

S-89713 Series

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MINI ANALOG SERIES LOW INPUT OFFSET VOLTAGE CMOS OPERATIONAL AMPLIFIER

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The mini-analog series is a group of ICs that incorporate a general purpose analog circuit in a small package. The S-89713 Series is an auto-zero operation, zero-drift operational amplifier that has input and output of low input offset voltage and Rail-to-Rail*¹. The S-89713 Series is suitable for applications requiring less offset voltage. The S-89713 Series is a dual operational amplifier (with 2 circuits).

*1. Rail-to-Rail is a trademark of Motorola, Inc.

■ Features

• Low input offset voltage: V_{IO} = 10 μ V max. (Ta = +25°C)

• Low voltage operation: $V_{DD} = 2.65 \text{ V}$ to 5.5 V

• Low current consumption: I_{DD} = 165 μA typ. (Per circuit, Ta = +25°C) I_{DD} = 330 μA typ. (2 circuits, Ta = +25°C)

• No external parts required for internal phase compensation

• Rail-to-Rail input and output

• Operation temperature range: $Ta = -40^{\circ}C \text{ to } +85^{\circ}C$

• Lead-free (Sn 100%), halogen-free

■ Applications

- · Various sensor interfaces
- · High-accuracy current detection
- Strain gauge amplifier
- Game
- · Various electric devices

■ Packages

- TMSOP-8
- SNT-8A

■ Block Diagram

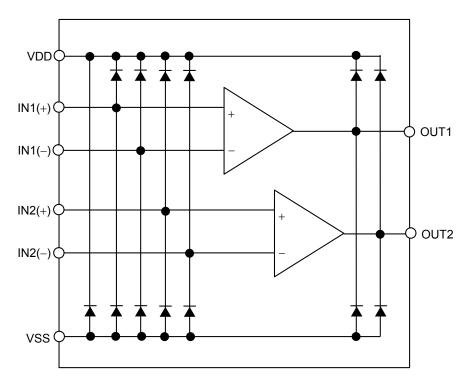


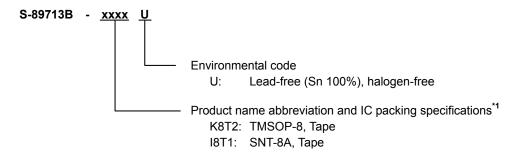
Figure 1

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■ Product Name Structure

Users can select the package type for the S-89713 Series. Refer to "1. Product name" regarding the contents of product name, "2. Packages" regarding the package drawings and "3. Product name list" regarding the product type.

1. Product name



^{*1.} Refer to the tape drawing.

2. Packages

Table 1 Package Drawing Codes

Package Name	Dimension	Tape	Reel	Land
TMSOP-8	FM008-A-P-SD	FM008-A-C-SD	FM008-A-R-SD	_
SNT-8A	PH008-A-P-SD	PH008-A-C-SD	PH008-A-R-SD	PH008-A-L-SD

3. Product name list

Table 2

Product Name	Package
S-89713B-K8T2U	TMSOP-8
S-89713B-I8T1U	SNT-8A

■ Pin Configurations

1. TMSOP-8

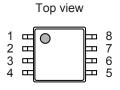


Figure 2

Table 3

Pin No.	Symbol	Description
1	OUT1	Output pin 1
2	IN1(-)	Inverted input pin 1
3	IN1(+)	Non-inverted input pin 1
4	VSS	GND pin
5	IN2(+)	Non-inverted input pin 2
6	IN2(-)	Inverted input pin 2
7	OUT2	Output pin 2
8	VDD	Positive power supply pin

2. SNT-8A

Figure 3

Table 4

Pin No.	Symbol	Description
1	OUT1	Output pin 1
2	IN1(-)	Inverted input pin 1
3	IN1(+)	Non-inverted input pin 1
4	VSS	GND pin
5	IN2(+)	Non-inverted input pin 2
6	IN2(-)	Inverted input pin 2
7	OUT2	Output pin 2
8	VDD	Positive power supply pin

