

6MHz CMOS Rail-to-Rail IO Opamps

General Description

The LMV82X have a high gain-bandwidth product of 6MHz, a slew rate of 4.2V/ μ s, and a quiescent current of 470 μ A per amplifier at 5V. The LMV82X are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for LMV82X. They are specified over the extended industrial temperature range (-40°C to +125°C). The operating range is from 2.1V to 5.5V. The LMV821 single is available in Green SC70-5, SOT23-5 and SOP-8 packages.

The LMV822 dual is available in Green SOP-8 and MSOP-8 packages. The LMV824 Quad is available in Green SOP-14 and TSSOP-14 packages.

Features

- Single-Supply Operation from +2.1V~+5.5V
- Rail-to-Rail Input / Output
- Gain-Bandwidth Product: 6MHz (Typ)
- Low Input Bias Current: 1pA (Typ)
- Low Offset Voltage: 3.5mV (Max)
- Quiescent Current: 470 μ A per Amplifier (Typ)
- Operating Temperature: -40°C ~ +125°C
- Small Package:
LMV821 Available in SOT23-5, SOP-8 and SC70-5 Packages
LMV822 Available in SOP-8 and MSOP-8 Packages
LMV824 Available in SOP-14 and TSSOP-14 Packages

Applications

- Sensors
- Active Filters
- Cellular and Cordless Phones
- Laptops and PDAs
- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters

Ordering Information

DEVICE	Package Type	MARKING	Packing	Packing Qty
LMV821IDBVRG	SOT23-5	V821	REEL	3000pcs/reel
LMV821IDCKRG	SC70-5	V821	REEL	3000pcs/reel
LMV821IDRG	SOP8	LMV821	REEL	2500pcs/reel
LMV822IDRG	SOP8	LMV822	REEL	2500pcs/reel
LMV822IDGKRG	MSOP8	V822	REEL	3000pcs/reel
LMV824IDRG	SOP14	LMV824	REEL	2500pcs/reel
LMV824IPWRG	TSSOP14	LMV824	REEL	2500pcs/reel

Pin Configuration

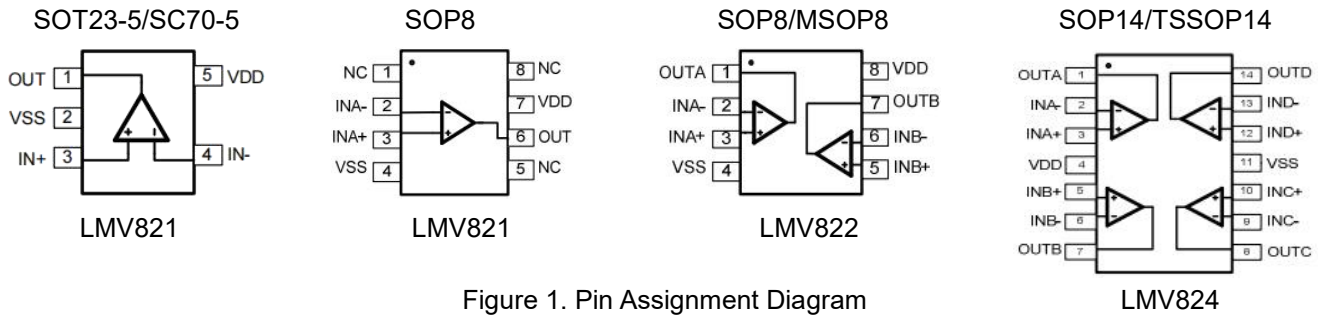


Figure 1. Pin Assignment Diagram

Absolute Maximum Ratings

Condition		Min	Max
Power Supply Voltage (VDD to Vss)		-0.5V	+7.5V
Analog Input Voltage (IN+ or IN-)		Vss-0.5V	VDD+0.5V
PDB Input Voltage		Vss-0.5V	+7V
Operating Temperature Range		-40°C	+125°C
Junction Temperature		-	+160°C
Storage Temperature Range		-55°C	+150°C
Lead Temperature (soldering, 10sec)		-	+260°C
Package Thermal Resistance (TA=+25°C)	SOP-8, θ_{JA}	-	125°C/W
	MSOP-8, θ_{JA}	-	216°C/W
	SOT23-5, θ_{JA}	-	190°C/W
	SC70-5, θ_{JA}	-	333°C/W
ESD Susceptibility	HBM	-	8KV
	MM	-	400V

