



TPS79101-EP, TPS79118-EP TPS79133-EP, TPS79147-EP

SGLS161A-APRIL 2003-REVISED JUNE 2008

low-dropout (LDO)

# ULTRALOW NOISE, HIGH PSRR, FAST RF, 100-mA LOW-DROPOUT LINEAR REGULATORS

DESCRIPTION

The TPS791xx

response time.

family

of

low-power linear voltage regulators features high

power supply rejection ratio (PSRR), ultralow noise,

fast start-up, and excellent line and load transient

responses in a small outline, SOT23, package. Each

device in the family is stable, with a small 1-µF ceramic capacitor on the output. The family uses an

advanced, proprietary BiCMOS fabrication process to

yield extremely low dropout voltages (e.g., 38 mV at

100 mA, TPS79147). Each device achieves fast

start-up times (approximately 63 µs with a 0.001-µF bypass capacitor) while consuming very low

quiescent current (170 µA typical). Moreover, when

the device is placed in standby mode, the supply current is reduced to less than 1 µA. The TPS79118

exhibits approximately 15  $\mu V_{RMS}$  of output voltage noise with a 0.1-µF bypass capacitor. Applications

with analog components that are noise sensitive,

such as portable RF electronics, benefit from the

high-PSRR and low-noise features as well as the fast

#### FEATURES

- 100-mA Low-Dropout Regulator With EN
- Available in 1.8-V, 3.3-V, 4.7-V, and Adjustable Versions
- High PSRR (70 dB at 10 kHz)
- Ultralow Noise (15 µV<sub>RMS</sub>) •
- Fast Start-Up Time (63 µs)
- Stable With Any 1-µF Ceramic Capacitor
- **Excellent Load/Line Transient**
- Very Low Dropout Voltage (38 mV at Full Load, TPS79147)
- 5-Pin SOT23 (DBV) Package
- **TPS792xx Provides EN Options**

### APPLICATIONS

- VCOs
- RF
- Bluetooth<sup>™</sup>, Wireless LAN

#### SUPPORTS DEFENSE. AEROSPACE. AND MEDICAL APPLICATIONS

- **Controlled Baseline**
- **One Assembly/Test Site**
- **One Fabrication Site**
- Available in Military (-55°C/125°C) Temperature Range<sup>(1)</sup>
- **Extended Product Life Cycle**
- **Extended Product-Change Notification**
- **Product Traceability**
- (1) Custom temperature ranges available

#### **ORDERING INFORMATION**<sup>(1)</sup>

TJ	OUTPUT VOLTAGE	PACK	AGE <sup>(2)</sup>	ORDERABLE PART NUMBER	TOP-SIDE MARKING
	1.2 to 5.5 V			TPS79101DBVREP	PEUE
40%C to 125%C	1.8 V		Reel of 3000	TPS79118DBVREP	PERE
–40°C to 125°C	3.3 V	SOT23 (DBV)		TPS79133DBVREP	PESE
	4.7 V			TPS79147DBVREP	PETE
–55°C to 125°C	3.3 V	SOT23 (DBV)	Reel of 250	TPS79133MDBVTEP	PIDM

For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI (1)web site at www.ti.com.

Package drawings, thermal data, and symbolization are available at www.ti.com/packaging. (2)



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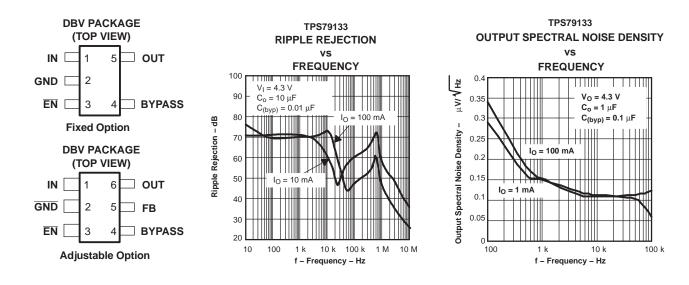
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AR AN

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.



#### ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

over operating free-air temperature range (unless otherwise noted)

		TPS79101, TPS79118, TPS79133, TPS79147
Input voltage range <sup>(2)</sup>		-0.3 V to 6 V
Voltage range at EN		-0.3 V to V <sub>1</sub> + 0.3 V
Voltage on OUT		-0.3 V to 6 V
Peak output current		Internally limited
ESD rating, HBM		2 kV
ESD rating, CDM	500 V	
Continuous total power dissipation		See Dissipation Rating Table
Operating virtual innotion temperature range. T	All others	-40°C to 150°C
Operating virtual-junction temperature range, $T_J$	TPS79133MDBVTEP	-55°C to 125°C
Operating ambient temperature range T	All others	-40°C to 120°C
Operating ambient temperature range, T <sub>A</sub>	TPS79133MBVTEP	-55°C to 125°C
Storage temperature range, T <sub>stg</sub>	· ·	-65°C to 150°C
	Low K	63.75°C/W
$R_{ ext{ heta}JC}^{(3)}$	High K	63.75°C/W
P (4)	Low K	256°C/W
$R_{ ext{ heta}JA}^{(4)}$	High K	178.3°C/W

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to network ground terminal.

(3) The JEDEC low-K (1s) board design used to derive this data was a 3-inch × 3-inch, two-layer board with 2-ounce copper traces on top of the board.

(4) The JEDEC high-K (2s2p) board design used to derive this data was a 3-inch × 3-inch, multilayer board with 1-ounce internal power and ground planes and 2-ounce copper traces on top and bottom of the board.



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#### **RECOMMENDED OPERATING CONDITIONS**

		MI N	MA X	UNI T		
Input voltage, V <sub>1</sub> <sup>(1)</sup>		2.7	5.5	V		
Continuous output current, I <sub>O</sub> <sup>(2)</sup>	Continuous output current, I <sub>O</sub> <sup>(2)</sup>					
Operating junction temperature T		-40	125	°C		
Operating junction temperature, T <sub>J</sub>	TPS79133MBVTEP			C		

(1) To calculate the minimum input voltage for your maximum output current, use the following formula: V<sub>I</sub>(min) = V<sub>O</sub>(max) + VDO (max load)

(2) Continuous output current and operating junction temperature are limited by internal protection circuitry, but it is not recommended that the device operate under conditions beyond those specified in this table for extended periods of time.

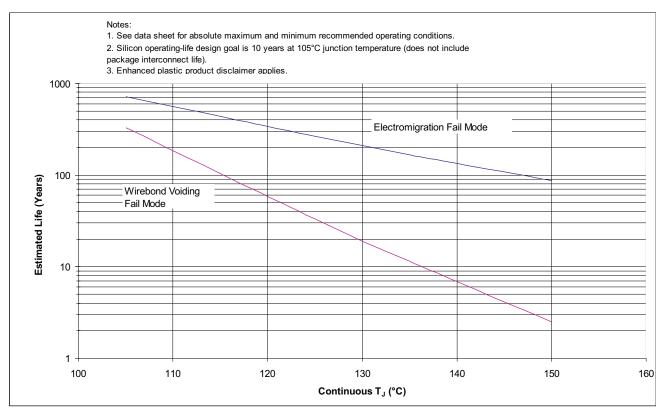


Figure 1. TPS79133 Operating Life Derating Chart

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#### ELECTRICAL CHARACTERISTICS

over recommended operating free-air temperature range, ( $T_J = -40^{\circ}C$  to 125°C),  $V_I = V_O(typ) + 1$  V,  $I_O = 1$  mA,  $\overline{EN} = 0$  V,  $C_o = 10 \ \mu$ F,  $C_o(byp) = 0.01 \ \mu$ F (unless otherwise noted)

