

Technical documentation







TCA4307 ZHCSLQ0 - AUGUST 2020

TCA4307 具有阻塞总线恢复功能的热插拔 I²C 总线和 SMBus 缓冲器

1 特性

- 支持 I²C 总线信号的双向数据传输 •
- 工作电源电压范围为 2.3V 至 5.5V
- -40°C 至 125°C 的 T_A 环境空气 温度范围
- 具有自动总线恢复功能的阻塞总线恢复
- 对所有 SDA 和 SCL 线路的 1V 预充电可防止带电 板插入过程中发生损坏
- 可适应标准模式及快速模式 I²C 器件
- 支持时钟展宽、仲裁及同步
- 断电高阻抗 I²C 引脚 •

2 应用

- 服务器
- 企业交换
- 电信交换设备
- 基站
- 工业自动化设备

3 说明

TCA4307 是一款热插拔 I²C 总线缓冲器,支持将 I/O 卡插入带电背板中,而不会损坏数据和破坏时钟线路。 控制电路可防止背板侧 I²C 线路 (输入) 与板卡侧 I²C 线路(输出)相连接(直到背板上出现停止命令或总线 空闲情况为止),而不会在板卡上发生总线争用的情 况。当建立连接时,该器件可提供双向缓冲,从而使背 板及板卡电容保持隔离。在插入过程中,会对 SDA 和 SCL 线路预充电至 1V,从而有效减小对器件的寄生电 容充电所需的电流。

TCA4307 具有阻塞总线恢复功能,当它检测到 SDAOUT 或 SCLOUT 处于低电平状态大约 40ms 时,将自动断开总线。总线断开之后,该器件会在 SCLOUT 上自动生成多达 16 个脉冲,以尝试复位使总 线保持低电平的器件。

当 I²C 总线空闲时,可通过将 EN 引脚设置为低电平将 TCA4307 置于关断模式,从而降低功耗。当 EN 被拉 高时,TCA4307将恢复正常运行。该器件还包括一个 开漏 READY 输出引脚,该引脚负责在背板与板卡侧相 连时发出指示信号。当 READY 引脚为高电平时, SDAIN 和 SCLIN 被连接至 SDAOUT 和 SCLOUT。当 两侧断开时,READY 引脚为低电平。

哭件信息

器件型号	封装 ⁽¹⁾	封装尺寸(标称值)	
TCA4307	VSSOP (8)	3.00mm × 3.00mm	

如需了解所有可用封装,请参阅数据表末尾的可订购产品附 (1) 录。

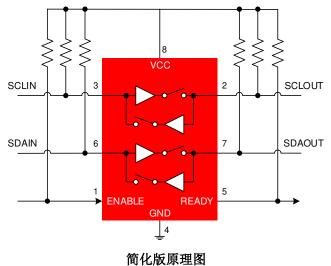




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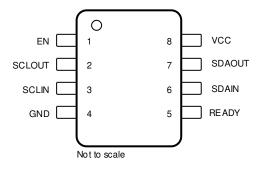
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4 Revision History 注:以前版本的页码可能与当前版本的页码不同

DATE	REVISION	NOTES
August 2020	*	Initial release.



5 Pin Configuration and Functions





Pin Functions

Р	PIN		DESCRIPTION		
NAME NO.		I/O	DESCRIPTION		
EN	1	I	Active-high chip enable pin. If EN is low, the TCA4307 is in a low current mode. It also disables the rise-time accelerators, disables the bus pre-charge circuitry (after the initial power up), drives READY low, isolates SDAIN from SDAOUT and isolates SCLIN from SCLOUT. EN should be high (at VCC) for normal operation. Connect EN to VCC if this feature is not being used.		
SCLOUT	2	I/O	Serial clock output. Connect this pin to the SCL bus on the card.		
SCLIN	3	I/O	Serial clock input. Connect this pin to the SCL bus on the backplane.		
GND	4	-	Supply ground		
READY	5	0	Connection flag/rise-time accelerator control. Ready is low when either EN is low or the start-up sequence has not been completed. READY goes high when EN is high and start-up is complete. Connect a 10-k Ω resistor from this pin to V _{CC} to provide the pull-up current.		
SDAIN	6	I/O	Serial data input. Connect this pin to the SDA bus on the backplane.		
SDAOUT	7	I/O	Serial data output. Connect this pin to the SDA bus on the card.		
vcc	8	Supply Power. Main input power supply from backplane. This is the supply voltage for the devices on the backplane l ² C buses. Connect pull-up resistors from SDAIN and SCLIN (and also from SDAOUT and SCLOUT) to this supply. It is recommended to place a bypass capacitor of 0.1 μ F close to this pin for best results.			

