# onsemi

## High Side Driver, Self Protected, Very Low Iq, with Analog Current Sense

## NCV84090

The NCV84090 is a fully protected single channel high side driver that can be used to switch a wide variety of loads, such as bulbs, solenoids, and other actuators. The device incorporates advanced protection features such as active inrush current management, over-temperature shutdown with automatic restart and an overvoltage active clamp. A dedicated Current Sense pin provides precision analog current monitoring of the output as well as fault indication of short to  $V_D$ , short circuit to ground and ON and OFF state open load detection. An active high Current Sense Enable pin allows all diagnostic and current sense features to be enabled.

#### Features

- Short Circuit Protection with Inrush Current Management
- CMOS (3 V / 5 V) Compatible Control Input
- Very Low Standby Current
- Very Low Current Sense Leakage
- Proportional Load Current Sense
- Current Sense Enable
- Off State Open Load Detection
- Output Short to V<sub>D</sub> Detection
- Overload and Short to Ground Indication
- Thermal Shutdown with Automatic Restart
- Undervoltage Shutdown
- Integrated Clamp for Inductive Switching
- Loss of Ground and Loss of V<sub>D</sub> Protection
- ESD Protection
- Reverse Battery Protection with External Components
- AEC–Q100 Grade 1 Qualified and PPAP Capable
- This is a Pb–Free Device

#### **Typical Applications**

- Switch a Variety of Resistive, Inductive and Capacitive Loads
- Can Replace Electromechanical Relays and Discrete Circuits
- Automotive / Industrial

#### FEATURE SUMMARY

Max Supply Voltage	VD	41	V
Operating Voltage Range	VD	4 to 28	V
$R_{DSon}$ (typical) $T_J = 25^{\circ}C$	R <sub>ON</sub>	90	mΩ
Output Current Limit (typical)	I <sub>lim</sub>	24	А
OFF-state Supply Current (max)	I <sub>D(off)</sub>	0.5	μΑ





CASE 751 STYLE 11



(Note: Microdot may be in either location)

#### PIN CONNECTIONS



#### ORDERING INFORMATION

Devic	e	Package	Shipping <sup>†</sup>
NCV84090[	DR2G	SOIC-8 (Pb-Free)	2500 / Tape & Reel

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specification Brochure, BRD8011/D.

#### **Block Diagram & Pin Configuration**





#### Table 1. SO8 PACKAGE PIN DESCRIPTION

Pin #	Symbol	Description
1	IN	Logic Level Input
2	CS_EN	Current Sense Enable
3	GND	Ground
4	CS	Analog Current Sense Output
5	VD	Supply Voltage
6	OUT	Output
7	OUT	Output
8	VD	Supply Voltage



Figure 2. Voltage and Current Conventions

Table 2. Connection suggestions	for unused and	d or unconnected	pins
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Connection	Input	Output	Current Sense	Current Sense Enable
Floating	Х	Х	Not Allowed	Х
To Ground	Through 10 k $\Omega$ resistor	Not Allowed	Through 1 k $\Omega$ Resistor	Through 10 k $\Omega$ resistor



Figure 3. Pin Configuration (top view)

#### **ELECTRICAL SPECIFICATIONS**

#### **Table 3. MAXIMUM RATINGS**

		Va		
Rating	Symbol	Min	Мах	Unit
DC Supply Voltage	VD	-0.3	41	V
Max Transient Supply Voltage (Note 1) Load Dump Suppresses	V <sub>PEAK</sub>	-	45	V
Input Voltage	V <sub>IN</sub>	-10	10	V
Input Current	I <sub>IN</sub>	-5	5	mA
Reverse Ground Pin Current	I <sub>GND</sub>	-	-200	mA
Output Current (Note 2)	I <sub>OUT</sub>	-6	Internally Limited	А
Reverse CS Current (Note 1)	I <sub>CS</sub>	-	-200	mA
CS Voltage	V <sub>CS</sub>	V <sub>D</sub> -41	V <sub>D</sub>	V
CS_EN Voltage	V <sub>CS_EN</sub>	-10	10	V
CS_EN Current	I <sub>CS_EN</sub>	-5	5	mA
Power Dissipation Tc = 25°C (Note 6)	P <sub>tot</sub>	1.9	95	W
Electrostatic Discharge (Note 3) (HBM Model 100 pF / 1500 Ω)	V <sub>ESD</sub>		-	DC
Input Current Sense		4 4		kV kV
Output		4		kV kV
VD		4		kV
Charge Device Model CDM-AEC-Q100-011		750		V
+Single Pulse Inductive Load Switching Energy (L = 5 mH, Vbat = 13.5 V; $I_L$ = 4.8 A, $T_{Jstart}$ = 150°C (Note 4)	E <sub>AS</sub>	_	81	mJ
Operating Junction Temperature	TJ	-40	+150	°C
Storage Temperature	T <sub>storage</sub>	-55	+150	°C

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected. 1. Load Dump Test B (with centralized load dump suppression) according to ISO16750–2 standard. Guaranteed by design. Not tested in

production. Passed Class C (or A, B) according to ISO16750–1.
Reverse Output current has to be limited by the load to stay within absolute maximum ratings and thermal performance.

