

# FULL DIFFERENTIAL ANALOG INPUT 24-BIT, 192-kHz STEREO A/D CONVERTER

# FEATURES

- 24-Bit Delta-Sigma Stereo A/D Converter
- **High Performance:** 
  - Dynamic Range: 112 dB (Typical)
  - SNR: 111 dB (Typical)
  - THD+N: -102 dB (Typical)
- **High-Performance Linear Phase Antialias Digital Filter:** 
  - Pass-Band Ripple: ±0.005 dB
  - Stop-Band Attenuation: –100 dB
- Fully Differential Analog Input: ±2.5 V
- Audio Interface: Master- or Slave-Mode Selectable
- Data Formats: Left-Justified, I<sup>2</sup>S, Standard 24-Bit, and DSD
- Function:
  - Peak Detection
  - High-Pass Filter (HPF): –3 dB at 1 Hz,  $f_s = 48 \text{ kHz}$
- Sampling Rate up to 192 kHz
- System Clock: 128 f<sub>S</sub>, 256 f<sub>S</sub>, 384 f<sub>S</sub>, 512 f<sub>s</sub>, or 768 f<sub>s</sub>
- Dual Power Supplies:
  - 5 V for Analog
  - 3.3 V for Digital
- Power Dissipation: 225 mW
- Small 28-Pin SSOP
- DSD Output: 1 Bit, 64 fs

# APPLICATIONS

- **AV Amplifier**
- **MD** Player •
- **Digital VTR** •
- **Digital Mixer**
- **Digital Recorder**

# DESCRIPTION

The PCM1804 is a high-performance, single-chip stereo A/D converter with fully differential analog voltage input. The PCM1804 uses a precision delta-sigma modulator and includes a linear phase antialias digital filter and high-pass filter (HPF) that removes dc offset from the input signal. The PCM1804 is suitable for a wide variety of mid- to high-grade consumer and professional applications, where excellent performance and 5-V analog supply and 3.3-V digital power-supply operation are required. The PCM1804 can achieve both PCM audio and DSD format due to the precision delta-sigma modulator. The PCM1804 is fabricated using an advanced CMOS process and is available in a small 28-pin SSOP package.



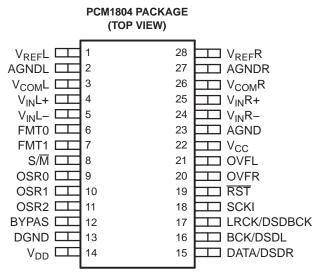
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This device contains circuits to protect its inputs and outputs against damage due to high static voltages or electrostatic fields. These circuits have been qualified to protect this device against electrostatic discharges (ESD) of up to 2 kV according to MIL-STD-883C, Method 3015; however, it is advised that precautions be taken to avoid application of any voltage higher than maximum-rated voltages to these high-impedance circuits. During storage or handling, the device leads should be shorted together or the device should be placed in conductive foam. In a circuit, unused inputs should always be connected to an appropriate logic voltage level, preferably either  $V_{CC}$  or ground. Specific guidelines for handling devices of this type are contained in the publication *Electrostatic Discharge (ESD)* (SSYA008), available from Texas Instruments.

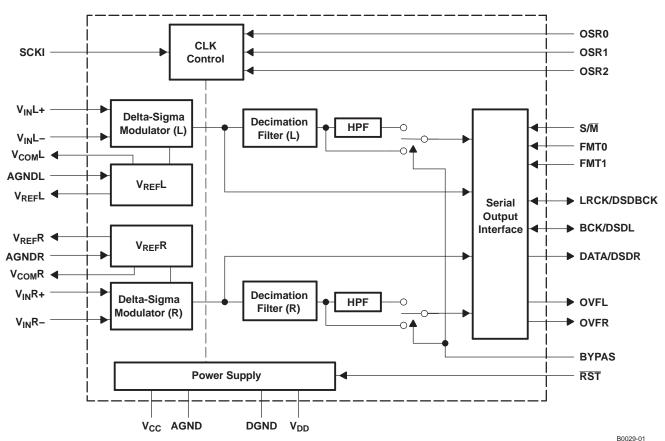
**PIN ASSIGNMENTS** 



P0007-02

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FUNCTIONAL BLOCK DIAGRAM



#### **Terminal Functions**

TERMINAL		1/0	DECODIDITIONO
NAME	PIN	I/O	DESCRIPTIONS
AGND	23		Analog ground
AGNDL	2	-	Analog ground for V <sub>REF</sub> L
AGNDR	27	-	Analog ground for V <sub>REF</sub> R
BCK/DSDL	16	I/O	Bit clock input/output in PCM mode. L-channel audio data output in DSD mode. (1)
BYPAS	12	Ι	HPF bypass control. High: HPF disabled, Low: HPF enabled <sup>(1)</sup>
DATA/DSDR	15	0	L-channel and R-channel audio data output in PCM mode. R-channel audio data output in DSD mode. (DSD output, when in DSD mode)
DGND	13	Ι	Digital ground
FMT0	6	Ι	Audio data format 0. See Table 5. (2)
FMT1	7	Ι	Audio data format 1. See Table 5. <sup>(2)</sup>
LRCK/DSDBCK	17	I/O	Sampling clock input/output in PCM and DSD modes. (1)
OSR0	9	Ι	Oversampling ratio 0. See Table 1 and Table 2. (2)
OSR1	10	Ι	Oversampling ratio 1. See Table 1 and Table 2. (2)
OSR2	11	Ι	Oversampling ratio 2. See Table 1 and Table 2. (2)
OVFL	21	0	Overflow signal of L-channel in PCM mode. This is available in PCM mode only.
OVFR	20	0	Overflow signal of R-channel in PCM mode. This is available in PCM mode only.
RST	19	Ι	Reset, power-down input, active-low <sup>(2)</sup>
SCKI	18	Ι	System clock input; 128 $f_S,$ 256 $f_S,$ 384 $f_S,$ 512 $f_S,$ or 768 $f_S.\ ^{(3)}$
S/M	8	Ι	Slave/master mode selection. See Table 4. (2)
V <sub>CC</sub>	22	-	Analog power supply
V <sub>COM</sub> L	3	Ι	L-channel analog common-mode voltage (2.5 V)
V <sub>COM</sub> R	26	-	R-channel analog common-mode voltage (2.5 V)
V <sub>DD</sub>	14	-	Digital power supply
V <sub>IN</sub> L-	5	Ι	L-channel analog input, negative pin
V <sub>IN</sub> L+	4	Ι	L-channel analog input, positive pin
V <sub>IN</sub> R–	24	Ι	R-channel analog input, negative pin
V <sub>IN</sub> R+	25	Ι	R-channel analog input, positive pin
V <sub>REF</sub> L	1	١	L-channel voltage reference output, requires capacitors for decoupling to AGND
V <sub>REF</sub> R	28	-	R-channel voltage reference output, requires capacitors for decoupling to AGND

(1)

Schmitt-trigger input Schmitt-trigger input with internal pulldown (51 kµ typically), 5-V tolerant. Schmitt-trigger input, 5-V tolerant. (2) (3)

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# ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	Supply voltage	V <sub>CC</sub>	–0.3 V to 6.5 V
	Supply voltage	V <sub>DD</sub>	–0.3 V to 4 V
	Ground voltage differences	AGND, AGNDL, AGNDR, DGND	±0.1 V
	Supply voltage difference	V <sub>CC</sub> , V <sub>DD</sub>	$V_{CC} - V_{DD} < 3 V$
	Digital input valtage	FMT0, FMT1, S/M, OSR0, OSR1, OSR2, SCKI, RST	–0.3 V to 6.5 V
	Digital input voltage	BYPAS, DATA/DSDR, BCK/DSDL, LRCK/DSDBCK, OVFL, OVFR	-0.3 V to (V <sub>DD</sub> + 0.3 V)
	Analog input voltage	V <sub>REF</sub> L, V <sub>REF</sub> R, V <sub>COM</sub> L, V <sub>COM</sub> R, V <sub>IN</sub> L+, V <sub>IN</sub> R+, V <sub>IN</sub> L–, V <sub>IN</sub> R–	–0.3 V to (V <sub>CC</sub> + 0.3 V)
	Input current (any pins excep	±10 mA	
T <sub>A</sub>	Ambient temperature under b	ias	-40°C to 125°C
T <sub>stg</sub>	Storage temperature		–55°C to 150°C
TJ	Junction temperature		150°C
	Lead temperature (soldering)		260°C, 5 s
	Package temperature (IR refl	260°C	

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

# **RECOMMENDED OPERATING CONDITIONS**

over operating free-air temperature range

		MIN	NOM	MAX	UNIT
Analog supply voltage, V <sub>CC</sub>		4.75	5	5.25	V
Digital supply voltage, V <sub>DD</sub>		3	3.3	3.6	V
Analog input voltage, full-scale (–0 dB), differential input			5		Vp-p
Digital input logic family		TTL	compatible		
	System clock	8.192		36.864	MHz
Digital input clock frequency	Sampling clock	32		36.864 192 10	kHz
Digital output load capacitance				10	pF
Dperating free-air temperature, T <sub>A</sub>		-10		70	°C



# **ELECTRICAL CHARACTERISTICS**

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 5 V$ ,  $V_{DD} = 3.3 V$ , master mode, single-speed mode,  $f_S = 48 \text{ kHz}$ , system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.

