DACB [

DACA 1 2

DATA 🛮 4

 V_{DD}

DACE ¶ 7

DACF [] 8

CLK Π 5

3

GND [

SLAS089E - NOVEMBER 1994 - REVISED APRIL 1997

16 DACC

15 DACD

14 REF1 13 LDAC

12 LOAD

11 | REF2

10 DACH

9 DACG

N OR DW PACKAGE (TOP VIEW)

- Eight 8-Bit Voltage Output DACs
- 5-V Single-Supply Operation
- Serial Interface
- High-Impedance Reference Inputs
- Programmable 1 or 2 Times Output Range
- Simultaneous Update Facility
- Internal Power-On Reset
- Low-Power Consumption
- Half-Buffered Output

applications

- Programmable Voltage Sources
- Digitally Controlled Amplifiers/Attenuators
- Mobile Communications
- Automatic Test Equipment
- Process Monitoring and Control
- Signal Synthesis

description

The TLC5628C and TLC5628I are octal 8-bit voltage output digital-to-analog converters (DACs) with buffered reference inputs (high impedance). The DACs produce an output voltage that ranges between either one or two times the reference voltages and GND and are monotonic. The device is simple to use, running from a single supply of 5 V. A power-on reset function is incorporated to ensure repeatable start-up conditions.

Digital control of the TLC5628C and TLC5628I are over a simple three-wire serial bus that is CMOS compatible and easily interfaced to all popular microprocessor and microcontroller devices. The 12-bit command word comprises eight bits of data, three DAC select bits, and a range bit, the latter allowing selection between the times 1 or times 2 output range. The DAC registers are double buffered, allowing a complete set of new values to be written to the device, then all DAC outputs are updated simultaneously through control of LDAC. The digital inputs feature Schmitt triggers for high-noise immunity.

The 16-terminal small-outline (D) package allows digital control of analog functions in space-critical applications. The TLC5628C is characterized for operation from 0°C to 70°C. The TLC5628I is characterized for operation from –40°C to 85°C. The TLC5628C and TLC5628I do not require external trimming.

AVAILABLE OPTIONS

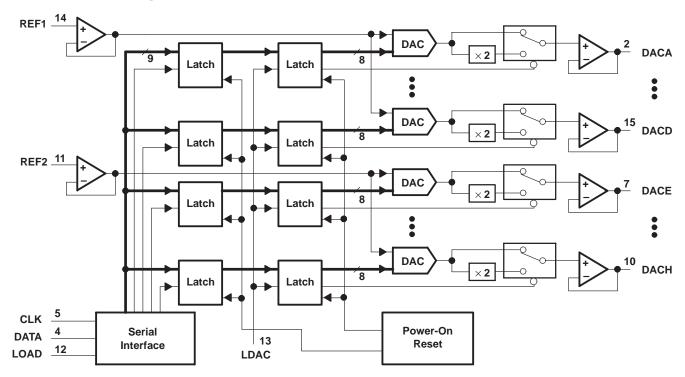
PACKAGE								
TA	SMALL OUTLINE (DW)	PLASTIC DIP (N)						
0°C to 70°C	TLC5628CDW	TLC5628CN						
-40°C to 85°C	TLC5628IDW	TLC5628IN						



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



functional block diagram



Terminal Functions

TERMIN	IAL	.,,	DECODINE
NAME	NO.	1/0	DESCRIPTION
CLK	5	I	Serial interface clock. The input digital data is shifted into the serial interface register on the falling edge of the clock applied to the CLK terminal.
DACA	2	0	DAC A analog output
DACB	1	0	DAC B analog output
DACC	16	0	DAC C analog output
DACD	15	0	DAC D analog output
DACE	7	0	DAC E analog output
DACF	8	0	DAC F analog output
DACG	9	0	DAC G analog output
DACH	10	0	DAC H analog output
DATA	4	ı	Serial interface digital data input. The digital code for the DAC is clocked into the serial interface register serially. Each data bit is clocked into the register on the falling edge of the clock signal.
GND	3	I	Ground return and reference terminal
LDAC	13	ı	Load DAC. When LDAC is high, no DAC output updates occur when the input digital data is read into the serial interface. The DAC outputs are only updated when LDAC is taken from high to low.
LOAD	12	ı	Serial interface load control. When LDAC is low, the falling edge of the LOAD signal latches the digital data into the output latch and immediately produces the analog voltage at the DAC output terminal.
REF1	14	Ι	Reference voltage input to DAC A B C D. This voltage defines the analog output range.
REF2	11	I	Reference voltage input to DAC E F G H. This voltage defines the analog output range.
V_{DD}	6	ı	Positive supply voltage