PARAMETER		TEST CO	MIN	TYP	MAX	UNIT	
	TD070404	$T_J = 25^{\circ}C, 1.22 \text{ V} \le V_O \le 5.2$		Vo			
	TPS79101	$0 \ \mu A < I_O < 100 \ mA^{(1)}, 1.22$	V ≤ V <sub>O</sub> ≤ 5.2 V	0.98 V <sub>O</sub>		1.02 V <sub>O</sub>	
	TPS79118	$T_J = 25^{\circ}C$		1.8			
Dutput voltage Quiescent current (GND curr oad regulation Dutput voltage line regulation Dutput noise voltage (TPS79	122/9110	0 μA < I <sub>O</sub> < 100 mA, 2.8 V <	: V <sub>I</sub> < 5.5 V	1.764		1.836	v
Output voltage	TPS79133	$T_J = 25^{\circ}C$			3.3		V
	122/9133	0 μA < I <sub>O</sub> < 100 mA, 4.3 V <	: V <sub>I</sub> < 5.5 V	3.234		3.366	
	TPS79147	$T_J = 25^{\circ}C$			4.7		
	1F379147	0 $\mu$ A < I <sub>O</sub> < 100 mA, 5.2 V <	4.606		4.794		
		$0 \ \mu A < I_O < 100 \ mA, T_J = 25$		170		۸	
Quiescent current (GND C	urrent)	0 μA < I <sub>O</sub> < 100 mA	μA < I <sub>O</sub> < 100 mA			250	μA
Load regulation		$0 \ \mu A < I_O < 100 \ mA, T_J = 25$	5°C		5		mV
Output values line regulation $(\Lambda) ( \Lambda ( \lambda (2))$		$V_{O} + 1 V < V_{I} \le 5.5 V, T_{J} = 2$	25°C		0.05		%/V
Output voltage lille regula		$V_{O} + 1 V < V_{I} \le 5.5 V$				0.12	/0/ V
			$C_{(byp)} = 0.001 \ \mu F$		32		
	\$70118)	BW = 100 Hz to 100 kHz,	$C_{(byp)} = 0.0047 \ \mu F$		17		
Output hoise voltage (11 c	575110)	$I_{O} = 100 \text{ mA}, \text{ T}_{J} = 25^{\circ}\text{C}$	$C_{(byp)} = 0.01 \ \mu F$	16			μV <sub>RMS</sub>
			$C_{(byp)} = 0.1 \ \mu F$		1.02 Vo         1.8         1.836         3.3         3.366         4.7         4.794         170         250         5         0.05         0.12         32         17         16         15         53         67         98         600		
		D 00 0 00 1 F	$C_{(byp)} = 0.001 \ \mu F$		53		
Time, start-up (TPS79133)		R <sub>L</sub> 33 Ω, CO = 1 μF, T <sub>1</sub> = 25°C	$C_{(byp)} = 0.0047 \ \mu F$		67		μs
		Ū	$C_{(byp)} = 0.01 \ \mu F$		98		
Output current limit		$V_{O} = 0 V^{(1)}$		285		600	mA
UVLO threshold		V <sub>CC</sub> rising		2.25		2.65	V
UVLO hysteresis		$T_J = 25^{\circ}C, V_{CC}$ rising		100		mV	

(1) The minimum IN operating voltage is 2.7 V or V<sub>O</sub>(typ) + 1 V, whichever is greater. The maximum IN voltage is 5.5 V. The maximum output current is 100 mA.

(2) If  $\dot{V_O} \le 1.8$  V then  $V_{Imin} = 2.7$  V,  $V_{Imax} = 5.5$  V:

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**NSTRUMENTS** 

**EXAS** 

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#### **ELECTRICAL CHARACTERISTICS (continued)**