		TEST CONDITIONS	PCM1804DB			UNIT	
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
	Resolution			24		Bits	
ΑΤΑ	FORMAT						
	Audio data interface format		Standard,	I <sup>2</sup> S, left-jus	stified		
	Audio data bit length			24		Bits	
	Audio data format			ISB first, plement, DS	SD		
DIGIT	AL INPUT/OUTPUT				,		
	Logic family		TTL	compatible			
/ <sub>IH</sub>	High-level input voltage	(1) (2)	2		5.5	Vdc	
'IH	r lightever linput voltage	(3)	2		$V_{\text{DD}}$	Vuc	
/ <sub>IL</sub>	Low-level input voltage	(1) (2) (3)			0.8	Vdc	
		$V_{IN} = V_{DD}$ <sup>(1)</sup>		65	100		
н	High-level input current	$V_{IN} = V_{DD}$ <sup>(2)</sup>			±10	μA	
		$V_{IN} = V_{DD}$ <sup>(3)</sup>			±100		
	Low lovel input ourrent	$V_{IN} = 0 V^{(1)}^{(2)}$			±10		
IL	Low-level input current	$V_{\rm IN} = 0 \ V^{(3)}$			±50	μA	
/ <sub>ОН</sub>	High-level output voltage	$I_{OH} = -1 \text{ mA}^{(4)}$	2.4			Vdc	
/ <sub>OL</sub>	Low-level output voltage	$I_{OL} = 1 \text{ mA}^{(4)}$			0.4	Vdc	
CLOC	K FREQUENCY						
S	Sampling frequency		32		192	kHz	
		256 f <sub>S</sub> , single rate <sup>(5)</sup>		12.288			
		384 f <sub>S</sub> , single rate <sup>(5)</sup>		18.432			
		512 f <sub>S</sub> , single rate <sup>(5)</sup>		24.576			
		768 f <sub>S</sub> , single rate <sup>(5)</sup>		36.864		N 41 1-	
	System clock frequency	256 f <sub>S</sub> , dual rate <sup>(6)</sup>		24.576		MHz	
		384 f <sub>S</sub> , dual rate <sup>(6)</sup>		36.864			
		128 f <sub>S</sub> , quad rate <sup>(7)</sup>		24.576			
		192 $f_S$ , quad rate $(7)$		36.864			
OC AC	CURACY		· · ·		1		
	Gain mismatch, channel- to-channel				±3	% of FS	
	Gain error ( $V_{IN} = -0.5 \text{ dB}$ )				±4	% of FS	
	Bipolar zero error	HPF bypass		±0.2		% of FS	

(1) Pins 6–11, 19: FMT0, FMT1, S/M, OSR0, OSR1, OSR2, RST [Schmitt-trigger input with internal pulldown (51 kµ typically), 5-V tolerant] Pin 18: SCKI (Schmitt-trigger input, 5-V tolerant) (2)

(3) Pins 12, 16–17: BYPAS, BCK/DSDL, LRCK/DSDBCK (in slave mode, Schmitt-trigger input)
 (4) Pins 15–17, 20, and 21: DATA/DSDR, BCK/DSDL, LRCK/DSDBCK (in master mode), OVFR, OVFL

(5) Single rate,  $f_S = 48 \text{ kHz}$ 

(6) Dual rate,  $f_S = 96$  kHz (7) Quad rate,  $f_S = 192$  kHz

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# ELECTRICAL CHARACTERISTICS (continued)

All specifications at  $T_A = 25^{\circ}$ C,  $V_{CC} = 5$  V,  $V_{DD} = 3.3$  V, master mode, single-speed mode,  $f_S = 48$  kHz, system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.

	DADAMETED		TEAT CONDITIONS	PC	PCM1804DB		UNIT	
	PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
DYNAM	IC PERFORMANCE <sup>(8)</sup>							
		$V_{IN} = -0.5 \text{ dB}$	$f_{-} = 48 \text{ kHz}$ system clock = 256 f_		-102	-95		
		$V_{IN} = -60 \text{ dB}$	$V_{IN} = -60 \text{ dB}$ $f_S = 48 \text{ kHz}$ , system clock = 256 $f_S$		-49			
		$V_{IN} = -0.5 \text{ dB}$	f = 06  kHz eveter eleck = 256 f		-101			
THD+N	Total harmonic distortion plus noise	$V_{IN} = -60 \text{ dB}$	$f_{S} = 96 \text{ kHz}$ , system clock = 256 $f_{S}$		-47		dB	
		$V_{IN} = -0.5 \text{ dB}$			-101			
		$V_{IN} = -60 \text{ dB}$	$f_{\rm S}$ = 192 kHz, system clock = 128 $f_{\rm S}$		-47			
		$V_{IN} = -0.5 \text{ dB}$	DSD mode		-100			
			$f_S = 48 \text{ kHz}$ , system clock = 256 $f_S$	106	112			
	Dynamic range (A-weighted)	$V_{IN} = -60 \text{ dB}$	$f_S = 96$ kHz, system clock = 256 $f_S$		112		15	
			$f_{\rm S}$ = 192 kHz, system clock = 128 $f_{\rm S}$		112		dB	
		DSD mode			112			
		f <sub>S</sub> = 48 kHz, s	ystem clock = 256 f <sub>S</sub>	105	111			
		f <sub>S</sub> = 96 kHz, s	ystem clock = 256 f <sub>S</sub>		111		15	
	SNR (A-weighted)	f <sub>S</sub> = 192 kHz,	system clock = 128 f <sub>S</sub>		111		dB	
		DSD mode			111			
	$f_{S} = 48$		ystem clock = 256 f <sub>S</sub>	97	109			
	Channel separation	f <sub>S</sub> = 96 kHz, s	ystem clock = 256 f <sub>S</sub>		107		dB	
		f <sub>S</sub> = 192 kHz,	system clock = 128 f <sub>S</sub>	107				
ANALO	G INPUT	L.		I		1		
	Input voltage	Differential inp	put		±2.5		V	
	Center voltage				2.5		Vdc	
	Input impedance	Single-ended			10		kμ	
DIGITAL	L FILTER PERFORMANCE	L.		I		1		
	Pass-band edge	Single rate, du	ial rate		0	.453 f <sub>S</sub>	Hz	
	Stop-band edge	Single rate, du	ial rate	0.547 f <sub>S</sub>			Hz	
	Pass-band ripple	Single rate, du	ial rate			±0.005	dB	
	Stop-band attenuation	Single rate, du	ial rate	-100			dB	
	Pass-band edge (–0.005 dB)	Quad rate			0	.375 f <sub>S</sub>	Hz	
	Pass-band edge (-3 dB)	Quad rate				0.49 f <sub>S</sub>	Hz	
	Stop-band edge	Quad rate		0.77 f <sub>S</sub>		-	Hz	
	Pass-band ripple	Quad rate		-		±0.005	dB	
	Stop-band attenuation	Quad rate		-135			dB	
	Group delay	Single rate, du	ial rate		37/f <sub>S</sub>		S	
	Group delay	Quad rate			9.5/f <sub>S</sub>		s	
	HPF frequency response	–3 dB		1	f <sub>s</sub> /48000		Hz	

(8) f<sub>IN</sub> = 1 kHz, using System Two<sup>™</sup> audio measurement system by Audio Precision<sup>™</sup> in RMS mode, with 20-kHz LPF and 400-Hz HPF in calculation for single rate, or with 40-kHz LPF in calculation for dual and quad rates.