6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
	VCC		- 0.5	7	V
Input Voltage	SDAIN, SCLIN, SDAOUT, SCLOUT		- 0.5	7	V
renage	EN, READY		- 0.5	7	V
I _{IK}	Input clamp current	V ₁ < 0		- 50	mA
I _{OK}	Output clamp current	V ₀ < 0		- 50	mA
I _O	Continuous output current	SDAIN, SDAOUT, SCLIN, SCLOUT, EN, READY		±50	mA
I _{CC}	Continuous current through VCC or GN	ND		±100	mA
TJ	Maximum junction temperature			130	°C
T _{stg}	Storage temperature		- 65	150	°C

(1) Stresses beyond those listed under Absolute Maximum Rating may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Condition. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/ JEDEC JS-001, all pins ⁽¹⁾	±3500	V
V _(ESD)		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	±1000	v

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process

6.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
V _{CC}	CC Supply voltage		2.3	5.5	
VI	Input voltage range	EN input	0	5.5	V
V _{IO}	Input/output voltage range	SDAIN, SCLIN, SDAOUT, SCLOUT	0	5.5	v
Vo	Output voltage range	READY	0	5.5	
T _A	Ambient temperature		- 40	125	°C

6.4 Thermal Information

		TCA4307	
	THERMAL METRIC ⁽¹⁾	DGK	UNIT
		8 Pin	
R _{0 JA}	Junction-to-ambient thermal resistance	177.1	°C/W
R ₀ JC(top)	Junction-to-case (top) thermal resistance	64.5	°C/W
R _{0 JB}	Junction-to-board thermal resistance	99.6	°C/W
ΨJT	Junction-to-top characterization parameter	9.5	°C/W
Ψ_{JB}	Junction-to-board characterization parameter	97.9	°C/W
R ₀ JC(bot)	Junction-to-case (bottom) thermal resistance	N/A	°C/W

(1) For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.



6.5 Electrical Characteristics

Over operating free-air temperature range (unless otherwise noted). Typical specifications are at $T_A = 25$ °C, $V_{CC} = 3.3$ V, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER	SUPPLY	1				
I _{CC}	Supply current	VCC = 5.5V SDAIN, SCLIN = 0V SDAOUT, SCLOUT = 10k R _{PU}		2.5	4.5	mA
I _{SD}	Supply current in shutdown mode through the $V_{CC}\ pin^{(1)}$	$ EN = 0 V \\ SDAIN, SCLIN, SDAOUT, SCLOUT = 0V \\ or V_{CC} \\ READY pin = Hi-Z \\ EN pulled low after bus connection event (disable precharge) $		10	30	μΑ
UVLO	Under voltage lockout (rising)	EN = V _{CC}		2.1		V
UVLO	Under voltage lockout (falling)	READY = 10 k Ω to V _{CC}		2		V
START-U	P CIRCUITRY	L (I	
V _{PRE}	Pre-charge voltage	SDA, SCL = Hi-Z	0.8	1	1.2	V
RISE-TIM	E ACCELERATORS					
I _{PU}	RTA pull-up current ⁽²⁾	Position transition on SDA, SCL $V_{SDA/SCL}$ = 0.6 V, Slew rate = 1.25 V/µs. VCC = 3.3 V	2	5		mA
INPUT-O	UTPUT CONNECTION					
ILI	Input pin leakage	$ SDA/SCL pins = 90\% V_{CC}, EN = V_{CC}, \\ GND \\ SDA/SCL pins = 10\% V_{CC}, EN = GND $	-1		1	μA
V _{OS}	Input-output offset voltage (SCLIN to SCLOUT, SCLOUT to SCLIN and SDAIN to SDAOUT, SDAOUT to SDAIN	R _{PU} for SDA/SCL = 10 k Ω		60	100	mV
I _{I_RDY}	Ready pin leakage	EN = VCC, READY = V _{CC} , Bus connected	-1		1	μA
DIGITAL	IO THRESHOLD					
V _{IH}	High-level input voltage	EN	0.7 × V _{CC}		V _{CC}	
V _{IL}	Low-level input voltage	EN	0		0.3 × V _{CC}	
V _{OL}	Low-level output voltage	SDAIN, SCLIN, SDAOUT, SCLOUT $I_{OL} = 4 \text{ mA}$ $V_{IN} = 0.1 \text{ V}$		0.15	0.4	V
		READY I _{OL} = 3 mA	0		0.4	
DYNAMIC	CCHARACTERISTICS					
C _{IN (EN)}	EN input capacitance	V _{EN} = 0 V or V _{CC} f = 400 kHz		1.6	4	
C _{IO} (READY)	READY output capacitance	V _{READY} = 0 V or V _{CC} f = 400 kHz		7	10	pF
C _{IO (SDA/} SCL)	SDA/SCL pin capacitance	V _{PIN} = 0 V or V _{CC} f = 400 kHz		5	10	
STUCK В	SUS RECOVERY					
t _{sтucквu} s	Stuck bus timer		25	40	65	ms
f _{SB_SCLO} UT	Stuck bus recovery clock frequency		5.5	8.5	14	kHz



6.5 Electrical Characteristics (continued)

Over operating free-air temperature range (unless otherwise noted). Typical specifications are at $T_A = 25$ °C, $V_{CC} = 3.3$ V, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
V _{OL}	Low level output during stuck bus clock output	I _{OL} = 4 mA	0		0.4	V

(1) In shutdown mode there will also be current flowing from V_{CC} through the ready pin as this pin is pulled down to indicate the bus is disconnected.