detailed description

The TLC5628 is implemented using eight resistor-string DACs. The core of each DAC is a single resistor with 256 taps, corresponding to the 256 possible codes listed in Table 1. One end of each resistor string is connected to GND and the other end is fed from the output of the reference input buffer. Monotonicity is maintained by use of the resistor strings. Linearity depends upon the matching of the resistor segments and upon the performance of the output buffer. Since the inputs are buffered, the DACs always present a high-impedance load to the reference sources. There are two input reference terminals; REF1 is used for DACA through DACD and REF2 is used by DACE through DACH.

Each DAC output is buffered by a configurable-gain output amplifier, that can be programmed to times 1 or times 2 gain.

On power up, the DACs are reset to CODE 0.

Each output voltage is given by:

$$V_O(DACA|B|C|D|E|F|G|H) = REF \times \frac{CODE}{256} \times (1 + RNG bit value)$$

where CODE is in the range 0 to 255 and the range (RNG) bit is a 0 or 1 within the serial control word.

D7	D6	D5	D4	D3	D2	D1	D0	OUTPUT VOLTAGE
0	0	0	0	0	0	0	0	GND
0	0	0	0	0	0	0	1	(1/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
0	1	1	1	1	1	1	1	(127/256) × REF (1+RNG)
1	0	0	0	0	0	0	0	(128/256) × REF (1+RNG)
•	•	•	•	•	•	•	•	•
•	•	•	•	•	•	•	•	•
1	1	1	1	1	1	1	1	(255/256) × REF (1+RNG)

Table 1. Ideal Output Transfer

data interface

With LOAD high, data is clocked into the DATA terminal on each falling edge of CLK. Once all data bits have been clocked in, LOAD is pulsed low to transfer the data from the serial input register to the selected DAC as shown in Figure 1. When LDAC is low, the selected DAC output voltage is updated when LOAD goes low. When LDAC is high during serial programming, the new value is stored within the device and can be transferred to the DAC output at a later time by pulsing LDAC low as shown in Figure 2. Data is entered most significant bit (MSB) first. Data transfers using two 8-clock cycle periods are shown in Figures 3 and 4.

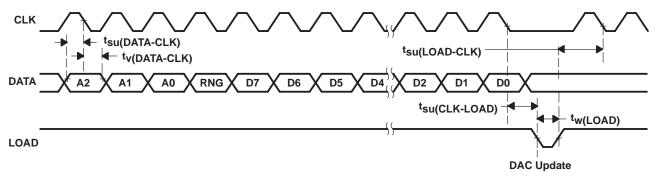


Figure 1. LOAD-Controlled Update (LDAC = Low)



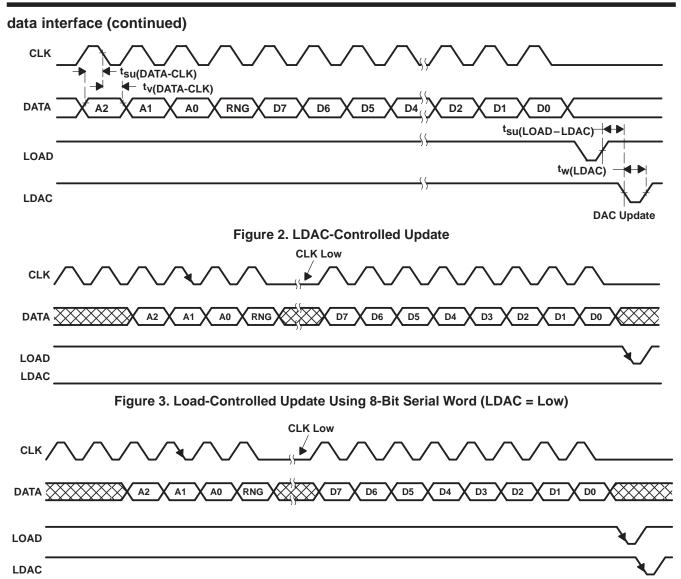


Figure 4. LDAC-Controlled Update Using 8-Bit Serial Word

Table 2 lists the A2, A1, and A0 bits and the selection of the updated DACs. The RNG bit controls the DAC output range. When RNG = low, the output range is between the applied reference voltage and GND, and when RNG = high, the range is between twice the applied reference voltage and GND.