3. This device series incorporates ESD protection and is tested by the following methods:

ESD Human Body Model tested per AEC-Q100-002 (JS-001-2017)

Field Induced Charge Device Model ESD characterization is not performed on plastic molded packages with body sizes smaller than 2 x 2 mm due to the inability of a small package body to acquire and retain enough charge to meet the minimum CDM discharge current waveform characteristic defined in JEDEC JS-002-2018

4. Not subjected to production testing.

#### **Table 4. THERMAL RESISTANCE RATINGS**

Parameter	Symbol	Max. Value	Units
Thermal Resistance Junction-to-Lead (Note 5) Junction-to-Ambient (Note 5) Junction-to-Ambient (Note 6)	R <sub>thJL</sub> R <sub>thJA</sub> R <sub>thJA</sub>	27.3 50 64	°C / W

5. 645 mm<sup>2</sup> pad size, mounted on four-layer 2s2p PCB - FR4, 2 oz. Cu thickness for top layer and 1 oz. Cu thickness for inner layers (planes not electrically connected)

6. 2 cm<sup>2</sup> pad size, mounted on single-layer 2s0p PCB - FR4, 2 oz. Cu thickness (planes not electrically connected)

**ELECTRICAL CHARACTERISTICS** ( $7 \le V_D \le 28 \text{ V}$ ;  $-40^{\circ}\text{C} < T_J < 150^{\circ}\text{C}$  unless otherwise specified) Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

#### Table 5. POWER

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Operating Supply Voltage	VD		4	-	28	V
Undervoltage Shutdown	V <sub>UV</sub>		-	3.5	4	V
Undervoltage Shutdown Hysteresis	V <sub>UV_hyst</sub>		-	0.4	-	V
On Resistance	R <sub>ON</sub>	$I_{OUT} = 2.5 \text{ A}, \text{ T}_{\text{J}} = 25^{\circ}\text{C}$	-	90	-	mΩ
		I <sub>OUT</sub> = 2.5 A, T <sub>J</sub> = 150°C	-	-	180	
		$I_{OUT}$ = 2.5 A, $V_{D}$ = 4.5 V, $T_{J}$ = 25°C	-	-	144	
Supply Current (Note 7)	۱ <sub>D</sub>	OFF-state: $V_D$ = 13 V, $V_{IN}$ = $V_{OUT}$ = 0 V, $T_J$ = 25°C	-	0.2	0.5	μΑ
		OFF-state: $V_D$ = 13 V, $V_{IN}$ = $V_{OUT}$ = 0 V, $T_J$ = 85°C (Note 8)	-	0.2	0.5	μΑ
		OFF-state: $V_D$ = 13 V, $V_{IN}$ = $V_{OUT}$ = 0 V, $T_J$ = 125°C	-	-	3	μΑ
		ON-state: $V_D = 13 V$ , $V_{IN} = 5 V$ , $I_{OUT} = 0 A$	-	1.9	3.5	mA
On State Ground Current	I <sub>GND(ON)</sub>	$V_D$ = 13 V, $V_{CS EN}$ = 5 V $V_{IN}$ = 5 V, $I_{OUT}$ = 1 A	-	-	6	mA
Output Leakage Current	١L	$V_{IN} = V_{OUT} = 0 V, V_D = 13 V, T_J = 25^{\circ}C$	_	_	0.5	μΑ
		$V_{IN} = V_{OUT} = 0 \text{ V}, V_D = 13 \text{ V}, T_J = 125^{\circ}\text{C}$	_	-	3	