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Electrical Characteristics

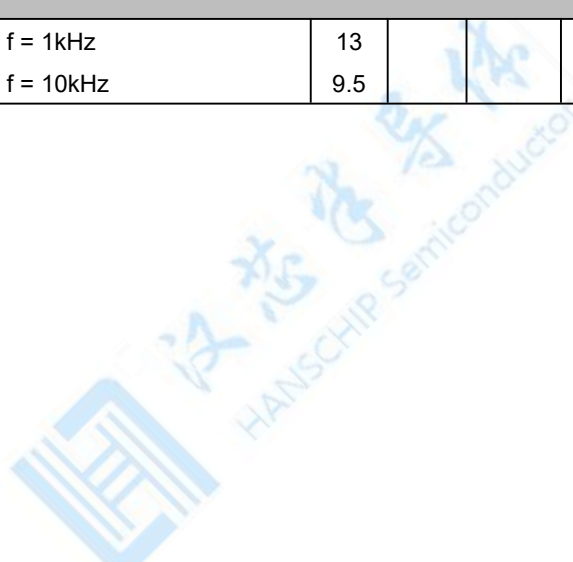
(At $V_S=5V$, $T_A = +25^\circ C$, $V_{CM} = V_S/2$, $R_L = 600\Omega$, unless otherwise noted.)

PARAMETER	CONDITIONS	LMV821/822/824							
		TYP	MIN/MAX OVER TEMPERATURE					UNITS	MIN / MAX
		+25°C	+25°C	0°C to 70°C	-40°C To 85°C	-40°C to 125°C			
INPUT CHARACTERISTICS									
Input Offset Voltage (VOS)	$V_S = 5.5V$	0.8	3.5	3.9	4.3	4.6	mV	MAX	
Input Bias Current (IB)		1					pA	TYP	
Input Offset Current (IOS)		1					pA	TYP	
Input Common Mode Voltage Range (VCM)		-0.1 to +5.6					V	TYP	
Common Mode Rejection Ratio (CMRR)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to $4V$	90	73	70	70	65	dB	MIN	
Open-Loop Voltage Gain (AOL)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to $5.6V$	83					dB	MIN	
	$R_L = 600\Omega$, $V_O = 0.15V$ to $4.85V$	97	90	87	86	79	dB	MIN	
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta T$)	$R_L = 10k\Omega$, $V_O = 0.05V$ to $4.95V$	108					dB	MIN	
		2.4					$\mu V/^\circ C$	TYP	
OUTPUT CHARACTERISTICS									
Output Voltage Swing from Rail	$R_L = 600\Omega$	0.1					V	TYP	
	$R_L = 10k\Omega$	0.015					V	TYP	
Output Current (IOUT)		53	49	45	40	35	mA	MIN	
Closed-Loop Output Impedance	$f = 200kHz$, $G = 1$	3					Ω	TYP	
POWER-DOWN DISABLE									
Turn-On Time		4					μs	TYP	
Turn-Off Time		1.2					μs	TYP	
POWER SUPPLY									
Operating Voltage Range			2.1	2.1	2.1	2.1	V	MIN	
			5.5	5.5	5.5	5.5	V	MAX	
Power Supply Rejection Ratio (PSRR)	$V_S = +2.5V$ to $+5.5V$, $V_{CM} = (-V_S) + 0.5V$, $I_{OUT} = 0$	91	74	72	72	68	dB μA	MIN	
Quiescent Current/Amplifier (IQ)		470	650	727	750	815		MAX	

Electrical Characteristics

 (At $V_s=5V$, $T_A = +25^\circ C$, $V_{CM} = V_S/2$, $R_L = 600\Omega$, unless otherwise noted.)

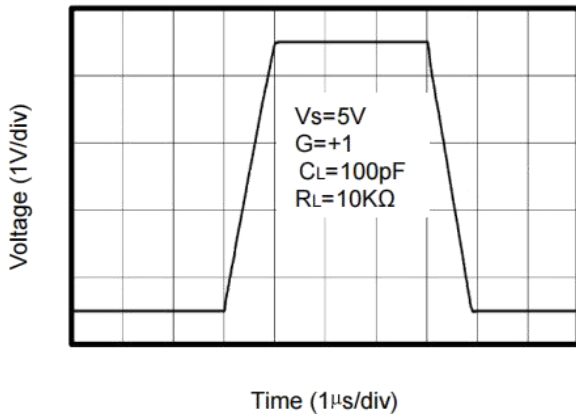
PARAMETER	CONDITIONS	LMV821/822/824							
		TYP	MIN/MAX OVER TEMPERATURE					UNITS	MIN/MAX
		+25°C	+25°C	0°C to 70°C	-40°C to 85°C	-40°C to 125°C			
DYNAMIC PERFORMANCE									
Gain-Bandwidth Product (GBP)	$R_L = 10k\Omega$, $C_L = 100pF$	6					MHz	TYP	
Phase Margin (ϕ_O)	$R_L = 10k\Omega$, $C_L = 100pF$	53					Degrees	TYP	
Full Power Bandwidth (BWP)	<1% distortion, $R_L = 600\Omega$	250					kHz	TYP	
Slew Rate (SR)	$G = +1$, 2V Step, $R_L = 10k\Omega$	4.2					V/ μs	TYP	
Settling Time to 0.1% (t_S)	$G = +1$, 2V Step, $R_L = 600\Omega$	0.4					μs	TYP	
Overload Recovery Time	$V_{IN} \cdot Gain = V_S$, $R_L = 600\Omega$	2.5					μs	TYP	
NOISE PERFORMANCE									
Voltage Noise Density (e_n)	$f = 1kHz$	13					$nV\sqrt{Hz}$	TYP	
	$f = 10kHz$	9.5					$nV\sqrt{Hz}$	TYP	