Table 2. Serial Input Decode

A2	A1	A0	DAC UPDATED
0	0	0	DACA
0	0	1	DACB
0	1	0	DACC
0	1	1	DACD
1	0	0	DACE
1	0	1	DACF
1	1	0	DACG
1	1	1	DACH



linearity, offset, and gain error using single-end supplies

When an amplifier is operated from a single supply, the voltage offset can still be either positive or negative. With a positive offset voltage, the output voltage changes on the first code change. With a negative offset the output voltage may not change with the first code depending on the magnitude of the offset voltage.

The output amplifier, therefore, attempts to drive the output to a negative voltage. However, because the most negative supply rail is ground, the output cannot drive below ground.

The output voltage remains at 0 V until the input code value produces a sufficient output voltage to overcome the inherent negative offset voltage, resulting in the transfer function shown in Figure 5.

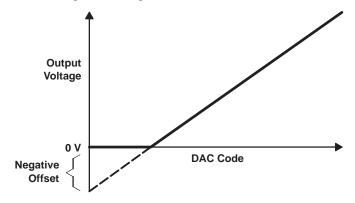


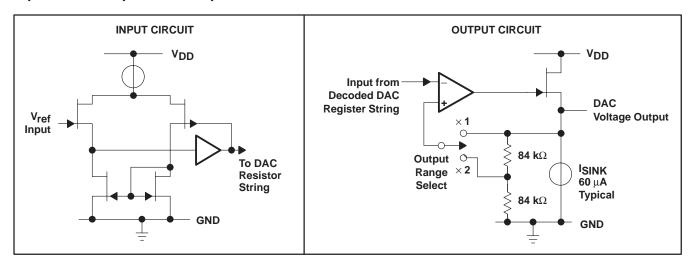
Figure 5. Effect of Negative Offset (Single Supply)

This offset error, not the linearity error, produces the breakpoint. The transfer function would have followed the dotted line if the output buffer could drive below ground.

For a DAC, linearity is measured between the zero-input code (all inputs are 0) and the full-scale code (all inputs are 1) after offset and full scale are adjusted out or accounted for in some way. However, single-supply operation does not allow for adjustment when the offset is negative due to the breakpoint in the transfer function. So the linearity in the unipolar mode is measured between full-scale code and the lowest code that produces a positive output voltage.

The code is calculated from the maximum specification for the negative offset voltage.

equivalent of inputs and outputs



absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage (V _{DD} – GND)	
Digital input voltage range, V _{ID}	GND $- 0.3 \text{ V to V}_{DD} + 0.3 \text{ V}$
Reference input voltage range	GND $- 0.3 \text{ V to V}_{DD}^{-} + 0.3 \text{ V}$
Operating free-air temperature range, T _A : TLC5628C	0°C to 70°C
TLC5628I	40°C to 85°C
Storage temperature range, T _{Stq}	–50°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

[†] Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V _{DD}		4.75		5.25	V
High-level voltage, VIH		0.8 V _{DD}			V
Low-level voltage, V _{IL}				0.8	V
Reference voltage, V _{ref} [A B C D E F G H]				V _{DD} -1.5	V
Analog full-scale output voltage, $R_L = 10 \text{ k}\Omega$			3.5		V
Load resistance, R _L		10			kΩ
Setup time, data input, t _{SU(DATA-CLK)} (see	Figures 1 and 2)	50			ns
Valid time, data input valid after CLK↓, t _{V(DA}	ATA-CLK) (see Figures 1 and 2)	50			ns
Setup time, CLK eleventh falling edge to LO	AD, t _{su(CLK-LOAD)} (see Figure 1)	50			ns
Setup time, LOAD↑ to CLK↓, t _{SU(LOAD-CL}	() (see Figure 1)	50			ns
Pulse duration, LOAD, t _{W(LOAD)} (see Figure	e 1)	250			ns
Pulse duration, LDAC, tw(LDAC) (see Figure	e 2)	250			ns
Setup time, LOAD↑ to LDAC↓, t _{Su(LOAD-L[}	DAC) (see Figure 2)	0			ns
CLK frequency				1	MHz
On creating free circumparature T.	TLC5628C	0		70	°C
Operating free-air temperature, T _A	TLC5628I	-40		85	°C



electrical characteristics over recommended operating free-air temperature range, V_{DD} = 5 V \pm 5%, $V_{ref} = 2 V_1 \times 1$ gain output range (unless otherwise noted)

