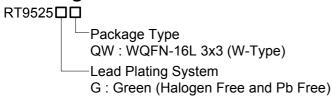


# Linear Single Cell Li-Ion Battery Charger with Auto Power Path

# **General Description**

The RT9525 is an integrated single cell Li-ion battery charger with Auto Power Path Management (APPM). No external MOSFETs are required. The RT9525 enters sleep mode when power is removed. Charging tasks are optimized by using a control algorithm to vary the charge rate including pre-charge mode, fast charge mode and constant voltage mode. For the RT9525, the charge current can also be programmed with an external resister. Additionally, the internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. The charging task will always be terminated in constant voltage mode when the charging current reduces to the termination current of 20%  $I_{CHG\_FAST}$ . Other features include under voltage protection and over voltage protection for the VIN supply.

# **Ordering Information**

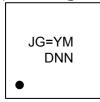


#### Note:

Richtek products are:

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

# **Marking Information**



JG = : Product Code YMDNN : Date Code

#### **Features**

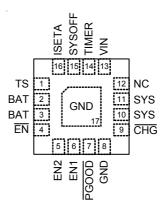
- 28V Maximum Rating for VIN Power
- Selectable Power Current Limit (0.1A / 0.5A / 1.5A)
- Integrated Power MOSFETs
- Auto Power Path Management (APPM)
- Programmable Charging Current Timer and Safe Charge Timer
- Under Voltage Protection
- Over Voltage Protection
- Power Good and Charger Status Indicator
- Optimized Charge Rate via Thermal Feedback
- 16-Lead WQFN Package
- RoHS Compliant and Halogen Free

# **Applications**

- Digital Cameras
- PDAs and Smart Phones
- Portable Instruments

# **Pin Configuration**

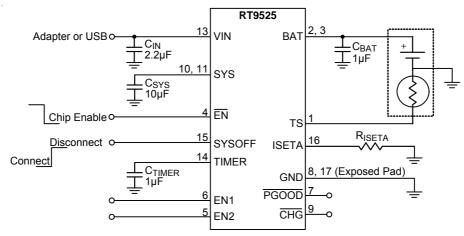
(TOP VIEW)



WQFN-16L 3x3



# **Typical Application Circuit**

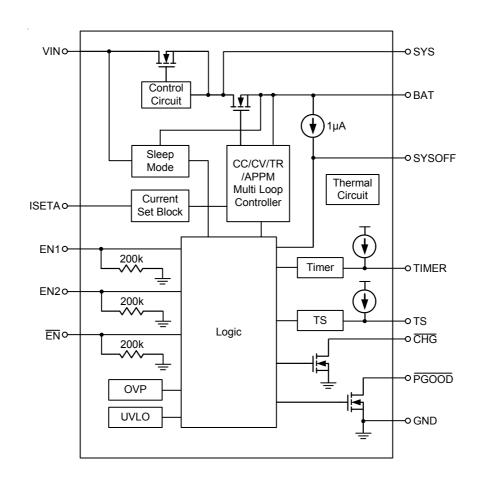


# **Functional Pin Description**

Pin No.	Pin Name	Pin Function	
1	TS	Thermistor Monitor Input. The TS pin connects to a battery's thermistor to determine if the battery is too hot or too cold to charge. If the battery's temperature is out of range, charging is paused until it re-enters the valid range. TS also detect whether the battery (with NTC) is present or not	
2, 3	BAT	Battery Connect Pin.	
4	ĒN	Charge Enable, Active-low input. 200kΩ pull low.	
5	EN2	Input Current Limit Configuration Setting	
6	EN1	Input Current Limit Configuration Setting.	
7	PGOOD	Power Good Status Output. Active-low, open-drain output.	
8, 17 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected t GND for maximum power dissipation.	
9	CHG	Charger Status Output. Active-low, open-drain output.	
10, 11	SYS	System Connect Pin. Connect this pin to system load with a minimum 10uF MLCC to GND.	
12	NC	No Internal Connection.	
13	VIN	Supply Voltage Input.	
14	TIMER	Safe Charge Timer Setting.	
15	SYSOFF	System Disconnect Pin. Pull SYSOFF high to disconnect SYS from bat connect to GND for normal operation. Internally pulled up by $1\mu$ A current source BAT.	
16	ISETA	Charge Current Set Input. Connect a resistor (R <sub>ISETA</sub> ) between ISET and GND.	



