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FLS0116

内置MOSFET具有功率因数校正功能的智能 LED 灯驱动器集成电路

特性

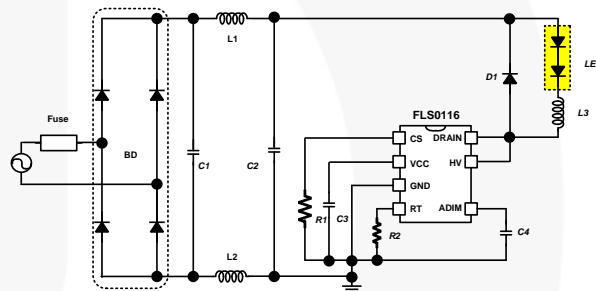
- 内置 MOSFET (1 A / 550 V)
- 数字实施有源PFC 功能
- 无达到高PF的额外电路
- 应用输入范围: 80 V_{AC} ~ 308 V_{AC}
- 内置HV电源电路: 自偏置
- 支持自动重启模式的AOCV功能
- 内置过温保护 (OTP)
- 逐周期限流
- 电流传感引脚开路保护
- 低工作电流: 0.85 mA (典型值)
- 带有 5 V 滞回的欠压锁定
- 可编程 振荡频率
- 可编程 LED电流
- 模拟调光功能
- 软启动功能
- 精确的内部参考: ±3%

应用

- 装饰照明用 LED 灯
- 低能耗照明器材用 LED 灯

说明

FLS0116 LED 灯驱动器是一个具有集成 MOSFET 功能和 PFC 功能的简单 IC。特殊的“已采用数字”技术可自动检测输入电压情况,并发送内部参考信号来实现高功率因数。当AC输入应用于IC时,系统会自动启用PFC功能。当DC输入应用于IC时,系统会自动禁用PFC功能。FLS0116 不需要大电解电容来维持供电的稳定性,这能有效延长 LED 灯寿命。



□ 1. 典型应用

订购信息

器件编号	工作温度范围	封装	包装方法
FLS0116MX	-40° C to +125° C	7 引脚小型封装集成电路 (SOIC), JEDEC MS-012, .150 英寸, 窄体	卷带和卷盘

框图

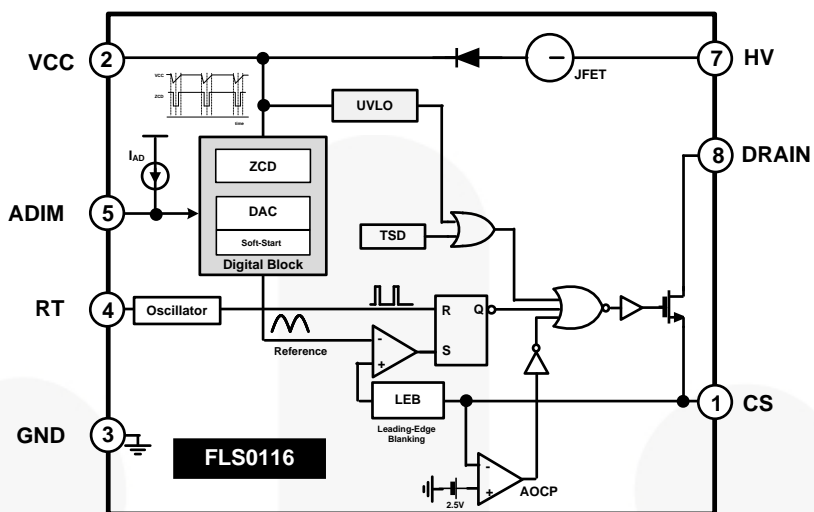


图2. 框图

引脚布局

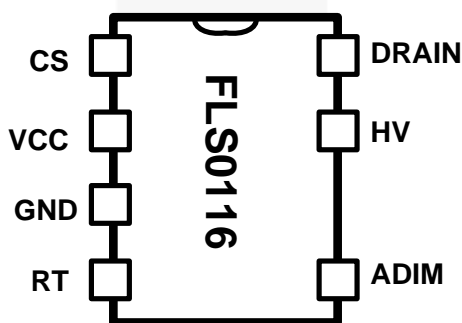


图3. 引脚布局

引脚定义

引脚号	名称	说明
1	CS	电流检测。 依靠检测电阻电压限制输出电流 CS引脚还用于设置 LED 电流调节
2	VCC	VCC. 电源引脚用于集成电路稳定运行；ZCD 信号检测用于精确的 PFC 功能。
3	GND	接地。 IC接地
4	RT	RT. 通过外部电阻可编程工作频率；当此引脚开路或悬空时，IC工作于预设固定的频率。
5	ADIM	模拟调光。 连接到内部电流源。通过外部电阻改变输出电流。如果 ADIM 没有使用，在 ADIM 和 GND 之间连接一个 0.1 μ F 的旁路电容器。
7	HV	高压。 连接高压线，给IC提供电流。
8	DRAIN	DRAIN. 内部 MOSFET 漏极引脚