# **ELECTRICAL CHARACTERISTICS (continued)**

All specifications at  $T_A = 25^{\circ}$ C,  $V_{CC} = 5$  V,  $V_{DD} = 3.3$  V, master mode, single-speed mode,  $f_S = 48$  kHz, system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.

		TEST CONDITIONS	PC	PCM1804DB		UNIT
PARAMETER		TEST CONDITIONS	MIN	ТҮР	MAX	UNIT
POWE	ER SUPPLY REQUIREMENT	S	t.			
$V_{CC}$			4.75	5	5.25	Vdc
$V_{DD}$	<ul> <li>Supply voltage range</li> </ul>		3	3.3	3.6	vac
I <sub>CC</sub>		$V_{CC} = 5 V^{(9)} (10) (11)$		35	45	
	Supply current	$V_{DD} = 3.3 V^{(9)} (12)$		15	20	
I <sub>DD</sub>		$V_{DD} = 3.3 V^{(10)} (12)$		27		mA
		$V_{DD} = 3.3 V^{(11)} (12)$		18		
		Operation, $V_{CC} = 5 \text{ V}, V_{DD} = 3.3 \text{ V}^{(9)}$ (12)		225	290	
-	Dewes dissingtion	Operation, $V_{CC} = 5 \text{ V}, V_{DD} = 3.3 \text{ V}^{(10)}$ (12)		265		
PD	Power dissipation	Operation, $V_{CC} = 5 \text{ V}, V_{DD} = 3.3 \text{ V}^{(11)}$ (12)		235		mW
		Power down, $V_{CC} = 5 V$ , $V_{DD} = 3.3 V$		5		
TEMP	ERATURE RANGE					
	Operation temperature		-10		70	°C
$\theta_{JA}$	Thermal resistance			100		°C/W

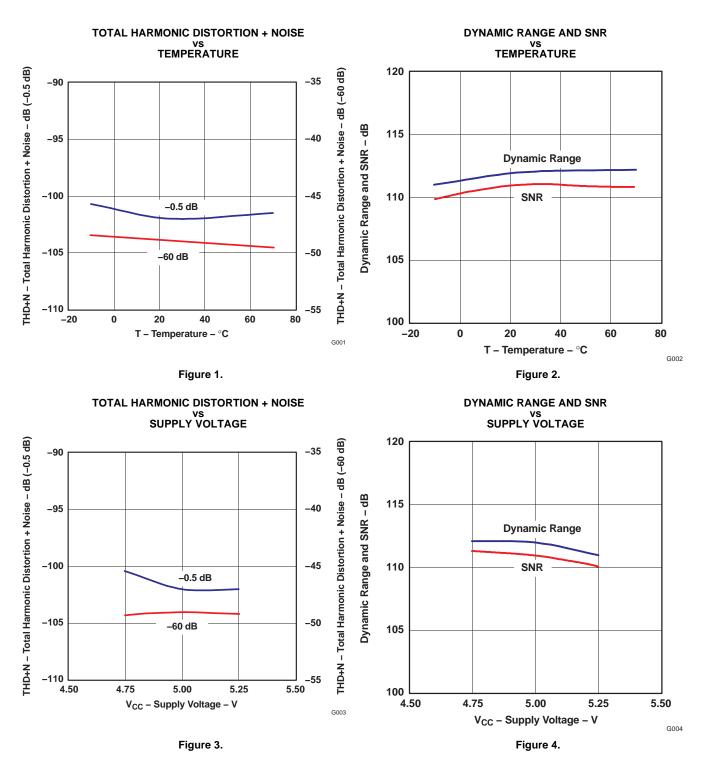
8



# TYPICAL PERFORMANCE CURVES - SINGLE RATE

EXAS RUMENTS ww.ti.com

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode,  $f_S = 48$  kHz, system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.





# **TYPICAL PERFORMANCE CURVES - SINGLE RATE (continued)**

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode,  $f_S = 48$  kHz, system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.

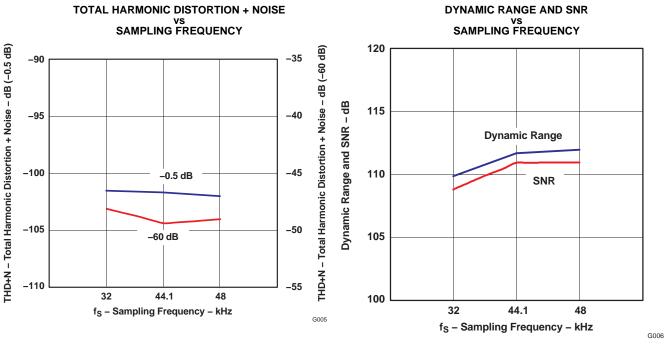
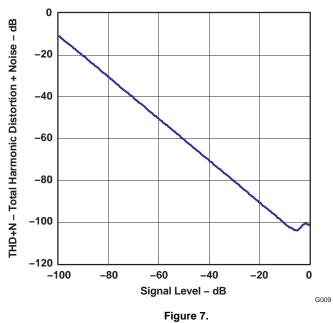


Figure 5.

Figure 6.

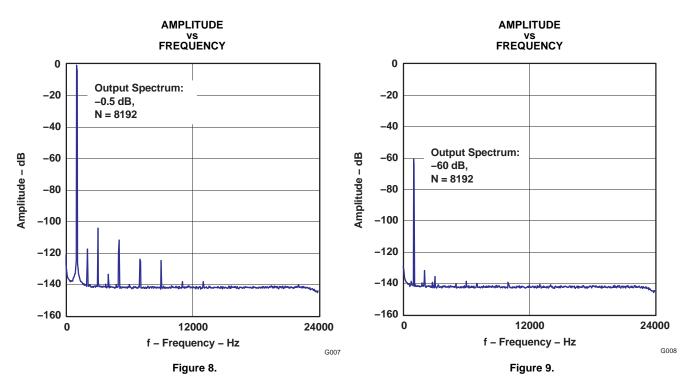
TOTAL HARMONIC DISTORTION + NOISE vs SIGNAL LEVEL





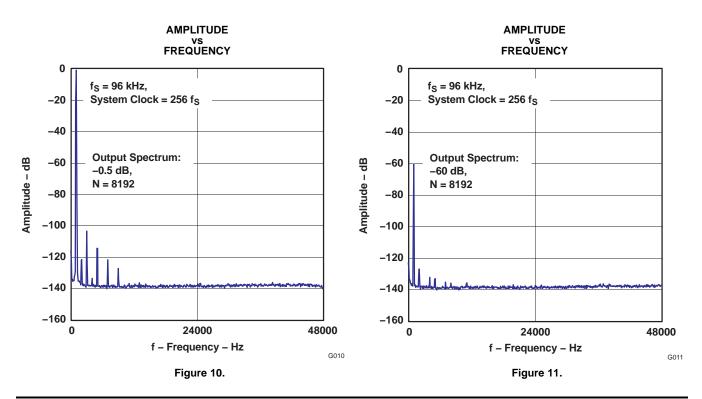
#### **TYPICAL PERFORMANCE CURVES - SINGLE RATE (continued)**

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode,  $f_S = 48$  kHz, system clock = 256  $f_S$ , 24-bit data, unless otherwise noted.



#### **TYPICAL PERFORMANCE CURVES - DUAL RATE**

All specifications at  $T_A = 25^{\circ}$ C,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode, and 24-bit data, unless otherwise noted.