	PARAMETER	TEST	CONDITIONS	MIN	TYP	MAX	UNIT
lΗ	High-level input current	$V_I = V_{DD}$				±10	μΑ
I _{IL}	Low-level input current	V _I = 0 V				±10	μΑ
I _{O(sink)}	Output sink current	Each DAC out	tout	20			μΑ
I _{O(source)}	Output source current	Each DAC ou	tput	2			mA
C.	Input capacitance				15		n.E
Ci	Reference input capacitance				15		pF
I _{DD}	Supply current	$V_{DD} = 5 V$				4	mA
I _{ref}	Reference input current	$V_{DD} = 5 V$,	V _{ref} = 2 V			±10	μΑ
EL	Linearity error (end point corrected)	$V_{ref} = 2 V$,	×2 gain (see Note 1)			±1	LSB
E _D	Differential-linearity error	$V_{ref} = 2 V$,	× 2 gain (see Note 2)			±0.9	LSB
EZS	Zero-scale error	$V_{ref} = 2 V$,	×2 gain (see Note 3)	0		30	mV
	Zero-scale-error temperature coefficient	V _{ref} = 2 V,	× 2 gain (see Note 4)		10		μV/°C
E _{FS}	Full-scale error	V _{ref} = 2 V,	× 2 gain (see Note 5)			±60	mV
	Full-scale-error temperature coefficient	V _{ref} = 2 V,	× 2 gain (see Note 6)		±25		μV/°C
PSRR	Power supply rejection ratio	See Notes 7 a	and 8		0.5		mV/V

- NOTES: 1. Integral nonlinearity (INL) is the maximum deviation of the output from the line between zero and full scale (excluding the effects of zero code and full-scale errors).
 - 2. Differential nonlinearity (DNL) is the difference between the measured and ideal 1 LSB amplitude change of any two adjacent codes. Monotonic means the output voltage changes in the same direction (or remains constant) as a change in the digital input code.
 - 3. Zero-scale error is the deviation from zero voltage output when the digital input code is zero.
 - 4. Zero-scale-error temperature coefficient is given by: $ZSETC = [ZSE(T_{max}) ZSE(T_{min})]/V_{ref} \times 10^6/(T_{max} T_{min})$.
 - 5. Full-scale error is the deviation from the ideal full-scale output ($V_{ref} 1$ LSB) with an output load of 10 k Ω .
 - $\text{6. Full-scale error temperature coefficient is given by: } \\ \text{FSETC} = [\text{FSE}(T_{max}) \text{FSE}\ (T_{min})] \\ \text{/V}_{ref} \times 10^6 \\ \text{/(}T_{max} T_{min}). \\ \text{(}T_{max} T_{min}) \\ \text{(}T_$
 - 7. Zero-scale-error rejection ratio (ZSE RR) is measured by varying the V_{DD} from 4.5 V to 5.5 V dc and measuring the proportion of this signal imposed on the zero-code output voltage.
 - 8. Full-scale-error rejection ratio (FSE RR) is measured by varying the VDD from 4.5 V to 5.5 V dc and measuring the proportion of this signal imposed on the full-scale output voltage.

operating characteristics over recommended operating free-air temperature range, $V_{DD} = 5 \text{ V} \pm 5\%$, $V_{ref} = 2 V_{ref} \times 1$ gain output range (unless otherwise noted)

	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Output slew rate	$C_L = 100 \text{ pF}, \qquad R_L = 10 \text{ k}\Omega$		1		V/μs
Output settling time	To ± 0.5 LSB, $C_L = 100$ pF, $R_L = 10$ k Ω , See Note 9		10		μs
Large signal bandwidth	Measured at -3 dB point		100		kHz
Digital crosstalk	CLK = 1-MHz square wave measured at DACA-DACD		-50		dB
Reference feedthrough	See Note 10		-60		dB
Channel-to-channel isolation	See Note 11		-60		dB
Reference input bandwidth	See Note 12		100		kHz

- NOTES: 9. Settling time is the time between a LOAD falling edge and the DAC output reaching full-scale voltage within ±0.5 LSB starting from an initial output voltage equal to zero.
 - 10. Reference feedthrough is measured at any DAC output with an input code = 00 hex with a V_{ref} input = 1 V dc + 1 V_{pp} at 10 kHz.
 - 11. Channel-to-channel isolation is measured by setting the input code of one DAC to FF hex and the code of all other DACs to 00 hex with V_{ref} input = 1 V dc + 1 V_{pp} at 10 kHz. 12. Reference bandwidth is the -3 dB bandwidth with an input at V_{ref} = 1.25 V dc + 2 V_{pp} and with a full-scale digital input code.