(2) Determined by design, not tested in production.

6.6 Timing Requirements

		MIN	NOM MA	X UNIT
f _{SCL_MAX}	Maximum SCL clock frequency	400		kHz
t _{BUF} ⁽¹⁾	Bus free time between a STOP and START condition	1.3		μs
t _{HD;STA} ⁽¹⁾	Hold time for a repeated START condition	0.6		μs
t _{SU;STA} (1)	Set-up time for a repeated START condition	0.6		μs
t _{SU;STO} (1)	Set-up time for a STOP condition	0.6		μs
t _{HD;DAT} (1)	Data hold time	0		ns
t _{SU;DAT} (1)	Data set-up time	100		ns
t _{LOW} (1)	LOW period of the SCL clock	1.3		μs
t _{HIGH} (1)	HIGH period of the SCL clock	0.6		μs
t _f (1)	Fall time of both SDA and SCL signals	20 × (V _{CC} /5.5 V)	30	0 ns
t _r (1)	Rise time of both SDA and SCL signals	20 × (V _{CC} /5.5 V)	30	0 ns

(1) These are system-level timing specs and are dependent upon bus capacitance and pull up resistor value. It is up to the system designer to ensure they are met

6.7 Switching Characteristics

Over operating free-air temperature range (unless otherwise noted). Typical specifications are at $T_A = 25$ °C, $V_{CC} = 3.3$ V, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
START-U	P CIRCUITRY					
t _{PRECHAR} GE	Time from V_{CC} to precharge enabled	SDA,SCL = Hi-Z EN = V _{CC} , GND		15	60	μs
t _{EN}	Time from V _{POR} to digital being ready	VCC transition from 0V to V_{CC} Time from V_{PORR} to earliest stop bit recongized		35	95	μs
t _{IDLE}	Bus idle time to READY active	SDA,SCL = 10 k Ω to V _{CC} EN = V _{CC} Measured at 0.5 × V _{CC}		95	150	μs
t _{DISABLE}	Time from EN high to low to READY low	$\begin{array}{l} \text{SDA,SCL} = 10 \text{ k} \Omega \text{ to } \text{V}_{\text{CC}} \\ \text{READY} = 10 \text{ k} \Omega \text{ to } \text{V}_{\text{CC}} \\ \text{Measured at } 0.5 \times \text{V}_{\text{CC}} \end{array}$		30	200	ns
t _{STOP}	SDAIN to READY delay after stop condition	$\begin{array}{l} \text{SDA,SCL} = 10 \text{ k} \Omega \text{ to } \text{V}_{\text{CC}} \\ \text{READY} = 10 \text{ k} \Omega \text{ to } \text{V}_{\text{CC}} \\ \text{Measured at } 0.5 \times \text{V}_{\text{CC}} \end{array}$		1.2	2	μs
t _{READY}	SCLOUT/SDAOUT to READY	$\begin{array}{l} \text{SDA,SCL} = 10 \text{ k} \Omega \ \text{to} \ \text{V}_{\text{CC}} \\ \text{READY} = 10 \text{ k} \Omega \ \text{to} \ \text{V}_{\text{CC}} \\ \text{Measured at} \ 0.5 \times \text{V}_{\text{CC}} \end{array}$		0.8	1.5	μs
INPUT-OL	JTPUT CONNECTION					



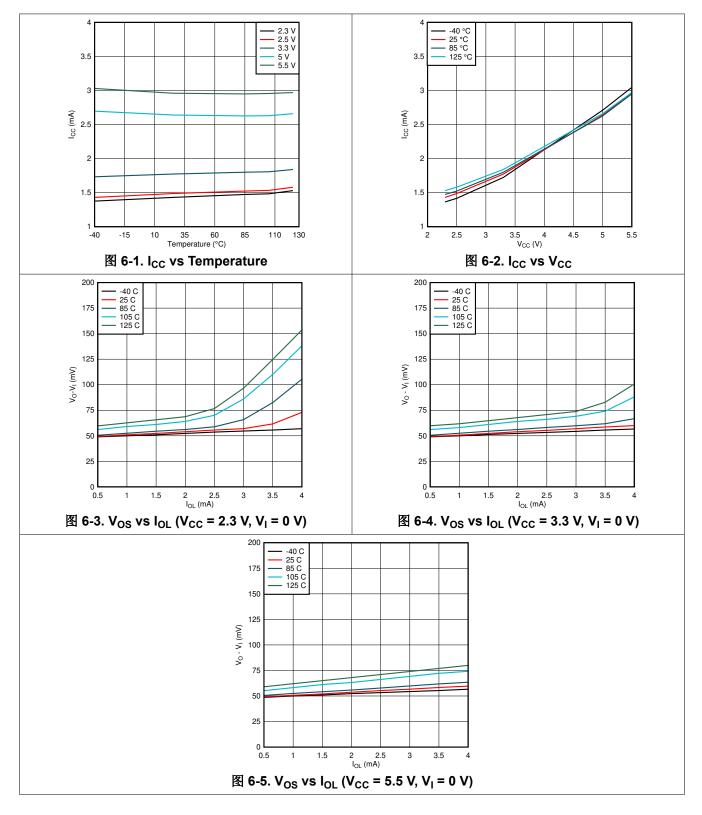
6.7 Switching Characteristics (continued)