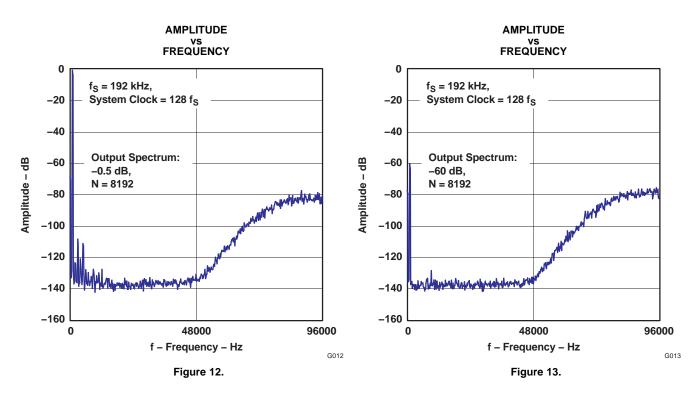


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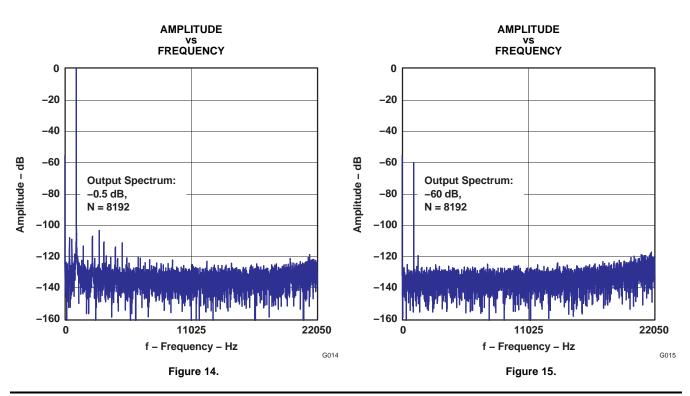
# TYPICAL PERFORMANCE CURVES - QUAD RATE

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode, 24-bit data, unless otherwise noted.



### **TYPICAL PERFORMANCE CURVES - DSD MODE**

All specifications at  $T_A = 25^{\circ}C$ ,  $V_{CC} = 3.3$  V,  $V_{DD} = 5$  V, master mode,  $f_S = 44.1$  kHz, system clock = 16.9344 MHz, unless otherwise noted.

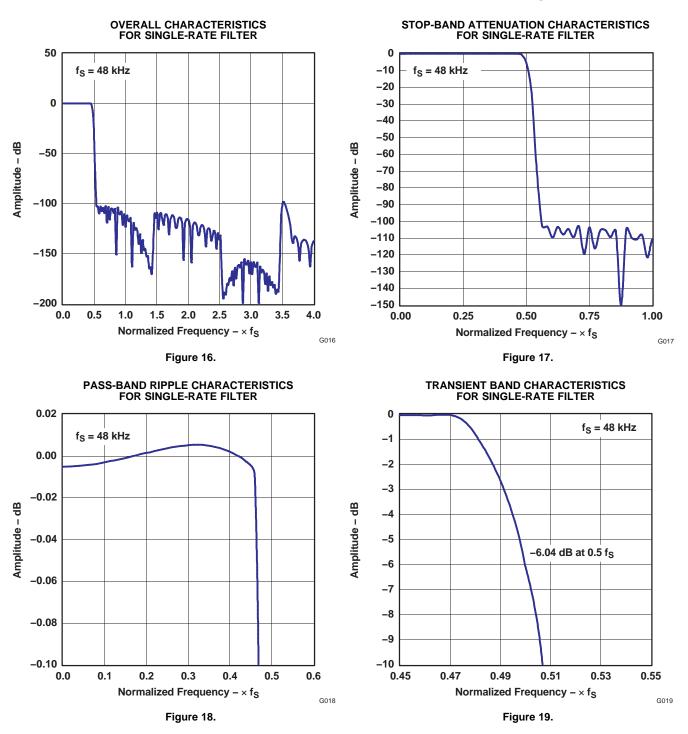


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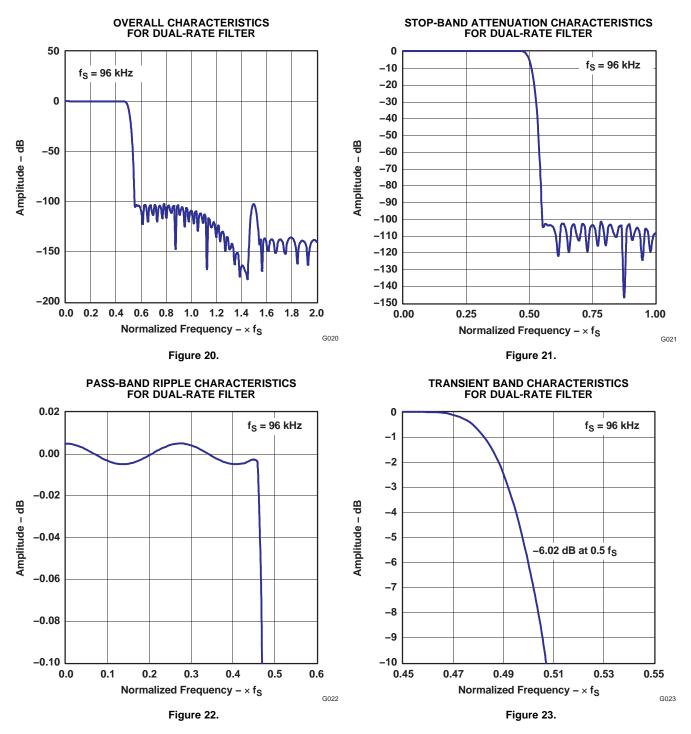
# TYPICAL PERFORMANCE CURVES OF INTERNAL FILTER

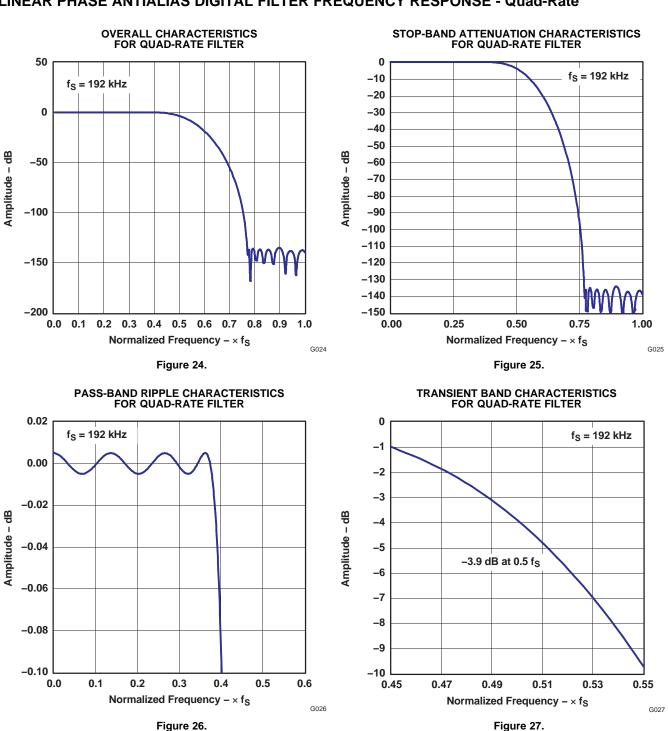
### LINEAR PHASE ANTIALIAS DIGITAL FILTER FREQUENCY RESPONSE - Single-Rate





# TYPICAL PERFORMANCE CURVES OF INTERNAL FILTER (continued) LINEAR PHASE ANTIALIAS DIGITAL FILTER FREQUENCY RESPONSE - Dual-Rate

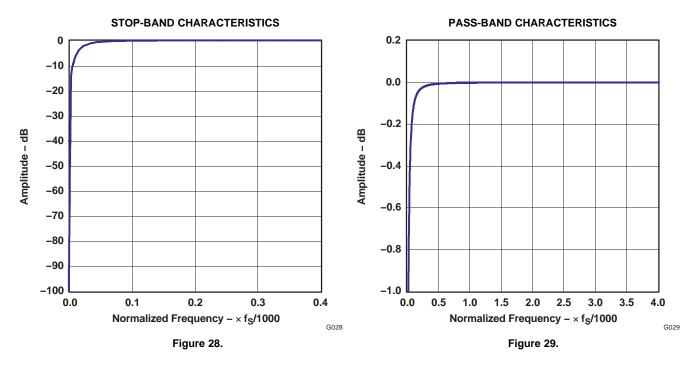




# TYPICAL PERFORMANCE CURVES OF INTERNAL FILTER (continued) LINEAR PHASE ANTIALIAS DIGITAL FILTER FREQUENCY RESPONSE - Quad-Rate



