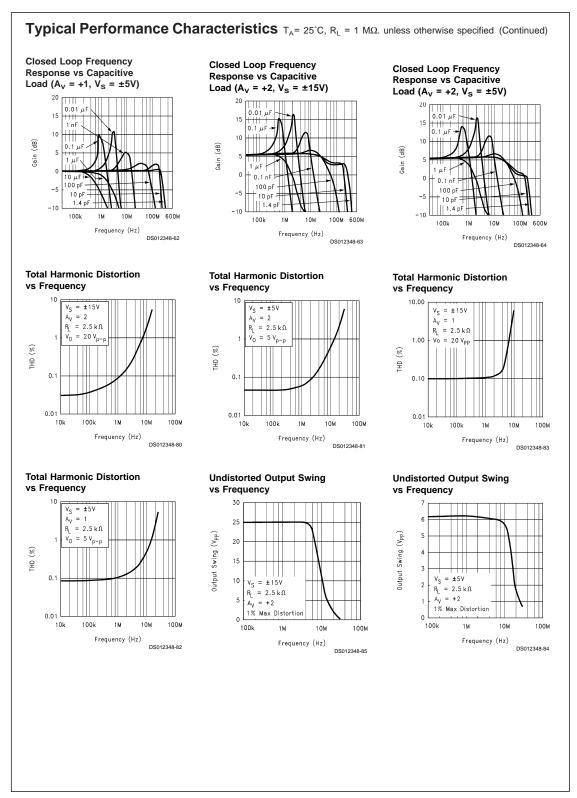


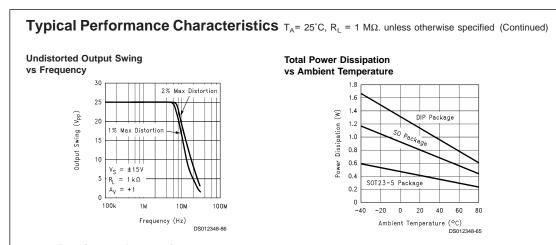












### **Application Information**

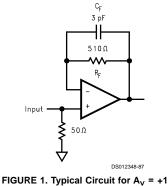
The table below, depicts the maximum operating supply voltage for each package type:

#### TABLE 1. Maximum Supply Voltage Values

	SOT23-5	SO-8
Single Supply	10V	30V
Dual Supplies	±5V	±15V

Stable unity gain operation is possible with supply voltage of 5V for all capacitive loads. This allows the possibility of using the device in portable applications with low supply voltages with minimum components around it.

Above a supply voltage of 6V (±3V Dual supplies), an additional resistor and capacitor (shown below) should be placed in the feedback path to achieve stability at unity gain over the full temperature range.



Operation ( $V_s \ge 6V$ )

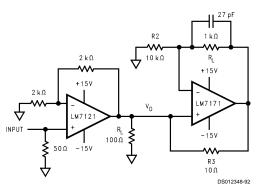
The package power dissipation should be taken into account when operating at high ambient temperatures and/or high power dissipative conditions. Refer to the power derating curves in the data sheet for each type of package. In determining maximum operable temperature of the device, make sure the total power dissipation of the device is considered; this includes the power dissipated in the device with a load connected to the output as well as the nominal dissipation of the op amp.

The device is capable of tolerating momentary short circits from its output to ground but prolonged operation in this mode will damage the device, if the maximum allowed junction temperation is exceeded.

#### APPLICATION CIRCUITS

#### **Current Boost Circuit**

The circuit in *Figure 2* can be used to achieve good linearity along with high output current capability.



#### FIGURE 2. Simple Circuit to Improve Linearity and Output Drive Current

By proper choice of  $R_3$ , the LM7121 output can be set to supply a minimal amount of current, thereby improving its output linearity.

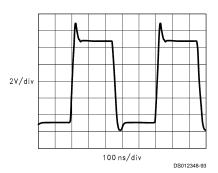
R<sub>3</sub> can be adjusted to allow for different loads:

 $R_3 = 0.1 R_L$ 

The circuit above has been set for a load of  $100\Omega$ .

Reasonable speeds (<30 ns rise and fall times) can be expected up to 120 mA<sub>PP</sub> of load current (see *Figure 3* for step response across the load).

## Application Information (Continued)

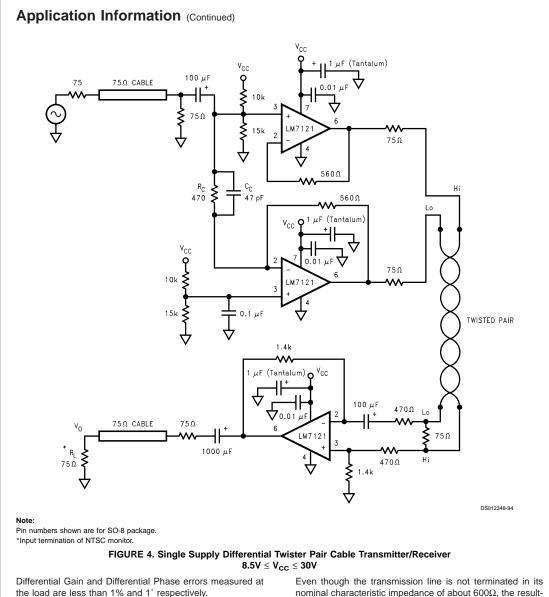


#### FIGURE 3. Waveform across a 100 $\Omega$ Load

It is very important to keep the lead lengths to a minimum and to provide a low impedance current path by using a ground-plane on the board.

**Caution:** If  $R_L$  is removed, the current balance at the output of LM7121 would be disturbed and it would have to supply the full amount of load current. This might damage the part if power dissipation limit is exceeded.

**Color Video on Twisted Pairs Using Single Supply** The circuit shown in *Figure 4* can be used to drive in excess of 25 meters length of twisted pair cable with no loss of resolution or picture definition when driving a NTSC monitor at the load end.



 $\rm R_G$  and  $\rm C_C$  can be adjusted for various cable lengths to compensate for the line losses and for proper response at the output. Values shown correspond to a twisted pair cable length of 25 meters with about 3 turns/inch (see *Figure 5* for step response).

The supply voltage can vary from 8.5V up to 30V with the output rise and fall times under 12 ns. With the component values shown, the overall gain from the input to the output is about 1.

Even though the transmission line is not terminated in its nominal characteristic impedance of about 600 $\Omega$ , the resulting reflection at the load is only about 5% of the total signal and in most cases can be neglected. Using 75 $\Omega$  termination instead, has the advantage of operating at a low impedance and results in a higher realizable bandwidth and signal fidelity.