绝对最大额定值

应力超过绝对最大额定值，可能会损坏器件。在超出推荐的工作条件的情况下，该器件可能无法正常工作，所以不建议让器件在这些条件下长期工作。此外，过度暴露在高于推荐的工作条件下，会影响器件的可靠性。绝对最大额定值仅是应力规格值。

符号	参数	最小值	最大值	单位
V _{CC}	IC 电源电压		20	V
HV	高压检测		550	V
DRAIN	内部漏极电压		550	V
V _{ADIM}	模拟调光		5	V
V _{RT}	设置频率引脚电压		5	V
V _{CS}	容许的电流检测电压		5	V
T _A	工作环境温度范围	-40	+125	°C
T _J	工作结温	-40	+150	°C
T _{STG}	存储温度范围	-65	+150	°C
θ _{JA}	结至环境热阻 ^(1,2)		135	°C/W
P _D	功耗		660	mW
ESD	静电放电能力	人体放电模式, JESD22-A114	2000	V
		组件充电模式, JESD22-C101	1000	

注意:

- 热阻测试板。尺寸: 76.2 mm x 114.3 mm x 1.6 mm (1SOP); JEDEC 标准: JESD51-2, JESD51-3.
- 假设无环境气流。

电气特性

典型值为 $T_A = +25^\circ\text{C}$ 时的值。 $-40^\circ\text{C} \sim 125^\circ\text{C}$ 的参数规格值由设计保证。

符号	参数	条件	最小值	典型值	最大值	单位
V_{CC}偏置部分						
V _{CC}	V _{CC} 稳压器输出电压	V _{HV} =100 V _{DC}	14.0	15.5	17.0	V
V _{CCST+}	UVLO 正向阈值	V _{CC} 增加	12	13	14	V
V _{CCST-}	UVLO 负向阈值	V _{CC} 降低	7	8	9	V
V _{CCHYS}	UVLO 滞回		4	5	6	V
I _{HV}	高压引脚电流	V _{HV} = 100 V _{DC} , RT = 开路		0.85	1.20	mA
I _{ST}	启动电流			120	150	μA
开关部分						
f _{OSC}	工作频率	R _T =5.95 kΩ	200	250	300	kHz
		R _T =87 kΩ	16	20	24	kHz
		R _T Open	40.5	45.0	49.5	kHz
t _{MIN}	最短导通时间 ⁽³⁾		400		ns	
D _{MAX}	最大占空比			50		%
t _{LEB}	前沿消隐时间 ⁽³⁾			350		ns
V _{RT}	RT引脚基准电压			1.5		V
软启动部分						
t _{SS}	软启动时间 ⁽³⁾	直流电模式	48	60	72	ms
		交流电模式		7		周期
基准部分						
V _{CS1}	CS引脚内部基准电压	直流电模式	0.354	0.365	0.376	V
V _{CS2}		交流电模式 ⁽³⁾	0.485	0.500	0.515	
保护部分						
OVP _{VCC}	VCC 引脚过压保护		17.7	18.7	19.7	V
V _{AACP}	CS引脚异常过流保护电位 ⁽³⁾			2.5		V
t _{AACP}	异常检测时间 ⁽³⁾			70		ns
T _{TSDH}	热关断阈值 ⁽³⁾		140	150		°C
T _{TSDHY}	热关断阈值滞回 ⁽³⁾			50		°C
调光部分						
V _{ADIM(ST+)}	模拟调光正向阈值 ⁽³⁾		3.15	3.50	3.85	V
V _{ADIM(ST-)}	模拟调光负向阈值 ⁽³⁾			0.50	0.75	V
I _{AD}	ADIM 引脚内部电流源		9	12	15	μA

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电气特性 (续)

典型值为 $T_A = +25^\circ\text{C}$ 时的值。 $-40^\circ\text{C} \sim 125^\circ\text{C}$ 的参数规格值由设计保证。