# TYPICAL PERFORMANCE CURVES OF INTERNAL FILTER (continued) HIGH-PASS FILTER (HPF) FREQUENCY RESPONSE

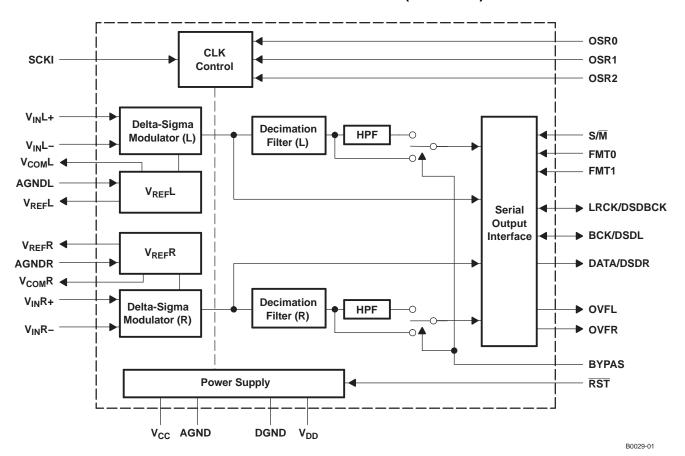


# PRINCIPLES OF OPERATION

# THEORY OF OPERATION

The PCM1804 consists of a band-gap reference, a delta-sigma modulator with full-differential architecture for L-channel and R-channel, a decimation filter with a high-pass filter, and a serial interface circuit. Figure 30 illustrates the total architecture of the PCM1804. An on-chip, high-precision reference with 10- $\mu$ F external capacitor(s) provides all the reference voltage needed in the PCM1804, and it defines the full-scale voltage range of both channels. Full-differential architecture provides a wide dynamic range and excellent power-supply rejection performance. The input signal is sampled at ×128, ×64, and ×32 oversampling rates according to the overasmpling ratio control, OSR[0:2]. The single rate, dual rate, and quad rate eliminate the external sample-hold amplifier. Figure 31 illustrates how for each oversampling ratio the PCM1804 decimates the modulator output down to PCM data when the modulator is running at 6.144 MHz. The delta-sigma modulation randomizes the modulator outputs and reduces the idle tone level. The oversampled data stream from the delta-sigma modulator is converted to a 1-f<sub>S</sub>, 24-bit digital signal, while removing high-frequency noise components using a decimation filter. The dc components of the signal are removed by the HPF, and the HPF output is converted to a time-multiplexed serial signal through the serial interface, which provides flexible serial formats and master/slave modes. The PCM1804 also has a DSD output mode. The PCM1804 can output the signal directly from the modulators to DSDL (pin 16) and DSDR (pin 15).





#### **PRINCIPLES OF OPERATION (continued)**



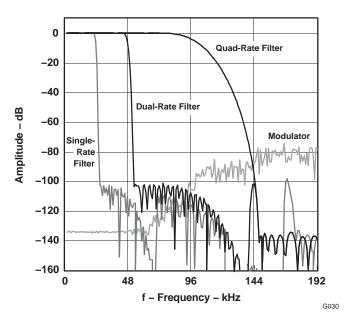


Figure 31. Spectrum of Modulator Output and Decimation Filter



#### **PRINCIPLES OF OPERATION (continued)**

#### SYSTEM CLOCK INPUT

The PCM1804 supports 128  $f_S$ , 192  $f_S$  (only in master mode at quad rate), 256  $f_S$ , 384  $f_S$ , 512  $f_S$ , and 768  $f_S$  as a system clock, where  $f_S$  is the audio sampling frequency. The system clock must be supplied on SCKI (pin 18). Table 3 shows the relationship of typical sampling frequency and the system clock frequency, and Figure 32 shows system clock timing. In master mode, the system clock rate is selected by OSR2 (pin 11), OSR1 (pin 10), and OSR0 (pin 9) as shown in Table 1. In slave mode, the system clock rate is automatically detected. In DSD mode, OSR2 (pin 11), OSR1 (pin 10), OSR0 (pin 9), and the system clock frequency are fixed as shown in Table 1 and Table 3.

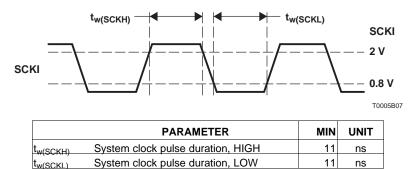


Figure 32. System Clock Input Timing

### **POWER-ON AND RESET FUNCTIONS**

The PCM1804 has both an internal power-on-reset circuit and  $\overline{RST}$  (pin 19). For internal power-on reset, initialization (reset) is performed automatically at the time when the power supply V<sub>DD</sub> exceeds 2 V (typical) and V<sub>CC</sub> exceeds 4 V (typical). RST accepts external forced reset, and a low level on  $\overline{RST}$  initiates the reset sequence. Because an internal pulldown resistor terminates  $\overline{RST}$ , no connection of  $\overline{RST}$  is equivalent to a low-level input. Because the system clock is used as a clock signal for the reset circuit, the system clock must be supplied as soon as power is supplied; more specifically, at least three system clocks are required prior to V<sub>DD</sub> > 2 V, V<sub>CC</sub> > 4 V, and  $\overline{RST}$  = high. While V<sub>DD</sub> < 2 V (typical), V<sub>CC</sub> < 4 V (typical), or  $\overline{RST}$  = low, and 1/f<sub>S</sub> (maximum) count after V<sub>DD</sub> > 2 V (typical), V<sub>CC</sub> > 4 V (typical) and  $\overline{RST}$  = high, the PCM1804 stays in the reset state and the digital output is forced to zero. The digital output is valid after the reset state is released and the time of 1116/f<sub>S</sub> has passed. Figure 33 and Figure 34 illustrate the internal power-on-reset and external-reset timing, respectively. Figure 35 illustrates the digital output for power-on reset and RST control. The PCM1804 needs RST = low when control pins are changed or in slave mode when SCKI, LRCK, and BCK are changed.

### **POWER-DOWN FUNCTION**

The PCM1804 has a power-down feature that is controlled by  $\overline{RST}$  (pin 19). Entering the power-down mode is done by keeping the  $\overline{RST}$  input level low for more than  $65536/f_S$ . In the master mode, the SCKI (pin 18) is used as the clock signal for the power-down counter. While in the slave mode, SCKI (pin 18) and LRCK (pin 17) are used as the clock signal. The clock(s) must be supplied until the power-down sequence completes. As soon as RST goes high, the PCM1804 starts the reset-release sequence described in the *Power-On and Reset Functions* section.

### **OVERSAMPLING RATIO**

The oversampling ratio is selected by OSR2 (pin 11), OSR1 (pin 10), and OSR0 (pin 9) as shown in Table 1 and Table 2. The PCM1804 needs  $\overline{\text{RST}}$  = low when logic levels on the OSR2, OSR1, and OSR0 pins are changed.