7. Includes PowerMOS leakage current.

8. Not Subject to production testing.

#### Table 6. LOGIC INPUTS (V\_D = 13.5 V; -40°C $\leq$ T\_J $\leq$ 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Input Voltage – Low	V <sub>in_low</sub>		-	-	0.9	V
Input Current – Low	I <sub>in_low</sub>	V <sub>in</sub> = 0.9 V	1	-	-	μΑ
Input Voltage – High	V <sub>in_high</sub>		2.1	-	-	V
Input Current – High	l <sub>in_high</sub>	V <sub>in</sub> = 2.1 V	-	-	10	μΑ
Input Hysteresis Voltage	V <sub>in_hyst</sub>		-	0.2	-	V
Input Clamp Voltage	V <sub>in_cl</sub>	l <sub>in</sub> = 1 mA	12	13	14	V
		I <sub>in</sub> = -1 mA	-14	-13	-12	
CS_EN Voltage – Low	V <sub>CSE_low</sub>		-	-	0.9	V
CS_EN Current – Low	I <sub>CSE_low</sub>	V <sub>CS_EN</sub> = 0.9 V	1	-	-	μA
CS_EN Voltage – High	V <sub>CSE_high</sub>		2.1	-	-	V
CS_EN Current – High	I <sub>CSE_high</sub>	V <sub>CS_EN</sub> = 2.1 V	-	-	10	μΑ
CS_EN Hysteresis Voltage	V <sub>CSE_Hys</sub>		-	0.2	-	V
CS_EN Clamp Voltage	V <sub>CSE_cl</sub>	I <sub>CS_EN</sub> = 1 mA	12	13	14	V
		I <sub>CS_EN</sub> = -1 mA	-14	-13	-12	1

#### Table 7. SWITCHING CHARACTERISTICS (Note 11) (V\_D = 13. V, $-40^{\circ}C < T_J < 150^{\circ}C$ )

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Turn-On Delay Time	t <sub>d_on</sub>	$V_{in}$ high to 20% Vout, $V_D$ = 13 V, $R_L$ = 6.5 $\Omega$ , $T_J$ = 25°C	5	70	120	μs
Turn-Off Delay Time	t <sub>d_off</sub>	$V_{in}$ low to 80% Vout, $V_D$ = 13 V, $R_L$ = 6.5 $\Omega,$ $T_J$ = 25°C	5	40	100	μs
Slew Rate On	dV <sub>out</sub> /dt <sub>on</sub>	20% to 80% Vout, V_D = 13 V, R_L = 6.5 $\Omega,$ T_J = 25°C	0.1	0.27	0.7	V / μs
Slew Rate Off	dV <sub>out</sub> /dt <sub>off</sub>	80% to 20% Vout, V_D = 13 V, R_L = 6.5 $\Omega,$ $T_J$ = 25°C	0.1	0.35	0.7	V / μs
Turn–On Switching Loss (Note 9)	E <sub>on</sub>	$V_D$ = 13 V, $R_L$ = 6.5 $\Omega$	-	0.15	0.22	mJ
Turn–Off Switching Loss (Note 9)	E <sub>off</sub>	$V_D$ = 13 V, $R_L$ = 6.5 $\Omega$	-	0.15	0.22	mJ
Differential Pulse Skew, (t <sub>OFF</sub> - t <sub>ON</sub> ) see Figure 4	t <sub>skew</sub>	$V_D$ = 13 V, $R_L$ = 6.5 $\Omega$	-50	-	50	μs

9. Not subjected to production testing.

#### Table 8. OUTPUT DIODE CHARACTERISTICS

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Forward Voltage	V <sub>F</sub>	$I_{out} = -1$ A, $V_F = V_{OUT} - VD$ , $T_J = 150^{\circ}C$	-	-	0.7	V

#### Table 9. PROTECTION FUNCTIONS (Note 10) (7 $\leq$ V\_D $\leq$ 18 V, -40°C < T\_J < 150°C)

				Value		
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Temperature Shutdown (Note 11)	T <sub>SD</sub>		150	175	200	°C
Temperature Shutdown Hysteresis (Note 11)	T <sub>SD_hyst</sub>		-	7	-	°C
Reset Temperature (Note 11)	T <sub>R</sub>		T <sub>RS</sub> +1	T <sub>RS</sub> +7	-	°C
Thermal Reset of Status (Note 11)	T <sub>RS</sub>		135	-	-	°C
Delta T Temperature Limit (Note 11)	T <sub>DELTA</sub>	$T_{\rm J} = -40^{\circ}$ C, $V_{\rm D} = 13$ V	-	60	-	°C
DC Output Current Limit	I <sub>limH</sub>	V <sub>D</sub> = 13 V	16	24	32	А
		4 V < V <sub>D</sub> < 18 V	-	-	32	А
Short Circuit Current Limit dur- ing Thermal Cycling (Note 11)	IlimTCycling	V <sub>D</sub> = 13 V T <sub>R</sub> < Tj < T <sub>TSD</sub>	-	8	-	A
Switch Off Output Clamp Voltage	V <sub>out_clamp</sub>	$I_{out} = 0.5 \text{ A}, V_{in} = 0 \text{ V}, L = 20 \text{ mH}$	V <sub>D</sub> – 41	V <sub>D</sub> - 46	V <sub>D</sub> – 52	V
Overvoltage Protection	V <sub>OV</sub>	V <sub>in</sub> = 0 V, I <sub>D</sub> = 20 mA	41	46	52	V
Output Voltage Drop Limitation	V <sub>DS_ON</sub>	$I_{out} = 0.2 \text{ A}, T_J = -40^{\circ}\text{C} \text{ to } 150^{\circ}\text{C}$	-	20	-	mV

10. To ensure long term reliability during overload or short circuit conditions, protection and related diagnostic signals must be used together with a fitting hardware & software strategy. If the device operates under abnormal conditions, this hardware & software solution must limit the duration and number of activation cycles.