PARAMETER MEASUREMENT INFORMATION

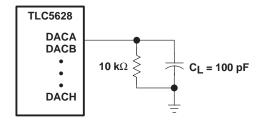


Figure 6. Slew, Settling Time, and Linearity Measurements

TYPICAL CHARACTERISTICS

POSITIVE RISE AND SETTLING TIME

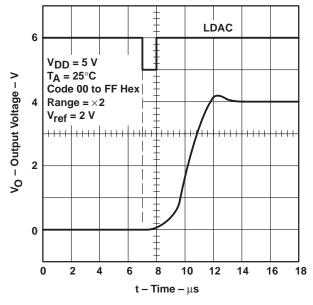


Figure 7

NEGATIVE FALL AND SETTLING TIME

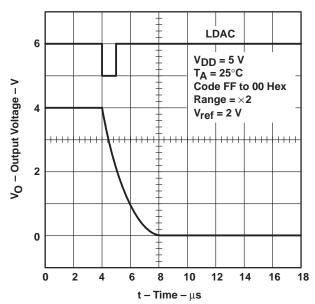
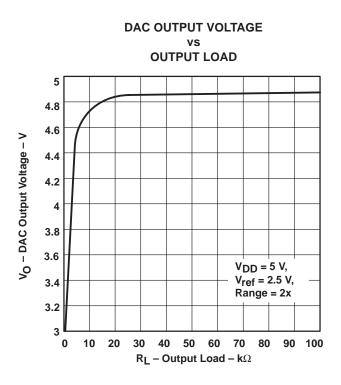


Figure 8

TYPICAL CHARACTERISTICS



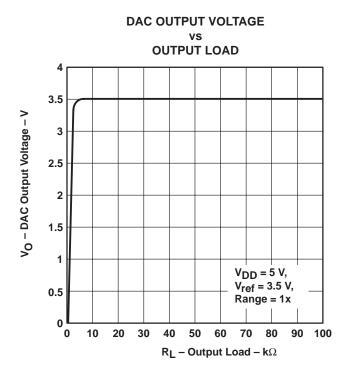
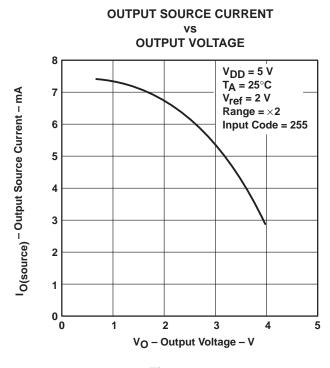


Figure 9





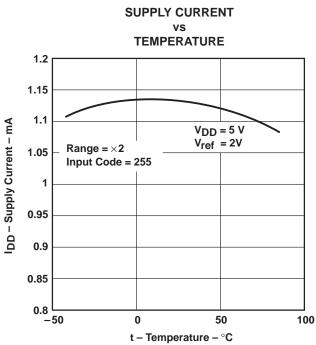
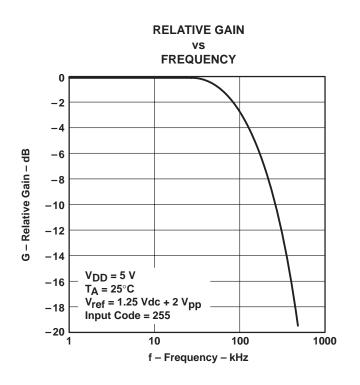


Figure 11

Figure 12

TYPICAL CHARACTERISTICS



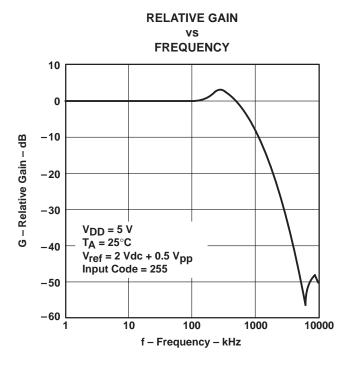
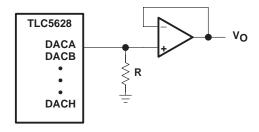