# **Functional Block Diagram**





# Absolute Maximum Ratings (Note 1)

• Supply Input Voltage, V <sub>IN</sub>	- −0.3V to 28V
• THG, PGOOD	- −0.3V to 28V
• Other Pins	- −0.3V to 6V
CHG, PGOOD Continuous Current	- 20mA
BAT Continuous Current (total in two pins) (Note 2)	- 2.5A
<ul> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> </ul>	
WQFN-16L 3x3	- 1.471W
Package Thermal Resistance (Note 3)	
WQFN-16L 3x3, $\theta_{JA}$	- 68°C/W
WQFN-16L 3x3, $\theta_{JC}$	- 7.5°C/W
• Lead Temperature (Soldering, 10 sec.)	- 260°C
• Junction Temperature	- 150°C
Storage Temperature Range	65°C to 150°C
• ESD Susceptibility (Note 4)	
HBM (Human Body Model)	
MM (Machine Model)	- 200V
Recommended Operating Conditions (Note 5)	
• Supply Input Voltage, V <sub>IN</sub> (EN2 = H, EN1 = L)	- 4.45V to 6V

### **Electrical Characteristics**

 $(V_{IN} = 5V, V_{BAT} = 4V, T_A = 25^{\circ}C, unless otherwise specification)$ 

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Supply Input						
VIN Operating Range			4.2		6	V
VIN Under Voltage Lockout Threshold	V <sub>UVLO</sub>		3.1	3.3	3.5	V
VIN Under Voltage Lockout Hysteresis	VUVLO _hys			240		mV
VIN Supply Current	lu.	I <sub>SYS</sub> = I <sub>BAT</sub> = 0mA, EN = L		1	2	m A
VIN Supply Current	lin	$I_{SYS} = I_{BAT} = 0mA, \overline{EN} = H$		0.8	1.5	mA
VIN Suspend Current	I <sub>SUS</sub>	V <sub>IN</sub> = 5V, EN2 = EN1 = H		195	333	μΑ
BAT Sleep Leakage Current	I <sub>SLEEP</sub>	$V_{BAT} > V_{IN}$ , $(V_{IN} = 0V)$		5	15	μА
VIN – BAT VOS Rising	Vos_H			200	300	mV
VIN – BAT VOS Falling	Vos_L		10	50		mV
Voltage Regulation	•					
Battery Regulation Voltage	V <sub>REG</sub>	0 to 85°C, I <sub>LOAD</sub> = 20mA	4.16	4.2	4.23	V
System Regulation Voltage	V <sub>SYS</sub>	V <sub>IN</sub> = 6V	5.3	5.5	5.7	V
APPM Regulation Voltage	V <sub>APPM</sub>	EN2 = L ,EN1 = H	4.2	4.3	4.4	V
DPM Regulation Voltage	V <sub>DPM</sub>	EN2 = L	4.35	4.5	4.63	V
VIN to VSYS MOSFET Ron	R <sub>DS(ON)</sub>	I <sub>LIM</sub> = 1000mA		0.2	0.35	Ω