11. Not subjected to production testing.

#### Table 10. OPEN–LOAD DETECTION (7 $\leq V_D \leq$ 18 V, –40°C < $T_J$ < 150°C)

			Value			
Rating	Symbol	Conditions	Min	Тур	Max	Unit
Open-load Off State Detection Threshold	V <sub>OL</sub>	V <sub>in</sub> = 0 V, V <sub>CS_EN</sub> = 5 V	2	-	4	V
Open-load Detection Delay at Turn Off	<sup>t</sup> d_OL_off		100	350	850	μs
Off State Output Current	I <sub>OLOFF1</sub>	$V_{in} = 0 V, V_{OUT} = V_{OL}$	-3	-	3	μΑ
Output rising edge to CS rising edge during open load	t <sub>d_OL</sub>	$V_{OUT}$ = 4 V, $V_{in}$ = 0 V $V_{CS}$ = 90% of $V_{CS}$ _High	-	5	30	μs

Table 11. CURRENT SENSE CHARACTERISTICS (7 $\leq$ V_D $\leq$ 18 V, -40°C $<$ T_J $<$ 150°	°C)
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			Value			
Rating	Symbol	Conditions	min	typ	max	Unit
Current Sense Ratio	K <sub>0</sub>	$I_{out}$ = 0.01 A, $V_{CS}$ = 0.5 V, $V_{CS}$ EN = 5 V	435	-	1250	
Current Sense Ratio	K <sub>1</sub>	$I_{out} = 0.025 \text{ A}, V_{CS} = 0.5 \text{ V}, V_{CS}_{EN} = 5 \text{ V}$ 400 800 1200		1200		
Current Sense Ratio Drift (Note 13)	$\Delta K_1 / K_1$	I <sub>out</sub> = 0.025 A, V <sub>CS_EN</sub> = 5 V	-20	-	15	%
Current Sense Ratio	K <sub>2</sub>	$I_{out}$ = 0.35 A, $V_{CS}$ = 4V, $V_{CS}$ _EN = 5 V	450	750	1000	
Current Sense Ratio Drift (Note 13)	$\Delta K_2 / K_2$	l <sub>out</sub> = 0.35 A, V <sub>CS_EN</sub> = 5 V	-15	-	15	%
Current Sense Ratio	K <sub>3</sub>	$I_{out}$ = 0.5 A, $V_{CS}$ = 4 V, $V_{CS}$ _EN = 5 V	530	750	950	
Current Sense Ratio Drift (Note 13)	$\Delta K_3 / K_3$	I <sub>out</sub> = 0.5 A, V <sub>CS_EN</sub> = 5 V	-15	-	10	%
Current Sense Ratio	K <sub>4</sub>	$I_{out}$ = 1 A, $V_{CS}$ = 4 V, $V_{CS}$ _EN = 5 V	625	750	835	
Current Sense Ratio Drift (Note 13)	$\Delta K_4 / K_4$	$I_{out} = 1 \text{ A}, V_{CS\_EN} = 5 \text{ V}$	-10	-	10	%
Current Sense Ratio	К <sub>5</sub>	$I_{out}$ = 3 A, $V_{CS}$ = 4 V, $V_{CS}$ _EN = 5 V	690	750	775	
Current Sense Ratio Drift (Note 13)	$\Delta K_5 / K_5$	$I_{out} = 3 \text{ A}, V_{CS_{EN}} = 5 \text{ V}$	-5	-	5	%
Current Sense Leakage Current	CS <sub>IIkg</sub>	$I_{out} = 0 \text{ A}, V_{CS} = 0 \text{ V}$ $V_{CS}_{EN} = 5 \text{ V}, V_{IN} = 0 \text{ V}$	-	-	1	μΑ
		$I_{out} = 0 \text{ A}, V_{CS} = 0 \text{ V}$ $V_{CS}_{EN} = 5 \text{ V}, V_{IN} = 5 \text{ V}$	-	-	2	
		$I_{out}$ = 2 A, $V_{CS}$ = 0 V $V_{CS}$ _EN = 0 V, $V_{IN}$ = 5 V,	-	-	0.5	
CS Max Voltage	CS <sub>Max</sub>	$\label{eq:VD} \begin{array}{l} V_{D} = 7 \; V,  V_{IN} = 5 \; V,  R_{CS} = 15 \; k\Omega, \\ I_{OUT} = 2 \; A,  V_{CS}{\_}EN = 5 \; V \end{array}$	5	-	7	V
Current Sense Voltage in Fault Condition (Note 12)	V <sub>CS_fault</sub>	$\label{eq:VD} \begin{array}{l} V_{D} = 13 \; V,  V_{IN} = 0 \; V,  R_{CS} = 1 \; k\Omega, \\ V_{OUT} = 4 \; V,  V_{CS\_EN} = 5 \; V \end{array}$	-	10	-	V
Current Sense Current in Fault Condition (Note 12)	I <sub>CS_fault</sub>	$V_{D}$ = 13 V, $V_{CS}$ = 5 V, $V_{IN}$ = 0 V, $V_{OUT}$ = 4 V, $V_{CS}$ _EN = 5 V	7	20	30	mA
Output Saturation Current (Note 13)	I <sub>OUT_sat</sub>	$V_{D} = 7 \text{ V}, V_{CS} = 4 \text{ V}, V_{IN} = 5 \text{ V}, T_{J} = 150^{\circ}\text{C}, V_{CS} = 15 \text{ V}$	3	-	-	A
CS_EN High to CS High Delay Time	t <sub>CS_High1</sub>		-	-	100	μs
CS_EN Low to CS Low Delay Time	t <sub>CS_Low1</sub>		-	5	25	μs
V <sub>in</sub> High to CS High Delay Time	t <sub>CS_High2</sub>	$V_{IN}$ = 0 to 5 V, $V_{CS EN}$ = 5 V, R <sub>CS</sub> = 1 kΩ, R <sub>L</sub> = 6.5 Ω		100	250	μs
V <sub>in</sub> Low to CS Low Delay Time	t <sub>CS_Low2</sub>			50	250	μs
Delay Time I <sub>D</sub> Rising Edge to Rising Edge of CS	$\Delta t_{CS_High2}$	$V_{IN}$ = 5 V, $V_{CS}$ EN = 5 V $R_{CS}$ = 1 k $\Omega$ , $I_{CS}$ = 90% of $I_{CS}$ Max	-	-	100	μs