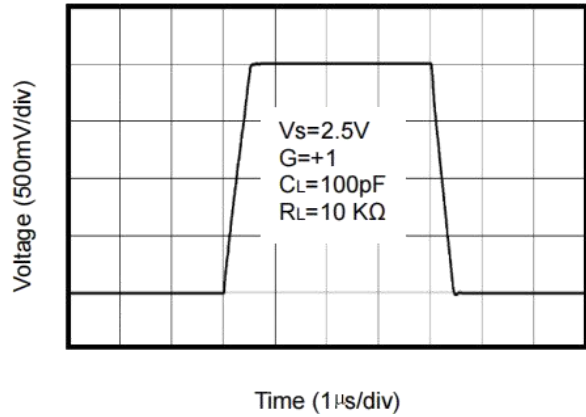
Typical Performance characteristics

(At $V_s=5V$, $T_A = +25^\circ C$, $V_{CM} = V_S/2$, $R_L = 600\Omega$, unless otherwise noted.)

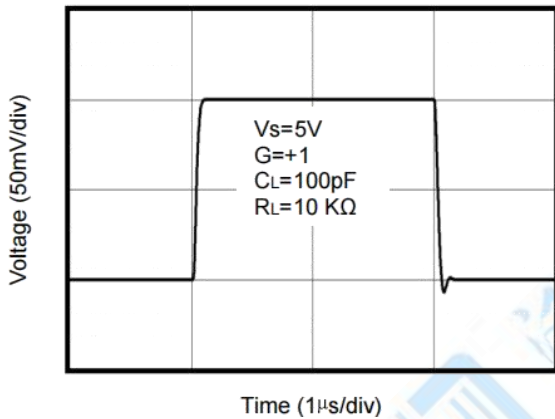
Large-Signal Step Response



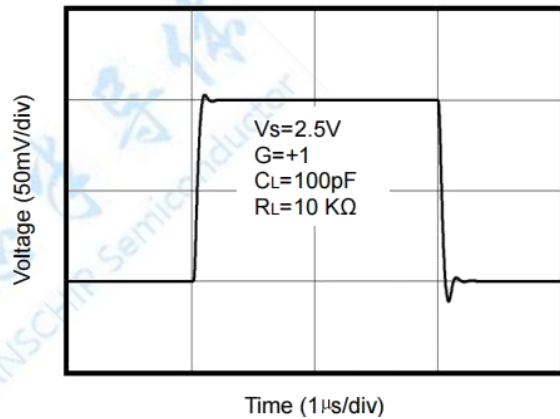
Large-Signal Step Response



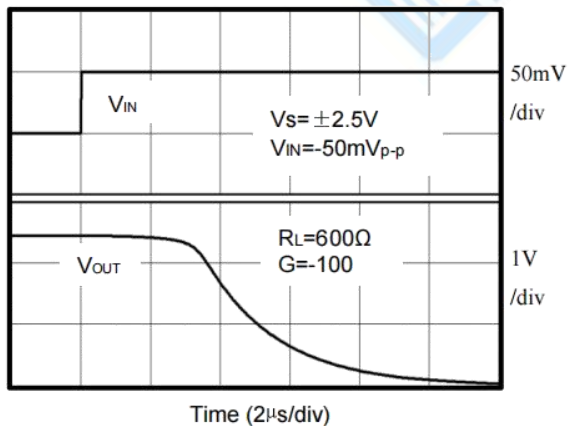
Small-Signal Step Response



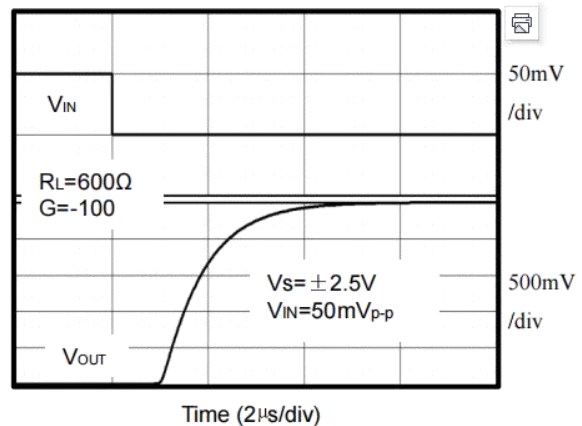
Small-Signal Step Response



Positive Overload Recovery