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Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
BAT to VSYS MOSFET Ron	R <sub>DS(ON)</sub>	V <sub>BAT</sub> = 4.2V,I <sub>SYS</sub> = 1A		0.05	0.1	Ω
Re-charge threshold	ΔV <sub>REGCGG</sub>	Battery Regulation – Recharge-level	120	200	280	mV
Current Regulation						
ISETA Set Voltage (Fast Charge Phase)	VISETA	V <sub>BATT</sub> = 4V, R <sub>ISETA</sub> = 1kΩ		2		V
VIN Charge Setting Range	I <sub>CHG</sub>		100		1200	mA
VIN Charge Current	I <sub>CHG</sub>	$V_{BATT} = 4V$ , $R_{ISETA} = 1k\Omega$	570	600	630	mA
		EN2 = H, EN1 = L (1.5A mode)	1.2	1.5	1.8	Α
VIN Current Limit	I <sub>LIM</sub>	EN2 = L, EN1 = H (500mA mode)	450	475	500	mA
		EN2 = L, EN1 = L (100mA mode)	80	90	100	mA
Pre-Charge						
BAT Pre-Charge Threshold	VPRECH	BAT Falling	2.75	2.85	2.95	V
BAT Pre-Charge Threshold Hysteresis	$\Delta V_{PRECH}$			200		mV
Pre-Charge Current	IPRECH	V <sub>BAT</sub> = 2V	5	10	15	%
Charge Termination Detection	n		_			
Termination Current Ratio to Fast Charge	I <sub>TERM</sub>		10	20	30	%
Termination Current Ratio to Fast Charge USB100mA	I <sub>TERM2</sub>	EN2 = L, EN1 = L		3.3		%
Login Input/Output						
CHG Pull Down Voltage	V <del>CHG</del>	I <sub>CHG</sub> = 5mA		200		mV
PGOOD Pull Down Voltage	V <sub>PGOOD</sub>	I <sub>PGOOD</sub> = 5mA		200		mV
EN, EN1,EN2, SYSOFF Pin	VIH		1.5		-	
Threshold	V <sub>IL</sub>				0.4	V
Protection	•		•	•		
Thermal Regulation	T <sub>REG</sub>			125		°C
Thermal Shutdown Temperature	T <sub>SD</sub>			155		°C
Thermal Shutdown Hysteresis	$\Delta T_{SD}$			20		°C
OVP SET Voltage	V <sub>OVP</sub>	V <sub>IN</sub> Rising	6.25	6.5	6.75	V
OVP Hysteresis	V <sub>OVP_hys</sub>			100		mV
Output Short Circuit Detection Threshold	V <sub>SHORT</sub>	V <sub>BAT</sub> – V <sub>SYS</sub>		300		mV
Time						
Pre-Charge Fault Time	tpchg	CTIMER = 1µF (1/8 x t <sub>FCHG</sub> )	1440	1800	2160	S
Fast charge Fault Time	t <sub>FCHG</sub>	C <sub>TIMER</sub> = 1μF	11520	14400	17280	S
PGOOD Deglitch Time	t <del>pgood</del>	Time measured from V <sub>IN</sub> : 0→5V 1μs rise-time to PGOOD = L		1.2		ms
Input Over Voltage Blanking Time	t <sub>OVP</sub>			50		μS

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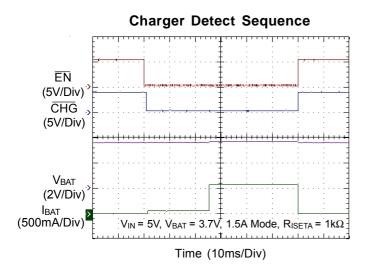


Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Pre-Charge to Fast-Charge Deglitch Time	tpf			25	1	ms
Fast-Charge to Pr-Charger Deglitch Time	t <sub>fp</sub>			25	1	ms
Termination Deglitch Time	ttermi			25	1	ms
Recharge Deglitch Time	trechg			100		ms
Input Power Loss to SYS LDO Turn-Off Delay Time	t <sub>NO-IN</sub>			25		ms
Packing Temperature Fault Detection Deglitch Time	tтs			25	1	ms
Short Circuit, Deglitch Time	tshort			250		μS
Short Circuit Recovery Time	tshort_r			64		ms
Other						
NTC Bias Current	I <sub>NTC</sub>	V <sub>IN</sub> > UVLO and V <sub>IN</sub> > V <sub>BAT</sub> + V <sub>OS</sub> _H	72	75	78	μА
High Temperature Trip Point	V <sub>HOT</sub>	V <sub>TS</sub> Falling	270	300	330	mV
High Temperature Trip Point Hysteresis	V <sub>HOT_hys</sub>	V <sub>TS</sub> Rising from V <sub>HOT</sub>		30		mV
Low Temperature Trip Point	V <sub>COLD</sub>	V <sub>TS</sub> Rising	2000	2100	2200	mV
Low Temperature Trip Point Hysteresis	V <sub>COLD_hys</sub>	V <sub>TS</sub> Falling from V <sub>COLD</sub>		300		mV