■ Absolute Maximum Ratings

Table 5

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Item		Symbol	Absolute Maximum Rating	Unit
Power supply voltage	је	V_{DD}	$V_{SS}-0.3$ to $V_{SS}+7.0$	V
Input voltage		$V_{IN(+)}$, $V_{IN(-)}$	$V_{SS}-0.3$ to $V_{DD}+0.3$	V
Output voltage		V _{OUT}	$V_{SS}-0.3$ to $V_{DD}+0.3$	V
Differential input vol	tage	V_{IND}	±5.5	V
Outside size successful		I _{SOURCE}	10.0	mA
Output pin current		I _{SINK}	10.0	mA
Dawer dissination	TMSOP-8		650 ^{*1}	mW
Power dissipation SNT-8A		P _D	450 ^{*1}	mW
Operation ambient temperature		Topr	−40 to +85	°C
Storage temperature	е	T _{stg}	−55 to +125	°C

*1. When mounted on board

[Mounted board]

(1) Board size: $114.3 \text{ mm} \times 76.2 \text{ mm} \times t1.6 \text{ mm}$ (2) Board name: JEDEC STANDARD51-7

Caution The absolute maximum ratings are rated values exceeding which the product could suffer physical damage. These values must therefore not be exceeded under any conditions.

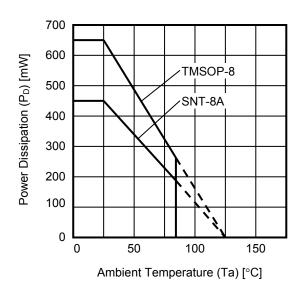


Figure 4 Power Dissipation of Package (When Mounted on Board)

■ Electrical Characteristics

Table 6

DC Electrical Characteristics $(V_{DD} = 3.0 \text{ V}, \text{ Ta} = +25^{\circ}\text{C} \text{ unless otherwise specified})$

DO LIECTICAI OHAIACTEHSTICS		(v DD —	0.0 V, Tu	120 0	urness oure	TWISC SP	Jeoniea ,
ltem	Symbol	Condition	Min.	Тур.	Max.	Unit	Test Circuit
Operation power supply voltage range	V_{DD}	-	2.65	3.00	5.50	V	1
Current consumption (for 2 circuits)	I _{DD}	$V_{CMR} = V_{OUT} = V_{DD} / 2$	_	330	380	μΑ	5
Input offset voltage	V _{IO}	$V_{CMR} = V_{DD} / 2$	-10	±1	+10	μV	1
Input offset voltage drift	$\frac{\Delta V_{10}}{\Delta Ta}$	V _{CMR} = V _{DD} / 2	_	±0.1	_	μV/°C	1
Input offset current	I _{IO}	-	_	±140	-	pА	_
Input bias current	I _{BIAS}	Т	_	±70	_	pА	_
Common-mode input voltage range	V _{CMR}	-	V _{SS} - 0.1	-	$V_{DD} + 0.1$	٧	2
Voltage gain (open loop)	A _{VOL}	$\begin{split} V_{SS} + 0.1 &~V \leq V_{OUT} \leq V_{DD} - 0.1 &~V, \\ V_{CMR} = V_{DD} / 2, &~R_L = 10 &~k\Omega \end{split} \label{eq:VSS}$	110	130	-	dB	8
Maximum autout autou valtana	V _{OH}	$R_L = 10 \text{ k}\Omega$	2.9	_	_	>	3
Maximum output swing voltage	V_{OL}	$R_L = 10 \text{ k}\Omega$	-	_	0.1	V	4
Common-mode input signal rejection ratio	CMRR	$V_{SS}-0.1~V \leq V_{CMR} \leq V_{DD}+0.1~V$	106	130	_	dB	2
Power supply voltage rejection ratio	PSRR	V _{DD} = 2.65 V to 5.50 V	106	120	-	dB	1
Source current	I _{SOURCE}	$V_{OUT} = V_{DD} - 0.1 V$	1.3	1.6	_	mA	6
Sink current	I _{SINK}	V _{OUT} = 0.1 V	1.6	2.0	_	mA	7

Table 7

AC Electrical Characteristics

 $(V_{DD} = 3.0 \text{ V}, \text{ Ta} = +25^{\circ}\text{C} \text{ unless otherwise specified})$

