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特性

ZHCS949B-JUNE 2012-REVISED OCTOBER 2012

带有预充电场效应晶体管 (FET) 的 16 通道、恒定电流发光二级管 (LED) 驱动器

查询样品: TLC59283

- 可进行开关控制的 16 通道、恒定电流灌电流输出
- 恒定电流吸收能力: 35mA (V_{CC} ≤3.6 V), 45mA (V_{CC}>3.6V)
- LED 电源电压: 高达 10V
- V_{CC}=3V 至 5.5V
- 恒定电流精度:
 通道到通道:
 ±1.4%(典型值),±3%(最大值)
 器件到器件:±2%(典型值),±4%(最大值)
- CMOS 逻辑电平 I/O
- 数据传输速率: 35MHz
- BLANK (空白) 脉冲宽度: 50ns
- 用于重影减少的预充电 FET
- 针对降噪的成组开关延迟
- 工作温度范围: -40°C 至 +85°C

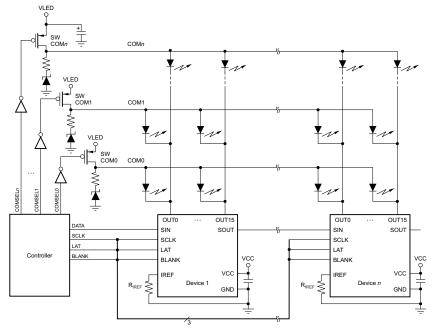
应用

- 视频显示器
- 留言板

说明

TLC59283 是一款 16 通道、恒定电流吸收发光二极管 (LED) 驱动器。每个通道可由一个与 3.3V 或者 5V CMOS 逻辑电平兼容(取决于运行 VCC)的简单串行 通信协议控制。当串行数据缓冲区被载入时,一个锁 存脉冲 (LAT) 上升边沿将数据转移到 OUT*n*输出。 BLANK 引脚可被用于关闭加电和输出数据锁存期间的 的所有 OUT*n*输出以防止此时间段内不必要的图像显 示。所有 16 个通道的恒定电流值有一个单一外部电 阻器设定。

每个恒定电流输出有一个预充电场效应晶体管 (FET), 此晶体管能够减少复用(动态)驱动 LED 显示时的重 影。可将多个 TLC59283 级联在一起以控制来自同一 处理器的额外的 LED。



典型应用电路(多菊花链 TLC59283)

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TLC59283



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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

PACKAGE AND ORDERING INFORMATION **							
PRODUCT	PACKAGE-LEAD	TRANSPORT MEDIA, QUANTITY					
		TLC59283DBQR	Tape and Reel, 2500				
TLC59283	SSOP-24 and QSOP-24	TLC59283DBQ	Tube, 50				
TLC59283		TLC59283RGER	Tape and Reel, 3000				
	QFN-24	TLC59283RGE	Tape and Reel, 250				

PACKAGE AND ORDERING INFORMATION⁽¹⁾

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the device product folder at www.ti.com.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾⁽²⁾

Over operating free-air temperature range, unless otherwise noted.

			VALUE		UNIT
			$\begin{array}{c cccc} -0.3 & + \\ \hline -0.3 & V_{CC} + 0. \\ \hline -0.3 & V_{CC} + 0. \\ \hline -0.3 & +1 \end{array}$	MAX	UNIT
	V _{CC}	Supply	-0.3	+6	V
Voltage	V _{IN}	Input range, SIN, SCLK, LAT, BLANK, IREF	-0.3	$V_{CC} + 0.3$	V
Voltage	V _{OUT}	Output range, SOUT	-0.3	$V_{CC} + 0.3$	V
		Output range, OUT0 to OUT15	-0.3	+11	V
Current	I _{OUT}	Output (dc), OUT0 to OUT15		+50	mA
Tomporofiuro	T _{J(MAX)}	Operating junction		+150	°C
Temperature	T _{stg}	Storage range	-55	+150	°C
Electro de la districtione de la company		Human body model (HBM)		3000	V
Electrostatic discharge ratings	ESD	Charged device model (CDM)		2000	V

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not supported.

(2) All voltage values are with respect to network ground terminal.



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RECOMMENDED OPERATING CONDITIONS

At $T_A = -40^{\circ}$ C to +85°C, unless otherwise noted.

			TLC59	283		
	PARAMETER	R	TEST CONDITIONS	MIN	MAX	UNIT
DC CHARA	ACTERISTICS (V _{CC} = 3	V to 5.5 V)	•		ţ	
V _{CC}	Supply voltage			3	5.5	V
Vo	Voltage applied to o	utput	OUT0 to OUT15		10	V
V _{IH}	Input voltage	High	SIN, SCLK, LAT, BLANK	$0.7 \times V_{CC}$	V _{CC}	V
V _{IL}		Low	SIN, SCLK, LAT, BLANK	GND	$0.3 \times V_{CC}$	V
I _{OH}	Output current	High	SOUT		-2	mA
I _{OL}		Low	SOUT		2	mA
1	Constant output sink		OUT0 to OUT15, 3 V \leq V _{CC} \leq 3.6 V	2	35	mA
I _{OLC}	Constant output sink	current	OUT0 to OUT15, 3.6 V < V _{CC} \leq 5.5 V	2	45	mA
T _A	Temperature range	Operating free-air		-40	+85	°C
TJ	Temperature range	Operating junction		-40	+125	°C
AC CHARA	ACTERISTICS (V _{CC} = 3	V to 5.5 V)				
f _{CLK (SCLK)}	Data shift clock freq	uency	SCLK		35	MHz
t _{WH0}			SCLK	10		ns
t _{WL0}			SCLK	10		ns
t _{WH1}	Pulse duration		LAT	20		ns
t _{WH2}			BLANK	100		ns
t _{WL2}			BLANK	50		ns
suo	Cotup time		SIN↑↓ – SCLK↑	4		ns
^t SU1	Setup time		LAT↓ – SCLK↑	10		ns
t _{H0}	Light time		SIN↑↓ – SCLK↑	4		ns
t _{H1}	Hold time		LAT↓ – SCLK↑	10		ns

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ELECTRICAL CHARACTERISTICS

All minimum and maximum specifications are at $T_A = -40^{\circ}$ C to +85°C and $V_{CC} = 3$ V to 5.5 V, unless otherwise noted. Typical specifications are at $T_A = +25^{\circ}$ C and $V_{CC} = 3.3$ V.