符号	参数	条件	最小值	典型值	最大值	单位
MOSFET 部分						
BV_{DSS}	击穿电压	$V_{GS}=0\text{ V}, I_D=250\ \mu\text{A}$	550			V
I_{LKMOS}	内部 MOSFET 泄漏电流	$V_{DS}=550\text{ V}_{DC}, V_{GS}=0\text{ V}$			250	μA
$R_{ON(ON)}$	漏源电阻 ⁽³⁾	$V_{GS}=10\text{ V}, V_{DS}=0\text{ V}, T_C=25^\circ\text{C}$		7.3	10.0	Ω
C_{ISS}	输入电容 ⁽³⁾	$V_{GS}=0\text{ V}, V_{DS}=25\text{ V}, f=1\text{ MHz}$		135		pF
C_{OSS}	输出电容 ⁽³⁾	$V_{GS}=0\text{ V}, V_{DS}=25\text{ V}, f=1\text{ MHz}$		21		pF
C_{RSS}	反向传输电容 ⁽³⁾	$V_{GS}=0\text{ V}, V_{DS}=25\text{ V}, f=1\text{ MHz}$		3.2		pF
$t_{d(ON)}$	导通延时 ⁽³⁾	$V_{DD}=350\text{ V}, I_D=1\text{ A}$		10		ns
t_r	上升时间 ⁽³⁾	$V_{DD}=350\text{ V}, I_D=1\text{ A}$		13.4		ns
$t_{d(OFF)}$	关断延时 ⁽³⁾	$V_{DD}=350\text{ V}, I_D=1\text{ A}$		14.9		ns
t_f	下降时间 ⁽³⁾	$V_{DD}=350\text{ V}, I_D=1\text{ A}$		36.8		ns

注:

3. 该参数由设计保证; 未经 100% 产品测试。

功能说明

FLS0116 是一种采用降压转换拓扑结构的基本PWM控制器，工作于连续导通模式 (CCM)，通过数字控制算法实现的智能功率因数校正功能。一个使用高压开关器件的内部自偏压电路。该IC不像典型反激控制集成电路或 PSR 产品系列需要辅助电源连接至 VCC 引脚。

当施加在 HV 引脚上的输入电压在工作范围内 (25 V 至 500 V) 时，FLS0116 在 VCC 引脚保持 15.5 V 的直流电压以维持稳定运行。UVLO 模块功能是，当 V_{CC} 电压高于 V_{CCST+} 时，内部 UVLO 模块启动，开始运行。否则，当 V_{CC} 降低至 V_{CCST-} 时，集成电路停止运行。通常来说，即使输入电压处于干扰极大或不稳定的情况下运行，滞后特性也可以保证稳定运行。

FLS0116 拥有“智能”内部数字单元来确定输入情况：交流或直流。当 50 Hz 或 60 Hz 的交流电源应用于 IC 时，IC 会自动转变内部基准信号，使其与输入信号相近，从而实现高功率因数。当直流电源连接集成电路时，内部基准会直接转变为直流。

软启动功能

FLS0116 具有内部软启动功能，以降低启动时的冲击电流。当 IC 依照内部时序开始工作时，内部基准会以预设的固定时间缓慢增加。在这个瞬态时期后，内部基准会到达一个稳定水平。此时，IC 不断试图从 VCC 引脚中获取相位信息。如果 IC 成功获取相位信息，它会自动在七个周期的过渡时间中产生的相似波形的参考电压。如果没能获取，IC 生成直流基准水平。

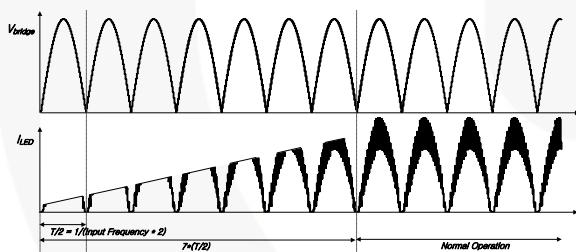


图4. 交流输入模式中的软启动功能

内部功率因数校正功能：如何实现高功率因数

FLS0116 具有一个简单“智能”的内部功率因数校正功能，无需额外的引脚来检测输入相位信息，也无需电解电容器来保证电源电压稳定。要实现高功率因数，FLS0116 在桥式二极管后不使用整流电容器 这是至关重要的，因为 IC 采用检测 VCC 引脚上的信号波动。基本上来说，为 IC 供电的 VCC 引脚具有和经桥式整流后的整流电压一样的电压纹波，电平值根据 V_{CC} 电容值不同而改

变。利用 VCC 引脚上的这种电压波动，IC 可以检测时间参考值，产生内部零电流检测信号。

为获得输入电压信号的精确可靠内部参考电压，FLS0116 使用数字技术（西格玛/增量调制）产生新的内部信号 (DAC_OUT)，该信号与输入电压具有相同的相位，如图 5 所示。该信号进入最终比较器，与来自检测电阻的电流信号进行比较。