over recommended operating free-air temperature range, ( $T_J = -40^{\circ}C$  to 125°C),  $V_I = V_O(typ) + 1$  V,  $I_O = 1$  mA,  $\overline{EN} = 0$  V,  $C_o = 10 \ \mu$ F,  $C_O(byp) = 0.01 \ \mu$ F (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
Standby current		<del>EN</del> = V <sub>I</sub> , 2.7 V < V <sub>I</sub> < 5.5 V		0.07	1	μΑ
High-level enable input voltag	е	2.7 V < V <sub>I</sub> < 5.5 V	2			V
Low-level enable input voltage	Э	2.7 V < V <sub>I</sub> < 5.5 V	< 5.5 V			
Input current (EN)		$\overline{EN} = V_I$	-1			
		$f = 100 \text{ Hz}, \text{ T}_{\text{J}} = 25^{\circ}\text{C}, \text{ I}_{\text{O}} = 10 \text{ mA}$		80		
	TPS79118	$f = 100 \text{ Hz}, \text{ T}_{\text{J}} = 25^{\circ}\text{C}, \text{ I}_{\text{O}} = 100 \text{ mA}$		75		
		f = 10 kHz, T <sub>J</sub> = 25°C, I <sub>O</sub> = 100 mA		72		
Dower ownels ringle rejection		f = 100 kHz, T <sub>J</sub> = 25°C, I <sub>O</sub> = 100 mA		45		aD
Power supply hpple rejection		f = 100 Hz, T <sub>J</sub> = 25°C, I <sub>O</sub> = 10 mA		70		dB
	TPS79133	$f = 100 \text{ Hz}, \text{ T}_{\text{J}} = 25^{\circ}\text{C}, \text{ I}_{\text{O}} = 100 \text{ mA}$		75		
	19579133	f = 10 kHz, T <sub>J</sub> = 25°C, I <sub>O</sub> = 100 mA		73		
		f = 100 kHz, T <sub>J</sub> = 25°C, I <sub>O</sub> = 100 mA		37		
	TD070400	I <sub>O</sub> = 100 mA, T <sub>J</sub> = 25°C		50		
December 11 - 11 - (3)	TPS79133	I <sub>O</sub> = 100 mA			90	
Diopoul vollage	TD070447	I <sub>O</sub> = 100 mA, T <sub>J</sub> = 25°C		38		mV
Power supply ripple rejection Tf Tf Dropout voltage <sup>(3)</sup>	TPS79147	I <sub>O</sub> = 100 mA			70	

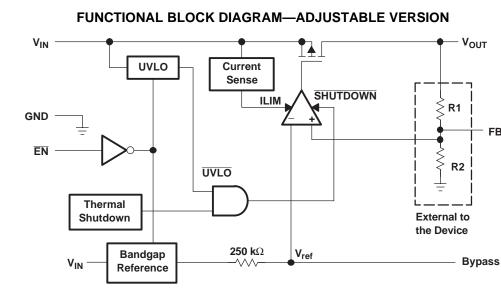
(3) IN voltage equals V<sub>O</sub>(typ) – 100 mV. The TPS79118 dropout voltage is limited by the input voltage range limitations.

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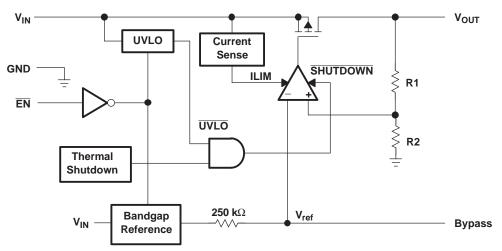
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FUNCTIONAL BLOCK DIAGRAM—FIXED VERSION



#### **TERMINAL FUNCTIONS**

	TERMINAL		I/O	DESCRIPTION
NAME	ADJ	FIXED	1/0	DESCRIPTION
BYPASS	4	4		An external bypass capacitor, connected to this terminal, in conjunction with an internal resistor, creates a low-pass filter to further reduce regulator noise.
ĒN	3	3	I	The $\overline{\text{EN}}$ terminal is an input which enables or shuts down the device. When $\overline{\text{EN}}$ is a logic high, the device will be in shutdown mode. When $\overline{\text{EN}}$ is a logic low, the device will be enabled.
FB	5	N/A	I	This terminal is the feedback input voltage for the adjustable device.
GND	2	2		Regulator ground
IN	1	1	I	The IN terminal is the input to the device.
OUT	6	5	0	The OUT terminal is the regulated output of the device.

