COEVER ANALOG 1A, 1.5MHz, High Efficiency Synchronous Buck Converter Datasheet

General Description

The EA8101 is a high switching frequency, high efficiency synchronous buck regulator, designed to operate from 2.5V to 5.5V input voltage range. Built-in low R_{DS(ON)} high/low side Power-MOSFETS not only reduce external components and has up to 95% efficiency, ideal for 1A output current applications. The EA8101 features 100% duty cycle low dropout operation, extending battery life in portable systems. Besides, this device is designed to take into account the light load mode operation and can provide high efficiency over a wide range of the load current. The internal compensation design not only allows users to more simplified application, and can reduce the cost of external components. The EA8101 is available in the SOT-23-5 packages and easy to use.

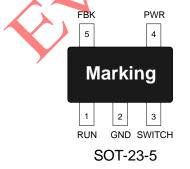
Features

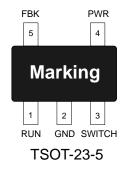
- Built-in Low R_{DS(ON)} Power-MOSFETS
- Efficiency Up to 95%
- 2.5V to 5.5V Input Voltage Range
- Adjustable Output Voltage Range from 0.6V to Vin
- Fixed 1.5MHz Switching Frequency
- 1A Continuous Load Current
- 100% Duty Cycle Low Dropout Operation
- 45uA Low Quiescent Current
- <1uA Shutdown Current</p>
- Internal Compensation
- Cycle-by-Cycle Current Limit
- Short Circuit Protection
- OTP Protection
- Available in SOT-23-5 and TSOT23-5 Packages

Applications

- Netcom Products
- Set-Top-Box
- LCD TVs and Flat TVs
- PDAs

Pin Configurations





1A, 1.5MHz High Efficiency Buck Converter

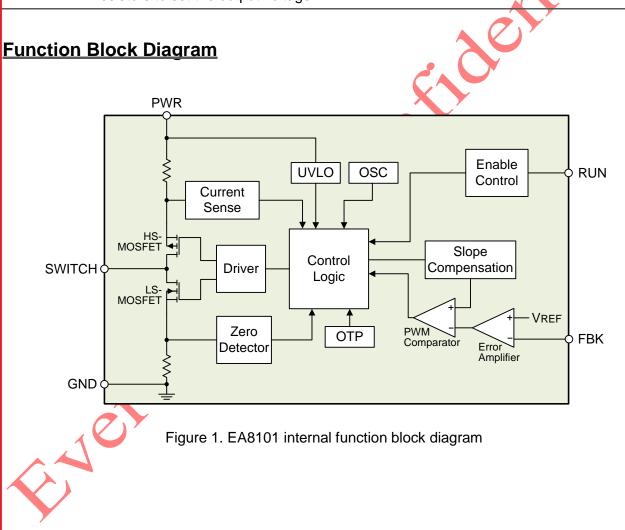
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Pin Description

Pin Name	Function Description			
RUN	The device turns on/turns off control input. Don't leave this pin floating.	1		
GND	Ground pin.	2		
SWITCH	Power switch output pin. Connect SWITCH pin to the switching node of the inductor.	3		
PWR	The EA8101 power input pin. It is recommended to use a 4.7uF MLCC capacitor between PWR pin and GND pin. The ceramic capacitor must be placed as close to the PWR pin as possible to avoid noise interference.	4		
FBK	Feedback input. Connect FBK pin and GND pin with voltage dividing resistors to set the output voltage.	5		

Function Block Diagram



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Absolute Maximum Ratings

Parameter	Value
Input Supply Voltage (V _{PWR})	-0.3V to +6.5V
RUN Pin Input Voltage (V _{RUN})	-0.3V to +6.5V
SWITCH Pin Voltage (V _{SWITCH})	-0.3V to (V _{PWR} +0.3V)
FBK Pin Voltage (V _{FBK})	-0.3V to +6.5V
Ambient Temperature operating Range (T _A)	-40°C to +85°C
Maximum Junction Temperature (T _{Jmax})	+150°C
Lead Temperature (Soldering, 10 sec)	+260°C
Storage Temperature Range (T _s)	-65°C to +150°C

Note (1):Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to "Absolute Maximum Ratings" conditions for extended periods may affect device reliability and lifetime.