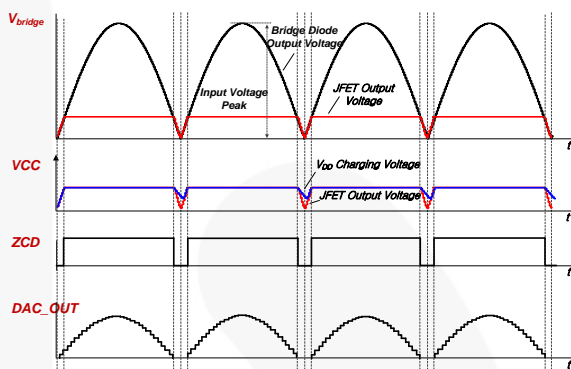


图5. 内部功率因数校正功能

自偏压功能

自偏压功能使用高压器件，可以向 IC 供应足够的工作电流，确保在整个输入电压范围内 (80 V~308 V_{AC}) 具有近似的启动时间。然而，在高压条件下，自偏压功能存在一个缺点。高压器件通常用作恒流源，因此当高输入电压连接高压引脚时，内部高压器件会有功率损耗。这些功率损耗与输入电压成正比。为了降低功率损耗，一个可行的解决办法是在输入电压源和高压引脚间额外安置一个电阻，如图所示图 6。

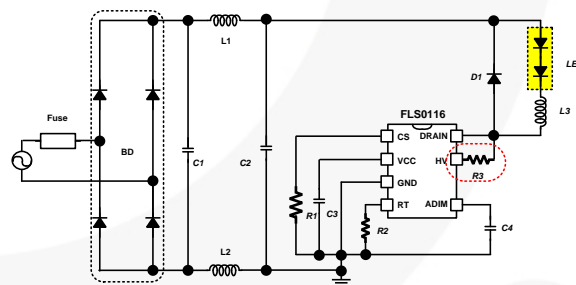


图6. 高压应用

调光功能

FLS0116 使用 ADIM 引脚来模拟或者使用电阻分压器在 0 V 至 10 V 间进行调光。内部参考电压的峰值（即图5DAC_OUT 信号），会被 V_{ADIM} 水平而改变，如图所示图7，根据不同的运行模式，有不同的峰值水平。

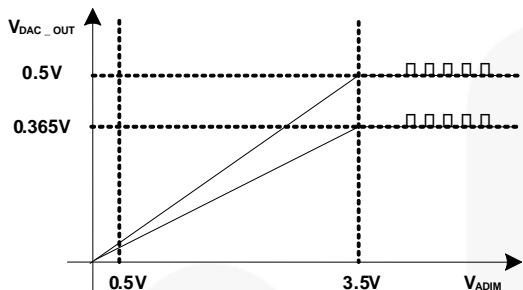


图7. V_{ADIM} 与 $V_{DAC_OUT(peak)}$ 对比

电感设计

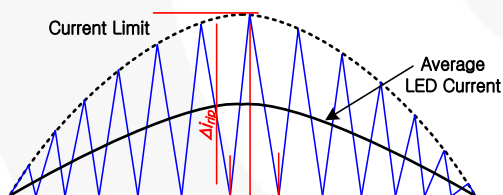
固定内部最大占空比低于 50%，最小导通时间为 400 ns。该范围取决于输入电压和串联 LED 数量。

最低占空比计算为：

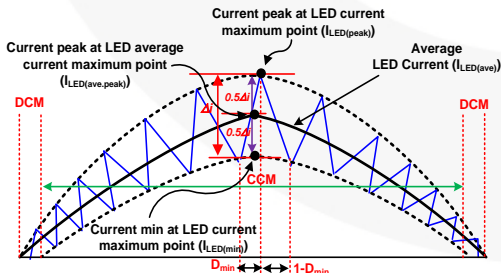
$$D_{min} = \frac{n \cdot V_f}{\eta \cdot V_{in(max)}} \quad (1)$$

其中，

- η = 系统效率；
- $V_{in(max)}$ = 最大输入电压；
- V_f = LED正向压降；且
- n = 串联 LED 数量。



(a) DCM Mode



(b) CCM Mode

图8.DCM 和 CCM 运行

在 DCM 模式下，电感为：

$$L_m = \frac{n \cdot V_f \cdot (1 - D_{min})}{f_s \cdot \Delta i_{rip}} [H] \quad (2)$$

