

3MHz CMOS Rail-to-Rail IO Opamps

Features:

Single-Supply Operation from +2.1V ~ +5.5V

Rail-to-Rail Input / Output

Gain-Bandwidth Product: 3MHz (Typ)

Low Input Bias Current: 1pA (Typ)

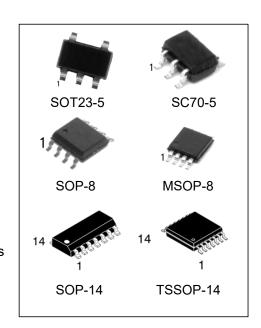
Low Offset Voltage: 3.5mV (Max)

Quiescent Current:250μA per Amplifier (Typ)

Operating Temperature:-40°C ~ +125°C

Small Package:

HGV8621 Available in SOT23-5, SOP-8 and SC70-5 Packages HGV8622 Available in SOP-8 and MSOP-8 Packages HGV8624 Available in SOP-14 and TSSOP-14 Packages



Ordering Information

DEVICE	Package Type	MARKING	Packing	Packing Qty
HGV8621M5/TR	SOT23-5	8621	REEL	3000pcs/reel
HGV8621M7/TR	SC70-5	8621	REEL	3000pcs/reel
HGV8621M/TR	SOP-8	V8621	REEL	2500pcs/reel
HGV8622M/TR	SOP-8	V8622	REEL	2500pcs/reel
HGV8622MM/TR	MSOP-8	V8622	REEL	3000pcs/reel
HGV8624M/TR	SOP-14	HGV8624	REEL	2500pcs/reel
HGV8624MT/TR	TSSOP-14	V8624	REEL	2500pcs/reel



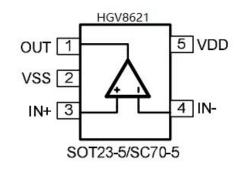
General Description

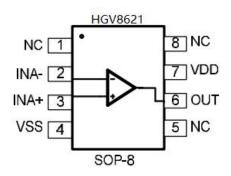
The HGV8621/22/24 have a high gain-bandwidth product of 3MHz, a slew rate of $1.66V/\mu s$, and a quiescent current of $250\mu A$ per amplifier at 5V. The HGV8621/22/24 are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for HGV8621/22/24. They are specified over the extended industrial temperature range (- $40^{\circ}C$ to $+125^{\circ}C$). The operating range is from 2.1V to 5.5V. The operating range is from 2.1V to 5.5V. The HGV8621 single is available in Green SC70-5, SOT23-5 and SOP-8 packages. The HGV8622 dual is available in Green SOP-14 and TSSOP-14 packages.

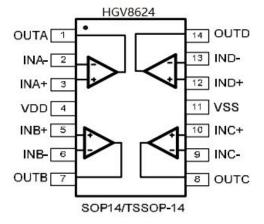
Applications

- Sensors
- Active Filters
- Cellular and Cordless Phones
- Laptops and PDAs
- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters

Pin Configurations







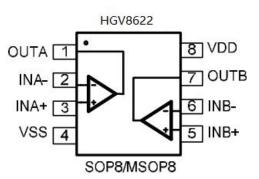


Figure 1. Pin Assignment Diagram



Absolute Maximum Ratings

CONDITIN	MIN	MAX				
Power Supply Voltage (V _{DD} to Vss)	-0.5V	+7.5V				
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V _{DD} +0.5V				
PDB Input Voltage	Vss-0.5V	+7V				
Operating Temperature Range	-40°C	+125°C				
Junction Temperature	+16	0°C				
Storage Temperature Range	-55°C	+150°C				
Lead Temperature (soldering, 10sec)	+26	0°C				
Package Thermal Resistance (Ta=+25℃)						
SOP-8, θ _{JA}	125°	C/W				
MSOP-8, θ _{JA}	216°	C/W				
SOT23-5, θ _{JA}	190°	C/W				
SC70-5, θ _{JA}	333°	C/W				
SOP-14, θ _{JA}	120°	C/W				
TSSOP-14, θ _{JA}	180°	C/W				
Esd Susceptibility						
НВМ	8KV					
MM	40	0V				

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.