						TLC59283		
	PARAMETER		TEST CONDITIONS		MIN	ТҮР	MAX	UNIT
V _{OH}		High	I _{OH} = -2 mA at SOUT		$V_{CC} - 0.4$		V _{CC}	V
V _{OL}	Output voltage	Low	I _{OL} = 2 mA at SOUT				0.4	V
V _{PCHG}	Pre-charged vol	tage	I _O = -10 μA		V _{CC} - 2.0	$V_{CC} - 1.4$	$V_{CC} - 0.8$	V
V _{IREF}	Reference voltage	ge output	R _{IREF} = 1.5 kΩ, T _A = +25°C			1.208		V
I _{IN}	Input current		V _{IN} = V _{CC} or GND at SIN and SCLK		-1		1	μA
I _{CC0}			SIN, SCLK, LAT = GND, BLANK = V _{OUTn} = V	_{CC} , R _{IREF} = open		1	2	mA
I _{CC1}			SIN, SCLK, LAT = GND, BLANK = $V_{OUTn} = V$ R _{IREF} = 3 k Ω (I _{OUT} = 17.6 mA target)	'сс,		3	4	mA
I _{CC2}	Supply current (V _{CC})	All OUT n = ON, SIN, SCLK, LAT, BLANK = C V _{OUTn} = 0.8 V, R _{IREF} = 3 k Ω		7	9	mA	
I _{CC3}	_		All OUT $n = ON$, SIN, SCLK, LAT, BLANK = GND, V _{OUT n} = 0.8 V, R _{IREF} = 1.5 k Ω (I _{OUT} = 35.3 mA target)			8	11	mA
lolc	Constant output current		All OUT $n = ON$, $V_{OUTn} = V_{OUTfix} = 0.8$ V, $R_{IREF} = 1.5$ k Ω , $T_A = +25^{\circ}C$ (see Figure 8)		32.9	35.3	37.7	mA
	Output leakage current		All OUT n = OFF, V _{OUTn} = V _{OUTfix} = 10 V, BLANK = V _{CC} , R _{IRFF} = 1.5 k Ω (see Figure 8) T _J = +85°C	$T_J = +25^{\circ}C$			0.1	μA
OLKG0						0.2	μA	
			$T_{J} = +125^{\circ}C$			0.07	0.5	μA
∆I _{OLC0}	Constant-	Channel-to- channel ⁽¹⁾	All OUT $n = ON$, $V_{OUTn} = V_{OUTfix} = 0.8$ V, R_{IRE} T _A = +25°C (see Figure 8)	_F = 1.5 kΩ,		±1.4	±3	%
∆I _{OLC1}	current error	Device-to- device ⁽²⁾	All OUT $n = ON$, $V_{OUTn} = V_{OUTfix} = 0.8$ V, R_{IRE} T _A = +25°C (see Figure 8)	_F = 1.5 kΩ,		±2	±4	%
∆I _{OLC2}	Line regulation ⁽³	3)	All OUT n = ON, V _{OUTn} = V _{OUTfix} = 0.8 V, R _{IREF} = 1.5 kΩ, V _{CC} = 3 V to 5.5 V			±0.05	±1	%/V
∆I _{OLC3}	Load regulation	d regulation ⁽⁴⁾ All OUT $n = ON$, $V_{OUTn} = 0.8 V$ to 3 V, $V_{OUTfix} = 0.8 V$, $R_{IREF} = 1.5 k\Omega$			±0.5	±1	%/V	
R _{PUP}	Decistor	Pull-up	BLANK		250	500	750	kΩ
R _{PDWN}	- Resistor	Pull-down	LAT		250	500	750	kΩ
R _{PCHG}	Pre-charge FET	on-resistance	VCC = 5.0 V, V_{OUTn} = 0 V, OUT0 to OUT15, BLANK = V_{CC} , T_A = +25°C			3	6	kΩ

(1) The deviation of each output from the average of OUT0 to OUT15 constant-current. Deviation is calculated by the formula:

$$\Delta (\%) = \left[\frac{l_{\text{OUTn}}}{\frac{(l_{\text{OUT0}} + l_{\text{OUT1}} + \dots + l_{\text{OUT14}} + l_{\text{OUT15}})}{16}} - 1 \right]$$

The deviation of the OUT0 to OUT15 constant-current average from the ideal constant-current value. Deviation is calculated by the following formula: (2)