如果峰值电流固定在 350 mA_{pk}，则峰值电流的计算公式为：

$$I_{LED(ave,peak)} = \Delta i_{con} + \frac{\Delta i_{rip}}{2} [A] \quad (3)$$

对于 FL7701，LED RMS 电流确定电感参数。为了能够获得 CCM 模式，需要首先确定 LED RMS 电流：

$$I_{LED(rms)} = \frac{I_{LED(ave,peak)}}{\sqrt{2}} [A] \quad (4)$$

将方程式 (2) 替换为方程式 (4)，可得电感量。

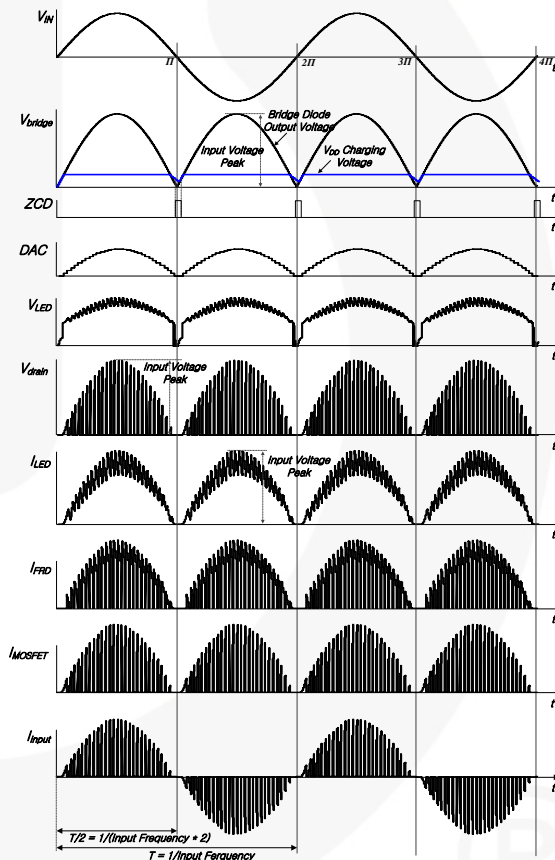


图9. 典型性能特征

应用电路实例

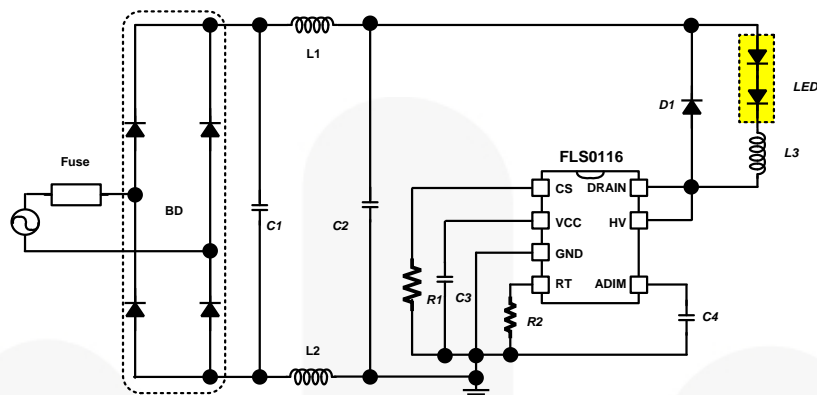


图10. 无电解电容器的应用电路

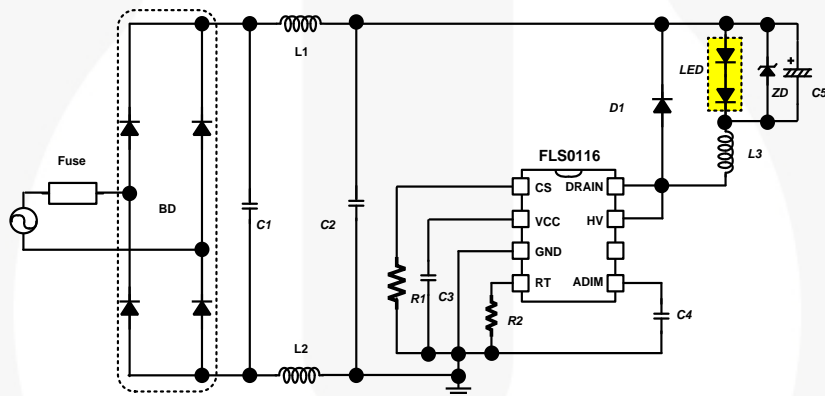


图11. 有电解电容器的应用电路

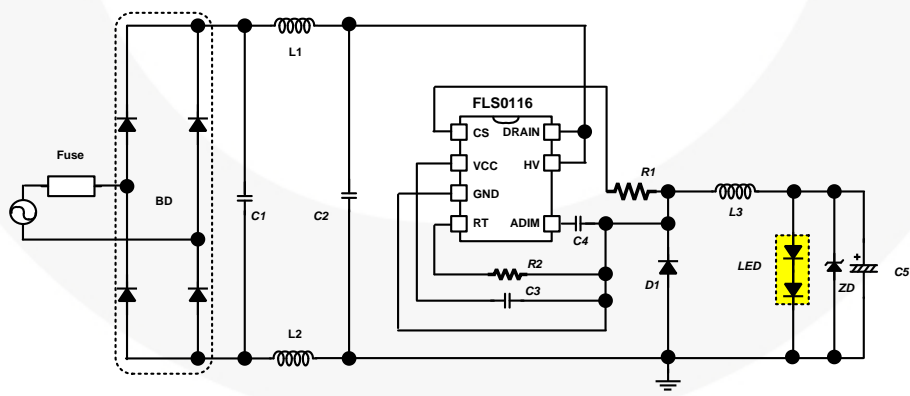