Package Thermal Characteristics

Value		
125°C/W		
250°C/W		
0.5W		
125°C/W		
250°C/W		
0.5W		

Note (1): P_{Dmax} is calculated according to the formula: $P_{DMAX}=(T_{JMAX}-T_A)/\theta_{JA}$.

Recommended Operating Conditions

Parameter	Value
Input Supply Voltage (V _{PWR})	+2.5V to +5.5V
Junction Temperature Range (T _J)	-40°C to +125°C

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Electrical Characteristics

 V_{PWR} =3.6V, V_{OUT} =1.8V, T_A =25°C, unless otherwise noted

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Voltage	V_{PWR}		2.5		5.5	V
Shutdown Supply Current	I _{SD}	$V_{RUN} = 0V$		0.1	1	uA
Quiescent Current	Ι _Q	$V_{RUN} = 2V, V_{FBK} =$ 105% $V_{REF}, I_{LOAD} =$ 0A		45	90	uA
UVLO Threshold	V_{UVLO}	V _{PWR} Rising	1.7	1.9	2.1	V
UVLO Hysteresis	$V_{\text{UV-HYST}}$			0.1	N Y	V
Output Load Current	I _{LOAD}				1	А
Reference Voltage	V_{REF}		0.588	0.6	0.612	V
Switching Frequency	F_{SW}	$I_{LOAD} = 100 \text{mA}$	1.2	1.5	1.8	MHz
Short Frequency	F _{SHORT}	$V_{OUT} = 0V$	240	300	360	KHz
PMOS On-Resistance	R _{DS(ON)-P}	I _{LOAD} = 100mA		250		mΩ
NMOS On-Resistance	R _{DS(ON)-N}	I _{LOAD} = 100mA		250		mΩ
PMOS Current Limit	I _{LIM-P}		1.35	1.8		А
SWITCH Leakage Current	I _{LEAK-SWITCH}	$V_{PWR} = 5V, V_{RUN} = 0V, V_{SWITCH} = 0V \text{ or}$ 5V	-1		1	uA
FBK Leakage Current	I _{LEAK-FBK}	V _{FB} = V _{PWR}	-1		1	uA
RUN Pin Input Low Voltage	V _{RUN-L}				0.4	V
RUN Pin Input High Voltage	V _{RUN-H}		1.5			V
Maximum Duty Cycle	D _{MAX}		100			%
Thermal Shutdown Threshold	T _{OTP}			160		°C

Note (1): MOSFET on-resistance specifications are guaranteed by correlation to wafer level measurements.