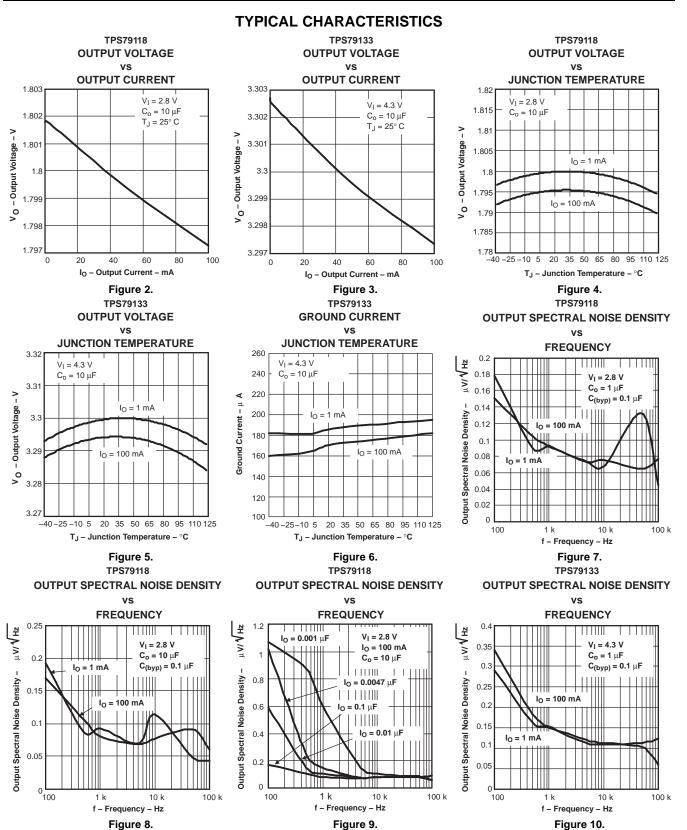
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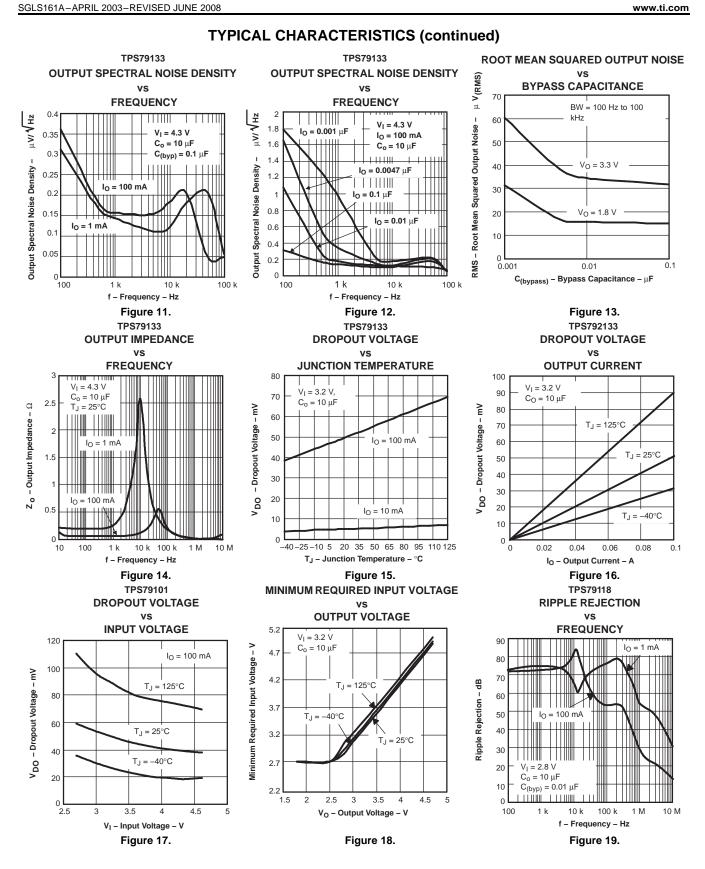