图12. 有电解电容器的高侧运行应用电路

典型特性

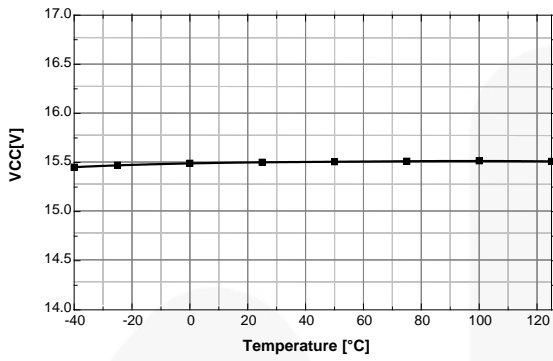


图13. V_{CC}与温度的关系

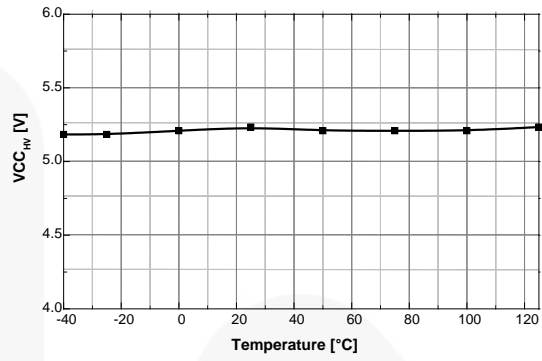


图14. V_{CC_{mv}}与温度的关系

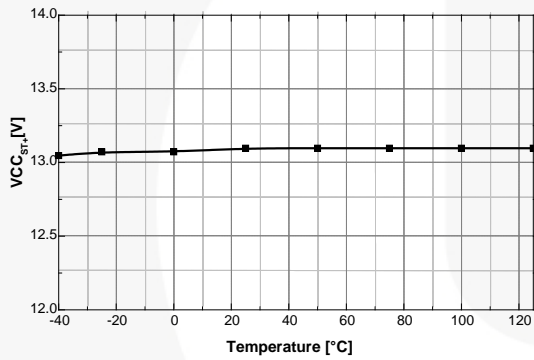


图15. V_{CC_{st+}}与温度的关系

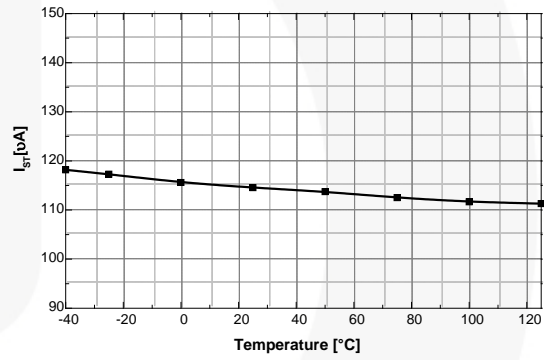


图16. I_{st}与温度的关系

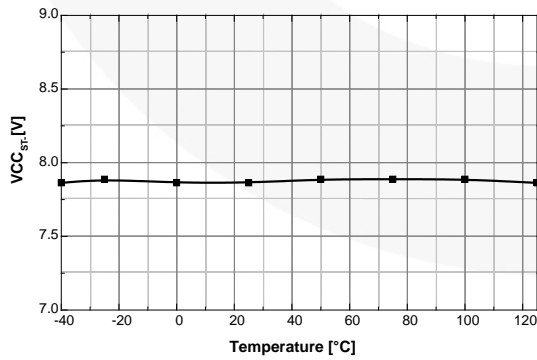


图17. V_{CC_{st-}}与温度的关系

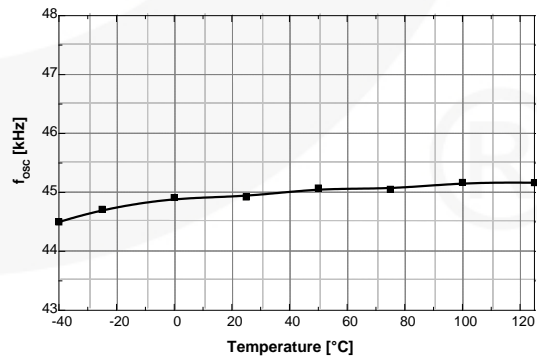


图18. f_{osc}与温度的关系 (RT= 开路)

典型特性

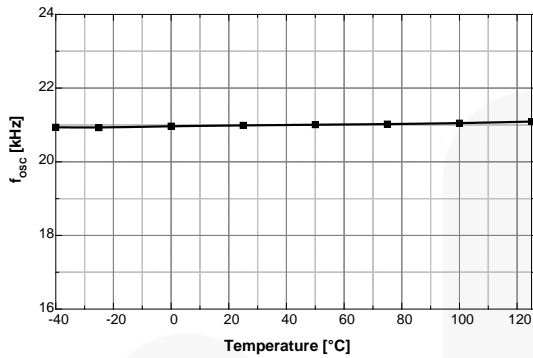


图19. f_{osc} 与温度的关系 (RT=87k Ω)

