UNISONIC TECHNOLOGIES CO., LTD

UC2843B

LINEAR INTEGRATED CIRCUIT

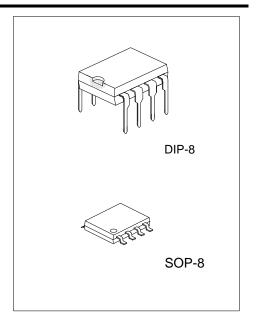
HIGH-PERFORMANCE CURRENT-MODE PWM CONTROLLERS

DESCRIPTION

UTC **UC2843B** DC-DC The provides off-line or fixed-frequency current-mode control design with minimum external components. Internally-implemented circuits include an under-voltage lockout (UVLO) and a precision reference with accuracy at the error amplifier input. The UTC UC2843B also contain internal circuits which include a pulse width modulation (PWM) comparator providing current-limit control, logic ensuring latched operation, and a totem-pole output stage designed to source or sink high-peak current. The output stage is low when it is in off-state condition and suitable for N-channel MOSFETs driving.

The UTC UC2843B also has following advantages: the start-up current lower than 0.5mA while the oscillator discharge current is specified to 8.3mA (Typ.). In UVLO conditions, the output has a maximum saturation voltage of 1.2V when sinking 10mA @ $V_{CC} = 5V$.

The typical UVLO threshold of the UTC UC2843B is 8.4V (on) and 7.6V (off) and can operate to duty cycles approximately 100%.

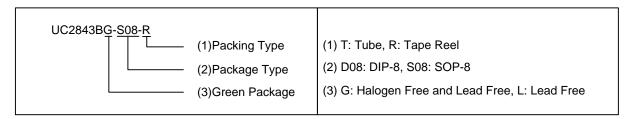


FEATURES

- * Current mode operation:500 kHz
- * Low start-up current value < 0.5 mA
- * Latching PWM for cycle-by-cycle current limiting
- * Trimmed oscillator discharge current
- * Automatic feed-forward compensation
- * Internally trimmed reference with UVLO
- * High-current totem-pole output UVLO with hysteresis
- * Double-pulse suppression

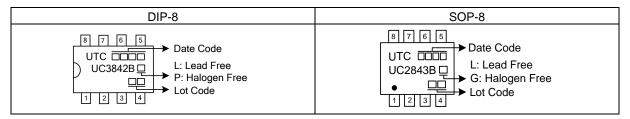
ORDERING INFORMATION

Ordering	Number	Dookses	Doolsing		
Lead Free	Halogen Free	Package	Packing		
UC2843BL-D08-T	UC2843BG-D08-T	DIP-8	Tube		
UC2843BL-S08-T	UC2843BG-S08-T	SOP-8	Tube		
UC2843BL-S08-R	UC2843BG-S08-R	SOP-8	Tape Reel		

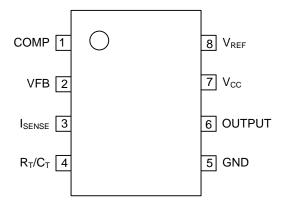


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MARKING



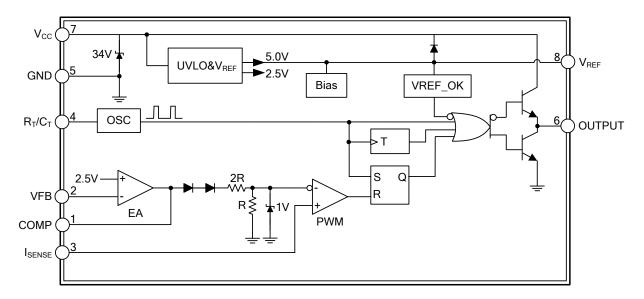
■ PIN CONFIGURATION



■ PIN DESCRIPTION

PIN NO.	PIN NAME	DESCRIPTION
1	COMP	This pin is the Error Amplifier output and is made available for loop compensation.
2	VFB	This is the inverting input of the Error Amplifier. It is normally connected to the switching power supply output through a resistor divider.
3	I _{SENSE}	A voltage proportional to inductor current is connected to this input. The PWM uses this information to terminate the output switch conduction.
4	R _T /C _T	The Oscillator frequency and maximum Output duty cycle are programmed by connecting resistor R_T to V_{REF} and capacitor C_T to ground. Operation to 500 kHz is possible.
5	GND	This pin is the combined control circuitry and power ground.
6	OUTPUT	This output directly drives the gate of a power MOSFET. Peak currents up to 1.0A are sourced and sunk by this pin.
7	Vcc	This pin is the positive supply of the control IC.
8	V_{REF}	This is the reference output. It provides charging current for capacitor C_{T} through resistor R_{T} .