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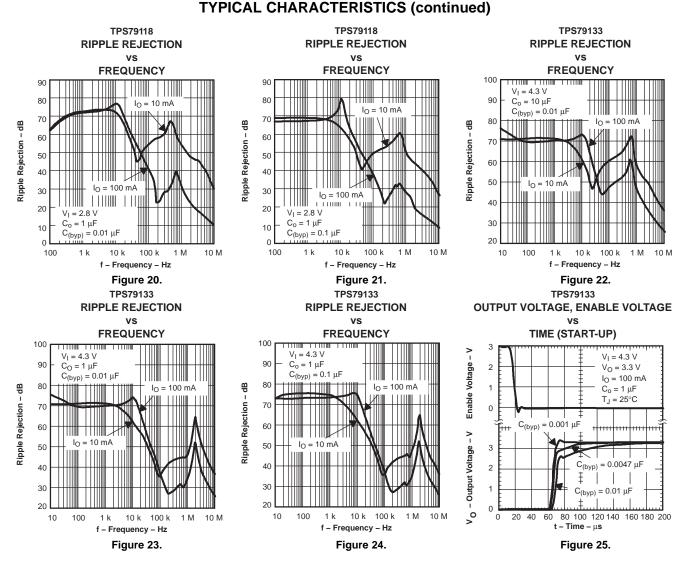
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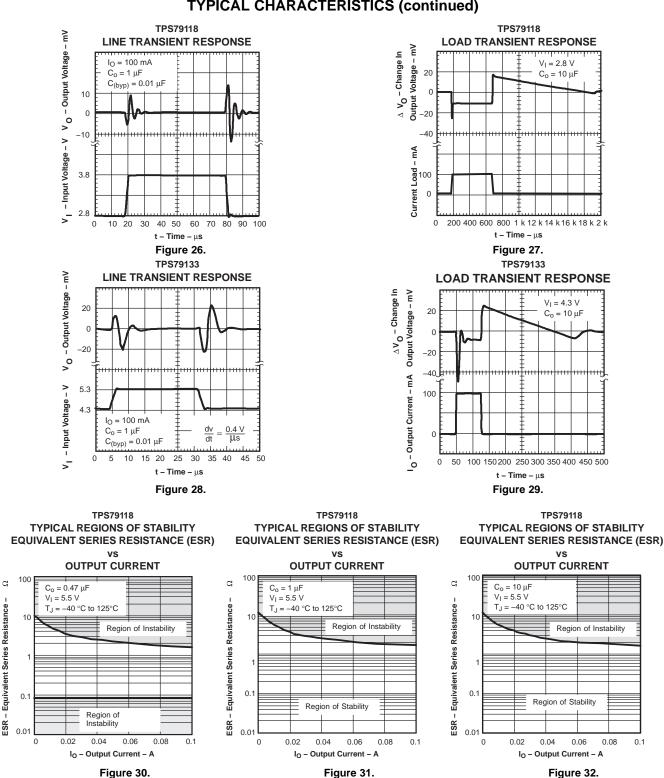




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#### **TYPICAL CHARACTERISTICS (continued)**

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#### **APPLICATION INFORMATION**

The TPS791xx family of low-dropout (LDO) regulators have been optimized for use in noise-sensitive battery-operated equipment. The device features extremely low dropout voltages, high PSRR, ultralow output noise, low quiescent current (170  $\mu$ A typically), and enable-input to reduce supply currents to less than 1  $\mu$ A when the regulator is turned off.

A typical application circuit is shown in Figure 33.

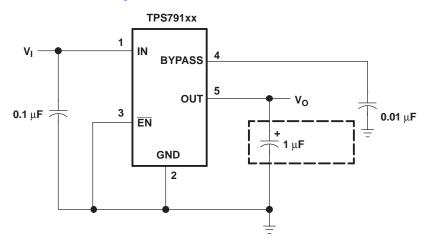


Figure 33. Typical Application Circuit

#### **External Capacitor Requirements**

A 0.1- $\mu$ F or larger ceramic input bypass capacitor, connected between IN and GND and located close to the TPS791xx, is required for stability and to improve transient response, noise rejection, and ripple rejection. A higher-value electrolytic input capacitor may be necessary if large, fast-rise-time load transients are anticipated and the device is located several inches from the power source.

Like all low dropout regulators, the TPS791xx requires an output capacitor connected between OUT and GND to stabilize the internal control loop. The minimum recommended capacitance is 1  $\mu$ F. Any 1- $\mu$ F or larger ceramic capacitor is suitable. The device is also stable with a 0.47- $\mu$ F ceramic capacitor with at least 75 m $\Omega$  of ESR.

