# SGM5532 **Dual Low-Noise Operational Amplifier**

# GENERAL DESCRIPTION

The SGM5532 is a high-performance operational amplifier combining excellent DC and AC characteristics. It features very low noise, high output-drive capability, high unity-gain and maximum output-swing bandwidths, low distortion, high slew rate, input protection diodes and output short-circuit protection. The operational amplifier is compensated internally for unity-gain operation.

The SGM5532 is available in a Green SOIC-8 package. It operates over an ambient temperature range of -40°C to +85°C.

## **FEATURES**

- Equivalent Input Noise Voltage:
  5nV/√Hz (TYP) at 1kHz
- Unity-Gain Bandwidth: 8.5MHz (TYP)
- Common Mode Rejection Ratio: 140dB (TYP)
- High DC Voltage Gain: 140dB (TYP)
- High Slew Rate: 18V/µs (TYP)
- -40°C to +85°C Operating Temperature Range
- Available in a Green SOIC-8 Package

# **APPLICATIONS**

**AV Receivers** 

**Embedded PCs** 

Netbooks

Video Broadcasting and Infrastructure: Scalable Platforms

**DVD Recorders and Players** 

Multichannel Video Transcoders

Pro Audio Mixers

# SIMPLIFIED SCHEMATIC

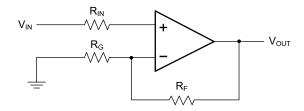
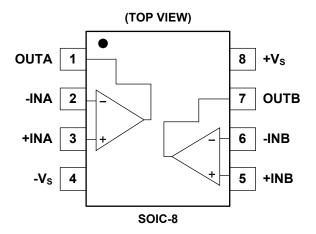


Figure 1. Simplified Schematic

# PIN CONFIGURATION



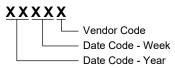
# PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION	
SGM5532	SOIC-8	-40°C to +85°C	SGM5532YS8G/TR	SGM 5532YS8 XXXXX	Tape and Reel, 4000	

NOTE: XXXXX = Date Code and Vendor Code.

#### MARKING INFORMATION

NOTE: XXXXX = Date Code and Vendor Code.



Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, +V <sub>S</sub> to -V <sub>S</sub>	40V
Input Common Mode Voltage Range	
(-V <sub>S</sub> ) - 0.3V to	$(+V_S) + 0.3V$
Junction Temperature	+150°C
Storage Temperature Range65°	°C to +150°C
Lead Temperature (Soldering, 10s)	+260°C
ESD Susceptibility	
HBM	5000V
MM	200V
CDM	1000V

#### RECOMMENDED OPERATING CONDITIONS

Supply Voltage, +V <sub>S</sub> to -V <sub>S</sub>	5V to 36V
Operating Temperature Range40	0°C to +85°C

#### **OVERSTRESS CAUTION**

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods

may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

#### **ESD SENSITIVITY CAUTION**

This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

#### **DISCLAIMER**

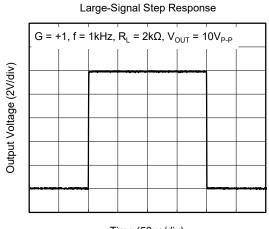
SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