Over operating free-air temperature range (unless otherwise noted). Typical specifications are at $T_A = 25$ °C, $V_{CC} = 3.3$ V, unless otherwise noted.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
t _{PLZ}	Low to high propagation delay	R_{PU} for SDA/SCL = 10 k Ω C _L = 100 pF per pin Measured at 0.5 × VCC		0	10	ns
t _{PZL}	High to low propagation delay	R_{PU} for SDA/SCL = 10 k Ω C _L = 100 pF per pin Measured at 0.5 × VCC		70	150	ns

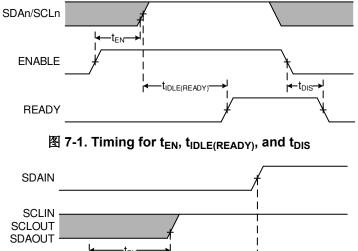


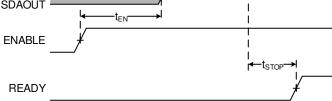
6.8 Typical Characteristics





7 Parameter Measurement Information









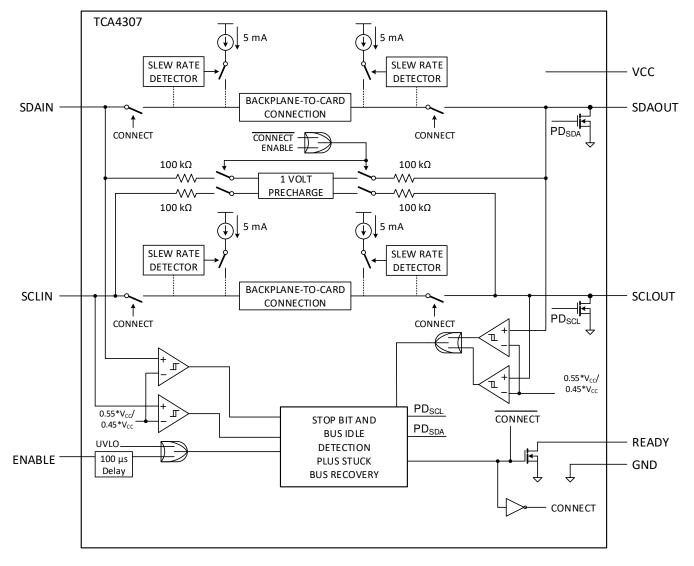
8 Detailed Description

8.1 Overview

The TCA4307 is a hot-swappable I²C bus buffer that supports I/O card insertion into a live backplane without corruption of the data and clock buses. Control circuitry prevents the backplane from being connected to the card until a stop command or bus idle condition occurs on the backplane without bus contention on the card. When the connection is made, this device provides bidirectional buffering, keeping the backplane and card capacitances isolated. During insertion, the SDA and SCL lines are pre-charged to 1 V to minimize the current required to charge the parasitic capacitance of the device.

The TCA4307 has stuck bus recovery, which will automatically disconnect the bus if it detects that SDAOUT or SCLOUT are low for about 40 ms. Once the bus is disconnected, the device will automatically generate up to 16 pulses on SCLOUT to attempt to free the bus from the device which is holding it low.

When the I²C bus is idle, the TCA4307 is put into shutdown mode by setting the EN pin low. When EN is high, the TCA4307 resumes normal operation. It also includes an open drain READY output pin, which indicates that the backplane and card sides are connected together. When READY is high, the SDAIN and SCLIN are connected to SDAOUT and SCLOUT. When the two sides are disconnected, READY is low.



8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Hot bus insertion

During a hot bus insertion event, the TCA4307 keeps the buses disconnected to ensure that no data corruption occurs on either bus. Once the buses are idle or a stop bit on the IN side is detected, the TCA4307 connects the buses and READY goes high.

8.3.2 Pre-charge voltage

Both the SDA and SCL pins feature a 1-V pre-charge circuit through an internal 100 k Ω resistor prior to the pins being connected to an I²C bus. This feature helps minimize disruptions as a result of a hot bus insertion event.