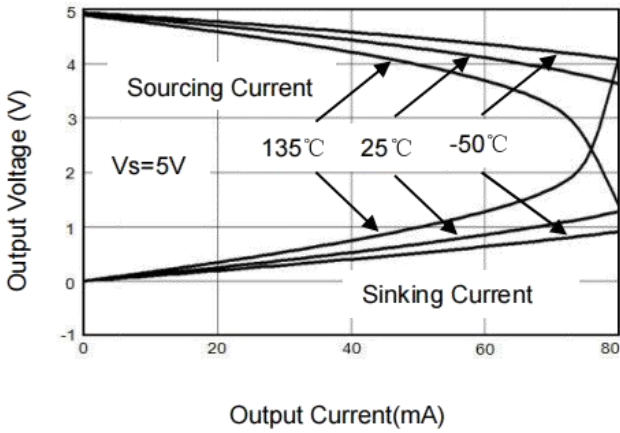
Negative Overload Recovery



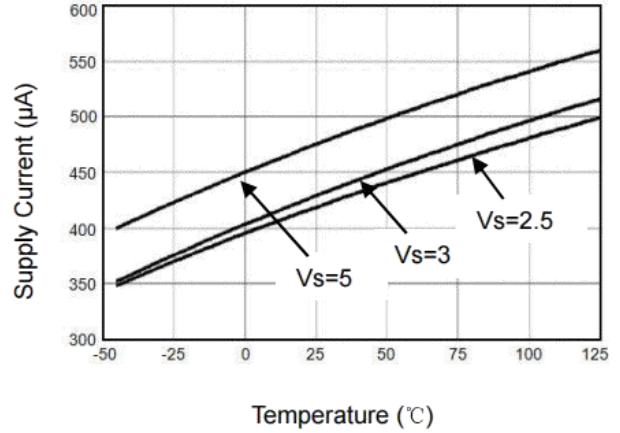
Typical Performance characteristics

(At $V_s=5V$, $T_A = +25^\circ C$, $V_{CM} = V_S/2$, $R_L = 600\Omega$, unless otherwise noted.)

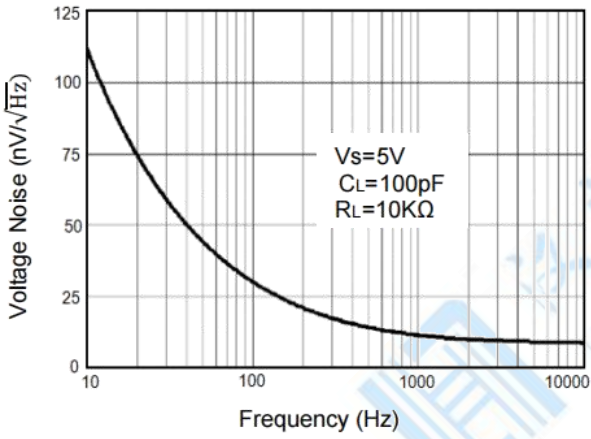
Output Voltage Swing vs. Output Current



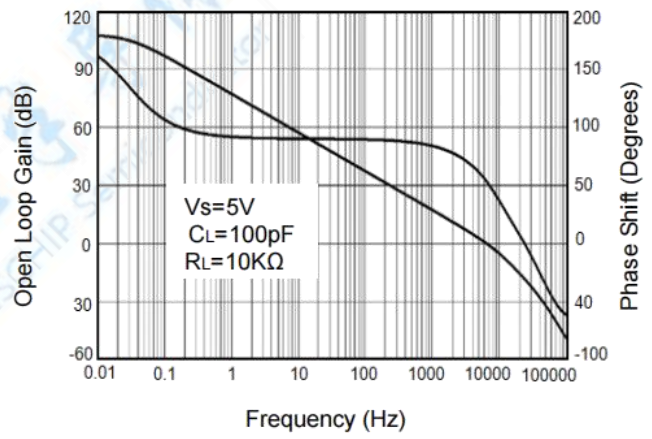
Supply Current vs. Temperature



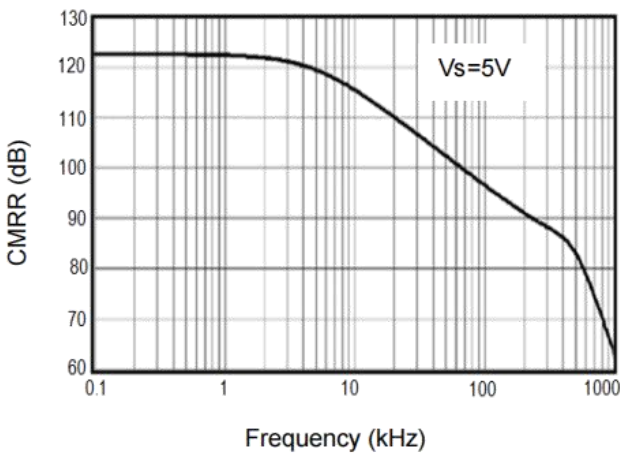
Input Voltage Noise Spectral Density vs. Frequency