× 100

$$\Delta (\%) = \left[\frac{\frac{(I_{OUT0} + I_{OUT1} + ... I_{OUT14} + I_{OUT15})}{16} - (Ideal Output Current)} \right] \times 100$$
Ideal current is calculated by the formula:
$$I_{OUT(IDEAL)} = 43.8 \times \left[\frac{1.208 \text{ V}}{\text{R}_{IREF}} \right]$$
(3) Line regulation is calculated by this equation:
$$\Delta (\%/\text{V}) = \left[\frac{(I_{OUTn} \text{ at } V_{CC} = 5.5 \text{ V}) - (I_{OUTn} \text{ at } V_{CC} = 3 \text{ V})}{(I_{OUTn} \text{ at } V_{CC} = 3 \text{ V})} \right] \times \frac{100}{5.5 \text{ V} - 3 \text{ V}}$$
(4) Load regulation is calculated by the equation:
$$\left[(I_{OUTn} \text{ at } V_{OUTn} = 3 \text{ V}) - (I_{OUTn} \text{ at } V_{OUTn} = 1 \text{ V}) \right] = 100$$

$$\Delta (\%/V) = \left[\frac{(I_{OUTn} \text{ at } V_{OUTn} = 3 \text{ V}) - (I_{OUTn} \text{ at } V_{OUTn} = 1 \text{ V})}{(I_{OUTn} \text{ at } V_{OUTn} = 1 \text{ V})} \right] \times \frac{100}{3 \text{ V} - 1 \text{ V}}$$



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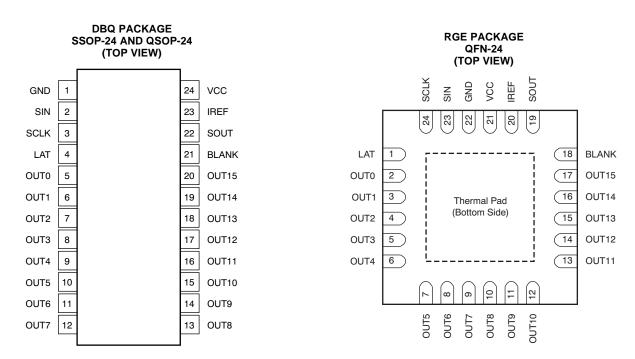
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SWITCHING CHARACTERISTICS

All minimum and maximum specifications are at $T_A = -40^{\circ}$ C to $+85^{\circ}$ C, $V_{CC} = 3$ V to 5.5 V, $C_L = 15$ pF, $R_L = 110 \Omega$, $R_{IREF} = 1.5 \text{ k}\Omega$, and $V_{LED} = 5.0$ V, unless otherwise noted. Typical values are at $T_A = +25^{\circ}$ C and $V_{CC} = 3.3$ V.

			TLC59283				
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
t _{R0}	Diag time	SOUT (see Figure 7)		3	10	ns	
t _{R1}	Rise time	OUT <i>n</i> (see Figure 6)		44		ns	
t _{F0}	- Fall time	SOUT (see Figure 7)		3	10	ns	
t _{F1}		OUT <i>n</i> (see Figure 6)		44		ns	
t _{D0}	Propagation delay time	SCLK↑ to SOUT↑↓		11	20	ns	
t _{D1}		LAT \uparrow or BLANK $\uparrow\downarrow$ to OUT0 on or off, T _A = +25°C		60	100	ns	
t _{D2}		Grouped OUT <i>n</i> on or off to next group on or off, $T_A = +25^{\circ}C$		2		ns	
t _{ON_ERR}	Output on-time error ⁽¹⁾	Output on or off latch data = all '1', 50-ns BLANK GND level pulse, V _{CC} = 3.3 V, T _A = +25°C	-45		45	ns	

(1) Output on-time error (t_{ON_ERR}) is calculated by the formula: t_{ON_ERR} (ns) = t_{OUT_ON} - BLANK low level one-shot pulse width (t_{WL2}). t_{OUT_ON} indicates the actual on-time of the constant-current output.



PIN CONFIGURATIONS

NOTE: Thermal pad is not connected to GND internally. The thermal pad must be connected to GND via the printed circuit board (PCB) pattern.

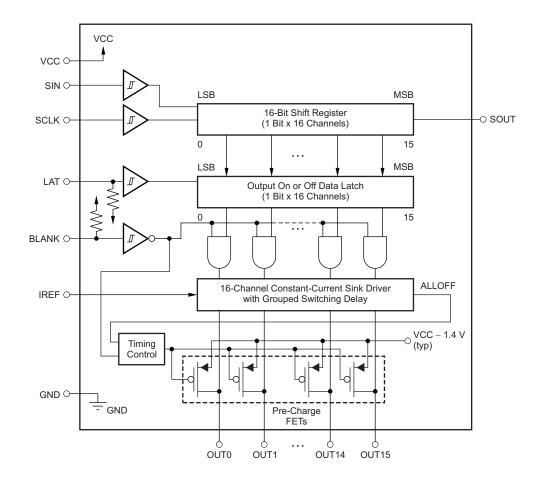
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	PIN			
	NUM	IBER		
NAME	DBQ	RGE	I/O	DESCRIPTION
BLANK	21	18	I	All outputs empty (blank); Schmitt buffer input. When BLANK is high, all constant-current outputs (OUT0 to OUT15) are forced off and all pre-charge FETs are turned on. When BLANK is low, all constant-current outputs are controlled by the data in the output on or off data latch and all pre-charge FETs are turned off. This pin is internally pulled up to V _{CC} with a 500-k Ω (typ) resistor.
GND	1	22	_	Power ground
IREF	23	20	I/O	Constant-current value setting, the OUT0 to OUT15 sink constant-current outputs are set to the desired values by connecting an external resistor between IREF and GND.
LAT	4	1	I	Level-triggered latch; Schmitt buffer input. The data in the 16-bit shift register continue to transfer to the output on or off data latch while LAT is high. Therefore, if the data in the 16-bit shift register are changed when LAT is high, the data in the data latch are also changed. The data in the data latch are held when LAT is low. This pin is internally pulled down to GND with a 500-k Ω (typ) resistor.
OUT0	5	2	0	Constant-current output. Each output can be tied together with others to increase the constant- current. Different voltages can be applied to each output.
OUT1	6	3	0	Constant-current output
OUT2	7	4	0	Constant-current output
OUT3	8	5	0	Constant-current output
OUT4	9	6	0	Constant-current output
OUT5	10	7	0	Constant-current output
OUT6	11	8	0	Constant-current output
OUT7	12	9	0	Constant-current output
OUT8	13	10	0	Constant-current output
OUT9	14	11	0	Constant-current output
OUT10	15	12	0	Constant-current output
OUT11	16	13	0	Constant-current output
OUT12	17	14	0	Constant-current output
OUT13	18	15	0	Constant-current output
OUT14	19	16	0	Constant-current output
OUT15	20	17	0	Constant-current output
SCLK	3	24	I	Serial data shift clock; Schmitt buffer input. All data in the 16-bit shift register are shifted toward the MSB by a 1-bit SCLK synchronization.
SIN	2	23	I	Serial data input for driver on or off control; Schmitt buffer input. When SIN is high, the LSB is set to '1' for only one SCLK input rising edge. If two SCLK rising edges are input while SIN is high, then the 16-bit shift register LSB and LSB+1 are set to '1'. When SIN is low, the LSB is set to '0' at the SCLK input rising edge.
SOUT	22	19	0	Serial data output. This output is connected to the 16-bit shift register MSB. SOUT data changes at the SCLK rising edge.
VCC	24	21		Power-supply voltage