8.3.3 Rise time accelerators

The TCA4307 features a rise time accelerator (RTA) on all I^2C pins that during a positive bus transition, switches on a current source to quickly slew the bus pins high. This allows the use of weaker pull-up resistors, which can lower V_{OL}s and lower power system level power consumption.

8.3.4 Bus ready output indicator

The READY pin is an open drain output that provides an indicator to whether the buses are connected and ready for traffic. This pin is pulled low when the connection between IN/OUT is high impedance. Once the bus is idle or a stop condition on the IN side is detected, and the connection between IN/OUT is made, the READY pin is released and pulled high by an external pull-up resistor, signaling that it is ready for traffic.

8.3.5 Powered-off high impedance for I²C and I/O pins

When the supply voltage is below the UVLO threshold, the I²C and digital I/O pins are a high impedance state to prevent leakage currents from flowing through the device. When the EN pin is taken low, the device enters an isolation state, presenting a high impedance on all bus pins and pulling the READY pin low.

8.3.6 Supports clock stretching and arbitration

The TCA4307 supports full clock stretching, and arbitration without lock up.

8.3.7 Stuck bus recovery

When SDAOUT or SCLOUT is low, an internal timer is started. After the timer expires, the TCA4307 will disconnect the IN/OUT buses and then clock the SCLOUT pin in an attempt to unstick the bus, generating up to 16 clock pulses. Once the clock pulses are complete, the device will generate a stop bit and release the bus. The device will then look for the same connection requirements as described in $\ddagger 8.4.2$ before reconnecting the IN/OUT buses.

8.4 Device Functional Modes

8.4.1 Start-up and precharge

When the TCA4307 first receives power on the VCC pin, either during power-up or during live insertion, it starts in an under voltage lockout (UVLO) state, ignoring any activity on the SDA and SCL pins until V_{CC} rises above UVLO.

Once the ENABLE pin goes high, the 'Stop Bit and Bus Idle' detect circuit is enabled and the device enters the bus idle state.

When V_{CC} rises above UVLO, the precharge circuitry will activate, which biases the bus pins on both sides to about 1 V through an internal 100 k Ω resistor.

8.4.2 Bus idle

After the Stop Bit and Bus Idle detect circuits are enabled the device enters the bus idle state. The pre-charge circuitry becomes active and forces 1 V through 100 k Ω nominal resistors to the SCL and SDA pins. The pre-charge circuitry minimizes the voltage differential seen by the SCL/SDA pins during a hot insertion event. This minimizes the amount of disturbance seen by the I/O card.

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The device waits for the SDAIN and SCLIN pins to be high for the bus idle time or a STOP condition to be observed on the IN pins. The SDAOUT and SCLOUT pins must be high and the SDAIN and SCLIN pins must meet 1 of the 2 qualifiers (idle timer or a STOP condition) before connecting SDAIN to SDAOUT and SCLIN to SCLOUT. Once the bus connections have been made, the pre-charge circuitry is disabled and the device enters the bus active state.

8.4.3 Bus active

In the bus active mode, the I²C IN and OUT pins are connected, and the input is passed bi-directionally from IN/OUT side of the bus to the OUT/IN side respectively. The buses remain connected until the EN pin is taken low.

When the bus is connected, the driven-low side of the device is reflected on the opposite side, but with a small offset voltage. For example, if the input is pulled low to 100 mV, the output side will be pulled to roughly 160 mV. This offset allows the device to determine which side is currently being driven and avoid getting stuck low.

For the TCA4307, once a stuck bus event is detected (about 40 ms), the bus disconnects, even if EN is high.

8.4.4 Bus stuck

Once a stuck bus condition has been detected on SDAOUT or SCLOUT, the TCA4307 disconnects the bus and begins a sequence to attempt to recover the bus. First, the OUT side is disconnected from the IN side. READY will go low to signal that the bus is disconnected. Second, the TCA4307 will begin generating clocks on SCLOUT, up to 16. It will constantly monitor the state of SDAOUT to see if it has been released. Clocking will continue until 16 clocks are generated, or the SDAOUT releases. Once the SDAOUT releases, the TCA4307 will stop clocking and will generate a stop condition to terminate the recovery sequence. The last step is to go back to the bus-idle state and wait for an idle bus on both sides or a stop condition to ensure it's safe to connect the bus.



9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The typical application is to place the TCA4307 on the card that is being inserted or connected to a live bus, rather than being placed on the live bus. The reason for this is to provide maximum benefit by ensuring that the bus stays disconnected until an idle condition or stop condition is seen.

9.2 Typical Application

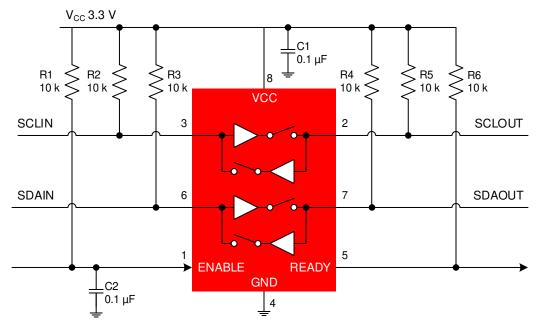


图 9-1. General Application Schematic

9.2.1 Design Requirements

9.2.1.1 Series connections

It is possible to have multiple buffers in series, but care must be taken when designing a system.