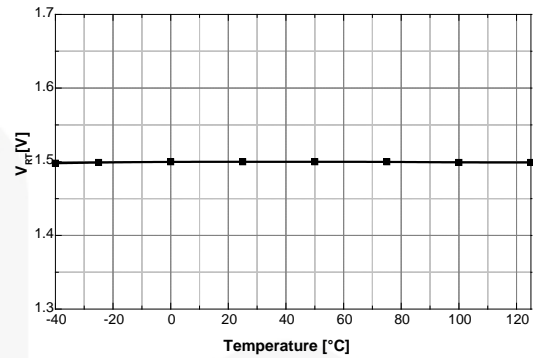


图20. V_{RT} 与温度的关系

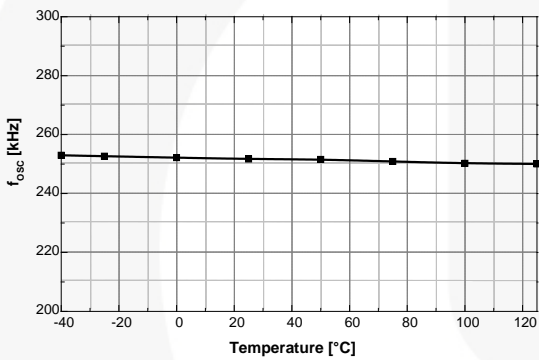


图21. f_{osc} 与温度的关系 (RT=5.95k Ω)