PIN DESCRIPTIONS

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FUNCTIONAL BLOCK DIAGRAM

PARAMETER MEASUREMENT INFORMATION

PIN-EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS

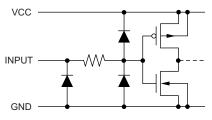
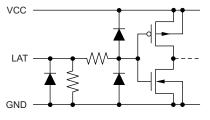


Figure 1. SIN and SCLK





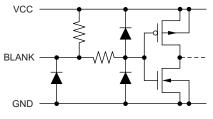
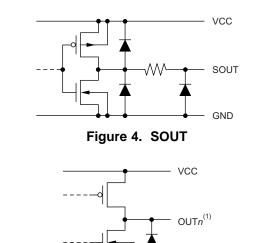


Figure 3. BLANK



(1) n = 0 to 15.

Figure 5. OUT0 Through OUT15

GND

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TEST CIRCUITS

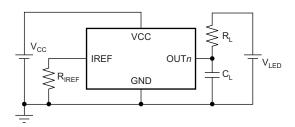


Figure 6. OUT*n* Rise and Fall Time Test Circuit

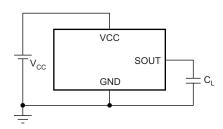


Figure 7. SOUT Rise and Fall Time Test Circuit

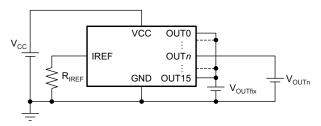


Figure 8. OUTn Constant-Current Test Circuit



TIMING DIAGRAMS

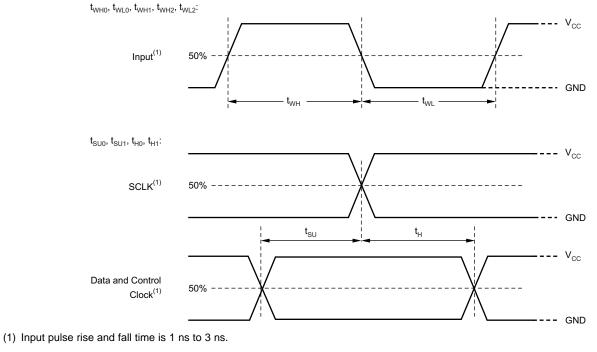
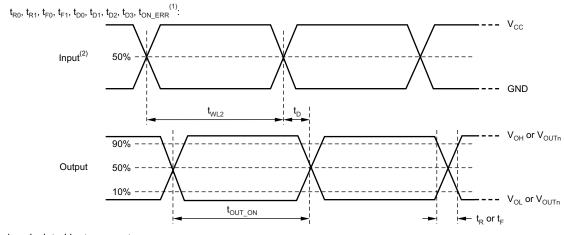
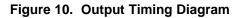


Figure 9. Input Timing Diagram

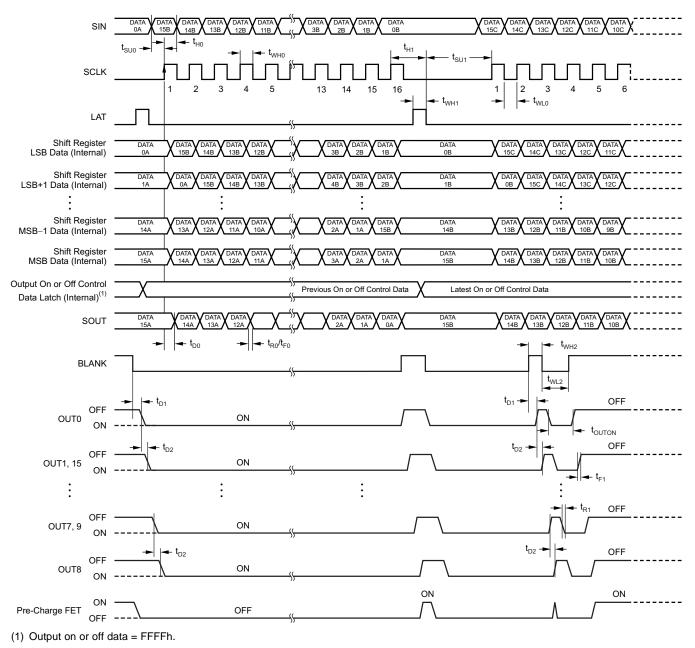


(1) t_{ON_ERR} is calculated by t_{OUTON} – $t_{\text{WL2}}.$

(2) Input pulse rise and fall time is 1 ns to 3 ns.