12. The following fault conditions are: Overtemperature, Power Limitation, and OFF State Open–Load Detection. 13. Not subjected to production testing. For more information, refer to the AND9733–D Applications Note.

#### Table 12. TRUTH TABLE

Conditions	Input	Output	CS (V <sub>CS_EN</sub> = 5V) (Note 14)
Normal Operation	L H	L H	0 I <sub>CS</sub> = I <sub>OUT</sub> /K <sub>NOMINAL</sub>
Overtemperature	L H	L	0 Vcs_High
Undervoltage	L H	L	0 0
Overload	H H	H (no active current mgmt) Cycling (active current mgmt)	$I_{CS} = I_{OUT}/K_{NOMINAL}$ VCS_High
Short circuit to Ground	L H	L	0 V <sub>CS_High</sub>
OFF State Open Load	L	Н	V <sub>CS_High</sub>

14. If V<sub>CS\_EN</sub> is low, the Current Sense output is at a high impedance, its potential depends on leakage currents and external circuitry.

#### WAVEFORMS AND GRAPHS



#### Figure 4. Switching Characteristics



Figure 5. Normal Operation with Current Sense Timing Characteristics



Figure 6. Delay Response from rising edge of IOUT and rising edge of CS (for CS\_EN = 5V)



Figure 7. OFF-State Open-Load Flag Delay Timing



Figure 8. Off-State Open-Load with added external components







Figure 10.  $I_{OUT}/I_{CS}$  vs.  $I_{OUT}$ 



Figure 11. Current Sense Ratio Drift vs. Load Current











Figure 14. Discontinuous Overload or Short to GND



Figure 15. Short Circuit from OUT to VD

#### **TYPICAL CHARACTERISTICS**





Figure 16. Output Leakage Current vs. V<sub>D</sub>



80

100

120

V<sub>D</sub> = 13 V

120

140

140







80

100

120

140

60

TEMPERATURE (°C)