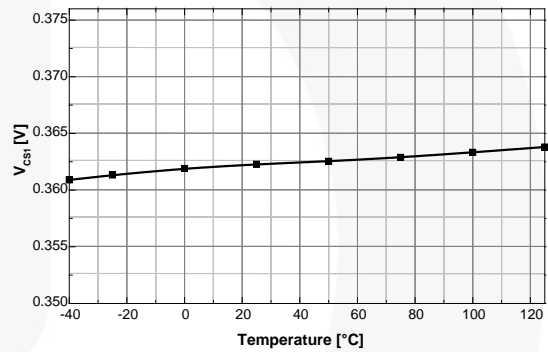


图22. V_{CS} 与温度的关系

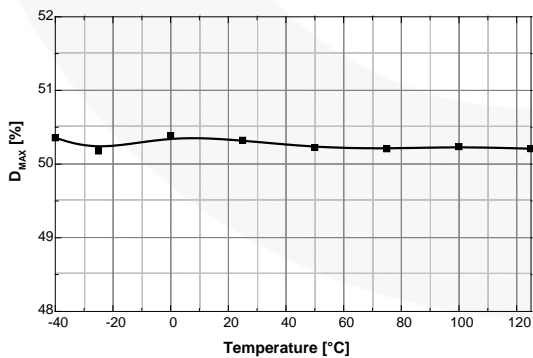


图23. D_{MAX} 与温度的关系

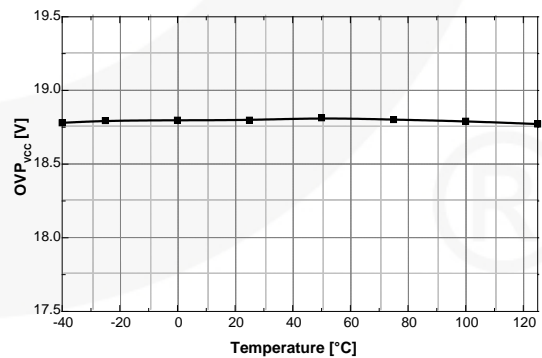


图24. OVP_{VCC} 与温度的关系

典型特性

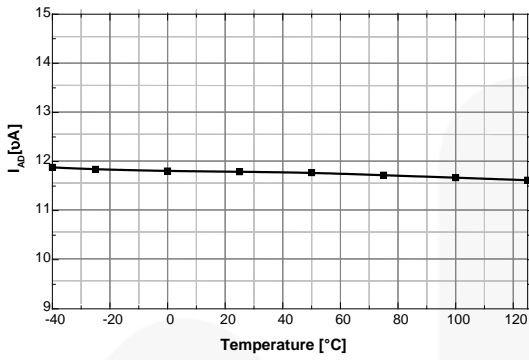


图25. I_{AD} 与温度的关系

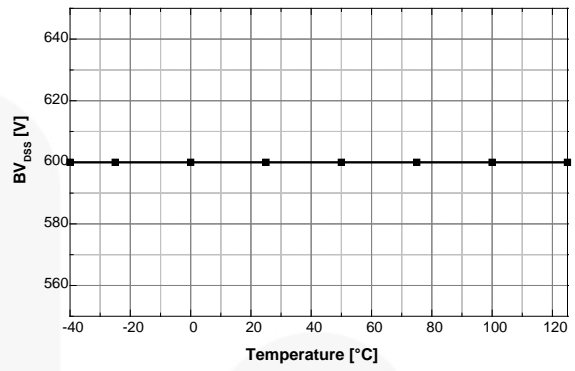


图26. BV_{DSS} 与温度的关系

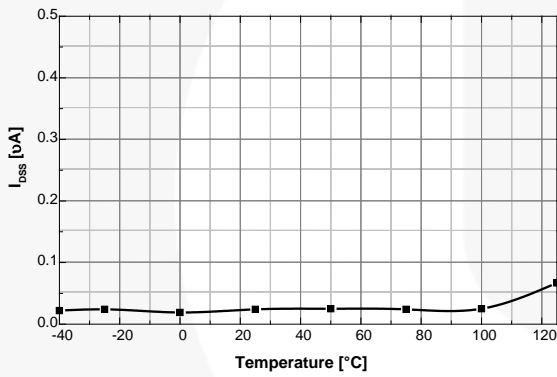


图27. I_{DSS} 与温度的关系



物理尺寸

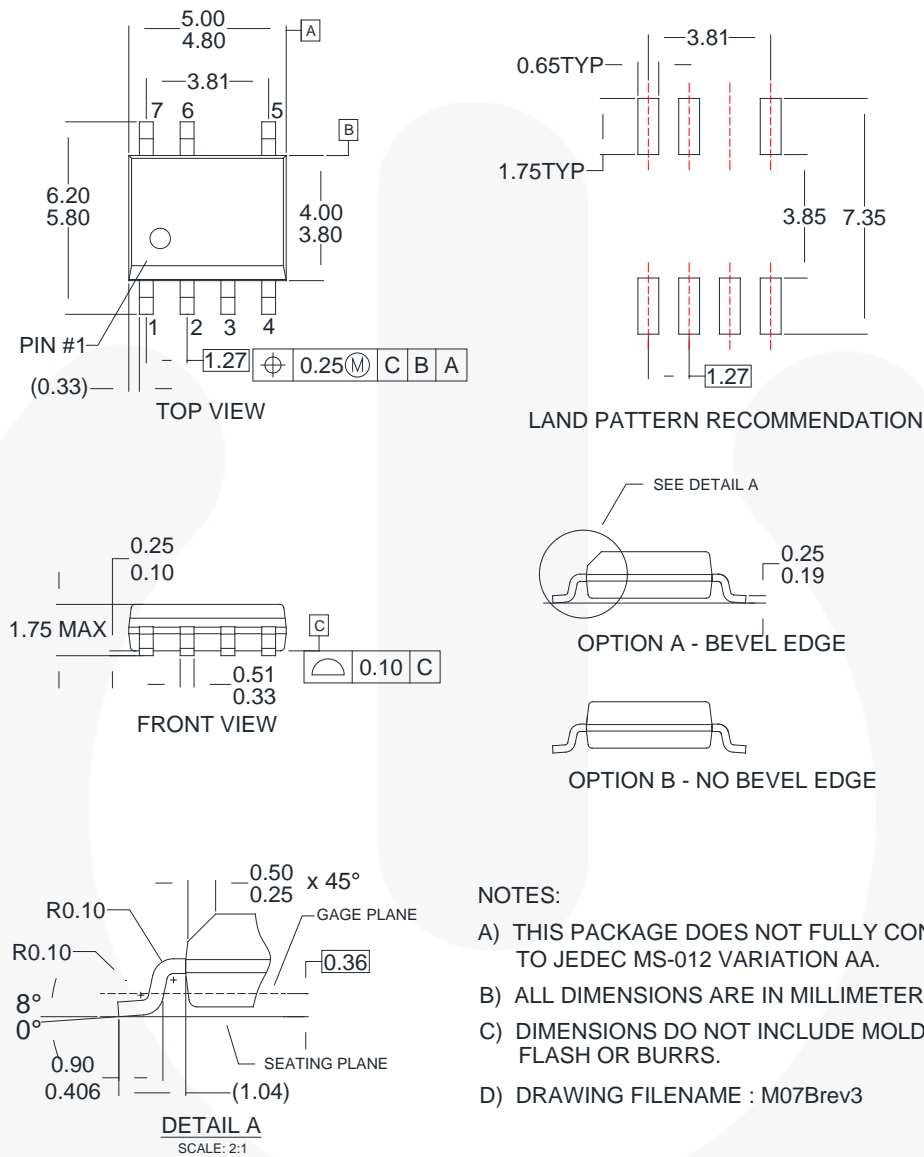


图28. 7 引线小型封装集成电路 (SOIC), JEDEC MS-012, .150 英寸窄体

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