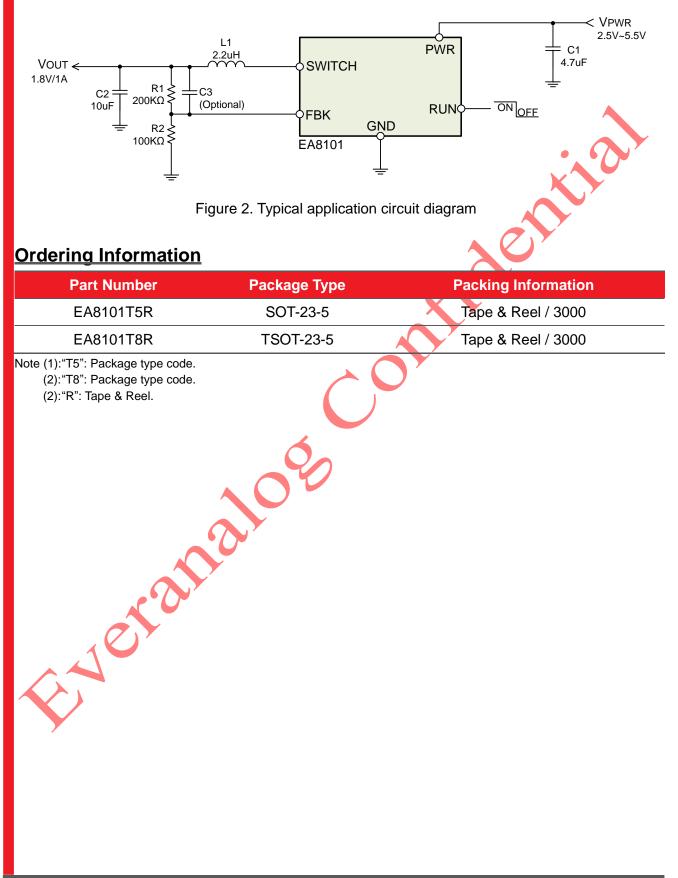
(2): Thermal shutdown specifications are guaranteed by correlation to the design and characteristics analysis.

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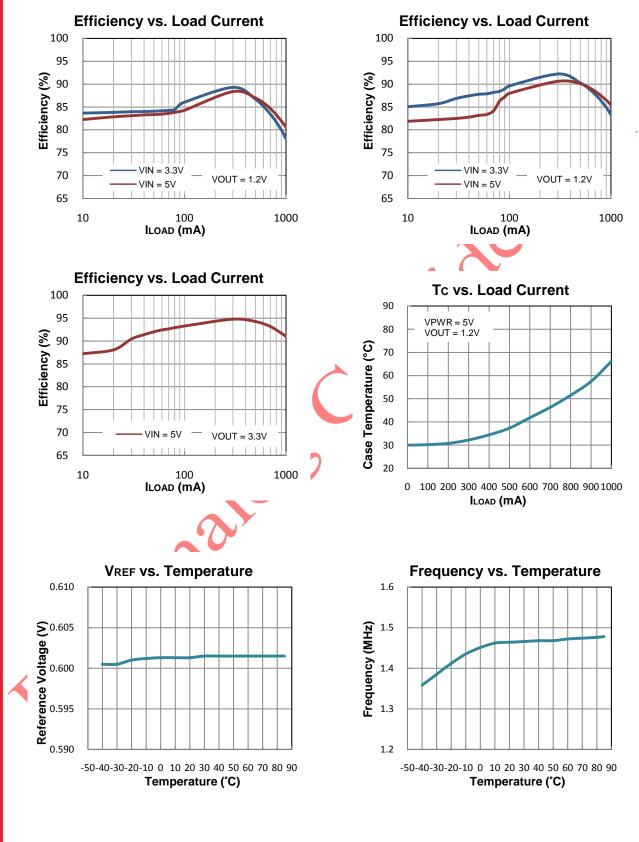
Application Circuit Diagram



1A, 1.5MHz High Efficiency Buck Converter

Typical Operating Characteristics

V_{PWR}=3.6V, V_{OUT}=1.8V, L1=2.2uH, C1=4.7uF, C2=10uF, T_A=25°C, unless otherwise noted