OSR2	OSR1	OSR0	OVERSAMPLING RATIO	SYSTEM CLOCK RATE			
Low	Low Low		Single rate (× 128 f <sub>S</sub> )	768 f <sub>S</sub>			
Low	Low	High	Single rate (× 128 f <sub>S</sub> )	512 f <sub>S</sub>			
Low	High	Low	Single rate (× 128 f <sub>S</sub> )	384 f <sub>S</sub>			
Low	High	High	Single rate (× 128 f <sub>S</sub> )	256 f <sub>S</sub>			
High	Low	Low	Dual rate (× 64 f <sub>S</sub> )	384 f <sub>S</sub>			
High	Low	High	Dual rate (× 64 f <sub>S</sub> )	256 f <sub>S</sub>			
High	High	Low	Quad rate (× 32 f <sub>S</sub> )	192 f <sub>S</sub>			
High	High	High	Quad rate (× 32 f <sub>S</sub> )	128 f <sub>S</sub>			
High	Low	Low	DSD mode (× 64 f <sub>S</sub> )	384 f <sub>S</sub>			
High	Low	High	DSD mode (× 64 f <sub>S</sub> )	256 f <sub>S</sub>			

#### Table 1. Oversampling Ratio in Master Mode

#### Table 2. Oversampling Ratio in Slave Mode

OSR2	OSR2 OSR1 OSR0		OVERSAMPLING RATIO	SYSTEM CLOCK RATE
Low	v Low Low		Single rate (× 128 f <sub>S</sub> )	Automatically detected
Low	Low	High	Dual rate (× 64 f <sub>S</sub> )	Automatically detected
Low	High	Low	Quad rate ( $\times$ 32 f <sub>S</sub> ) <sup>(1)</sup>	Automatically detected
Low	High	High	Reserved	-
High	Low	Low	Reserved	-
High	Low	High	Reserved	-
High	High	Low	Reserved	-
High	High	High	Reserved	-

(1) Only at the 128- $f_S$  system clock rate

#### Table 3. Sampling Frequency and System Clock Frequency

OVERSAMPLING RATIO	SAMPLING		SYST	EM CLOCK	M CLOCK FREQUENCY (MHz)			
OVERSAMPLING RATIO	FREQUENCY (kHz)	128 f <sub>S</sub>	192 f <sub>S</sub> <sup>(1)</sup>	256 f <sub>S</sub>	384 f <sub>S</sub>	512 f <sub>S</sub>	768 f <sub>S</sub>	
	32	-	-	8.192	12.288	16.384	24.576	
Single rate <sup>(2)</sup>	44.1	-	-	11.2896	16.9344	22.5792	33.8688	
	48	-	-	12.288	18.432	24.576	36.864	
Dual rate <sup>(3)</sup>	88.2	-	-	22.5792	33.8688	-	-	
Duarrale	96	-	-	24.576	36.864	-	-	
Quad rate <sup>(4)</sup>	176.4	22.5792	33.8688	-	-	-	-	
	192	24.576	36.864	-	-	-	-	
DSD mode <sup>(3)</sup>	44.1	-	-	11. 2896	16.9344	-	-	

(1) Only available in master mode at the quad rate

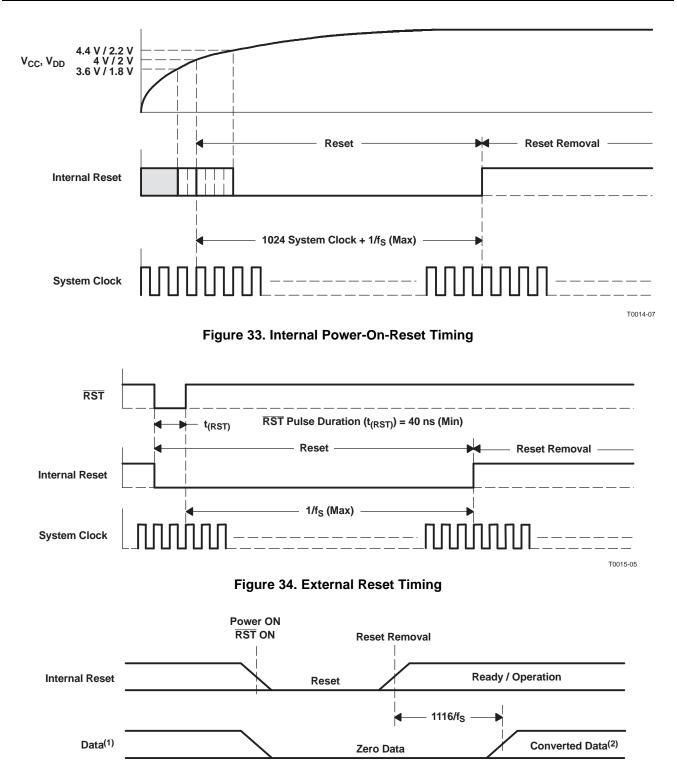
Modulator is running at 128 f<sub>S</sub>. (2) (3) (4)

Modulator is running at  $64 \text{ f}_{\text{S}}$ . Modulator is running at  $32 \text{ f}_{\text{S}}$ .

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(1) In the DSD mode, DSDL is also controlled like DSDR.

(2) The HPF transient response appears initially.

#### Figure 35. ADC Digital Output for Power-On Reset and RST Control

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### AUDIO DATA INTERFACE

The PCM1804 interfaces the audio system through BCK/DSDL (pin 16), LRCK/DSDBCK (pin 17), and DATA/DSDR (pin 15). The PCM1804 needs  $\overline{RST}$  = low when in the interface mode and/or the data format are changed.

#### INTERFACE MODE

The PCM1804 supports master mode and slave mode as interface modes, which are selected by  $S/\overline{M}$  (pin 8) as shown in Table 4. In master mode, the PCM1804 provides the timing of the serial audio data communications between the PCM1804 and the digital audio processor or external circuit. While in slave mode, the PCM1804 receives the timing for data transfer from an external controller. Slave mode is not available for DSD.

S/M	MODE						
Low	Master mode						
High	Slave mode						

Table 4. Interface Mode

**DATA FORMAT** 

The PCM1804 supports four audio data formats in both master and slave modes, and these data formats are selected by FMT0 (pin 6) and FMT1 (pin 7) as shown in Table 5.

FMT1	FMT0	FORMAT	MASTER	SLAVE
Low	Low	PCM, left-justified, 24-bit	Yes	Yes
Low	High	PCM, I <sup>2</sup> S, 24-bit	Yes	Yes
High	Low	PCM, standard, 24-bit	Yes	Yes
High	High	DSD	Yes	-

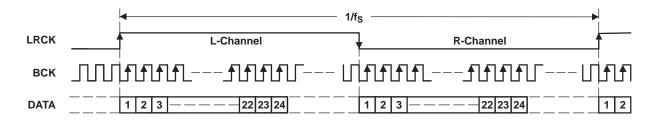
#### Table 5. Data Format



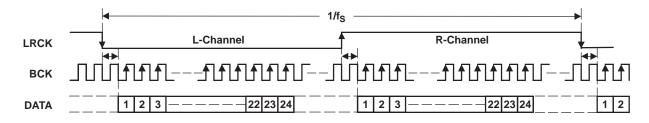
# INTERFACE TIMING FOR PCM

Figure 36 through Figure 38 illustrate the interface timing for PCM.

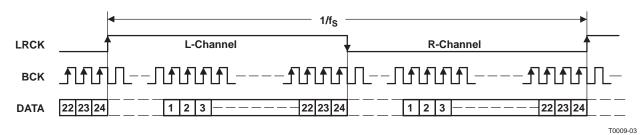
#### (1) Left-Justified Data Format; L-Channel = High, R-Channel = Low



#### (2) I<sup>2</sup>S Data Format; L-Channel = Low, R-Channel = High



(3) Standard Data Format; L-Channel = High, R-Channel = Low



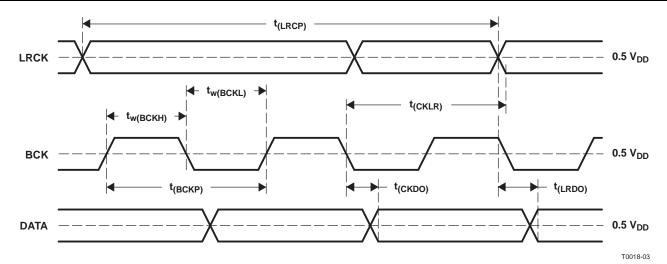
NOTE: LRCK and BCK work as outputs in master mode and as inputs in slave mode.

#### Figure 36. Audio Data Format for PCM



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PARAMETERS MIN TYP UNIT MAX 1/(64 f<sub>S</sub>)<sup>(3)</sup> BCK period t<sub>(BCKP)</sub> BCK pulse duration, HIGH 32 tw(BCKH) ns BCK pulse duration, LOW 32 ns tw(BCKL) Delay time, BCK falling edge to LRCK valid -5 15 ns  $t_{(CKLR)}$ LRCK period 1/f<sub>S</sub> t(LRCP) Delay time, BCK falling edge to DATA valid -5 15 t<sub>(CKDO)</sub> ns Delay time, LRCK edge to DATA valid -5 15 ns t<sub>(LRDO)</sub> tr Rising time of all signals<sup>(1)(2)</sup> 10 ns Falling time of all signals<sup>(1)(2)</sup> t<sub>f</sub> 10 ns

(1) Rising and falling times are measured from 10% to 90% of IN/OUT signal swing.