# **ELECTRICAL CHARACTERISTICS**

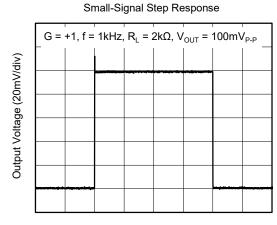
(At  $T_A$  = +25°C, Full = -40°C to +85°C,  $V_S$  = ±15V,  $R_L$  = 2k $\Omega$  connected to 0V, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS	
Input Characteristics								
			+25°C		100	500	1	
Input Offset Voltage	Vos	$V_{CM} = 0V$	Full			620	μV	
Input Offset Voltage Drift	ΔV <sub>OS</sub> /ΔT		Full		0.6		μV/°C	
Janua Dias Cumant		\( - 0\)	+25°C		550	750	nA	
Input Bias Current	I <sub>B</sub>	V <sub>CM</sub> = 0V	Full			900		
land Office Comment		V - 0V	+25°C		10	70	nA	
Input Offset Current	los	V <sub>CM</sub> = 0V	Full			100		
Input Common Mode Voltage Range	V <sub>CM</sub>		Full	-13		13	V	
Common Mode Rejection Ratio	CMRR	$V_S = \pm 15V$ , -13V < $V_{CM}$ < 13V	+25°C	128	140		dB	
Common wode Rejection Ratio	CIVILLIA	VS - 113V, -13V \ VCM \ 13V	Full	124			uБ	
		$V_S = \pm 15V, V_{OUT} = \pm 10V, R_L = 2k\Omega$	+25°C	128	140			
Open-Loop Voltage Gain	Δ	V <sub>S</sub> - ±13V, V <sub>OUT</sub> - ±10V, IN <sub>L</sub> - 2K12	Full	120			dB	
Open-Loop voltage Gain	A <sub>OL</sub>	$V_S = \pm 15V$ , $V_{OUT} = \pm 10V$ , $R_L = 600\Omega$	+25°C	112	128		_ ub	
		V <sub>S</sub> - ±13V, V <sub>OUT</sub> - ±10V, I <sub>L</sub> - 000Ω	Full	108			]	
Output Characteristics								
		$V_S = \pm 15V$ , $R_L = 2k\Omega$	+25°C		150	185	mV	
Output Voltage Swing from Rail	V <sub>out</sub>	VS - 113V, INL - 2K12	Full			230		
		$V_S = \pm 15V, R_L = 600\Omega$	+25°C		550	660		
			Full			840		
Output Short-Circuit Current	I <sub>SC</sub>	V <sub>S</sub> = ±15V	+25°C	±27	±36		mA	
Power Supply	_							
Operating Voltage Range	Vs		Full	5		36	V	
Quiescent Current	ΙQ	I <sub>OUT</sub> = 0	+25°C		8.5	17.5	mA	
Quicocont Gunont			Full			18		
Power Supply Rejection Ratio	PSRR	V <sub>S</sub> = 5V to 36V	+25°C	122	138		dB	
Tomor Supply Hojesdon Hadis		15 01 10 001	Full	119			45	
Dynamic Performance	_	,					_	
Gain-Bandwidth Product	GBP	f = 10kHz			20		MHz	
Slew Rate	SR				18		V/µs	
Overload Recovery Time	ORT	$V_{IN} \times G = V_{S}$			1.2		μs	
Maximum Output-Swing Bandwidth	B <sub>OM</sub>	$V_S = \pm 15V, V_{OUT} = \pm 10V, R_L = 600\Omega$			280		kHz	
Unity-Gain Bandwidth	B <sub>1</sub>	$R_L = 600\Omega$			8.5		MHz	
Total Harmonic Distortion + Noise	THD+N	$V_S = \pm 15V$ , $V_{OUT} = 10V_{P-P}$ , $f = 1kHz$ , $G = +1$ , $R_L = 600\Omega$			0.00005		%	
Noise	•		•		•		•	
Input Voltage Noise		f = 0.1Hz to 10Hz			0.3		μV <sub>P-P</sub>	
Input Voltage Naige Density		f = 30Hz			15		n\// ==	
Input Voltage Noise Density	e <sub>n</sub>	f = 1kHz			5		mV/ √HZ	
Input Current Noise Density	:	f = 30Hz			3		pA/ √HZ	
Impar Current Noise Density	i <sub>n</sub>	f = 1kHz			1		Prv √HZ	

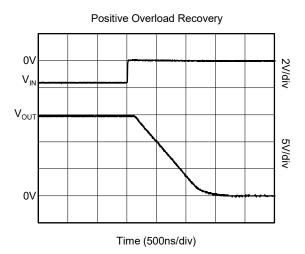
# TYPICAL PERFORMANCE CHARACTERISTICS



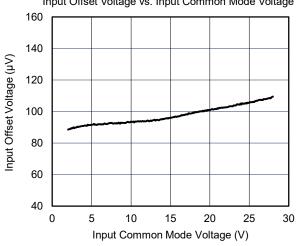




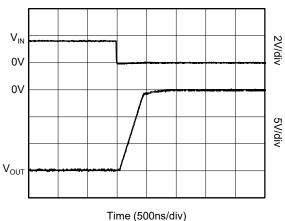
Time (50µs/div)