710 =1001110411 01141141011011		(*BB eie i,		0 0000	01110111100	
Item	Symbol	Condition	Min.	Тур.	Max.	Unit
Slew rate	SR	$R_L = 1.0 \text{ M}\Omega, C_L = 15 \text{ pF}$ (Refer to Figure 13)	_	0.16	_	V/μs
Gain-bandwidth product	GBP	$C_1 = 0 pF$	_	240	_	kHz

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■ Test Circuits (Per circuit)

1. Power supply voltage rejection ratio, input offset voltage

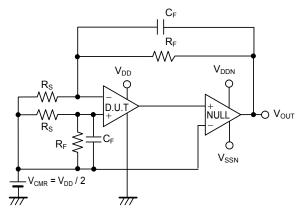


Figure 5 Test Circuit 1

• Power supply voltage rejection ratio (PSRR)

The power supply voltage rejection ratio (PSRR) can be calculated by the following expression, with V_{OUT} measured at each V_{DD} .

Test conditions:

$$V_{DD} = 2.65 \text{ V}$$
: $V_{DD} = V_{DD1}$, $V_{OUT} = V_{OUT1}$
 $V_{DD} = 5.5 \text{ V}$: $V_{DD} = V_{DD2}$, $V_{OUT} = V_{OUT2}$

$$PSRR = 20 log \left(\left| \frac{V_{DD1} - V_{DD2}}{\left(V_{OUT1} - \frac{V_{DD1}}{2}\right) - \left(V_{OUT2} - \frac{V_{DD2}}{2}\right)} \right| \times \frac{R_F + R_S}{R_S} \right)$$

• Input offset voltage (V_{IO})

$$V_{IO} = \left(V_{OUT} - \frac{V_{DD}}{2}\right) \times \frac{R_S}{R_F + R_S}$$

2. Common-mode input signal rejection ratio, common-mode input voltage range

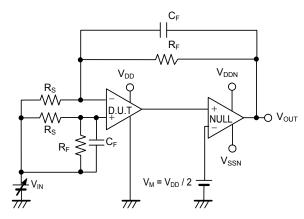


Figure 6 Test Circuit 2

• Common-mode input signal rejection ratio (CMRR)

The common-mode input signal rejection ratio (CMRR) can be calculated by the following expression, with V_{OUT} measured at each V_{IN} .

Test conditions:

$$V_{IN} = V_{CMR \ Max.}$$
: $V_{IN} = V_{IN1}$, $V_{OUT} = V_{OUT1}$
 $V_{IN} = V_{CMR \ Min.}$: $V_{IN} = V_{IN2}$, $V_{OUT} = V_{OUT2}$

CMRR = 20 log
$$\left(\left| \frac{V_{\text{IN1}} - V_{\text{IN2}}}{(V_{\text{OUT1}} - V_{\text{IN1}}) - (V_{\text{OUT2}} - V_{\text{IN2}})} \right| \times \frac{R_{\text{F}} + R_{\text{S}}}{R_{\text{S}}} \right)$$

• Common-mode input voltage range (V_{CMR})

The common-mode input voltage range is the range of V_{IN} in which V_{OUT} satisfies the common-mode input signal rejection ratio specifications.

3. Maximum output swing voltage

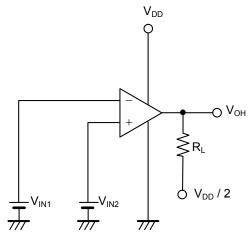


Figure 7 Test Circuit 3

• Maximum output swing voltage (V_{OH})

Test conditions:

Test conditions:

$$V_{IN1} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

$$V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

$$R_{I} = 10 \text{ kg}$$

$$V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$$

$$R_1 = 10 \text{ kO}$$

4. Maximum output swing voltage

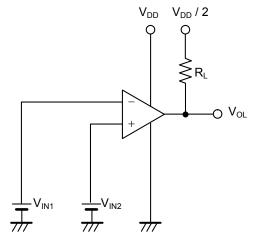


Figure 8 Test Circuit 4

• Maximum output swing voltage (Vol)