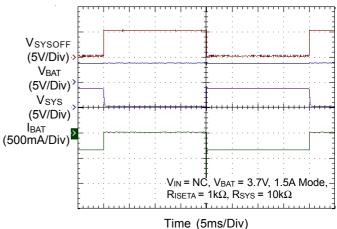
- **Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2. Guaranteed by design.
- Note 3.  $\theta_{JA}$  is measured at  $T_A$  = 25°C on a high effective thermal conductivity four-layer test board per JEDEC 51-7.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 4. Devices are ESD sensitive. Handling precaution is recommended.
- Note 5. The device is not guaranteed to function outside its operating conditions.

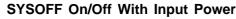


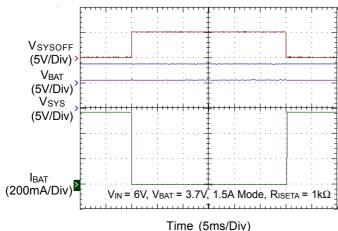
# **Typical Operating Characteristics**



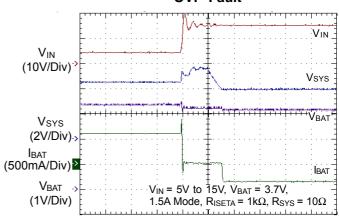
# SYSOFF On/Off Without Input Power

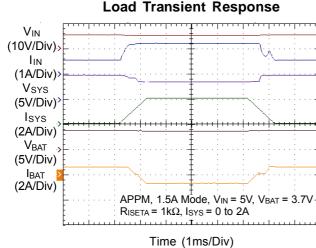






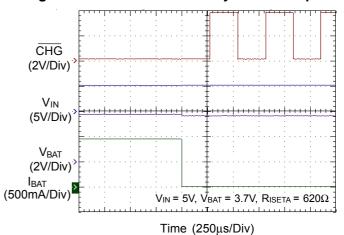
**OVP Fault** 





#### Charger CHG Status After Safety Timers Expired

Time (50µs/Div)

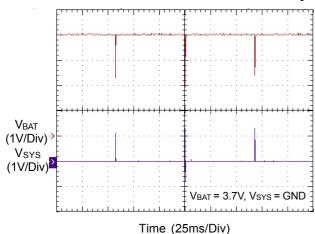


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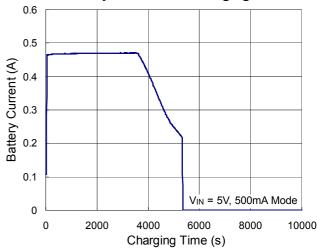
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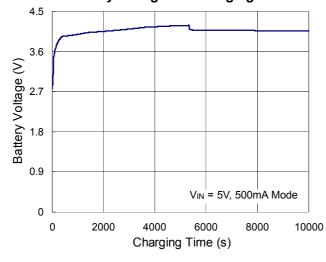




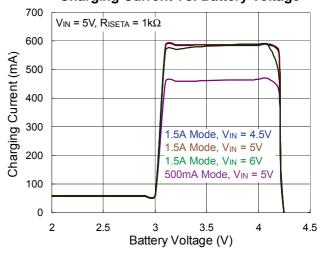
### **Battery Current vs. Charging Time**



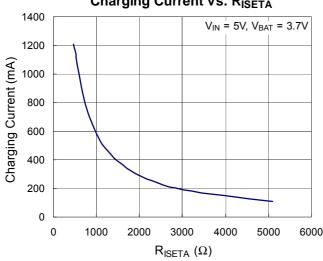
### **Battery Voltage vs. Charging Time**



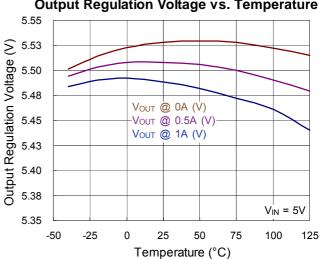
#### **Charging Current vs. Battery Voltage**