■ BLOCK DIAGRAM



■ **ABSOLUTE MAXIMUM RATING** (T_A=25°C, unless otherwise specified)

PARAMETER		SYMBOL	RATINGS	UNIT
Supply Voltage (Low impedance source)		V _{CC}	30	V
Analog Input Voltage (VFB and ISENSE)		V _{IN}	-0.3~+6.3	V
Supply Current		I _{CC}	30	mA
Error Amplifier Output Sink Current		I _{O(SINK)}	10	mA
Output Current		I _{OUT}	±1	Α
Danie dia dia	SOP-8	P _D	800	\^/
Power Dissipation	DIP-8		1250	mW
Output Energy (Capacitive load)		W	5	μJ
Junction Temperature		TJ	150	°C
Operating Temperature		T _{OPR}	-40~+85	°C
Storage Temperature		T _{STG}	-65~+150	°C

Notes: 1. Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

■ THERMAL DATA

PARAMETER		SYMBOL	RATINGS	UNIT
Thermal Resistance Junction to Ambient	SOP-8	0	156	°C/W
	DIP-8	θ_{JA}	100	°C/W

■ RECOMMENDED OPERATING CONDITIONS

PARAMETER		SYMBOL	MIN	TYP	MAX	UNIT
Supply Voltage		V _{CC}			30	V
lanut Valtage	R _T /C _T	V _{IN}	0		5.5	V
Input Voltage	VFB and I _{SENSE}		0		5.5	V
Output Voltage (OUTPUT)		V_{OUT}	0		30	V
Supply Current, Externally Limited		I _{CC}			25	mA
Output Current		I _{OUT}			200	mA
Reference Output Current		I _{O(REF)}			-20	mA
Oscillator Frequency		fosc		100	500	kHz
Operating Temperature		T _A	-40		+85	°C

ELECTRICAL CHARACTERISTICS

($V_{CC}=15V$, $R_T=10k\Omega$, $C_T=3.3nF$, $T_J=25$ °C, unless otherwise specified)

(VCC-13V, IX) - 10K22, C1-3.3III , 1J-2	J C, uniess on	nerwise specified)				
PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
REFERENCE SECTION						
Reference Output Voltage	V_{REF}	I _{OUT} =1mA, T _J =25°C	4.95	5	5.05	V
Line Regulation	ΔV_{OUT}	V _{CC} = 12V~25V		6	20	mV
Load Regulation	ΔV_{OUT}	I _{OUT} = 1mA~20mA		6	25	mV
Average Temperature Coefficient Of Output Voltage	Ts			0.2	0.4	mV/°C
Total Output Variation	V_{REF}	V _{CC} =12V~25V, I _{OUT} =1mA~20mA	4.9		5.1	V
Output Noise Voltage	e _N	f = 10Hz~10kHz, T _J =25°C		50		μV
Long Term Stability		T _J =25°C For 1000 hours		5	25	mV
Output Short Circuit Current	I _{SC}		-30	-100	-180	mA

^{2.} All voltages are concerning the device GND terminal.

■ ELECTRICAL CHARACTERISTICS (Cont.)