Test conditions:

$$V_{\text{IN1}} = \frac{V_{\text{DD}}}{2} + 0.1 \text{ V}$$

$$V_{\text{IN2}} = \frac{V_{\text{DD}}}{2} - 0.1 \text{ V}$$

$$R_{\text{L}} = 10 \text{ k}\Omega$$

$$V_{IN2} = \frac{V_{DD}}{2} - 0.1 \text{ V}$$

$$R_1 = 10 \text{ k}\Omega$$

5. Current consumption

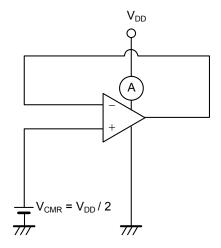


Figure 9 Test Circuit 5

6. Source current

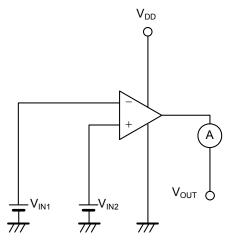


Figure 10 Test Circuit 6

• Current consumption (I_{DD})

• Source current (I_{SOURCE})

Test conditions: $V_{OUT} = V_{DD} - 0.1 \text{ V}$ $V_{IN1} = \frac{V_{DD}}{2} - 0.1 \text{ V}$ $V_{IN2} = \frac{V_{DD}}{2} + 0.1 \text{ V}$

7. Sink current

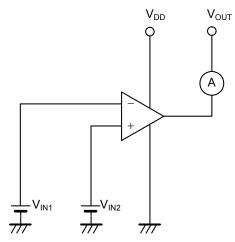


Figure 11 Test Circuit 7

• Sink current (I_{SINK})

Test conditions: V_{OUT} = 0.1 V V_{IN1} = $\frac{V_{DD}}{2}$ + 0.1 V V_{IN2} = $\frac{V_{DD}}{2}$ - 0.1 V

8. Voltage gain

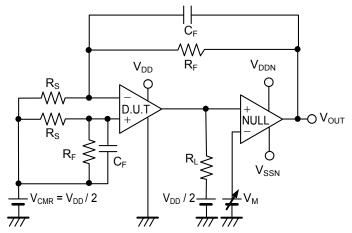


Figure 12 Test Circuit 8

• Voltage gain (open loop) (A_{VOL})

The voltage gain (A $_{VOL}$) can be calculated by the following expression, with V_{OUT} measured at each V_{M} .

Test conditions:

$$V_M = V_{DD} - 0.1 \text{ V: } V_M = V_{M1}, V_{OUT} = V_{OUT1}$$

 $V_M = 0.1 \text{ V: } V_M = V_{M2}, V_{OUT} = V_{OUT2}$

$$\begin{split} A_{VOL} &= 20 \ log \ \left(\left| \frac{V_{M1} - V_{M2}}{V_{OUT1} - V_{OUT2}} \right| \ \times \ \frac{R_F + R_S}{R_S} \right) \\ R_L &= 10 \ k\Omega \end{split}$$

9. Slew rate

Measured by the voltage follower circuit.

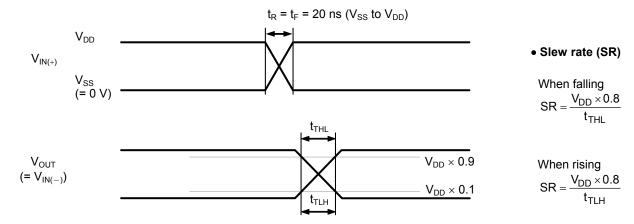


Figure 13

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■ Usage Example

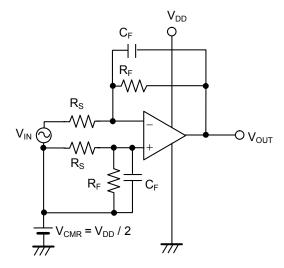


Figure 14 Differential Amplifier Circuit

[Example of Gain = 1000 times] $R_S = 1 \text{ k}\Omega$ $R_F = 1 \text{ M}\Omega$ $C_F = 1000 \text{ pF}$

[Example of Gain = 100 times] $R_S = 1 \text{ } k\Omega$ $R_F = 100 \text{ } k\Omega$ $C_F = 1000 \text{ } pF$