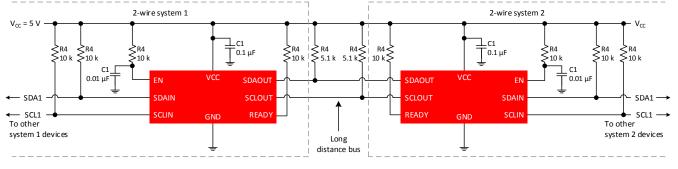


图 9-2. Series Buffer Connections



Each buffer adds approximately 60 mV of offset. Maximum offset (V_{OFFSET}) should be considered. The low level at the signal origination end is dependent upon bus load. The I²C-bus specification requires that a 3 mA current produces no larger than a 0.4 V V_{OL}. As an example, if the V_{OL} at the master is 0.1 V, and there are 4 buffers in series (each adding about 60 mV), then the V_{OL} at the farthest buffer is approximately 0.34 V. This device has a rise time accelerator (RTA) that activates at 0.6 V. With great care, a system with 4 buffers may work, but as the V_{OL} moves up, it may be possible to trigger the RTA, creating a false edge on the clock.

It is recommended to limit the number of buffers in series to two, and to keep the load light to minimize the offset.

Another special consideration of series connections is the effect on round-trip-delay. This is the sum of propagation delays through the buffers and any effects on rise time. It is possible that fast mode speeds (400 kHz) are not possible due to delays and bus loading.

9.2.1.2 Multiple connections to a common node

It is possible to have multiple buffers in connect to a common node, but care must be taken when designing a system.

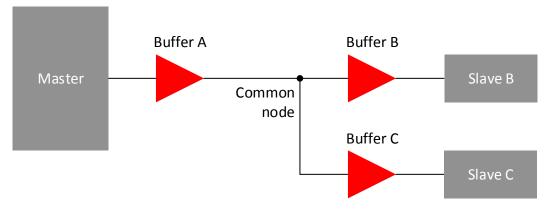


图 9-3. Connections to Common Node

It is important to try and avoid common node architectures. The multiple nodes sharing a common node can create glitches if the output voltage from a master slave device plus the offset voltage of the buffer are high enough to trip the RTA. Also keep in mind that the V_{OS} must be crossed in order for a device to begin to regulate the other side.

Consider a system with three buffers connected to a common node and communication between the Master and Slave B that are connected at either end of buffer A and buffer B in series as shown in 🗏 9-3. Consider if the V_{OL} at the input of buffer A is 0.3 V and the V_{OL} of Slave B (when acknowledging) is 0.36 V with the direction changing from Master to Slave B and then from Slave B to Master. Before the direction change the user should observe V_{II} at the input of buffer A of 0.3 V and its output, the common node, is ~0.36 V. The output of buffer B and buffer C would be ~0.42 V, but Slave B is driving 0.4 V, so the voltage at Slave B is 0.4 V. The output of buffer C is ~0.52 V. When the Master pull-down turns off, the input of buffer A rises and so does its output, the common node, because it is the only part driving the node. The common node rises to ~0.5 V before the buffer B output turns on, if the pull-up is strong the node may bounce. If the bounce goes above the threshold for the rising edge accelerator ~0.6 V, the accelerators on both buffer A and buffer C will fire, contending with the output of buffer B. The node on the input of buffer A goes high as will the input node of buffer C. After the common node voltage is stable for a while, the rising edge accelerators turn off, and the common node returns to ~0.5 V because the buffer B is still on. The voltage at both the Master and Slave C nodes then fall to ~0.6 V until Slave B turned off. This does not cause a failure on the data line as long as the return to 0.5 V on the common node (~0.56 V at the Master and Slave C) occurred before the data setup time. If this were the SCL line, the parts on buffer A and buffer C would see a false clock rather than a stretched clock, which causes a system error.

9.2.1.3 Propagation delays

The delay for a rising edge is determined by the combined pull-up current from the bus resistors and the rise time accelerator current source and the effective capacitance on the lines. If the pull-up currents are the same,



any difference in rise time is directly proportional to the difference in capacitance between the two sides. The t_{PLH} may be negative if the output capacitance is less than the input capacitance and would be positive if the output capacitance is larger than the input capacitance, when the currents are the same.

The t_{PHL} can never be negative because the output does not start to fall until the input is below 0.7 × V_{CC}, the output turn on has a non-zero delay, and the output has a limited maximum slew rate. Even if the input slew rate is slow enough that the output catches up, it would still lag the falling voltage of the input by the offset voltage. The maximum t_{PHL} occurs when the input is driven low with a very fast slew rate and the output is still limited by its turn-on delay and the falling edge slew rate.