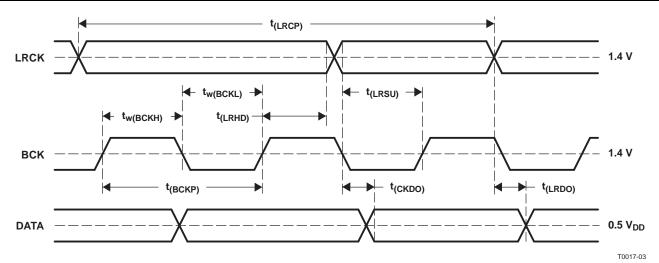
(2) Load capacitance of all signals is 10 pF.

(3)  $t_{(BCKP)}$  is fixed at 1/(64 f<sub>S</sub>) in case of master mode.

#### Figure 37. Audio Data Interface Timing for PCM (Master Mode: LRCK and BCK Work as Outputs)

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	PARAMETERS	MIN	TYP	MAX	UNIT
t <sub>(BCKP)</sub>	BCK period	1/(64 f <sub>S</sub> )		1/(48 f <sub>S</sub> )	
t <sub>w(BCKH)</sub>	BCK pulse duration, HIGH	32			ns
t <sub>w(BCKL)</sub>	BCK pulse duration, LOW	32			ns
t <sub>(LRSU)</sub>	LRCK setup time to BCK rising edge	12			ns
t <sub>(LRHD)</sub>	LRCK hold time to BCK rising edge	12			ns
t <sub>(LRCP)</sub>	LRCK period		1/f <sub>S</sub>		
t <sub>(CKDO)</sub>	Delay time, BCK falling edge to DATA valid	5		25	ns
t <sub>(LRDO)</sub>	Delay time, LRCK edge to DATA valid	5		25	ns
t <sub>r</sub>	Rising time of all signals <sup>(1)(2)</sup>			10	ns
t <sub>f</sub>	Falling time of all signals <sup>(1)(2)</sup>			10	ns

(1) Rising and falling times are measured from 10% to 90% of IN/OUT signals swing.

(2) Load capacitance of DATA/DSDR signal is 10 pF.

#### Figure 38. Audio Data Interface Timing for PCM (Slave Mode: LRCK and BCK Work as Inputs)

### INTERFACE TIMING FOR DSD

Figure 39 and Figure 40 illustrate the interface timing for DSD.

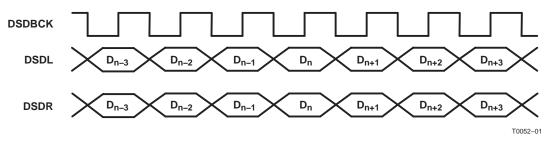
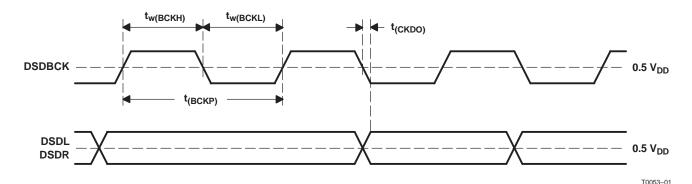


Figure 39. Audio Data Format

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	PARAMETERS	MIN	TYP	MAX	UNIT
t <sub>(BCKP)</sub>	DSDBCK period		354		ns
t <sub>w(BCKH)</sub>	DSDBCK pulse duration, HIGH		177		ns
t <sub>w(BCKL)</sub>	DSDBCK pulse duration, LOW		177		ns
t <sub>(CKDO)</sub>	Delay time DSDBCK falling edge to DSDL, DSDR valid	-5		15	ns
t <sub>r</sub>	Rising time of all signals <sup>(1)(2)</sup>			10	ns
t <sub>f</sub>	Falling time of all signals <sup>(1)(2)</sup>			10	ns

(1) Rising and falling times are measured from 10% to 90% of IN/OUT signal swing.

(2) Load capacitance of DSDBCK/DSDL/DSDR signal is 10 pF.

#### Figure 40. Audio Data Interface Timing for DSD (Master Mode Only)

#### SYNCHRONIZATION WITH DIGITAL AUDIO SYSTEM FOR PCM

In slave mode, the PCM1804 operates under LRCK synchronized with the system clock SCKI. The PCM1804 does not need a specific phase relationship between LRCK and SCKI, but does require the synchronization of LRCK and SCKI.

If the relationship between LRCK and SCKI changes more than  $\pm 6$  BCK during one sample period due to LRCK or SCKI jitter, internal operation of the ADC halts within  $1/f_S$  and digital output is forced into BPZ code until resynchronization between LRCK and SCKI is completed.

In case of changes less than ±5 BCK, resynchronization does not occur and the previously described digital output control and discontinuity do not occur.

Figure 41 illustrates ADC digital output for loss of synchronization and resynchronization. During undefined data, the PCM1804 may generate some noise in the audio signal. Also, the transitions of normal to undefined data and undefined or zero data to normal cause a discontinuity of data on the digital output. This can generate noise in the audio signal. In master mode, synchronization loss never occurs.

#### HIGH-PASS FILTER (HPF) BYPASS CONTROL FOR PCM

The built-in function for dc component rejection can be bypassed by BYPAS (pin 12) control. In bypass mode, the dc component of the input analog signal and the internal dc offset are also converted and output in the digital output data.

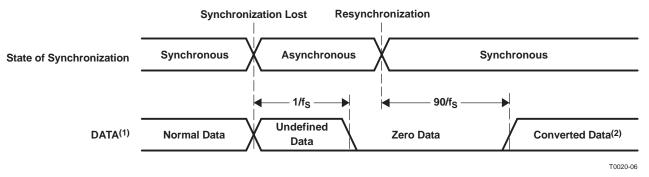
BYPAS PIN	HPF MODE
Low	Normal (high-pass) mode
High	Bypass (through) mode

#### **HPF Bypass Control**



### OVERFLOW FLAG FOR PCM

The PCM1804 has two overflow flag pins, OVFR (pin 20) and OVFL (pin 21). The pins go to high as soon as the analog input goes across the full-scale range. The high level is held for 1.016 s at maximum, and returns to low if the analog input does not go across the full-scale range for the period.



(1) Applies only for slave mode; the loss of synchronization never occurs in master mode.

(2) The HPF transient response appears initially.

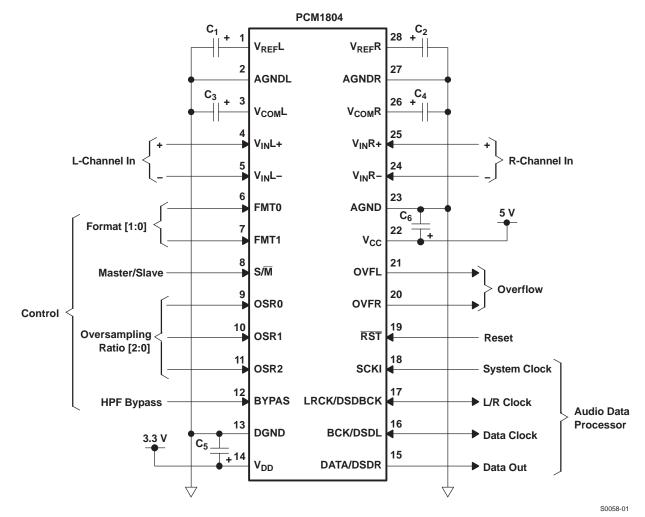
#### Figure 41. ADC Digital Output for Loss of Synchronization and Resynchronization





# TYPICAL CIRCUIT CONNECTION DIAGRAM

Figure 42 illustrates a typical circuit connection diagram in the PCM data format operation.