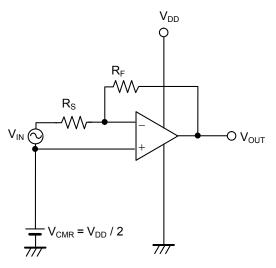


Figure 15 Inverting Amplifier Circuit

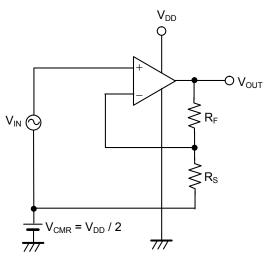


Figure 16 Non-inverting Amplifier Circuit

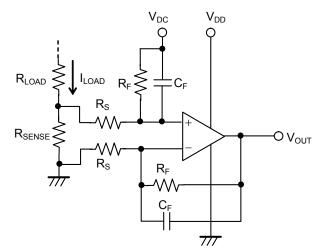


Figure 17 Low-side Current Detection Circuit

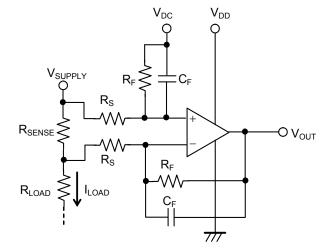


Figure 18 High-side Current Detection Circuit

Caution The above connection diagram and constant will not guarantee successful operation. Perform through evaluation using the actual application to set the constant.

■ Precautions

- Generally an operational amplifier may cause oscillation depending on the selection of external parts. Perform thorough evaluation using the actual application to set the constant.
- Do not apply an electrostatic discharge to this IC that exceeds performance ratings of the built-in electrostatic protection circuit.
- ABLIC Inc. claims no responsibility for any disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.
- Use this IC with the output current of 10 mA or less.
- When the output voltage is used in the range of V_{DD} 100 mV or more, or V_{SS} + 100 mV or less, the operation
 may become unstable depending on the circuit configuration. Contact our sales office for details.
- When using the voltage follower circuit (Gain = 1 time), insert a resistor of 470Ω or more for the stable operation, as shown in **Figure 19**. The operation may be unstable depending on the value of the load capacitance connected to the output pin, even when the voltage follower circuit is not used. Use the product under thorough evaluation.

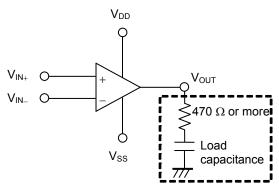


Figure 19

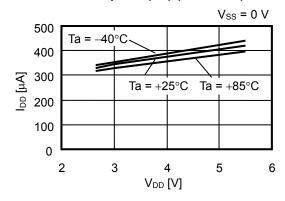
Caution The above connection diagram and constant will not guarantee successful operation.

Perform through evaluation using the actual application to set the constant.

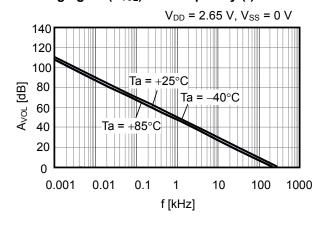
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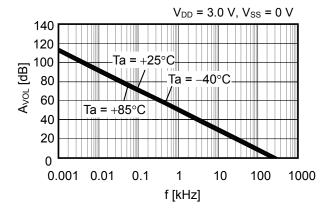
■ Characteristics (Typical Data)

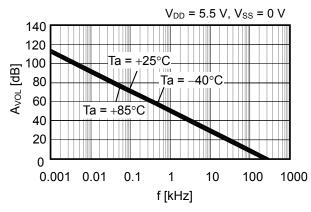
1. Current consumption (I_{DD}) (2 circuits) vs. Power supply voltage (V_{DD})



2. Voltage gain (A_{VOL}) vs. Frequency (f)

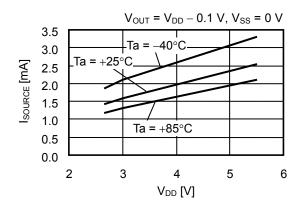




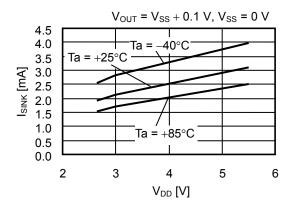


3. Output current

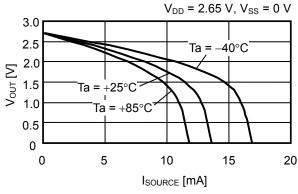
3. 1 Source current (I_{SOURCE}) vs. Power supply voltage (V_{DD})