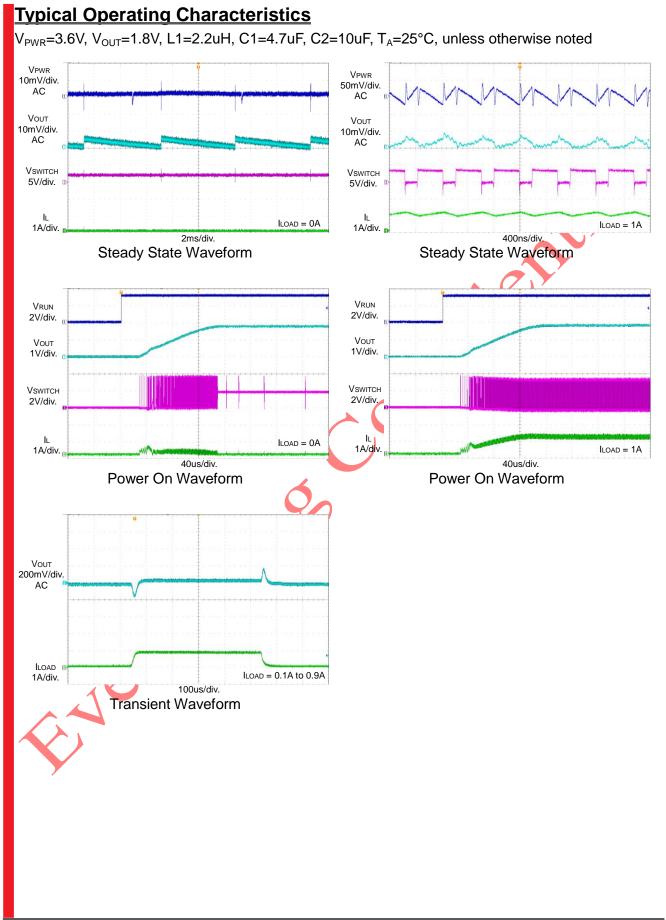




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R1 Resistance	R2 Resistance	Tolerance
510ΚΩ	110ΚΩ	1%
200ΚΩ	100ΚΩ	1%
100ΚΩ	100ΚΩ	1%
68KΩ	100ΚΩ 📐 🤇	1%
	510ΚΩ 200ΚΩ 100ΚΩ	510KΩ 110KΩ 200KΩ 100KΩ 100KΩ 100KΩ

EA8101 1A, 1.5MHz High Efficiency Buck Converter

Application Information

Output Voltage Setting

The EA8101 output voltage can be set via a resistor divider (R1, R2). The output voltage is calculated by following equation:

$$V_{\text{OUT}} = 0.6 \times \frac{R1}{R2} + 0.6 \text{ V}$$

The following table lists common output voltage and the corresponding R1, R2 resistance value for roforonco

Input / Output Capacitors Selection

The input capacitors are used to suppress the noise amplitude of the input voltage and provide a stable and clean DC input to the device. Because the ceramic capacitor has low ESR characteristic, so it is suitable for input capacitor use. It is recommended to use X5R or X7R MLCC capacitors in order to have better temperature performance and smaller capacitance tolerance. In order to suppress the output voltage ripple, the MLCC capacitor is also the best choice. The suggested part numbers of input / output capacitors are as follows:

Vendor	Part Number	Capacitance	Edc	Parameter	Size
TDK	C2012X5R1A475K	4.7uF	10V	X5R	0805
TDK	C3216X7R1A106K	10uF	10V	X7R	1206

Output Inductor Selection

The output inductor selection mainly depends on the amount of ripple current through the inductor ΔI_{L} . Large ΔI_{L} will cause larger output voltage ripple and loss, but the user can use a smaller inductor to save cost and space. On the contrary, the larger inductance can get smaller ΔI_{\perp} and thus the smaller output voltage ripple and loss. But it will increase the space and the cost. The inductor value can be calculated as:

$$L = \frac{V_{PWR} - V_{OUT}}{\Lambda I_{L} \times F_{SW}} \times \frac{V_{OUT}}{V_{PWR}}$$