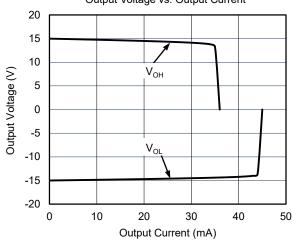
Input Offset Voltage vs. Input Common Mode Voltage



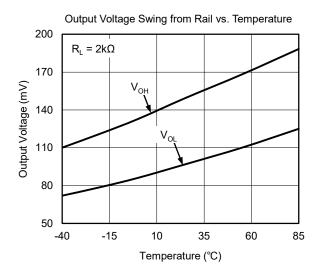
Negative Overload Recovery

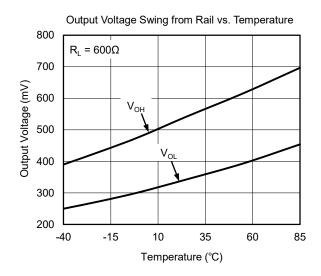


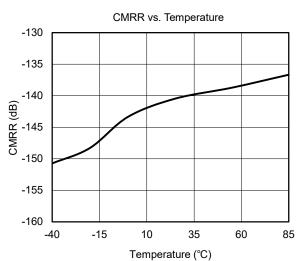
Output Voltage vs. Output Current

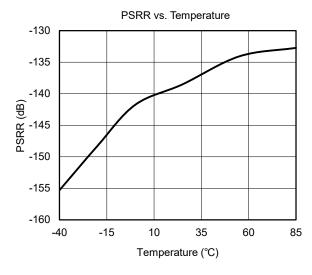


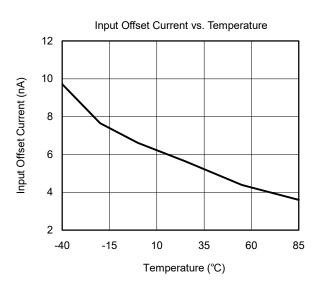
# **TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

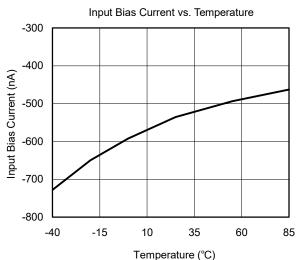




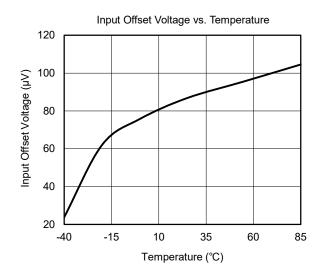


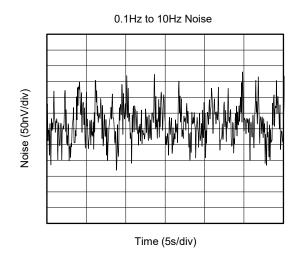


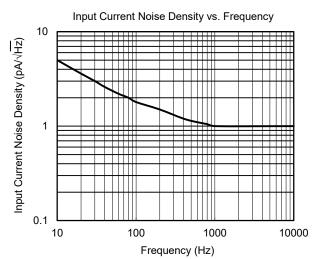


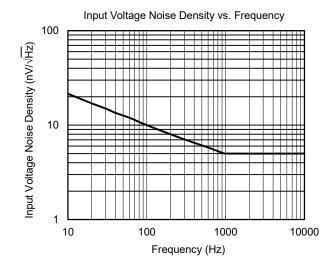


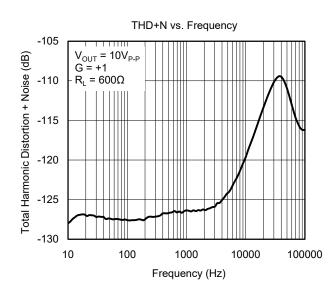
# **TYPICAL PERFORMANCE CHARACTERISTICS (continued)**