Figure 20.  $\rm V_{IN}$  Threshold High vs. Temperature

40

1.3 1.2

-40 -20 0

20



#### **TYPICAL CHARACTERISTICS**





Test Severity Levels					
ISO 7637-2:2011 Test Pulse		IV	Delays and Impedance	# of Pulses or Test Time	Pulse / Burst Bep, Time
1	-112	-150	2 ms. 10 Q	500 pulses	0.5 s
2a	55	112	0.05 ms 2.0	500 pulses	0.5 s
3a	-165	-220	0.1 μs. 50 Ω	1 h	100 ms
3b	112	150	0.1 μs, 50 Ω	1 h	100 ms
ISO 7637-2:2011	Test R	esults			
Test Pulse	III	IV			
1		Α			
2a		С			
3a		Α			
3b		Α			
Class	Functional Status				
A	All functions of a device perform as designed during and after exposure to disturbance.				
В	All functions of a device perform as designed during exposure. However, one or more of them can go beyond specified tolerance. All functions return automatically to within normal limits after exposure is removed. Memory functions shall remain class A.				
С	One or more functions of a device do not perform as designed during exposure but return automatically to normal operation after exposure is removed.				
D	One or mor operation u	One or more functions of a device do not perform as designed during exposure and do not return to normal operation until exposure is removed and the device is reset by simple "operator/use" action.			
Е	One or more functions of a device do not perform as designed during and after exposure and cannot be re- turned to proper operation without replacing the device.				

#### Table 13. ISO 7637-2: 2011(E) PULSE TEST RESULTS





#### **Protection Features**

In application, the device can be subject to stressful conditions which are outside of normal operating range. To prevent damage and destruction of the device from these fault conditions, several protection functions are integrated in device design. It is important to diagnose and remove any fault condition that may exist since the protection functions cannot prevent damage for continuous and repetitive faulty operation.

#### Loss of Ground Protection

When device or ECU ground connection is lost and load is still connected to ground, the device will turn the output OFF. In loss of ground state, the output stage is held OFF independent of the state of the input. Input resistors are recommended between the device and microcontroller.

#### **Undervoltage Protection**

The device has two under-voltage threshold levels,  $V_{D\_MIN}$  and  $V_{UV}$ . Switching function (ON/OFF) requires supply voltage to be at least  $V_{D\_MIN}$ . The device features a lower supply threshold  $V_{UV}$ , above which the output can remain in ON state. While all protection functions are guaranteed when the switch is ON, diagnostic functions are operational only within nominal supply voltage range  $V_{D}$ .



Figure 34. Undervoltage Behavior

#### **Overvoltage Protection:**

The NCV84090 has two Zener diodes  $Z_{VD}$  and  $Z_{CS}$ , which provide integrated overvoltage protection.  $Z_{VD}$ protects the logic block by clamping the voltage between supply pin  $V_D$  and ground pin GND to  $V_{ZVD}$ .  $Z_{CS}$  limits voltage at current sense pin CS to  $V_D - V_{ZCS}$ . The output power MOSFET's output clamping diodes provide protection by clamping the voltage across the MOSFET (between  $V_D$  pin and OUT pin) to  $V_{CLAMP}$ . During overvoltage protection, current flowing through  $Z_{VD}$ ,  $Z_{CS}$ and the output clamp must be limited. Load impedance  $Z_L$ limits the current in the body diode  $Z_{Body}$ . In order to limit the current in  $Z_{VD}$  a resistor,  $R_{GND}$  (150  $\Omega$ ), is required in the GND path. External resistors  $R_{CS}$  and  $R_{SENSE}$  limit the current flowing through  $Z_{CS}$  and out of the CS pin into the micro-controller I/O pin. With RGND, the GND pin voltage is elevated to  $V_D - V_{ZVD}$  when the supply voltage  $V_D$  rises above  $V_{ZVD}$ . ESD diodes  $Z_{ESD}$  pull up the voltage at logic pins IN, CS\_EN close to the GND pin voltage  $V_D - V_{ZVD}$ . External resistors  $R_{IN}$ , and  $R_{CS}$  are required to limit the current flowing out of the logic pins into the micro-controller I/O pins. During overvoltage exposure, the device transitions into a self-protection state, with automatic recovery after the supply voltage comes back to the normal operating range. The specified parameters as well as short circuit robustness and energy capability cannot be guaranteed during overvoltage exposure.

#### **Reverse Battery Protection**

Solution 1: Resistor in the GND line only (no parallel Diode)

The following calculations are true for any type of load. In the case for no diode in parallel with  $R_{GND}$ , the calculations below explain how to size the resistor.

Consider the following parameters:

 $-I_{GND}$  Maximum = 200 mA for up to  $-V_D$  = 32 V.