Electrical Characteristics

(At Vs=5V, T_A = +25°C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

(At VS-5V, TA - +25 C, VCM - V	5/2, 112 00012, dillious cities		otou.,	Н	GV8621	/22/4		
PARAMETER	CONDITIONS	TYP		MIN/MA	AX OVE	R TEMPI	ERATUR	E
PARAMETER	CONDITIONS	+25℃	+25 ℃	0℃ to 70℃	-40℃ to 85℃	-40℃ to 125℃	UNITS	MIN / MAX
INPUT CHARACTERISTICS		_						
Input Offset Voltage (Vos)		0.8	3.5	3.9	4.3	4.6	mV	MAX
Input Bias Current (I _B)		1					pА	TYP
Input Offset Current (Ios)		1					pА	TYP
Input Common Mode Voltage Range (V _{CM})	V _S = 5.5V	-0.1 to +5.6					V	TYP
Common Mode Rejection Ratio (CMRR)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 4V	82	65	64	64	63	dB	MIN
	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6V	71					dB	MIN
Open-Loop Voltage Gain (A _{OL})	$R_L = 600\Omega, V_O = 0.15V \text{ to } 4.85V$	90	80	76	75	68	dB	MIN
	$R_L = 10k\Omega, V_O = 0.05V \text{ to } 4.95V$	100					dB	MIN
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta_T$)		2.4					μV/°C	TYP
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	$R_L = 600\Omega$	0.1					V	TYP
	R _L = 10kΩ	0.015					V	TYP
Output Current (Іоит)		53	49	45	40	35	mA	MIN
Closed-Loop Output Impedance	f = 100kHz, G = 1	10					Ω	TYP
POWER-DOWN DISABLE								
Turn-On Time		4					μs	TYP
Turn-Off Time		1.2					μs	TYP
DISABLE Voltage-Off			0.8				V	MAX
DISABLE Voltage-On			2				V	MIN
POWER SUPPLY								
Operating Voltage Range			2.1 5.5	2.1 5.5	2.1 5.5	2.1 5.5	V V	MIN MAX
	V _S = +2.5V to +5.5V							
	$V_{CM} = (-V_S) + 0.5V$			7.0		000		
Power Supply Rejection Ratio (PSRR)	I _{OUT} = 0	91	74	72	72	68	dB 	MIN
Quiescent Current/Amplifier (I _Q)		250	350	427	450	515	μA	MAX



Electrical Characteristics

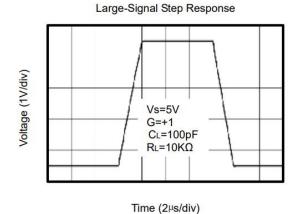
(At Vs=5V, T_A = +25°C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

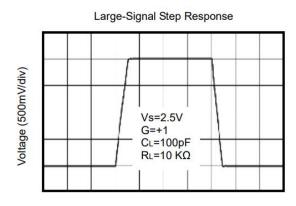
	·	HGV8621/22/4										
PARAMETER	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE									
		+25 ℃	+25 ℃	0℃ to 70℃	-40℃ to 85℃	-40℃to 125℃	UNITS	MIN / MAX				
DYNAMIC PERFORMANCE												
Gain-Bandwidth Product (GBP)	$R_L = 10k\Omega$, $C_L = 100pF$	3					MHz	TYP				
Phase Margin (φ ₀)	$R_L = 10k\Omega, C_L = 100pF$	50					Degrees	TYP				
Full Power Bandwidth (BWP)	$< 1\%$ distortion, R _L = 600Ω	50					kHz	TYP				
Slew Rate (SR)	G = +1, 2V Step, R _L = 10kΩ	1.66					V/µs	TYP				
Settling Time to 0.1% (t _S)	G = +1, 2V Step, R _L = 600Ω	0.5					μs	TYP				
Overload Recovery Time	V _{IN} ·Gain = VS, R _L = 600Ω	4.5					μs	TYP				
NOISE PERFORMANCE	NOISE PERFORMANCE											
Voltage Noise Density (e _n)	f = 1kHz	18					nV /√Hz	TYP				
Current Noise Density (in)	f = 1kHz	4.5					fA /√Hz	TYP				



Typical Performance characteristics

(At Vs=5V, T_A = +25°C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)







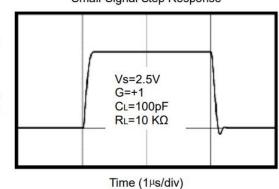
Small-Signal Step Response

Vs=5V
G=+1
CL=100pF
RL=10 KΩ

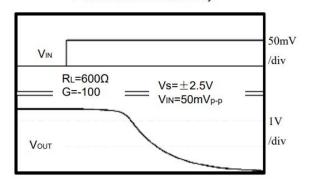
Time (1μs/div)