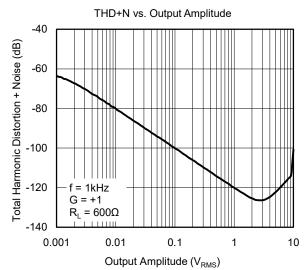




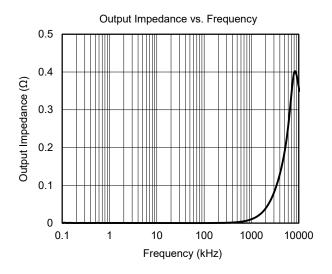


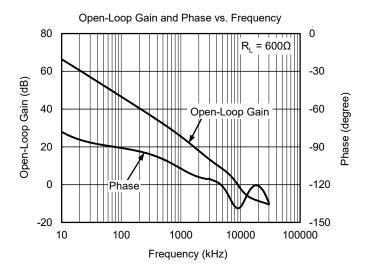


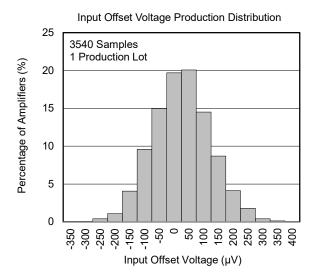




# **TYPICAL PERFORMANCE CHARACTERISTICS (continued)**







# **DETAILED DESCRIPTION**

The SGM5532 is a high-performance operational amplifier combining excellent DC and AC characteristics. It features very low noise, high output-drive capability, high unity-gain and maximum-output-swing bandwidths, low distortion, high slew rate, input-protection diodes and output short-circuit protection. The operational amplifier is compensated internally for unity-gain operation.

# **Unity-Gain Bandwidth**

The unity-gain bandwidth is the frequency up to which an amplifier with a unity gain may be operated without greatly distorting the signal. The SGM5532 device has an 8.5MHz unity-gain bandwidth.

## **Common Mode Rejection Ratio**

The common mode rejection ratio (CMRR) of an amplifier is a measure of how well the device rejects unwanted input signals common to both input leads. It

is found by taking the ratio of the change in input common mode voltage to the change in input offset voltage and converting to decibels. The CMRR of the SGM5532 device is 140dB.

#### Slew Rate

The slew rate is the rate at which an operational amplifier can change its output when there is a change on the input. The SGM5532 device has an 18V/µs slew rate.

#### **Device Functional Modes**

The SGM5532 device is powered on when the supply is connected. The device can be operated as a single-supply operational amplifier or dual-supply amplifier depending on the application.

# POWER SUPPLY RECOMMENDATIONS

The SGM5532 device is specified for operation over the range of  $\pm 2.5 \text{V}$  to  $\pm 18 \text{V}$ . Place  $0.1 \mu\text{F}$  bypass capacitors close to the power supply pins to reduce errors coupling in from noisy or high impedance power supplies. For more detailed information on bypass capacitor placement, refer to the Layout Guidelines.

#### Caution

Supply voltages outside of the ±18V range can permanently damage the device (see the Absolute Maximum Ratings).

# **APPLICATION INFORMATION**

Some applications require differential signals. Figure 2 shows a simple circuit to convert a single-ended input of 2V to 10V into differential output of ±8V on a single 15V supply. The output range is intentionally limited to maximize linearity. The circuit is composed of two

amplifiers. One amplifier acts as a buffer and creates a voltage,  $V_{\text{OUT+}}$ . The second amplifier inverts the input and adds a reference voltage to generate  $V_{\text{OUT-}}$ . Both  $V_{\text{OUT+}}$  and  $V_{\text{OUT-}}$  range from 2V to 10V. The difference,  $V_{\text{DIFF}}$ , is the difference between  $V_{\text{OUT+}}$  and  $V_{\text{OUT-}}$ .