The internal voltage reference is a key source of noise in an LDO regulator. The TPS791xx has a BYPASS pin which is connected to the voltage reference through a 250-k $\Omega$  internal resistor. The 250-k $\Omega$  internal resistor, in conjunction with an external bypass capacitor connected to the BYPASS pin, creates a low pass filter to reduce the voltage reference noise and, therefore, the noise at the regulator output. In order for the regulator to operate properly, the current flow out of the BYPASS pin must be at a minimum because any leakage current creates an IR drop across the internal resistor thus creating an output error. Therefore, the bypass capacitor must have minimal leakage current.

For example, the TPS79118 exhibits approximately 15  $\mu$ V<sub>RMS</sub> of output voltage noise using a 0.1- $\mu$ F ceramic bypass capacitor and a 1- $\mu$ F ceramic output capacitor. Note that the output starts up slower as the bypass capacitance increases due to the RC time constant at the bypass pin that is created by the internal 250 k $\Omega$  resistor and external capacitor.

#### Board Layout Recommendation To Improve PSRR And Noise Performance

To improve ac measurements like PSRR, output noise, and transient response, it is recommended that the board be designed with separate ground planes for  $V_{IN}$  and  $V_{OUT}$ , with each ground plane connected only at the ground pin of the device. In addition, the ground connection for the bypass capacitor should connect directly to the ground pin of the device.

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#### **Power Dissipation and Junction Temperature**

Specified regulator operation is assured to a junction temperature of 125°C; the maximum junction temperature should be restricted to 125°C under normal operating conditions. This restriction limits the power dissipation the regulator can handle in any given application. To ensure the junction temperature is within acceptable limits, calculate the maximum allowable dissipation, P<sub>D(max)</sub>, and the actual dissipation, P<sub>D</sub>, which must be less than or equal to P<sub>D(max)</sub>.

The maximum power-dissipation limit is determined using the following equation:

$$P_{D(max)} = \frac{T_{J}max - T_{A}}{R_{\theta JA}}$$
(1)

Where:

T<sub>J</sub>max is the maximum allowable junction temperature.

R<sub>0.JA</sub> is the junction-to-ambient thermal resistance for the package (see the dissipation rating table).

T<sub>A</sub> is the ambient temperature.

The regulator dissipation is calculated using:

$$\mathsf{P}_{\mathsf{D}} = \left(\mathsf{V}_{\mathsf{I}} - \mathsf{V}_{\mathsf{O}}\right) \times \mathsf{I}_{\mathsf{O}}$$

Power dissipation resulting from quiescent current is negligible. Excessive power dissipation triggers the thermal protection circuit.

#### Programming the TPS79101 Adjustable LDO Regulator

The output voltage of the TPS79101 adjustable regulator is programmed using an external resistor divider as shown in Figure 34. The output voltage is calculated using:

$$V_{O} = V_{ref} \times \left(1 + \frac{R1}{R2}\right)$$
(3)

Where:

 $V_{ref}$  = 1.2246 V typ (the internal reference voltage)

Resistors R1 and R2 should be chosen for approximately 50-µA divider current. Lower value resistors can be used for improved noise performance, but the solution consumes more power. Higher resistor values should be avoided as leakage current into/out of FB across R1/R2 creates an offset voltage that artificially increases/decreases the feedback voltage and thus erroneously decreases/increases Vo. The recommended design procedure is to choose R2 = 30.1 k $\Omega$  to set the divider current at 50  $\mu$ A, C1 = 15 pF for stability, and then calculate R1 using:

$$R1 = \left(\frac{V_{O}}{V_{ref}} - 1\right) \times R2$$
(4)

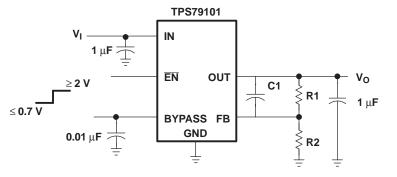
In order to improve the stability of the adjustable version, it is suggested that a small compensation capacitor be placed between OUT and FB. For voltages <1.8 V, the value of this capacitor should be 100 pF. For voltages > 1.8 V, the approximate value of this capacitor can be calculated as:

$$C1 = \frac{(3 \times 10^{-7}) \times (R1 + R2)}{(R1 \times R2)}$$
(5)

The suggested value of this capacitor for several resistor ratios is shown in the table below. If this capacitor is not used (such as in a unity-gain configuration) or if an output voltage < 1.8 V is chosen, then the minimum recommended output capacitor is 2.2  $\mu$ F instead of 1  $\mu$ F.