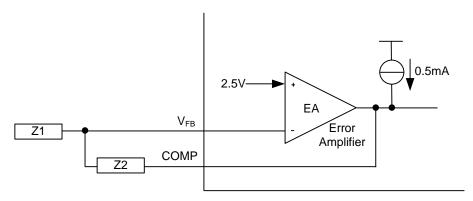
	1		1	1	1	г
		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OSCILLATOR SECTION	T		1	1	ı	ī
		$T_J = 25$ °C, $R_T = 62$ k Ω , $C_T = 1$ nF,	49	52	55	kHz
Frequency	f _{OSC}	Min =225 kHz, Max =275 kHz	48			
		T _J = Full range			56	kHz
Frequency Change with Voltage	$\frac{\Delta fosc}{\Delta V}$	V _{CC} = 12V ~ 25V		0.2	1	%
Frequency Change with	Δfosc	T. Full range		5		%
Temperature	ΔΤ	T _J = Full range		3		/0
Oscillator Voltage Swing	Vosc	Peak to peak		1.7		V
Discharge Current	lavaa	$T_J = 25$ °C, $R_T/C_T = 2V$	7.8	8.3	8.8	mA
Discharge Current	I _{DISC}	$R_T/C_T = 2V$	7.5		8.8	mA
ERROR-AMPLIFIER SECTION				ı		
Voltage Feedback Input	V_{FB}	COMP =2.5V	2.45	2.5	2.55	V
Input Bias Current	I _{I(BIAS)}			-0.3	-1	μΑ
Open Loop Voltage Gain	G_{VO}	V _{OUT} =2V~4V	65	90		dB
Unity Gain Bandwidth	G _{BW}		0.7	1		MHz
Power Supply Rejection Ratio	PSRR	V _{CC} =12V~25V	60	70		dB
Output Sink Current	I _{SINK}	V _{FB} =2.7V, COMP=1.1V	2	6		mA
Output Source Current	I _{SOURCE}	V _{FB} =2.3V, COMP =5V	-0.5	-0.8		mA
Output Valtage Cuing High State	V _{OH}	V_{FB} =2.3V, R_L =15k Ω to GND	5	6		V
Output Voltage Swing High State	V _{OL}	$V_{FB} = 2.7V$, $R_L = 15k\Omega$ to GND		0.7	1.1	V
CURRENT-SENSE SECTION	•	•				<u>, </u>
Current Sense Input Voltage Gain	G _V	(Note 2,3)	2.85	3	3.15	V/V
Maximum Current Sense Input		COMP. 51/ (Note 2)	0.0	4	4.4	V
Threshold	V_{TH}	COMP =5V (Note 2)	0.9	1	1.1	V
Power Supply Rejection Ratio	PSRR	V _{CC} =12V~25V (Note 2)		70		dB
Input Bias Current	I _{I(BIAS)}			-2	-10	μΑ
Propagation Delay	t _D	V _{FB} =0V~2V		150	300	ns
OUTPUT SECTION						
15.1.1.10.4.17.16	.,	I _{OH} =-20mA	13	13.5		V
High-Level Output Voltage	V _{OH}	I _{OH} =-200mA	12	13.5		V
	V _{OL}	I _{OL} =20mA		0.1	0.4	V
Low-Level Output Voltage		I _{OL} =200mA		1.5	2.2	V
Under-Voltage Lockout Output Voltage	V _{UVLO}	V _{CC} =5V, I _{OL} =1mA		0.7	1.2	V
Output Voltage Rise Time	t _R	C _L =1nF, T _J = 25°C		50	150	ns
Output Voltage Fall Time	t _F	C _L =1nF, T _J = 25°C		50	150	ns
UNDERVOLTAGE-LOCKOUT SECT			1			
Startup Threshold	V _{TH}		7.8	8.4	9	V
Minimum Operating Voltage						
After Start-Up	V _{CC(MIN)}		7	7.6	8.2	V
PULSE-WIDTH MODULATOR SECT	TION		I.	I	I	
Maximum Duty Cycle	D _{C(MAX)}		94	96	100	%
Minimum Duty Cycle	D _{C(MIN)}				0	%
SUPPLY VOLTAGE	- O(MIN)	1	ı	ı	. <u>-</u>	
Power Startup Supply Current	Icc+Ic			0.3	0.5	mA
Power Operating Supply Current	I _{CC} +I _C	V _{FB} and I _{SENSE} at 0V		11	17	mA
Power Supply Zener Voltage	Vz	I _{CC} =25mA	30	34		V
. S. O Supply Lonor Voltage	v ∠	1.00 -2011/1	50	U-T	ı	

Notes: 1. Adjust V_{CC} above the start threshold before setting it to 15V.

- 2. Measured at the trip point of the latch, with VFB at 0V.
- 3. Measured between I_{SENSE} and COMP, with the input changing from 0V ~ 0.8V.