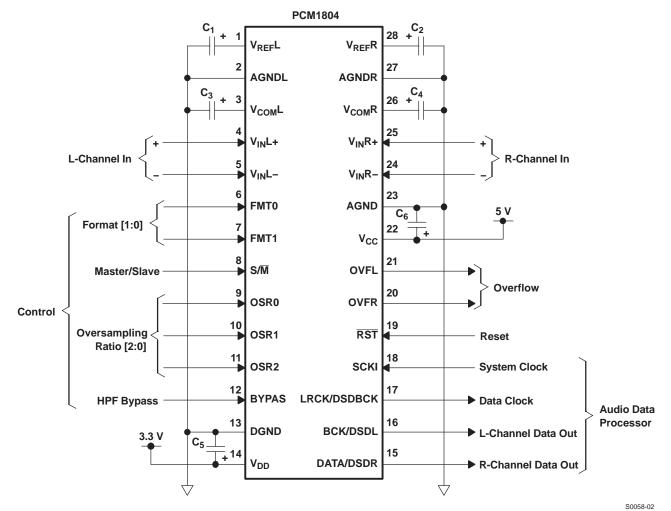
- A. C1, C2, C5, and C6: Bypass capacitors, 0.1-µF ceramic and 10-µF tantalum, depending on layout and power supply
- B. C3, C4: Bypass capacitor, 0.1-µF tantalum, depending on layout and power supply

Figure 42. Typical Circuit Connection Diagram for PCM

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Figure 43 illustrates a typical circuit connection diagram in the DSD data format operation.



A. C1, C2, C5, and C6: Bypass capacitors, 0.1-µF ceramic and 10-µF tantalum, depending on layout and power supply

C3 and C4: Bypass capacitors, 0.1-µF tantalum, depending on layout and power supply

Figure 43. Typical Circuit Connection Diagram for DSD

### **APPLICATION INFORMATION**

#### **BOARD DESIGN AND LAYOUT CONSIDERATIONS**

#### V<sub>CC</sub>, V<sub>DD</sub> Pins

The digital and analog power supply lines to the PCM1804 should be bypassed to the corresponding ground pins with 0.1- $\mu$ F ceramic and 10- $\mu$ F tantalum capacitors placed as close to the pins as possible to maximize the dynamic performance of the ADC. Although the PCM1804 has two power lines to maximize the potential of dynamic performance, using one common power supply is recommended to avoid unexpected power-supply trouble like latch-up or power-supply sequence.

#### V<sub>IN</sub> Pins

Use of 0.01- $\mu$ F film capacitors between V<sub>IN</sub>L+ and V<sub>IN</sub>L– and between V<sub>IN</sub>R+ and V<sub>IN</sub>R– is strongly recommended to remove higher-frequency noise from the delta-sigma input section.

#### V<sub>REF</sub>X, V<sub>COM</sub>X Inputs

Use 0.1- $\mu$ F ceramic and 10- $\mu$ F tantalum capacitors between V<sub>REF</sub>L, V<sub>REF</sub>R, and corresponding AGNDx, to ensure low-source impedance at ADC references. Use 0.1- $\mu$ F tantalum capacitors between V<sub>COM</sub>L, V<sub>COM</sub>R and corresponding AGNDx to ensure low source impedance of common voltage. These capacitors should be located as close as possible to the V<sub>REF</sub>L, V<sub>REF</sub>R, V<sub>COM</sub>L, and V<sub>COM</sub>R pins to reduce dynamic errors on references and common voltage. The dc voltage level of these pins is 2.5 V.

#### DATA/DSDR, BCK/DSDL, and LRCK/DSDBCK Pins

The DATA/DSDR, BCK/DSDL, and LRCK/DSDBCK pins in master mode have large load drive capability. Locating the buffer near the PCM1804 and minimizing the load capacitance, minimizes the digital-analog crosstalk and maximizes the dynamic performance of the ADC.

#### System Clock

The quality of the system clock can influence dynamic performance, as the PCM1804 operates based on a system clock. Therefore, it might be necessary to consider the system clock duty, jitter, and the time difference between system clock transition and BCK/DSDL or LRCK/DSDBCK transition in slave mode.

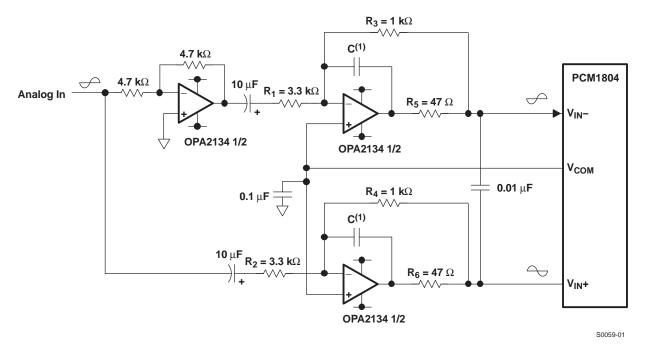
#### Reset Control

If capacitors larger than 10  $\mu$ F are used on V<sub>REF</sub>L and V<sub>REF</sub>R, an external reset control with a delay time corresponding to the V<sub>REF</sub>L and V<sub>REF</sub>R response is required. Also, it works as a power-down control.

#### APPLICATION CIRCUIT FOR SINGLE-ENDED INPUT

An application diagram for a single-ended input circuit is shown in Figure 44. The maximum signal input voltage and differential gain of this circuit is designed as Vinmax = 8.28 Vpp, Ad = 0.3. Differential gain (Ad) is given by R3/R1(R4/R2) in a circuit configured as a normal inverted-gain amplifier. Resistor R5(R6) in the feedback loop gives low-impedance drive operation and noise filtering for the analog input of the PCM1804. The circuit technique using R5(R6) is recommended.





(1) A capacitor value of 1800 pF is recommended, unless an input signal greater than -6 dBFS at 100 kHz or higher is applied in the DSD mode. In that case, 3300 pF is recommended.

Figure 44. Application Circuit for Single-Ended Input Circuit (PCM)

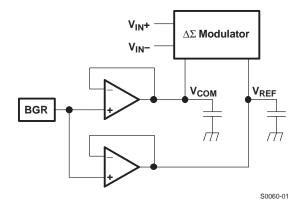


Figure 45. Equivalent Circuit of Internal Reference (V<sub>COM</sub>, V<sub>REF</sub>)



### PACKAGING INFORMATION

Orderable Device	Status	Package Type	•	Pins	•	Eco Plan	Lead finish/	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	Ball material	(3)		(4/5)	
							(6)				
PCM1804DB	ACTIVE	SSOP	DB	28	47	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	-10 to 70	PCM1804	Samples
PCM1804DBG4	ACTIVE	SSOP	DB	28	47	RoHS & Green	Call TI	Level-1-260C-UNLIM	-10 to 70	PCM1804	Samples
PCM1804DBR	ACTIVE	SSOP	DB	28	2000	RoHS & Green	Call TI   NIPDAU	Level-1-260C-UNLIM	-10 to 70	PCM1804	Samples
PCM1804DBRG4	ACTIVE	SSOP	DB	28	2000	RoHS & Green	Call TI	Level-1-260C-UNLIM	-10 to 70	PCM1804	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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#### OTHER QUALIFIED VERSIONS OF PCM1804 :

• Automotive : PCM1804-Q1

NOTE: Qualified Version Definitions:

• Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects



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# TAPE AND REEL INFORMATION





#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are non	ninal
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Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
PCM1804DBR	SSOP	DB	28	2000	330.0	16.4	8.2	10.5	2.5	12.0	16.0	Q1



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# PACKAGE MATERIALS INFORMATION

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
PCM1804DBR	SSOP	DB	28	2000	356.0	356.0	35.0

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# TUBE



# - B - Alignment groove width

#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	Τ (μm)	B (mm)
PCM1804DB	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DB	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DB	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DBG4	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DBG4	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DBG4	DB	SSOP	28	47	530	10.5	4000	4.1
PCM1804DBR	DB	SSOP	28	2000	530	10.5	4000	4.1
PCM1804DBRG4	DB	SSOP	28	2000	530	10.5	4000	4.1

# **DB0028A**



# **PACKAGE OUTLINE**

# SSOP - 2 mm max height

SMALL OUTLINE PACKAGE



NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M. 2. This drawing is subject to change without notice. 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-150.



# DB0028A

# **EXAMPLE BOARD LAYOUT**

# SSOP - 2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



# DB0028A

# **EXAMPLE STENCIL DESIGN**

# SSOP - 2 mm max height

SMALL OUTLINE PACKAGE



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

9. Board assembly site may have different recommendations for stencil design.



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