Figure 13

Figure 14

APPLICATION INFORMATION



NOTE A: Resistor R \geq 10 k Ω

Figure 15. Output Buffering Scheme

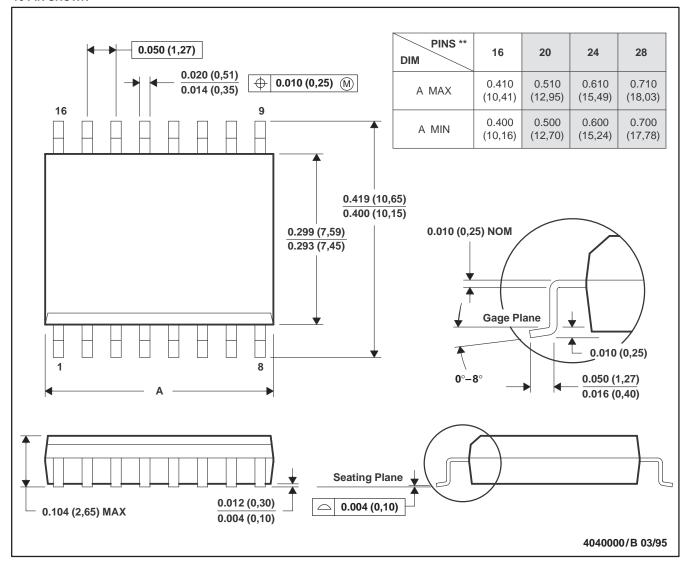


MECHANICAL DATA

DW (R-PDSO-G**)

16 PIN SHOWN

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

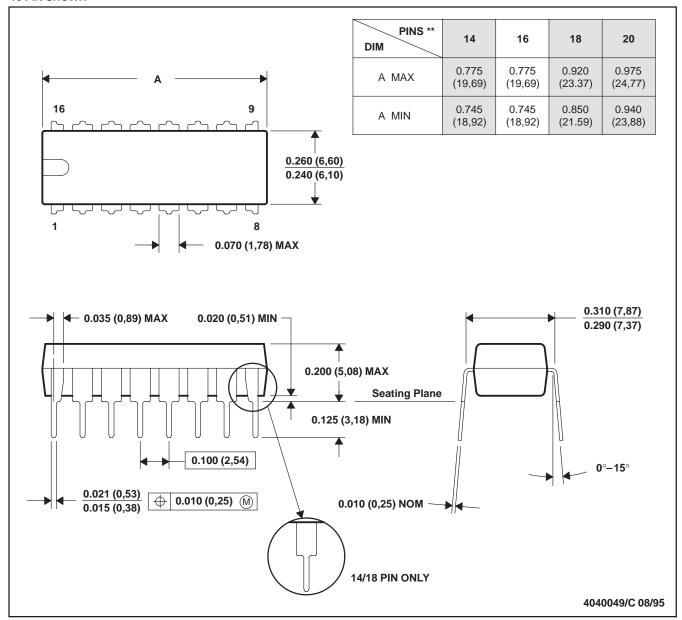
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15).
- D. Falls within JEDEC MS-013

MECHANICAL DATA

N (R-PDIP-T**)

16 PIN SHOWN

PLASTIC DUAL-IN-LINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 (20-pin package is shorter than MS-001)

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
TLC5628CDW	ACTIVE	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC5628C	Samples
TLC5628CDWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	0 to 70	TLC5628C	Samples
TLC5628CN	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type	0 to 70	TLC5628CN	Samples
TLC5628IDW	ACTIVE	SOIC	DW	16	40	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC5628I	Samples
TLC5628IDWR	ACTIVE	SOIC	DW	16	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	TLC5628I	Samples
TLC5628IN	ACTIVE	PDIP	N	16	25	RoHS & Green	NIPDAU	N / A for Pkg Type	-40 to 85	TLC5628IN	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

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- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.



PACKAGE OPTION ADDENDUM

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC5628CDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1
TLC5628IDWR	SOIC	DW	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC5628CDWR	SOIC	DW	16	2000	350.0	350.0	43.0
TLC5628IDWR	SOIC	DW	16	2000	350.0	350.0	43.0

PACKAGE MATERIALS INFORMATION

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TUBE



*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
TLC5628CDW	DW	SOIC	16	40	506.98	12.7	4826	6.6
TLC5628CN	N	PDIP	16	25	506	13.97	11230	4.32
TLC5628IDW	DW	SOIC	16	40	506.98	12.7	4826	6.6
TLC5628IN	N	PDIP	16	25	506	13.97	11230	4.32

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