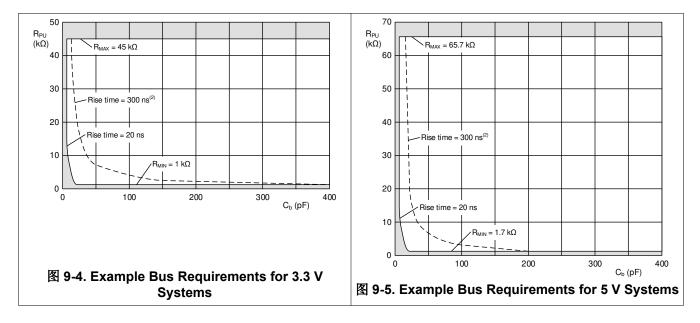
9.2.2 Detailed Design Procedure

The system pull-up resistors must be strong enough to provide a positive slew rate of 1.25 V/ μ s on the SDA and SCL pins, in order to activate the boost pull-up currents during rising edges. Choose maximum resistor value using the formula given in β 程式 1.

$$R \le 800 \times 10^{3} \left(\frac{V_{CC(MIN)} - 0.6}{C} \right)$$
(1)

where R is the pull-up resistor value in Ω , $V_{CC(MIN)}$ is the minimum V_{CC} voltage in volts, and C is the equivalent bus capacitance in picofarads (pF).

In addition, regardless of the bus capacitance, always choose $R_{PU} \le 65.7 \text{ k}\Omega$ for V_{CC} = 5.5 V, $R_{PU} \le 45 \text{ k}\Omega$ for V_{CC} = 3.3 V, and $R_{PU} \le 33 \text{ k}\Omega$ for V_{CC} = 2.5 V. The start-up circuitry requires logic HIGH voltages on SDAOUT and SCLOUT to connect the backplane to the card, and these pull-up values are needed to overcome the precharge voltage.



9.2.3 Application Curves



9.2.4 Typical Application on a Backplane

As shown in [8] 9-6, the TCA4307 is used in a backplane connection. The TCA4307 is placed on the I/O peripheral card and connects the I²C devices on the card to the backplane safely upon a hot insertion event. Note that if the I/O cards were plugged directly into the backplane, all of the backplane and card capacitances would add directly together, making rise time and fall time requirements difficult to meet. Placing a bus buffer on the edge of each card; however, isolates the card capacitance from the backplane. For a given I/O card, the TCA4307 drives the capacitance of everything on the card and the backplane must drive only the capacitance of the bus buffer, which is less than 10 pF, the connector, trace, and all additional cards on the backplane.

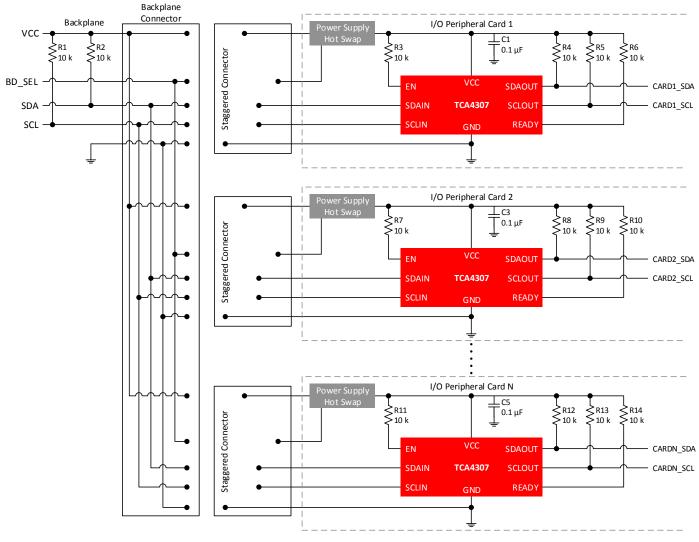


图 9-6. Backplane Application Schematic

9.2.4.1 Design Requirements

There are a few considerations when using these hot swap buffers. It is NOT recommended to place the TCA4307 on the backplane connector as it cannot isolate the cards from one another which will possibly result in disturbing on-going I^2C transactions. Instead, place the TCA4307 on the I/O peripheral card to maximize benefit.

9.2.4.2 Detailed Design Procedure

The design procedure is the same as outlined in \ddagger 9.2.2.



10 Power Supply Recommendations

10.1 Power Supply Best Practices

In order for the pre-charge circuitry to dampen the effect of hot-swap insertion of the TCA4307 into an active I^2C bus, V_{CC} must be applied before the SCL and SDA pins make contact to the main I^2C bus. This is essential when the TCA4307 is placed on the add-on card circuit board, as in $\ddagger 9.2.4$. Although the pre-charge circuitry exists on both the -IN and -OUT side, the example in $\ddagger 9.2.4$ shows SCLIN and SDAIN connecting to the main bus. The supply voltage to VCC can be applied early by ensuring that the VCC and GND pin contacts are physically longer than the contacts for the SCLIN and SDAIN pins. If a voltage supervisor will also be used to control the voltage supply on the add-on card, additional delay will exist before the 1 V pre-charge voltage is present on the SCL and SDA pins.