### Charging Current vs. RISETA

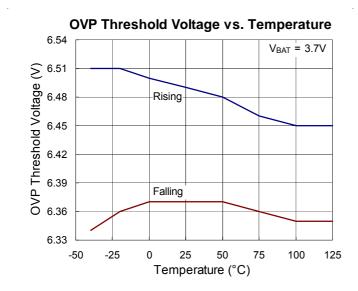


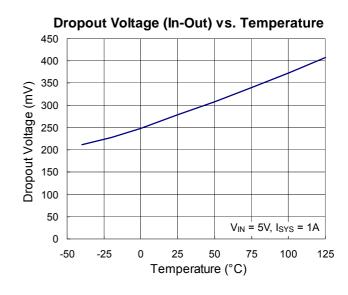
**Output Regulation Voltage vs. Temperature** 

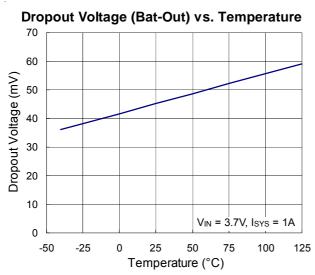


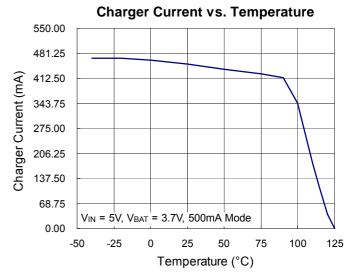
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# **Application Information**

The RT9525 is a fully integrated single cell Li-ion battery charger ideal for portable applications. The internal thermal feedback circuitry regulates the die temperature to optimize the charge rate for all ambient temperatures. Other features include under voltage protection and over voltage protection.

#### **Pre-Charge Mode**

When the output voltage is lower than 2.8V, the charging current will be reduced to a fast charge current ratio set by R<sub>ISETA</sub> to protect the battery life time.

#### **Fast Charge Mode**

When the output voltage is higher than 3V, the charging current will be equal to the fast charge current set by RISETA.

### **Constant Voltage Mode**

When the output voltage is near 4.2V and the charging current falls below the termination current, after a deglitch time check of 25ms, the charger will become disabled and CHG will go from L to H.

#### **Re-Charge Mode**

When the chip is in charge termination mode, the charging current will gradually go down to zero. However, once the voltage of the battery drops to below 4V, there will be a deglitch time of 100ms, and then the charging current will resume again.

#### **Charging Current Decision**

The charge current can be set according to the following equations:

$$I_{CHG\_FAST} = V_{ISETA} / R_{ISETA} \times 300$$
 (A)

$$I_{CHG\_PRE} = 10\% \times I_{CHG\_FAST}$$
 (A)

where  $V_{ISETA}$  unit = V;  $R_{ISETA}$  unit =  $\Omega$ 

#### Time Fault

The Fast Charge Fault Time can be set according to the following equations:

Fast Charge Fault Time :  $t_{FCHG} = 14400 \times C_{TIMER}$  (s) Pre-Charge Fault Time :  $t_{PCHG} = 1/8 x t_{FCHG}$  (s)

where  $C_{TIMER}$  unit is  $\mu F$ .

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During the fast charge phase, several events may increase the timer duration.

For example, the system load current may have activated the APPM loop which reduces the available charging current, the device has entered thermal regulation because the IC junction temperature has exceeded T<sub>REG</sub>.

During each of these events, if  $3V < V_{BAT} < 4V$ , the internal timers are slowed down proportionately to the reduction in charging current. However, once the duration exceeds the fault time, the CHG output will flash at approximately 2Hz to indicate a fault condition and a charger current ~ 1mA.

$$t_{FCHG\_true} = t_{FCHG} \times \frac{2}{V_{ISETA}}$$

t<sub>FCHG</sub> true: modified timer in fast charge

t<sub>FCHG</sub>: original timer in fast charge

$$t_{\text{FCHG}}$$
 = 14400 sec x (  $\frac{C_{\text{TIMER}}}{1\mu\text{F}}$  )  $t_{\text{PCHG}}$  =  $\frac{t_{\text{FCHG}}}{8}$ 

t<sub>PCHG</sub>: timer in pre-charge

Time fault release:

- (1) Re-plug power
- (2) Toggle /EN
- (3) Enter/Exit USB suspend mode
- (4) Removes Battery
- (5) OVP

Note that the fast charge fault time is independent of the charge current.