Where  $-I_{GND}$  is the DC reverse current through the GND pin and  $-V_D$  is the DC reverse battery voltage.

$$-I_{GND} = \frac{-V_D}{R_{GND}}$$
 (eq. 1)

Since this resistor can be used amongst multiple High–Side devices, please take note the sum of the maximum active GND currents ( $I_{GND(On)max}$ ) for each device when sizing the resistor. Please note that if the microprocessor GND is not shared by the device GND, then  $R_{GND}$  produces a shift of ( $I_{GND(On)max} \times R_{GND}$ ) in the input thresholds and CS output values. If the calculated power dissipation leads to too large of a resistor size or several devices have to share the same resistor, please look at the second solution for Reverse Battery Protection. Refer to Figure 35 for selecting the proper  $R_{GND}$ .

#### NCV84090 Reverse Battery Considerations Normal Operation VIN = 5 V, Reverse Battery = 32 V



Figure 35. Reverse Battery R<sub>GND</sub> Considerations

#### **Overload Protection**

Current limitation as well as overtemperature shutdown mechanisms are integrated into NCV84090 to provide protection from overload conditions such as bulb inrush or short to ground.

#### **Current Limitation**

In case of overload, NCV84090 limits the current in the output power MOSFET to a safe value. Due to high power dissipation during current limitation, the device's junction temperature increases rapidly. In order to protect the device, the output driver is shut down by one of the two overtemperature protection mechanisms. The output current limit is dependent on the device temperature, and will fold back once the die reaches thermal shutdown. If the input remains active during the shutdown, the output power MOSFET will automatically be re-activated after a minimum OFF time or when the junction temperature returns to a safe level.

#### **Output Clamping with Inductive Load Switch Off:**

The output voltage  $V_{OUT}$  drops below GND potential when switching off inductive loads. This is because the inductance develops a negative voltage across the load in response to a decaying current. The integrated clamp of the device clamps the negative output voltage to a certain level relative to the supply voltage  $V_{BAT}$ . During output clamping with inductive load switch off, the energy stored in the inductance rapidly dissipated in the device resulting in high power dissipation. This is a stressful condition for the device and the maximum energy allowed for a given load inductance should not be exceeded in any application.



Figure 37. Maximum Switch–Off Current vs. Load Inductance, V<sub>D</sub> = 13.5 V; R<sub>L</sub> = 0  $\Omega$ 

#### Inverse Current:

When the output voltage  $V_{OUT}$  rises above the supply voltage  $V_D$ , the output power MOSFET's integral body diode will be forward biased causing a current flow from the OUT pin to the  $V_D$  pin. The device does not provide any protection function such as current limitation or overtemperature shutdown.

#### Underload Detection in ON State

An underload condition in ON state is indicated by reducing the sense output current to a very minimal current. In order to detect an underload condition, NCV84090 performs a real-time monitoring of the load current. In case the output current falls below a specified threshold level  $(I_{OL})$ , the current sense output current is reduced to a very low value  $(I_{OL})$ . This mechanism helps to overcome a high absolute tolerance of the current sense signal at very low load current and to implement an accurate underload detection threshold.

#### **Open Load Detection in OFF State**

Open load diagnosis in OFF state can be performed by activating an external resistive pull–up path ( $R_{PU}$ ) to  $V_{BAT}$ . To calculate the pull–up resistance, external leakage currents (designed pull–down resistance, humidity–induced leakage etc) as well as the open load threshold voltage  $V_{OL\ OFF}$  have to be taken into account.



Figure 38. Open Load Detection in Off State

#### **Current Sense in PWM Mode**

When operating in PWM mode, the current sense functionality can be used, but the timing of the input signal and the response time of the current sense need to be considered. When operating in PWM mode, the following performance is to be expected. The CS\_EN pin should be left high to eliminate any unnecessary delay time to the circuit. When VIN switches from low to high, there will be a typical delay (tCS\_High2) before the current sense responds. Once this timing delay has passed, the rise time of the current sense output ( $\Delta tCS_High2$ ) also needs to be considered. When VIN switches from high to low a delay time (tCS\_Low1) needs to be considered. As long as these timing delays are allowed, the current sense pin can be operated in PWM mode.

#### PACKAGE AND PCB THERMAL DATA



Figure 39. Junction-to-Ambient Transient Thermal Impedance (645 mm<sup>2</sup> Cu Area)



Figure 40. Junction-to-Ambient Transient Thermal Impedance (2 cm<sup>2</sup> Cu Area)

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\*For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

#### STYLES ON PAGE 2

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#### SOIC-8 NB CASE 751-07 **ISSUE AK**