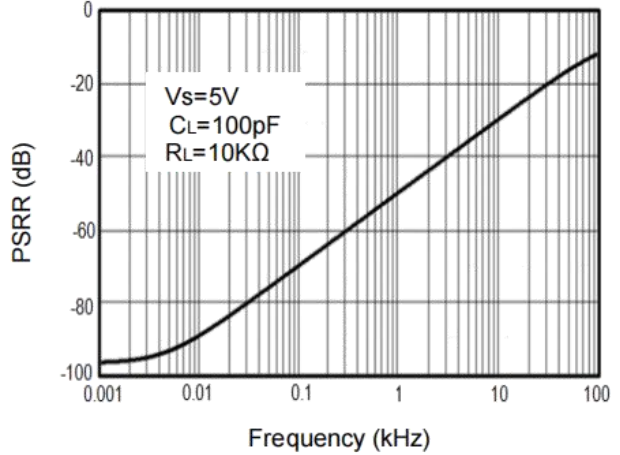
Open Loop Gain, Phase Shift vs. Frequency



CMRR vs. Frequency



PSRR vs. Frequency



Application Note

Size

LMV82X series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the LMV82X series packages save space on printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

LMV82X series operates from a single 2.1V to 5.5V supply or dual $\pm 1.05\text{V}$ to $\pm 2.75\text{V}$ supplies. For best performance, a $0.1\mu\text{F}$ ceramic capacitor should be placed close to the VDD pin in single supply operation. For dual supply operation, both VDD and VSS supplies should be bypassed to ground with separate $0.1\mu\text{F}$ ceramic capacitors.

Low Supply Current

The low supply current (typical $470\mu\text{A}$ per channel) of LMV82X series will help to maximize battery life. They are ideal for battery powered systems.

Operating Voltage

LMV82X series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from $-40\text{ }^{\circ}\text{C}$ to $+125\text{ }^{\circ}\text{C}$. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

Rail-to-Rail Input

The input common-mode range of LMV82X series extends 100mV beyond the supply rails ($\text{VSS}-0.1\text{V}$ to $\text{VDD}+0.1\text{V}$). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of LMV82X series can typically swing to less than 2mV from supply rail in light resistive loads ($>100\text{k}\Omega$), and 60mV of supply rail in moderate resistive loads ($10\text{k}\Omega$).

Capacitive Load Tolerance

The LMV82x family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain.

Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

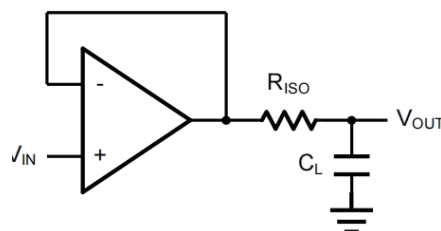


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the RISO resistor value, the more stable VOUT will be. However, if there is a resistive load RL in parallel with the capacitive load, a voltage divider (proportional to RISO/RL) is formed, this will result in a gain error. The circuit in Figure 3 is an improvement to the one in Figure 2. RF provides the DC accuracy by feed-forward the VIN to RL. CF and RISO serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of CF. This in turn will slow down the pulse response.

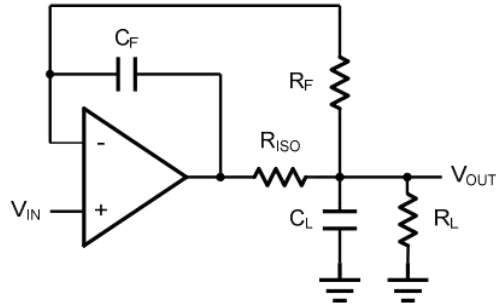


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common to the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4.

shown the differential amplifier using LMV82X.

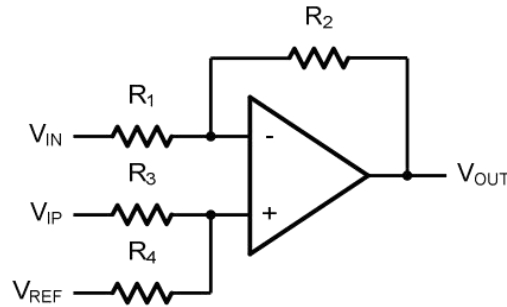


Figure 4. Differential Amplifier

$$V_{OUT} = \left(\frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_4}{R_1} V_{IN} - \frac{R_2}{R_1} V_{IP} + \left(\frac{R_1 + R_2}{R_3 + R_4} \right) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal (i.e. $R_1=R_3$ and $R_2=R_4$), then

$$V_{OUT} = \frac{R_2}{R_1} (V_{IP} - V_{IN}) + V_{REF}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_C=1/(2\pi R_3 C_1)$.