#### **Power Good**

VIN Power Good (PGOOD = L)

Input State	PGOOD Output
V <sub>IN</sub> < V <sub>UVLO</sub>	High impedance
V <sub>UVLO</sub> < V <sub>IN</sub> < V <sub>BAT</sub> + V <sub>OS</sub> _H	High impedance
V <sub>BAT</sub> + V <sub>OS_H</sub> < V <sub>IN</sub> < V <sub>OVP</sub>	Low impedance
V <sub>IN</sub> > V <sub>OVP</sub>	High impedance

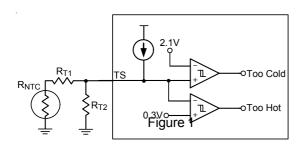


#### **Charge State Indicator**

Charge State	CHG Output	
Charging	I am (fan finat alaana	
Charging suspended by thermal loop	Low (for first charge cycle)	
Safety timers expired	2Hz flash	
Charging done		
Recharging after termination	High impedance	
IC disabled or no valid input power	i ngri impedance	

#### **Battery Pack Temperature Monitoring**

The RT9525 features an external battery pack temperature monitoring input. The TS input connects to the NTC thermistor in the battery pack to monitor battery temperature and prevent dangerous over temperature conditions. If at any time the voltage at TS falls outside of the operating range, charging will be suspended. The timers maintain their values but suspend counting. When charging is suspended due to a battery pack temperature fault, the CHG pin remains low and continues to indicate charging.



When temperature reaches at "Too Cold" state,

$$R_{NTC} = R_{COLD}$$

$$\frac{(R_{T1} + R_{COLD}) \times R_{T2}}{(R_{T1} + R_{COLD}) + R_{T2}} \times I_{NTC} = 2.1V \quad (V)$$
 (1)

where  $I_{NTC} = 75\mu A$  (typ.)

When temperature reaches at "Too Hot" state,  $(R_{T1} + R_{HOT}) \times R_{T2} \times I_{NTC} = 0.3V$  (V)

From (1), (2), the  $R_{T1}$  and  $R_{T2}$  can be calculated by the following equations :

$$R_{T1} = \frac{-1500\mu \times (R_{HOT} + R_{COLD})}{3000\mu} +$$

$$\frac{20\sqrt{5625\mu^2} \text{ x } (\mathsf{R}_{\mathsf{COLD}} - \mathsf{R}_{\mathsf{HOT}})^2 + 105\mu \times (\mathsf{R}_{\mathsf{COLD}} - \mathsf{R}_{\mathsf{HOT}})}{3000\mu}$$

$$R_{T2} = \frac{(R_{T1} + R_{HOT})}{250\mu \times R_{T1} + 250\mu \times R_{HOT} - 1}$$

#### **Charge Enable**

When  $\overline{EN}$  is low, the charger turns on. When  $\overline{EN}$  is high, the charger turn off.  $\overline{EN}$  is pulled low for initial condition.

#### **VIN Input Current Limit**

EN2	EN1	VIN Input Current Limit
L	L	90mA
L	Н	475mA
Н	L	1.5A
Н	Н	Suspend Mode

#### Suspend Mode

Set EN1 = EN2 = H, and the charger will enter Suspend Mode. In Suspend Mode, <u>CHG</u> is in high impedance and  $I_{SUS(MAX)}$  < 330 $\mu$ A.

#### **Power Switch**

For the RT9525, there are three power scenarios:

- (1) When a battery and an external power supply(USB or adapter) are connected simultaneously :
  - If the system load requirements exceed that of the input current limit, the battery will be used to supplement the current to the load. However, if the system load requirements are less than that of the input current limit, the excess power from the external power supply will be used to charge the battery.
- (2) When only the battery is connected to the system :

  The battery provides the power to the system.
- (3) When only an external power supply is connected to the system :
- (2) The external power supply provides the power to the system.