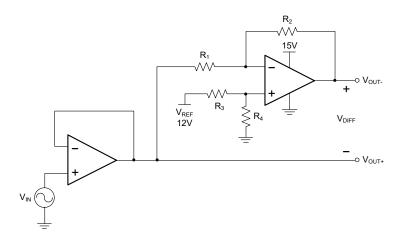


Figure 2. Schematic for Single-Ended Input to Differential Output Conversion

# **Detailed Design Procedure**

The circuit in Figure 2 takes a single-ended input signal,  $V_{\text{IN}}$ , and generates two output signals,  $V_{\text{OUT+}}$  and  $V_{\text{OUT-}}$  using two amplifiers and a reference voltage,  $V_{\text{REF}}$ .  $V_{\text{OUT+}}$  is the output of the first amplifier and is a buffered version of the input signal,  $V_{\text{IN}}$  in Equation 1.  $V_{\text{OUT-}}$  is the output of the second amplifier which uses  $V_{\text{REF}}$  to add an offset voltage to  $V_{\text{IN}}$  and feedback to add inverting gain. The transfer function for  $V_{\text{OUT-}}$  is Equation 2.

$$V_{OLIT_{+}} = V_{IN} \tag{1}$$

$$V_{OUT-} = V_{REF} \times \left(\frac{R_4}{R_3 + R_4}\right) \times \left(1 + \frac{R_2}{R_1}\right) - V_{IN} \times \frac{R_2}{R_1}$$
 (2)

The differential output signal,  $V_{DIFF}$ , is the difference between the two single-ended output signals,  $V_{OUT+}$  and  $V_{OUT-}$ . Equation 3 shows the transfer function for  $V_{DIFF}$ . By applying the conditions that  $R_1 = R_2$  and  $R_3 = R_4$ , the transfer function is simplified into Equation 6. Using this configuration, the maximum input signal is equal to the reference voltage and the maximum output of each amplifier is equal to the  $V_{REF}$ . The differential output range is 2 ×  $V_{REF}$ . Furthermore, the common mode voltage will be one half of  $V_{REF}$  (see Equation 7).

$$V_{\text{DIFF}} = V_{\text{OUT-}} - V_{\text{OUT-}} = V_{\text{IN}} \times \left(1 + \frac{R_2}{R_1}\right) - V_{\text{REF}} \times \left(\frac{R_4}{R_3 + R_4}\right) \left(1 + \frac{R_2}{R_1}\right) \quad \text{(3)}$$

$$V_{OLIT_{+}} = V_{IN} \tag{4}$$

$$V_{\text{OUT-}} = V_{\text{REF}} - V_{\text{IN}} \tag{5}$$

$$V_{\text{DIFF}} = 2 \times V_{\text{IN}} - V_{\text{RFF}} \tag{6}$$

$$V_{CM} = \left(\frac{V_{OUT_{+}} + V_{OUT_{-}}}{2}\right) = \frac{1}{2}V_{REF}$$
 (7)

## **Amplifier Selection**

Linearity over the input range is key for good DC accuracy. The input common mode range and the output swing limitations determine the linearity. In general, an amplifier with rail-to-rail input and output swing is required. Bandwidth is a key concern for this design. Since the SGM5532 has a bandwidth of 8.5MHz, this circuit will only be able to process signals with frequencies of less than 8.5MHz.

## **Passive Component Selection**

Because the transfer function of  $V_{\text{OUT}}$  is heavily reliant on resistors ( $R_1$ ,  $R_2$ ,  $R_3$ , and  $R_4$ ), use resistors with low tolerances to maximize performance and minimize error. This design used resistors with resistance values of  $36k\Omega$  with tolerances measured to be within 2%. But, if the noise of the system is a key parameter, the user can select smaller resistance values to keep the overall system noise low. This ensures that the noise from the resistors is lower than the amplifier noise.

# **LAYOUT**

# **Layout Guidelines**

For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole and the operational amplifier. Bypass capacitors are used to reduce the coupled noise by providing low impedance power sources local to the analog circuitry.
- Connect low-ESR,  $0.1\mu F$  ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from  $+V_S$  to ground is applicable for single-supply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes. A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital and analog grounds, paying attention to the flow of the ground current.
- To reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If it is not possible to keep them separate, it is much

better to cross the sensitive trace perpendicular as opposed to in parallel with the noisy trace.