Small-Signal Step Response

Time (2µs/div)

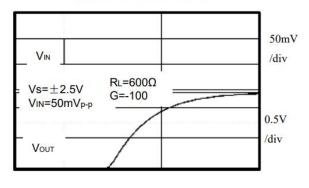


Positive Overload Recovery



Time (5µs/div)

Negative Overload Recovery



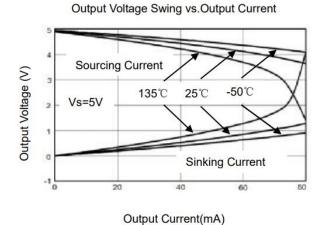
Time (5µs/div)

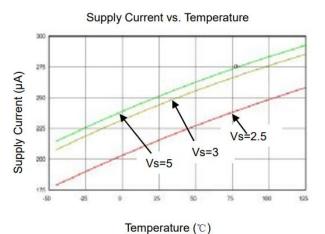
Voltage (50mV/div)

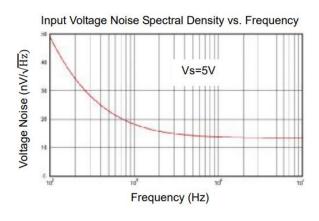


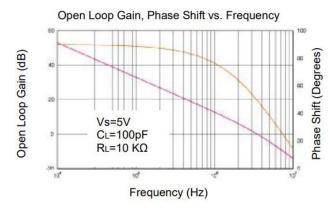
Typical Performance characteristics

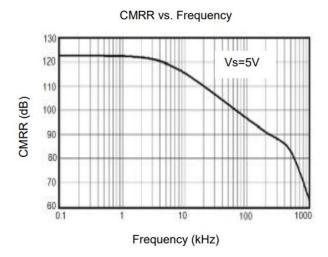
(At Vs=5V, T_A = +25 °C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

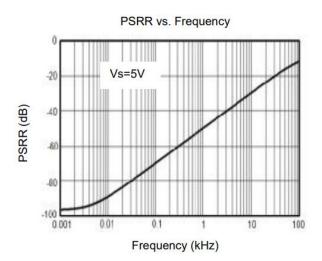














Application Note

Size

HGV8621/22/24 series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the HGV8621/22/24 series packages save space on printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

HGV8621/22/24 series operates from a single 2.1V to 5.5V supply or dual ± 1.05 V to ± 2.75 V supplies. For best performance, a $0.1\mu F$ ceramic capacitor should be placed close to the VDD pin in single supply operation. For dual supply operation, both VDD and VSS supplies should be bypassed to ground with separate $0.1\mu F$ ceramic capacitors.

Low Supply Current

The low supply current (typical 250µA per channel) of HGV8621/22/24 series will help to maximize battery life. They are ideal for battery powered systems.

Operating Voltage

HGV8621/22/24 series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

Rail-to-Rail Input

The input common-mode range of HGV8621/22/24 series extends 100mV beyond the supply rails (VSS-0.1V to VDD+0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of HGV8621/22/24 series can typically swing to less than 2mV from supply rail in light resistive loads (>100k Ω), and 60mV of supply rail in moderate resistive loads (10k Ω).



Capacitive Load Tolerance

The HGV8621/22/24 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are using a small resistor in series with the amplifier's output and the load capacitance and reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

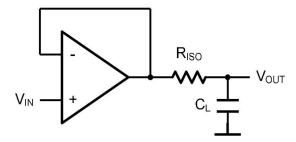


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the $R_{\rm ISO}$ resistor value, the more stable $V_{\rm OUT}$ will be. However, if there is a resistive load $R_{\rm L}$ in parallel with the capacitive load, a voltage divider (proportional to $R_{\rm ISO}/R_{\rm L}$) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R_F provides the DC accuracy by feed-forward the VIN to RL. C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased byincreasing the value of C_F . This in turn will slow down the pulse response.

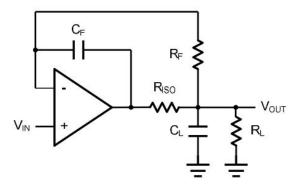


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using HGV8621/22/24.

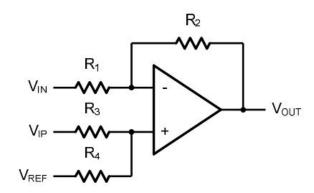


Figure 4. Differential Amplifier

If the resistor ratios are equal (i.e. $R_1=R_3$ and $R_2=R_4$), then

$$V_{OUT} = \frac{R2}{R1} (V_{IP} - V_{IN}) + V_{REF}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_C=1/(2\pi R_3C_1)$.