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#### **Input DPM Mode**

For the RT9525, the input voltage is monitored when the USB100 or USB500 is selected. If the input voltage is lower than V<sub>DPM</sub>, the input current limit will be reduced to stop the input voltage from dropping any further. This can prevent the IC from damaging improperly configured or inadequately designed USB sources.

#### **APPM Mode**

Once the sum of the charging and system load currents becomes higher than the maximum input current limit, the SYS pin voltage will be reduced. When the SYS pin voltage is reduced to V<sub>APPM</sub>, the RT9525 will automatically operate in APPM mode. In this mode, the charging current is reduced while the SYS current is increased to maintain system output. In APPM mode, the battery termination function is disabled.

#### **Battery Supplement Mode Short Circuit Protect**

In APPM mode, the SYS voltage will continue to drop if the charge current is zero and the system load increases beyond the input current limit. When the SYS voltage decreases below the battery voltage, the battery will kick in to supplement the system load until the SYS voltage rises above the battery voltage.

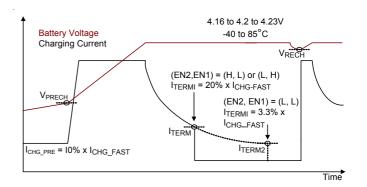
While in supplement mode, there is no battery supplement current regulation. However, a built in short circuit protection feature is available to prevent any abnormal current situations. While the battery is supplementing the load, if the difference between the battery and SYS voltage becomes more than the short circuit threshold voltage. SYS will be disabled. After a short circuit recovery time, t<sub>SHORT R</sub>, the counter will be restarted. In supplement mode, the battery termination function is disabled. Note that for the battery supply mode exit condition, V<sub>BAT</sub> –  $V_{SYS} < 0V$ .

#### **Battery Disconnect (SYSOFF input)**

The RT9525 features a SYSOFF input that allows the user to turn off the switch to disconnect the battery from the SYS pin.

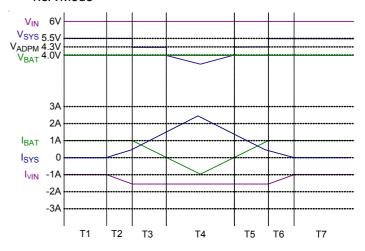
#### Thermal Regulation and Thermal Shutdown

The RT9525 provides a thermal regulation loop function to monitor the device temperature. If the die temperature rises above the regulation temperature, T<sub>REG</sub>, the charge current will automatically be reduced to lower the die temperature. However, in certain circumstances (such as high VIN, heavy system load, etc.) even with the thermal loop in place, the die temperature may still continue to increase. In this case, if the temperature rises above the thermal shutdown threshold, T<sub>SD</sub> the internal switch between VIN and SYS will be turned off. The switch between the battery and SYS will remain on, however, to allow continuous battery power to the load. Once the die temperature decreases by  $\Delta T_{SD}$ , the internal switch between VIN and SYS will be turned on again and the device returns to normal thermal regulation.



#### **APPM Profile**

#### 1.5A Mode



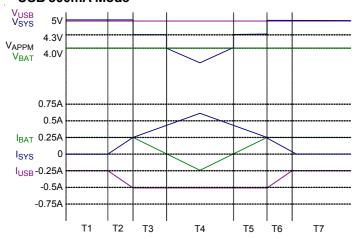
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	Isys	V <sub>SYS</sub>
T1, T7	0	SYS Regulation Voltage
T2, T6	< I <sub>VIN_OC</sub> - CHG_MAX	SYS Regulation Voltage
T3, T5	> I <sub>VIN_OC</sub> - CHG_MAX < I <sub>VIN_OC</sub>	Auto Charge Voltage Threshold
T4	> I <sub>VIN_OC</sub>	V <sub>BAT</sub> – I <sub>BAT</sub> x R <sub>DS(ON)</sub>