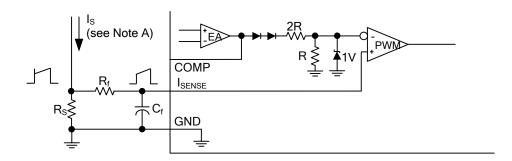
APPLICATION INFORMATION

Error amplifier (EA) configuration circuit:



Note: Error amplifier can source or sink up to 0.5mA.

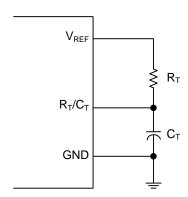
Current-sense circuit:



Notes: 1. Peak current (Is) is determined by the formula: $I_{S(max)} = 1V/R_S$

2. A small RC filter formed by resistor R_{F} and capacitor C_{F} may be required to suppress switch transients.

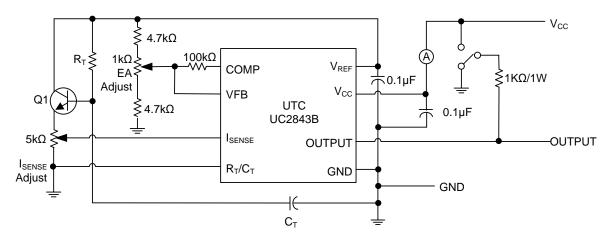
The oscillator frequency is set using the circuit:



■ APPLICATION INFORMATION (Cont.)

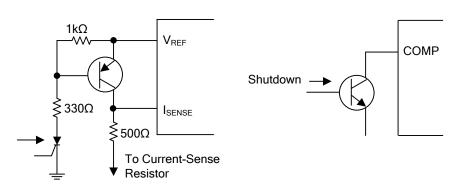
Open-Loop Laboratory Test Fixture

In the open-loop laboratory test fixture, high peak currents and loads need grounding techniques. The transistor and $5-k\Omega$ potentiometer sample the oscillator waveform, applying an adjustable ramp to the I_{SENSE} terminal. Timing and bypass capacitors should be connected closely to the GND terminal in a single-point ground.



Shutdown Technique

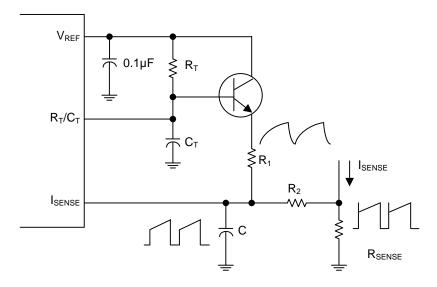
The PWM controller can be shut down through two methods: the one is raising voltage (above 1 V) at I_{SENSE} , the other is pulling the COMP terminal below a voltage two diode drops above ground. Either method can leave the output of the PWM comparator high (refer to block diagram). To reset the PWM latch is dominant so the output can stay low in the case of the next clock cycle is coming and the COMP or I_{SENSE} terminal is removed beyond this shutdown condition. For example, an externally-latched shutdown can be accomplished by adding an SCR reset by cycling V_{CC} below the lower UVLO threshold. So the reference turns off then allows the SCR to reset at this condition.



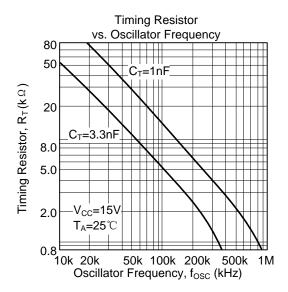
■ APPLICATION INFORMATION (Cont.)

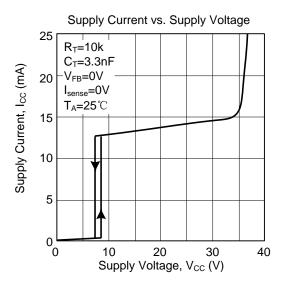
Shutdown Technique (cont.)

A fraction of the triangular-wave oscillator can be summed resistively with the current-sense signal providing slope compensation for converters, which requiring duty cycles over 50%. Please note that capacitor C forms a filter with R2 to suppress the leading-edge switch spikes.



■ TYPICAL CHARACTERISTICS





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