- ullet Place the external components as close to the device as possible. Keeping  $R_F$  and  $R_G$  close to the inverting input minimizes parasitic capacitance, as shown in Figure 3 and Figure 4.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.

## **Layout Example**

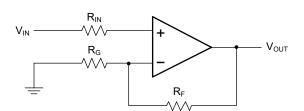


Figure 3. Operational Amplifier Schematic for Non-Inverting Configuration

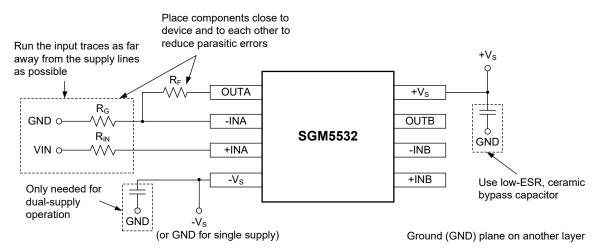


Figure 4. Operational Amplifier Board Layout for Non-Inverting Configuration

# SGM5532

# **Dual Low-Noise Operational Amplifier**

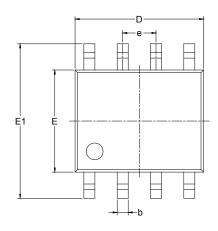
# **REVISION HISTORY**

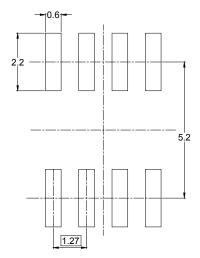
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

APRIL 2018 – REV.A to REV.A.1	Page
Updated Electrical Characteristics section	3
Changes from Original (DECEMBER 2017) to REV.A	Page
Changed from product preview to production data	All

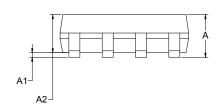


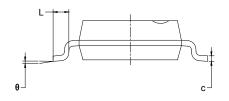
# PACKAGE OUTLINE DIMENSIONS SOIC-8





RECOMMENDED LAND PATTERN (Unit: mm)

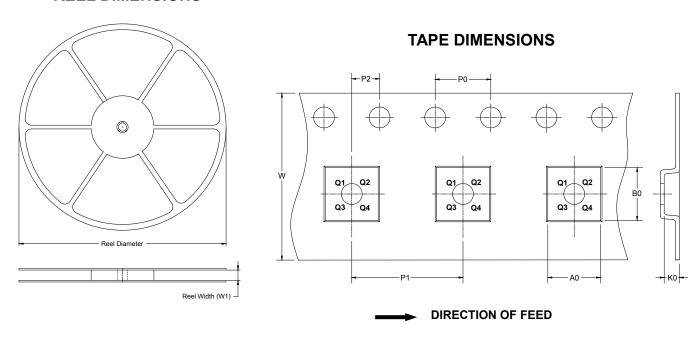




Symbol		nsions meters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
Α	1.350	1.350 1.750		0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350		0.053	0.061	
b	0.330		0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
Е	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.27	BSC	0.050	BSC	
L	0.400	1.270	0.016	0.050	
θ	0°	8°	0°	8°	

# TAPE AND REEL INFORMATION

# **REEL DIMENSIONS**

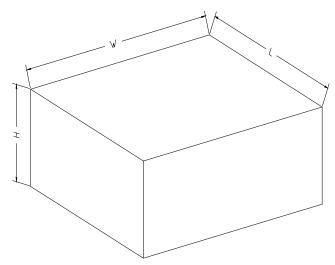


NOTE: The picture is only for reference. Please make the object as the standard.

# **KEY PARAMETER LIST OF TAPE AND REEL**

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
SOIC-8	13"	12.4	6.40	5.40	2.10	4.0	8.0	2.0	12.0	Q1

# **CARTON BOX DIMENSIONS**



NOTE: The picture is only for reference. Please make the object as the standard.

# **KEY PARAMETER LIST OF CARTON BOX**

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton		
13"	386	280	370	5	200002	