	I <sub>VIN</sub>	I <sub>BAT</sub>
T1, T7	CHG_MAX	CHG_MAX
T2, T6	I <sub>SYS</sub> + CHG_MAX	CHG_MAX
T3, T5	VIN_OC	VIN_OC - I <sub>SYS</sub>
T4	VIN_OC	Isys - IVIN_OC

#### **USB 500mA Mode**



	I <sub>SYS</sub>	V <sub>SYS</sub>
T1, T7	0	SYS Regulation Voltage
T2, T6	<usb_oc - CHG_MAX</usb_oc 	SYS Regulation Voltage
T3, T5	> USB_OC - CHG_MAX < USB_OC	Auto Charge Voltage Threshold
T4	> USB_OC	V <sub>BAT</sub> – I <sub>BAT</sub> x R <sub>DS(ON)</sub>

	$I_{USB}$	I <sub>BAT</sub>
T1, T7	CHG_MAX	CHG_MAX
T2, T6	I <sub>SYS</sub> + CHG_MAX	CHG_MAX
T3, T5	USB_OC	USB_OC - I <sub>SYS</sub>
T4	USB_OC	I <sub>SYS</sub> - USB_OC

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For WQFN-16L 3x3 packages, the thermal resistance,  $\theta_{JA}$ , is 68°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at  $T_A$ = 25°C can be calculated by the following formula :

$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (68^{\circ}C/W) = 1.471W$$
 for WQFN-16L 3x3 package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . The derating curve in Figure 2 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

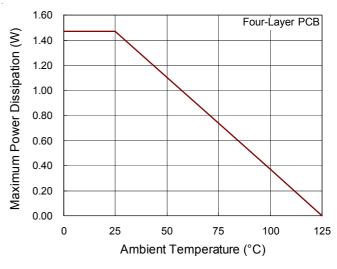


Figure 2. Derating Curve of Maximum Power Dissipation

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#### **Layout Considerations**

The RT9525 is a fully integrated low cost single cell Li-Ion battery charger ideal for portable applications. Careful PCB layout is necessary. For best performance, place all peripheral components as close to the IC as possible. A short connection is highly recommended. The following guidelines should be strictly followed when designing a PCB layout for the RT9525.

- ▶ Input and output capacitor should be placed close to IC and connected to ground plane. The trace of input in the PCB should be placed far away from the sensitive devices AND shielded by the ground.
- ▶ The GND and exposed pad should be connected to a strong ground plane for heat sinking and noise protection.
- ► The connection of R<sub>ISETA</sub> should be isolated from other noisy traces. A short wire is recommended to prevent EMI and noise coupling.

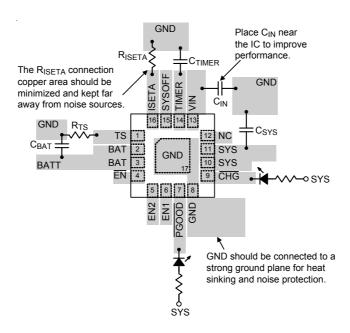
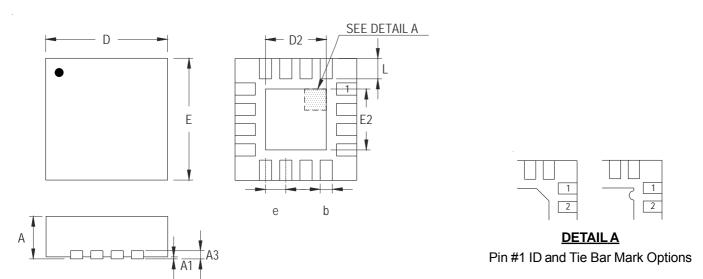


Figure 3. PCB Layout Guide



### **Outline Dimension**



Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
Α	0.700	0.800	0.028	0.031
A1	0.000	0.050	0.000	0.002
A3	0.175	0.250	0.007	0.010
b	0.180	0.300	0.007	0.012
D	2.950	3.050	0.116	0.120
D2	1.300	1.750	0.051	0.069
Е	2.950	3.050	0.116	0.120
E2	1.300	1.750	0.051	0.069
е	0.500		0.020	
L	0.350	0.450	0.014	0.018

W-Type 16L QFN 3x3 Package

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