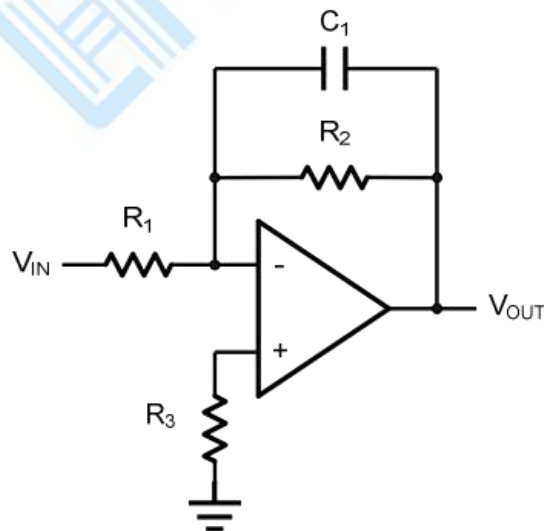


Figure 5. Low Pass Active Filter

Instrumentation Amplifier

The triple LMV82X can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

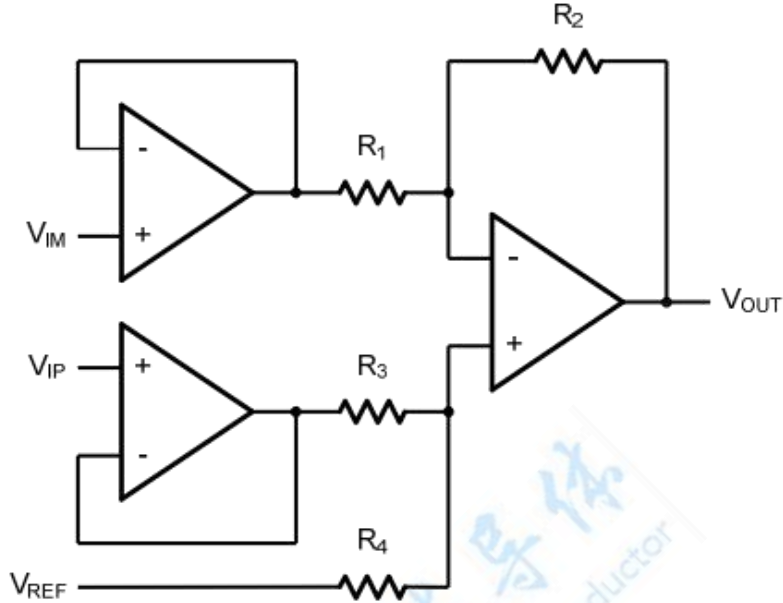
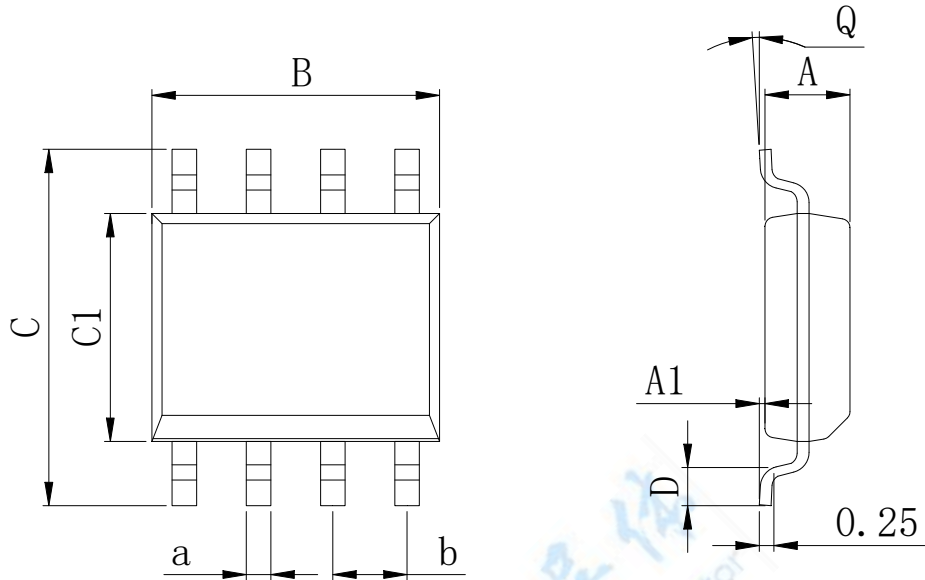