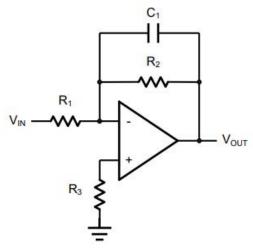


Figure 5. Low Pass Active Filter



Instrumentation Amplifier

The triple HGV8621/22/24 can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

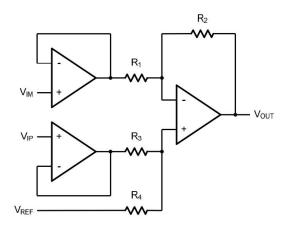
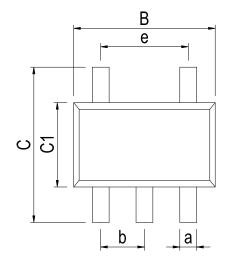


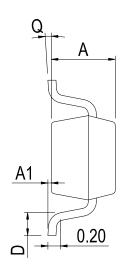
Figure 6. Instrument Amplifier



Physical Dimensions

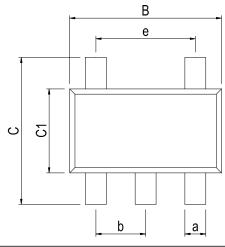
SOT23-5

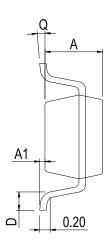




Dimensions In	Dimensions In Millimeters(SOT23-5)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	е	
Min:	1.05	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC	1.90 BSC	
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.40	0.83 630		

SC70-5



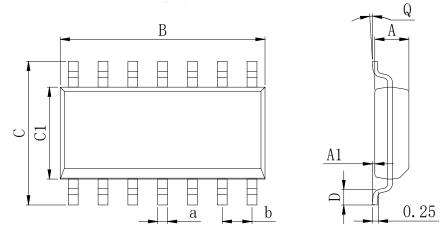


Dimensions In Millimeters(SC70-5)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	е
Min:	0.90	0.00	2.00	2.15	1.15	0.26	0°	0.30	0.65	1 20 PSC
Max:	1.00	0.15	2.20	2.45	1.35	0.46	8°	0.40	BSC	1.30 BSC



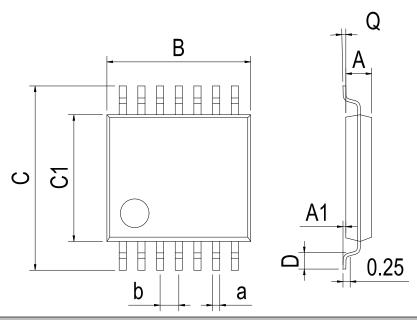
Physical Dimensions

SOP14



Dimensions In Millimeters(SOP14)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	1 27 DCC	
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	1.27 BSC	

TSSOP14

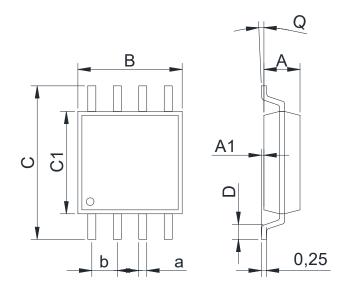


Dimensions In Millimeters(TSSOP14)										
Symbol:	Α	A1	В	С	C1	D	Q	а	b	
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0 65 000	
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25	0.65 BSC	



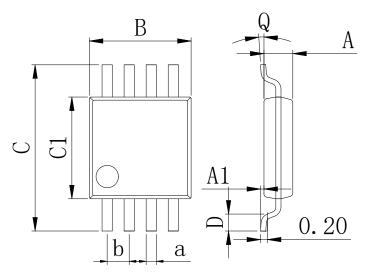
Physical Dimensions

SOP8 (208MIL)



Dimensions In Millimeters(SOP8)										
Symbol:	А	A1	В	С	C1	D	Q	а	b	
Min:	1.70	0.05	5.18	7.70	5.18	0.5	0°	0.35	4 27 BSC	
Max:	1.91	0.25	5.38	8.70	5.38	0.8	8°	0.48	1.27 BSC	

MSOP8



Dimensions In Millimeters(MSOP8)										
Symbol:	А	A1	В	С	C1	D	Q	а	b	
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65.050	
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	- 0.65 BSC	



Revision History

DATE	REVISION	PAGE
2018-2-25	New	1-16



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