(2)





#### OUTPUT VOLTAGE PROGRAMMING GUIDE

,	OUTPUT VOLTAGE	R1	R2	C1
	2.5 V	31.6 kΩ	30.1 kΩ	22 pF
	3.3 V	51 kΩ	30.1 kΩ	15 pF
Γ	3.6 V	59 kΩ	30.1 kΩ	15 pF

#### Figure 34. TPS79101 Adjustable LDO Regulator Programming

#### **Regulator Protection**

The TPS791xx PMOS-pass transistor has a built-in back diode that conducts reverse current when the input voltage drops below the output voltage (e.g., during power down). Current is conducted from the output to the input and is not internally limited. If extended reverse voltage operation is anticipated, external limiting might be appropriate.

The TPS791xx features internal current limiting and thermal protection. During normal operation, the TPS791xx limits output current to approximately 400 mA. When current limiting engages, the output voltage scales back linearly until the overcurrent condition ends. While current limiting is designed to prevent gross device failure, care should be taken not to exceed the power dissipation ratings of the package or the absolute maximum voltage ratings of the device. If the temperature of the device exceeds approximately 165°C, thermal-protection circuitry shuts it down. Once the device has cooled down to below approximately 140°C, regulator operation resumes.



### PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
	(1)				,	(2)	(6)	(3)		(4/3)	
TPS79101DBVREP	ACTIVE	SOT-23	DBV	6	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PEUE	Samples
TPS79133DBVREP	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PESE	Samples
TPS79133MDBVTEP	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	PIDM	Samples
TPS79147DBVREP	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PETE	Samples
V62/03644-01YE	ACTIVE	SOT-23	DBV	6	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PEUE	Samples
V62/03644-03XE	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PESE	Samples
V62/03644-04XE	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PETE	Samples
V62/03644-05XE	ACTIVE	SOT-23	DBV	5	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-55 to 125	PIDM	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <= 1000ppm threshold. Antimony trioxide based flame retardants must also meet the <= 1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.



10-Dec-2020

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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# PACKAGE MATERIALS INFORMATION

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#### TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS79101DBVREP	SOT-23	DBV	6	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS79133DBVREP	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS79133MDBVTEP	SOT-23	DBV	5	250	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3
TPS79147DBVREP	SOT-23	DBV	5	3000	180.0	9.0	3.15	3.2	1.4	4.0	8.0	Q3

TEXAS INSTRUMENTS

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## PACKAGE MATERIALS INFORMATION

29-Aug-2017



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS79101DBVREP	SOT-23	DBV	6	3000	182.0	182.0	20.0
TPS79133DBVREP	SOT-23	DBV	5	3000	182.0	182.0	20.0
TPS79133MDBVTEP	SOT-23	DBV	5	250	182.0	182.0	20.0
TPS79147DBVREP	SOT-23	DBV	5	3000	182.0	182.0	20.0

# **DBV0005A**



# **PACKAGE OUTLINE**

### SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice. 3. Refernce JEDEC MO-178.

- 4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.



# DBV0005A

# **EXAMPLE BOARD LAYOUT**

### SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

5. Publication IPC-7351 may have alternate designs.

6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



# DBV0005A

# **EXAMPLE STENCIL DESIGN**

### SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

8. Board assembly site may have different recommendations for stencil design.



# **DBV0006A**



# **PACKAGE OUTLINE**

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.2. This drawing is subject to change without notice.3. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.25 per side.

- 4. Leads 1,2,3 may be wider than leads 4,5,6 for package orientation. 5. Refernce JEDEC MO-178.



# **DBV0006A**

# **EXAMPLE BOARD LAYOUT**

### SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



# **DBV0006A**

# **EXAMPLE STENCIL DESIGN**

### SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.



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