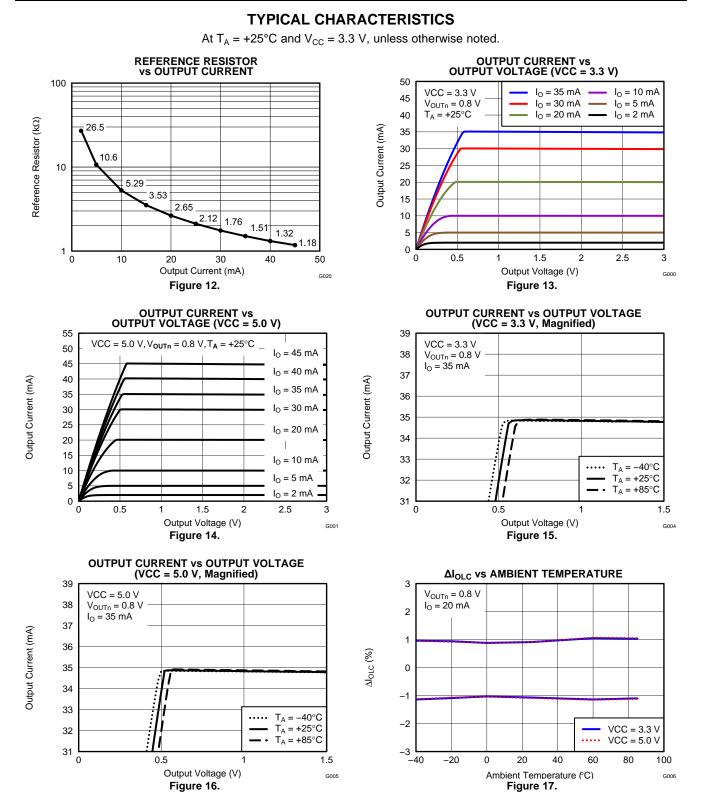


(2) $t_{ON_ERR} = t_{OUTON} - t_{WL2}$.



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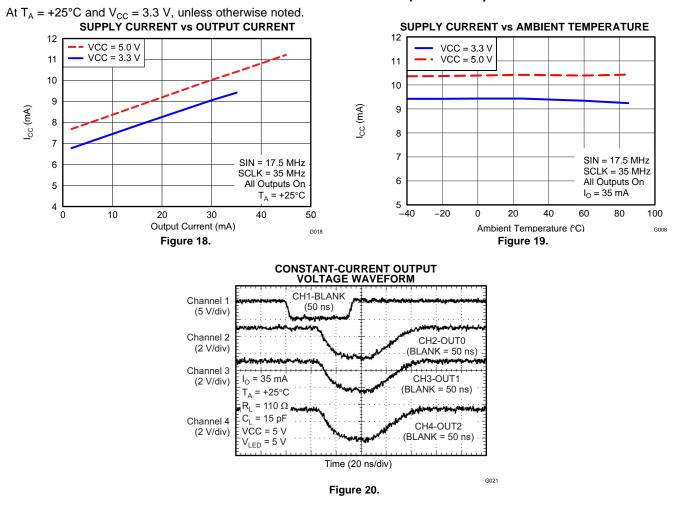
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TYPICAL CHARACTERISTICS (continued)



DETAILED DESCRIPTION

CONSTANT SINK CURRENT VALUE SETTING

The constant-current values are determined by an external resistor (R_{IREF}) placed between IREF and GND. The resistor (R_{IREF}) value is calculated by Equation 1.

$$\mathsf{R}_{\mathsf{IREF}}(\mathsf{k}\Omega) = \frac{\mathsf{V}_{\mathsf{IREF}}(\mathsf{V})}{\mathsf{I}_{\mathsf{OLC}}(\mathsf{m}\mathsf{A})} \times 43.8$$

Where:

 V_{IREF} = the internal reference voltage on the IREF pin (typically 1.208 V)

 I_{OLC} must be set in the range of 2 mA to 35 mA when V_{CC} is less than 3.6 V. Also, when V_{CC} is equal to 3.6 V or greater, I_{OLC} must be set in the range of 2 mA to 45 mA. The constant sink current characteristic for the external resistor value is illustrated in Figure 12. Table 1 describes the constant-current output versus external resistor value.

Table 1. Constant-Current Output versus External Resistor Value							
I _{OLC} (mA)	R _{IREF} (kΩ, Typical)						
45 (V _{CC} > 3.6 V only)	1.18						
40 (V _{CC} > 3.6 V only)	1.32						
35	1.51						
30	1.76						
25	2.12						
20	2.65						
15	3.53						
10	5.29						
5	10.6						
2	26.5						

Table 1. Constant-Current Output versus External Resistor Value

CONSTANT-CURRENT DRIVER ON OR OFF CONTROL

When BLANK is low, the corresponding output is turned on if the data in the on or off control data latch are '1' and remains off if the data are '0'. When BLANK is high, all outputs are forced off. This control is shown in Table 2.