10.2 Power-on Reset Requirements

In order to ensure that the part starts up in the correct state, it is recommended that the power supply ramp rates meet the below requirements.

	Parameter		Parameter MIN MAX				UNIT
t _{RT}	Rise rate		0.1	1000	ms		
t _{FT}	Fall rate		0.1	1000	ms		

表 10-1. Recommended supply ramp rates



11 Layout

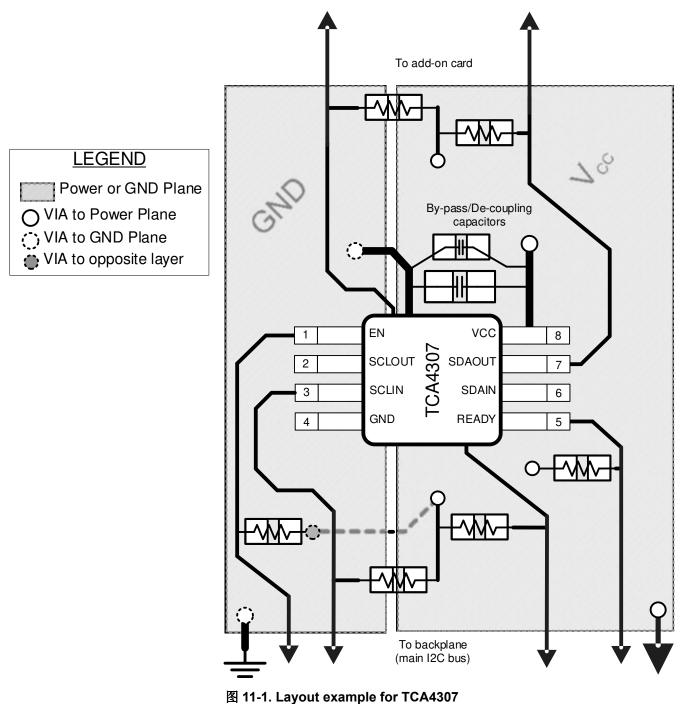
11.1 Layout Guidelines

For printed circuit board (PCB) layout of the TCA4307, common PCB layout practices should be followed but additional concerns related to high-speed data transfer such as matched impedances and differential pairs are not a concern for I^2C signal speeds. In all PCB layouts, it is a best practice to avoid right angles in signal traces, to fan out signal traces away from each other upon leaving the vicinity of an integrated circuit (IC), and to use thicker trace widths to carry higher amounts of current that commonly pass through power and ground traces. By-pass and de-coupling capacitors are commonly used to control the voltage on the VCC pin, using a larger capacitor to provide additional power in the event of a short power supply glitch and a smaller capacitor to filter out high frequency ripple. These capacitors should be placed as close to the TCA4307 as possible. These best practices are shown in $\ddagger 11.2$.

The layout example provided in \ddagger 11.2 shows a 4 layer board, which is preferable for boards with higher density signal routing. On a 4 layer PCB, it is common to route signals on the top and bottom layer, dedicate one internal layer to a ground plane, and dedicate the other internal layer to a power plane. In a board layout using planes or split planes for power and ground, vias are placed directly next to the surface mount component pad which needs to attach to V_{CC} or GND and the via is connected electrically to the internal layer or the other side of the board. Vias are also used when a signal trace needs to be routed to the opposite side of the board, shown in the \ddagger 11.2 for the VCC side of the resistor connected to the EN pin; however, this routing and via is not necessary if V_{CC} and GND are both full planes as opposed to the partial planes depicted.



11.2 Layout Example





12 Device and Documentation Support

12.1 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.2 支持资源

TI E2E[™] 支持论坛是工程师的重要参考资料,可直接从专家处获得快速、经过验证的解答和设计帮助。搜索现有 解答或提出自己的问题,获得所需的快速设计帮助。

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ESD 的损坏小至导致微小的性能降级,大至整个器件故障。精密的集成电路可能更容易受到损坏,这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

12.5 术语表

TI术语表 本术语表列出并解释了术语、首字母缩略词和定义。

Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TCA4307DGKR	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAUAG	Level-1-260C-UNLIM	-40 to 125	4307	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

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⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE MATERIALS INFORMATION

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Texas Instruments

TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nor	ninal
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Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TCA4307DGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TEXAS INSTRUMENTS

www.ti.com

PACKAGE MATERIALS INFORMATION

23-Aug-2020



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TCA4307DGKR	VSSOP	DGK	8	2500	364.0	364.0	27.0

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.

- D Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.



DGK (S-PDSO-G8)

PLASTIC SMALL OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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