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STYLE 2: PIN 1. COLLECTOR, DIE, #1 2. COLLECTOR, #1 COLLECTOR, #2 3. 4 COLLECTOR, #2 BASE, #2 5. EMITTER, #2 6. 7 BASE #1 EMITTER, #1 8. STYLE 6: PIN 1. SOURCE 2. DRAIN 3. DRAIN SOURCE 4. SOURCE 5. 6. GATE GATE 7. 8. SOURCE STYLE 10: GROUND PIN 1. BIAS 1 OUTPUT 2. З. GROUND 4. 5. GROUND 6 BIAS 2 INPUT 7. 8. GROUND STYLE 14: PIN 1. N-SOURCE 2. N-GATE 3 P-SOURCE P-GATE 4. P-DRAIN 5 6. P-DRAIN N-DRAIN 7. N-DRAIN 8. STYLE 18: PIN 1. ANODE ANODE 2. SOURCE 3. GATE 4. 5. DRAIN 6 DRAIN CATHODE 7. CATHODE 8. STYLE 22 PIN 1. I/O LINE 1 2. COMMON CATHODE/VCC 3 COMMON CATHODE/VCC 4. I/O LINE 3 COMMON ANODE/GND 5. 6. I/O LINE 4 7. I/O LINE 5 8. COMMON ANODE/GND STYLE 26: PIN 1. GND 2 dv/dt З. ENABLE 4. ILIMIT 5. SOURCE SOURCE 6. SOURCE 7. 8. VCC STYLE 30: DRAIN 1 PIN 1. DRAIN 1 2 GATE 2 З. SOURCE 2 4 SOURCE 1/DRAIN 2 SOURCE 1/DRAIN 2 5.

6.

7.

8 GATE 1

SOURCE 1/DRAIN 2

STYLE 3: PIN 1. DRAIN, DIE #1 DRAIN, #1 2. DRAIN, #2 З. DRAIN, #2 4. GATE, #2 5. SOURCE, #2 6. 7 GATE #1 8. SOURCE, #1 STYLE 7: PIN 1. INPUT 2. EXTERNAL BYPASS THIRD STAGE SOURCE GROUND З. 4. 5. DRAIN 6. GATE 3 SECOND STAGE Vd 7. FIRST STAGE Vd 8. STYLE 11: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. З. GATE 2 4. 5. DRAIN 2 6. DRAIN 2 DRAIN 1 7. 8. DRAIN 1 STYLE 15: PIN 1. ANODE 1 2. ANODE 1 ANODE 1 3 ANODE 1 4. 5. CATHODE, COMMON CATHODE, COMMON CATHODE, COMMON 6. 7. CATHODE, COMMON 8. STYLE 19: PIN 1. SOURCE 1 GATE 1 SOURCE 2 2. 3. GATE 2 4. 5. DRAIN 2 6. MIRROR 2 7. DRAIN 1 MIRROR 1 8. STYLE 23: PIN 1. LINE 1 IN COMMON ANODE/GND COMMON ANODE/GND 2. 3 LINE 2 IN 4. LINE 2 OUT 5. COMMON ANODE/GND COMMON ANODE/GND 6. 7. 8. LINE 1 OUT STYLE 27: PIN 1. ILIMIT OVI O 2 UVLO З. 4. INPUT+ 5. 6. SOURCE SOURCE SOURCE 7. 8 DRAIN

#### DATE 16 FEB 2011

STYLE 4: PIN 1. 2. ANODE ANODE ANODE З. 4. ANODE ANODE 5. 6. ANODE 7 ANODE COMMON CATHODE 8. STYLE 8: PIN 1. COLLECTOR, DIE #1 2. BASE, #1 З. BASE #2 COLLECTOR, #2 4. COLLECTOR, #2 5. 6. EMITTER, #2 EMITTER, #1 7. 8. COLLECTOR, #1 STYLE 12: PIN 1. SOURCE SOURCE 2. 3. GATE 4. 5. DRAIN 6 DRAIN DRAIN 7. 8. DRAIN STYLE 16 EMITTER, DIE #1 PIN 1. 2. BASE, DIE #1 EMITTER, DIE #2 3 BASE, DIE #2 4. 5. COLLECTOR, DIE #2 6. COLLECTOR, DIE #2 COLLECTOR, DIE #1 7. COLLECTOR, DIE #1 8. STYLE 20: PIN 1. SOURCE (N) GATE (N) SOURCE (P) 2. 3. 4. GATE (P) 5. DRAIN 6. DRAIN DRAIN 7. 8. DRAIN STYLE 24: PIN 1. BASE EMITTER 2. 3 COLLECTOR/ANODE COLLECTOR/ANODE 4. 5. CATHODE 6. CATHODE COLLECTOR/ANODE 7. 8. COLLECTOR/ANODE STYLE 28: PIN 1. SW\_TO\_GND 2. DASIC OFF DASIC\_SW\_DET З. 4. GND 5. 6. V MON VBULK 7. VBULK 8 VIN

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7.

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COLLECTOR, #1

COLLECTOR, #1

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