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OUTPUT ON OR OFF DATA	CONSTANT-CURRENT OUTPUT STATUS
0	Off
1	On

When the device is initially powered on, the data in the 16-bit shift register and output on or off data latch are not set to default values. Therefore, the output on or off data must be written to the data latch before turning the constant-current output on. **BLANK should be high when powered on because the constant-current may be turned on as a result of random data in the output on or off data latch.**

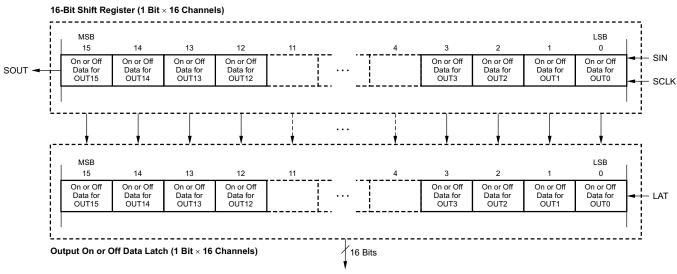


(1)



REGISTER CONFIGURATION

The TLC59283 has a 16-bit shift register and an output on or off data latch. Both the shift register and data latch are 16 bits long and are used to turn the constant-current outputs on and off. Figure 21 shows the shift register and data latch configuration. The data at the SIN pin are shifted into the 16-bit shift register LSB at the rising edge of the SCLK pin; SOUT data change at the SCLK rising edge.



To Constant-Current Driver Control Block

Figure 21. 16-Bit Shift Register and Output On or Off Data Latch Configuration

The output on or off data in the 16-bit shift register continue to transfer to the output on or off data latch while LAT is high. Therefore, if the data in the 16-bit shift register are changed when LAT is high, the data in the data latch are held when LAT is low. When the device initially powers on, the data in the output on or off shift register and latch are not set to default values; on or off control data must be written to the on or off control data latch before turning the constant-current output on. All constant-current outputs are forced off when BLANK is high. The OUT*n* on or off outputs are controlled by the data in the output on or off data truth table and timing diagram are shown in Table 3 and Figure 22, respectively.

	•								
SCLK	LAT	BLANK	SIN	OUT0OUT7OUT15	SOUT				
<u>↑</u>	High	Low	Dn	DnDn – 7Dn – 15	Dn – 15				
<u>↑</u>	Low	Low	Dn + 1	No change	Dn – 14				
↑	High	Low	Dn + 2	Dn + 2Dn – 5Dn – 13	Dn – 13				
Ļ	—	Low	Dn + 3	Dn + 2Dn – 5Dn – 13	Dn – 13				
\downarrow	—	High	Dn + 3	Off	Dn – 13				

Table 3. Truth Table in Operation

TLC59283

TEXAS INSTRUMENTS

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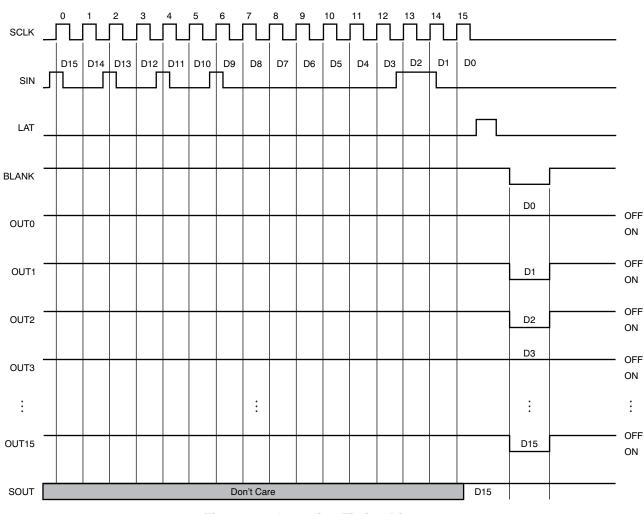


Figure 22. Operation Timing Diagram

NOISE REDUCTION

Large surge currents may flow through the device and board if all 16 outputs turn on or off simultaneously. These large current surges can induce detrimental noise and electromagnetic interference (EMI) into other circuits. The TLC59283 independently turns on or off the outputs for each group with a 1-ns (typ) delay time; see Figure 11. The 16 outputs are grouped into nine groups of either one or two outputs: group 1 (OUT0), group 2 (OUT1 and OUT15), group 3 (OUT2 and OUT14), group 4 (OUT3 and OUT13), group 5 (OUT4 and OUT12), group 6 (OUT5 and OUT11), group 7 (OUT6 and OUT10), group 8 (OUT7 and OUT9), and group 9 (OUT9). Both turn-on and turn-off times are delayed when BLANK transitions from low to high or high to low. Also when output-on and -off data are changed at the LAT rising edge while BLANK is low, both turn-on and turn-off times are delayed. However, the state of each output is controlled by the data in the output on or off data latch and the BLANK level.



TLC59283

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Internal Pre-Charge FET

The internal pre-charge FET prevents ghosting of multiplexed LED modules. One cause of this phenomenon is the parasitic capacitance charging current of the constant-current outputs (OUT*n*) and PCB wiring connected to OUT*n* through the LED. One of the mechanisms is shown in Figure 23.

In Figure 23, the constant-current driver turns LED0-0 on at (1) and off at (2). After LED0-0 is turned off, the OUT0 voltage is pulled up to V_{CHG} by LED0-0. This OUT0 node has some parasitic capacitance (such as the constant-current driver output capacitance and the board layout capacitance shown as C0-2). After LED0-0 turns off, SWPMOS0 is turned off, SWNMOS0 is turned on for COM0, and COM0 is pulled down to GND. Because there is a parasitic capacitance between COM0 and OUT0, the OUT0 voltage is also pulled down to GND. Afterwards, SWPMOS1 is turned on for the next common line (COM1). When SWPMOS1 turns on, the OUT0 voltage is pulled up from the ground voltage to $V_{LED} - V_F$. The charge current (I_{CHRG}) flows to the parasitic capacitor (C0) through LED1-0, causing the LED to briefly turn on and creating a ghosting effect of LED1-0.