Figure 6. Instrument Amplifier

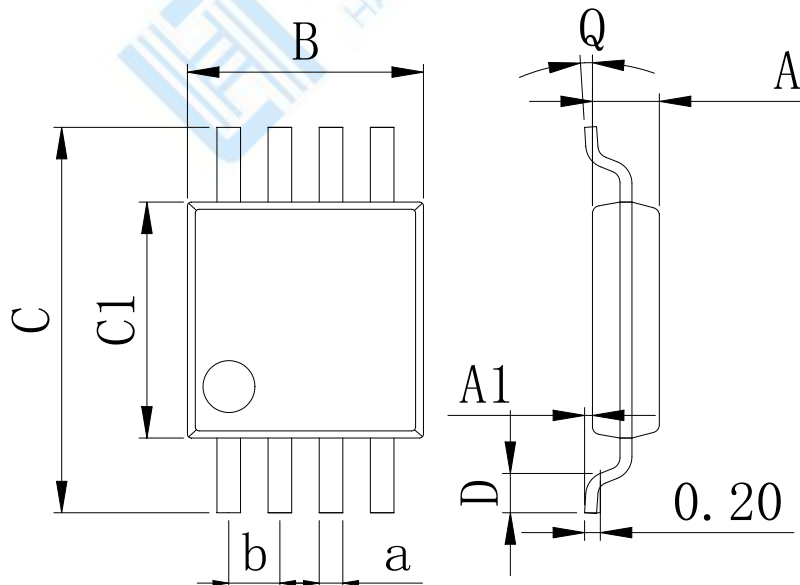
Physical Dimensions

SOP8 (150mil)

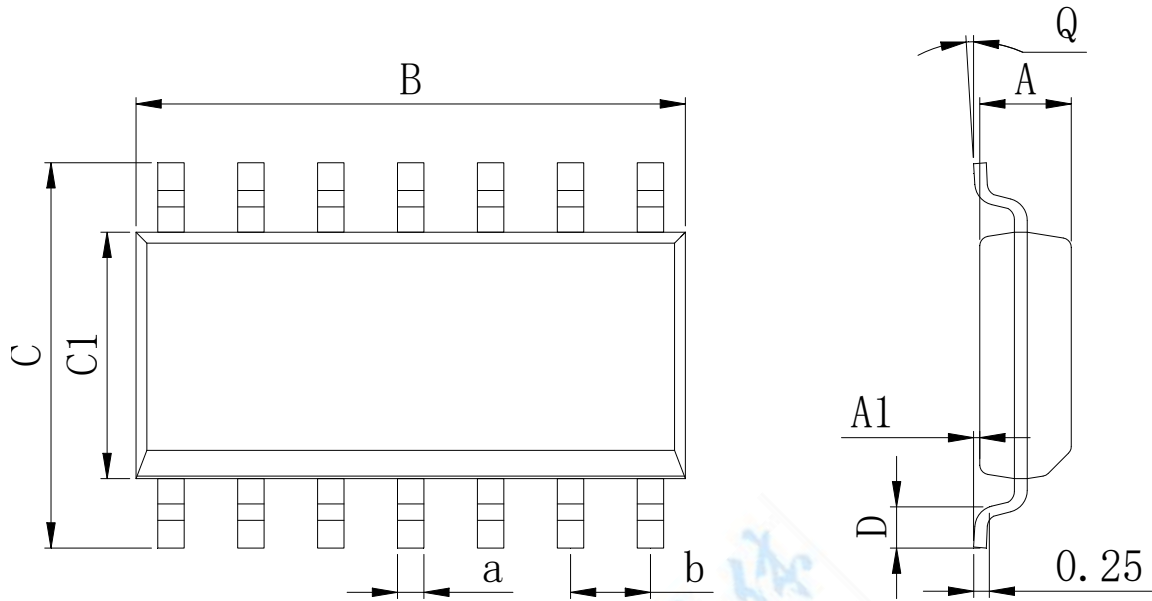

Dimensions In Millimeters(SOP8)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	

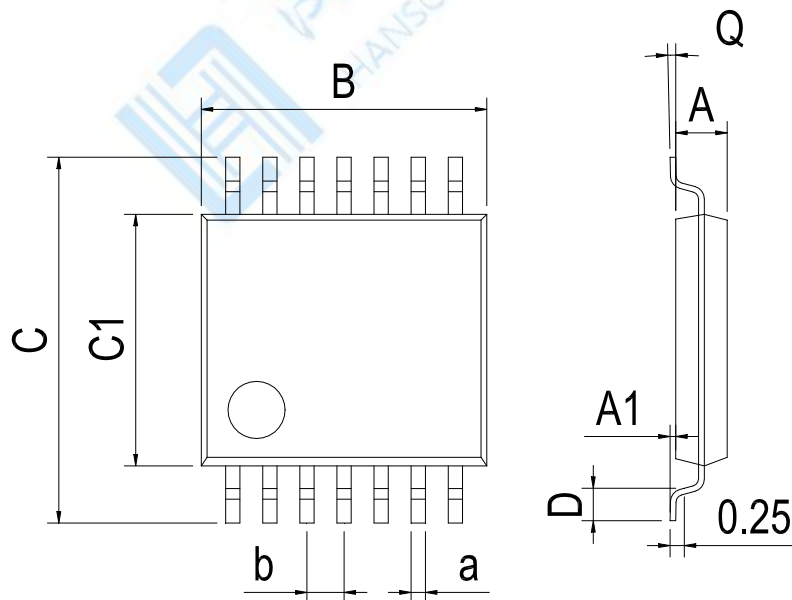
MSOP8


Dimensions In Millimeters(MSOP8)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65 BSC
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	