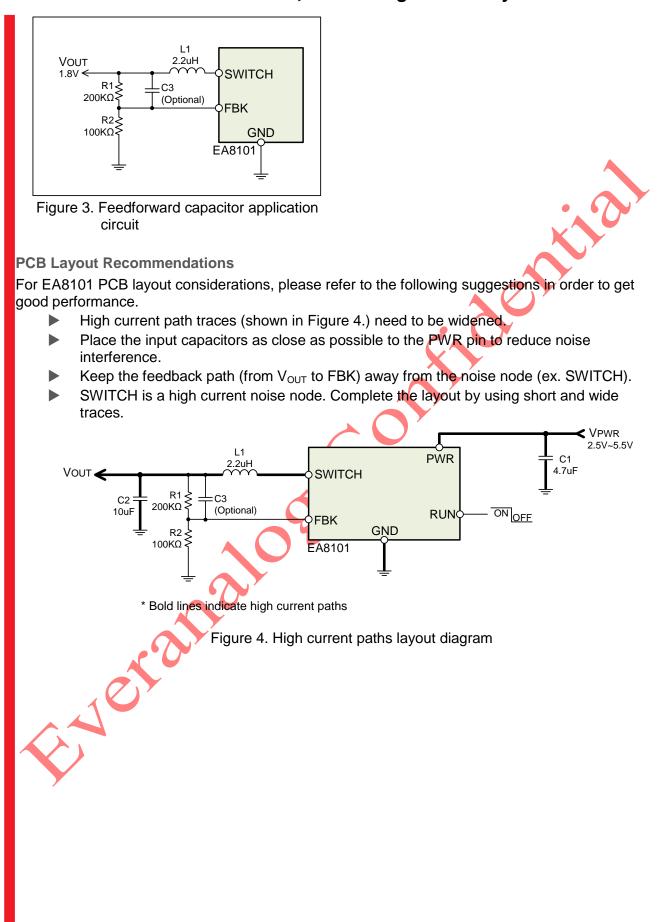
For most applications, 1uH to 4.7uH inductors are suitable for EA8101.

Feedforward Capacitor Selection

The ceramic capacitor C3 which is placed across the feedback resistor R1 is called a feedforward capacitor (shown in Figure 3). It can be used to adjust close loop stability and improve transient response. Generally the suitable capacitance range is from 22pF to 1nF.



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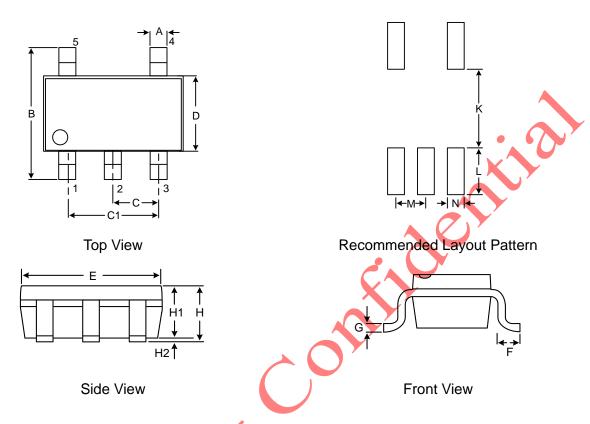
Package Information SOT-23-5 Package 4 B D I←M→I →I N **Recommended Layout Pattern Top View** Е H1 Н H2 Side View Front View Unit: mm Dimension Dimension Symbol Symbol Min Max Тур А 0.30 0.52 K 1.40 2.59 3.01 1.40 В L С 1.05 0.85 Μ 0.95 1.70 C1 2.10 Ν 0.65 1.80 D 1.40 Е 2.70 3.10 F 0.30 0.61 G 0.08 0.25 H 0.89 1.35 H1 0.89 1.20 H2 0.00 0.15



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Package Information



TSOT-23-5 Package

Unit: mm

Symbol	Dime Min	nsion Max	Symbol	Dimension Typ
A	0.30	0.52	K	1.40
В	2.59	3.01	L	1.40
С	0.85	1.05	М	0.95
C1	1.70	2.10	Ν	0.65
D	1.40	1.80		
E	2,70	3.10		
F	0.30	0.61		
G	0.08	0.25		
₩	0.70	1.00		
H1	0.70	0.90		
H2	0.00	0.10		