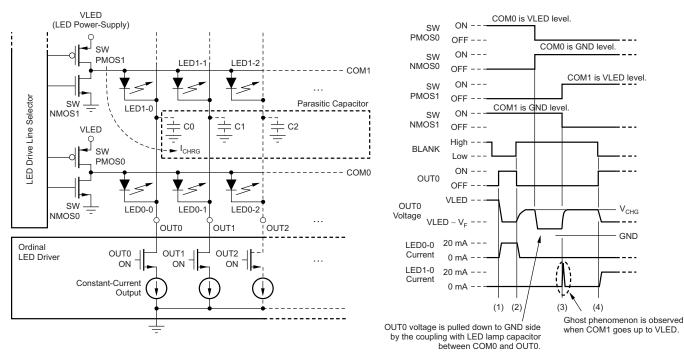


Figure 23. LED Ghost-Lighting Phenomenon Mechanism

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The TLC59283 has an internal pre-charge FET to prevent ghosting, as shown in Figure 24. When a small delay after PWM control for a single common line completes, the FET pulls OUTn up to V_{CC} . The charge current does not flow to C0 through LED1-0 when SWMOS1 is turned on and the ghosting is eliminated at (3). However, depending on the LED anode voltage, the number of LEDs in series, the LED forward voltage, and the TLC59283 V_{CC} supply voltage, there may not be a great enough ghost-canceling effect.

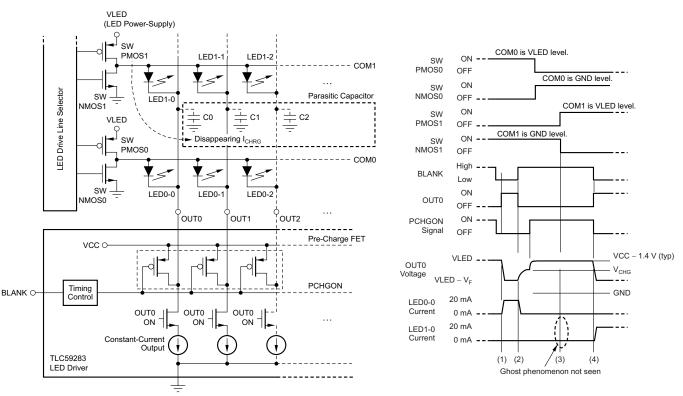


Figure 24. LED Ghost-Lighting Mechanism by Pre-Charge FET



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HISTORY TABLE

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	Changes from Revision A (June 2012) to Revision B F					
•	Changed HBM ESD rating maximum specification in the Absolute Maximum Ratings table	2				
•	Changed I _{CC2} typical and maximum specifications in Electrical Characteristics table	4				
•	Changed I _{CC3} typical specification in Electrical Characteristics table	4				



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLC59283DBQ	ACTIVE	SSOP	DBQ	24	50	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TLC59283	Samples
TLC59283DBQR	ACTIVE	SSOP	DBQ	24	2500	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TLC59283	Samples
TLC59283RGER	ACTIVE	VQFN	RGE	24	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TLC 59283	Samples
TLC59283RGET	ACTIVE	VQFN	RGE	24	250	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	TLC 59283	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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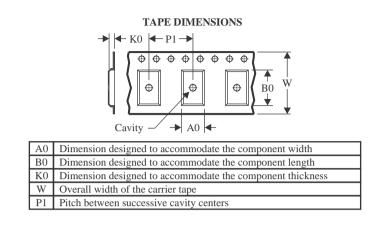


Texas

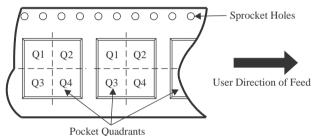
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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nor	ninal											
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLC59283DBQR	SSOP	DBQ	24	2500	330.0	16.4	6.5	9.0	2.1	8.0	16.0	Q1
TLC59283RGER	VQFN	RGE	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
TLC59283RGET	VQFN	RGE	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2



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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

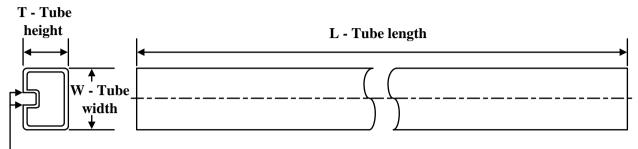
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLC59283DBQR	SSOP	DBQ	24	2500	356.0	356.0	35.0
TLC59283RGER	VQFN	RGE	24	3000	367.0	367.0	35.0
TLC59283RGET	VQFN	RGE	24	250	210.0	185.0	35.0

TEXAS INSTRUMENTS

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TUBE



- B - Alignment groove width

*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
TLC59283DBQ	DBQ	SSOP	24	50	506.6	8	3940	4.32

GENERIC PACKAGE VIEW

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

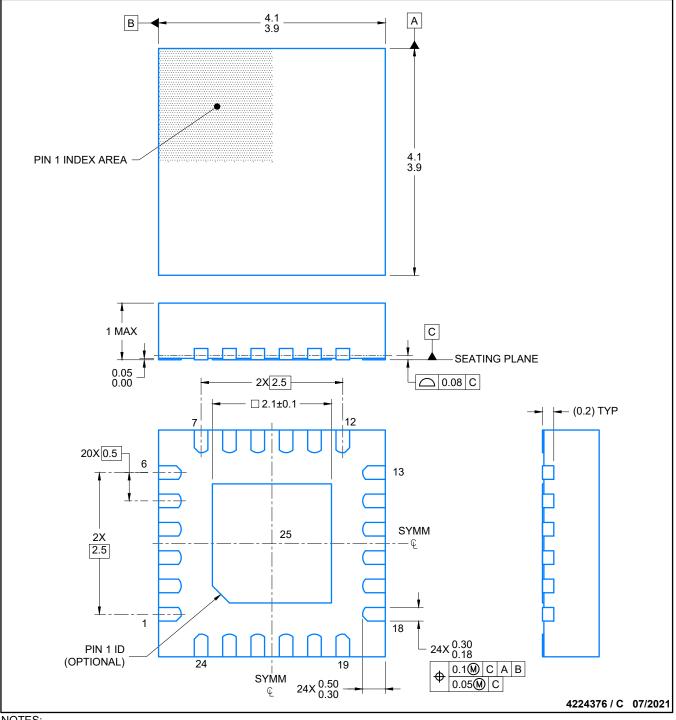


RGE0024C

PACKAGE OUTLINE

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK- NO LEAD



NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing 1. per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