Physical Dimensions
SOP14

Dimensions In Millimeters(SOP14)

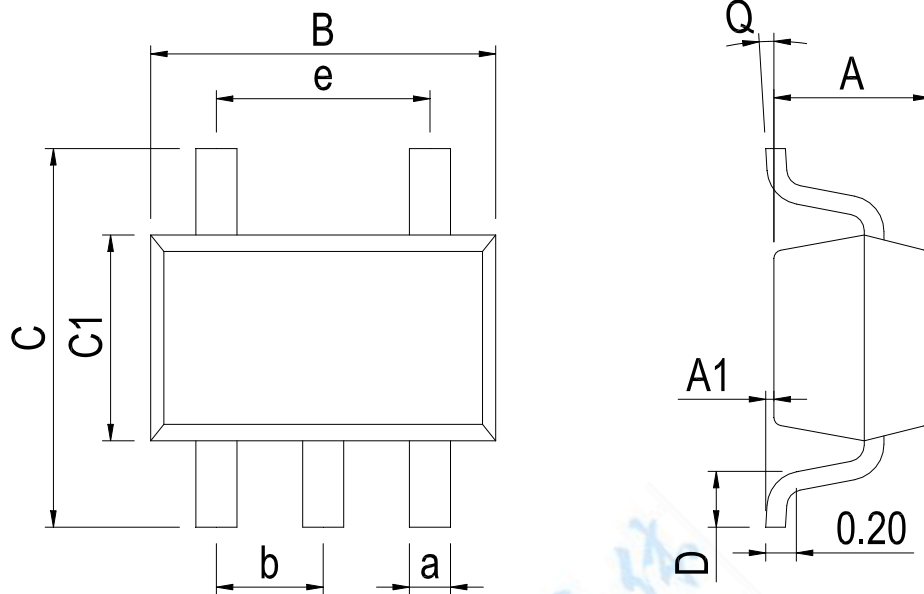
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	

TSSOP14

Dimensions In Millimeters(TSSOP14)

Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0.65 BSC
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25	

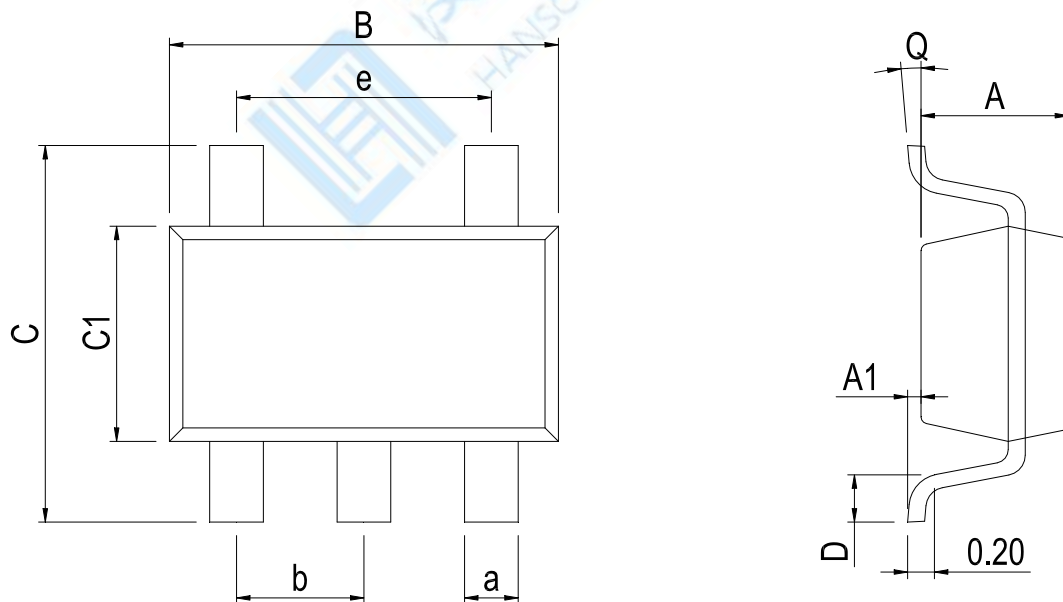
Physical Dimensions

SOT23-5


Dimensions In Millimeters(SOT23-5)

Symbol:	A	A1	B	C	C1	D	Q	a	b	e
Min:	1.05	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC	1.90 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.40		

SC70-5


Dimensions In Millimeters(SC70-5)

Symbol:	A	A1	B	C	C1	D	Q	a	b	e
Min:	0.90	0.00	2.00	2.15	1.15	0.26	0°	0.30	0.65	1.30 BSC
Max:	1.00	0.15	2.20	2.45	1.35	0.46	8°	0.40	BSC	

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