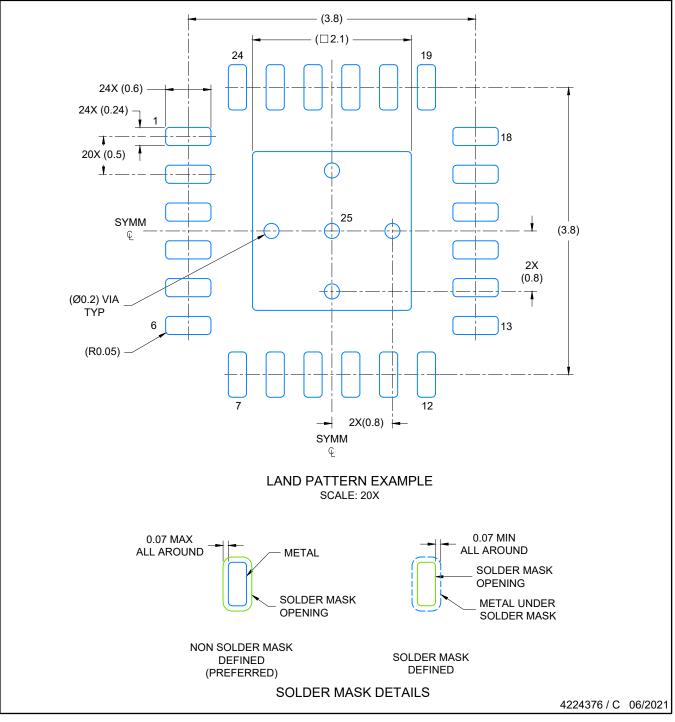


RGE0024C

EXAMPLE BOARD LAYOUT

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK- NO LEAD



NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

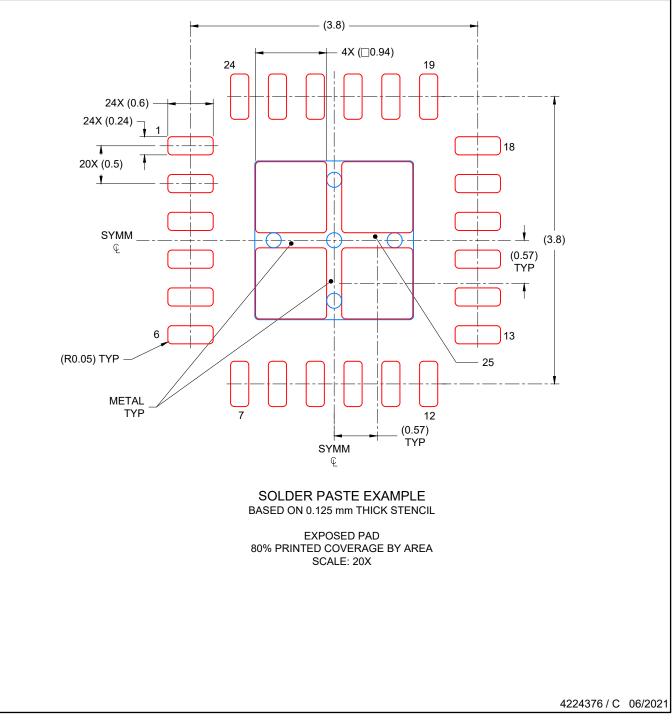


RGE0024C

EXAMPLE STENCIL DESIGN

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK- NO LEAD



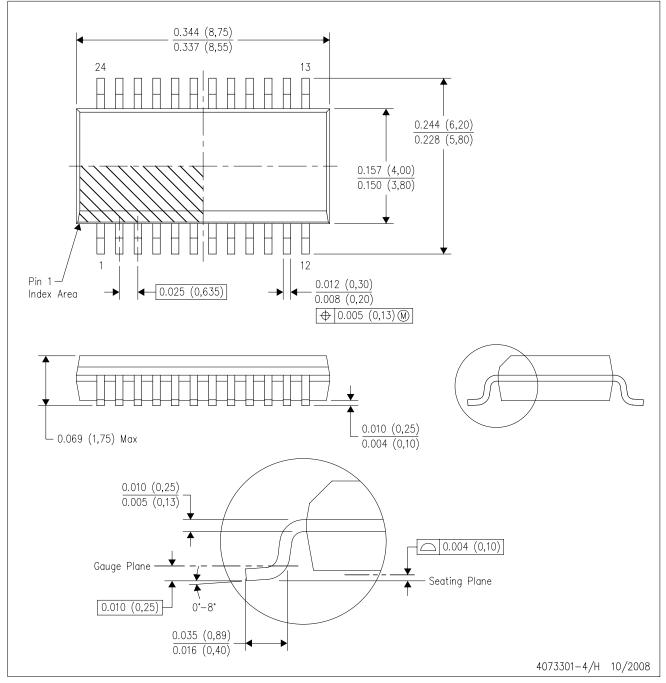
NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations..



DBQ (R-PDSO-G24)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in inches (millimeters).

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15) per side.

D. Falls within JEDEC MO-137 variation AE.



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