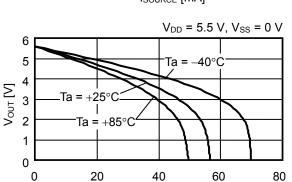


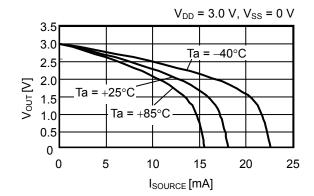
3. 2 Sink current (I_{SINK}) vs. Power supply voltage(V_{DD})



3. 3 Output voltage (Vout) vs. Source current (Isource)



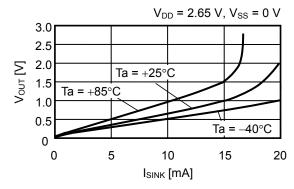


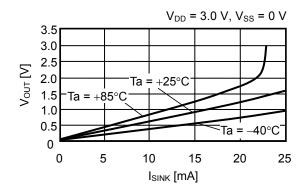


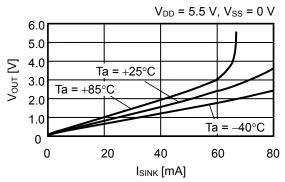
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I_{SOURCE} [mA]

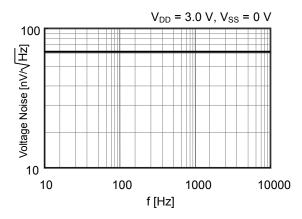
3. 4 Output voltage (V_{OUT}) vs. Sink current (I_{SINK})





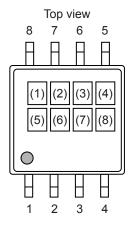


4. Input-referred noise voltage vs. Frequency (f)



■ Marking Specifications

1. TMSOP-8



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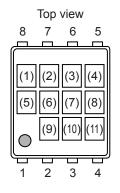
(2) to (4): Product code (Refer to Product name vs. Product code)

(5): Blank(6) to (8): Lot number

Product name vs. Product code

Product Name	Product Code			
Floduct Name	(2)	(3)	(4)	
S-89713B-K8T2U	Z	Υ	С	

2. SNT-8A



(1): Blank

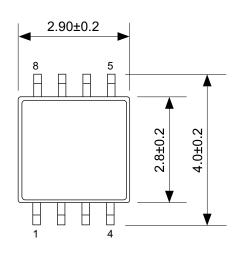
(2) to (4): Product code (Refer to **Product name vs. Product code**)

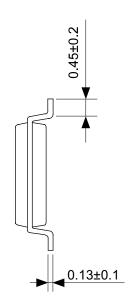
(5), (6): Blank

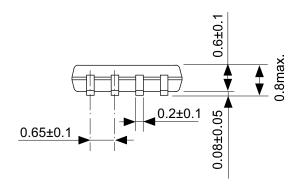
(7) to (11): Lot number

Product name vs. Product code

Product Name	Product Code			
Product Name	(2)	(3)	(4)	
S-89713B-I8T1U	Z	Υ	С	

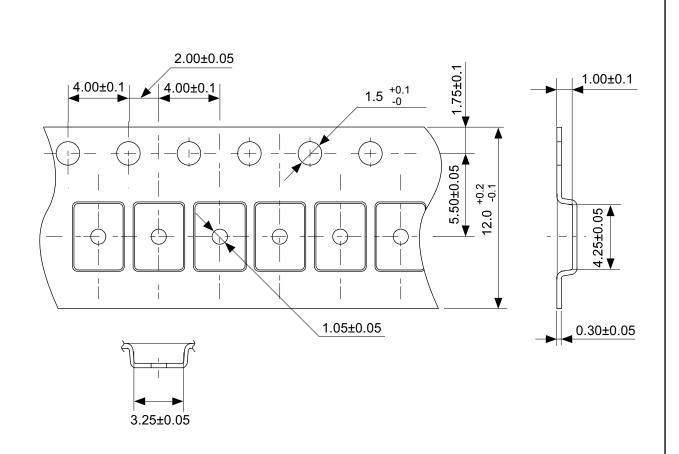


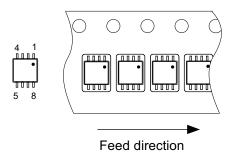




No. FM008-A-P-SD-1.2

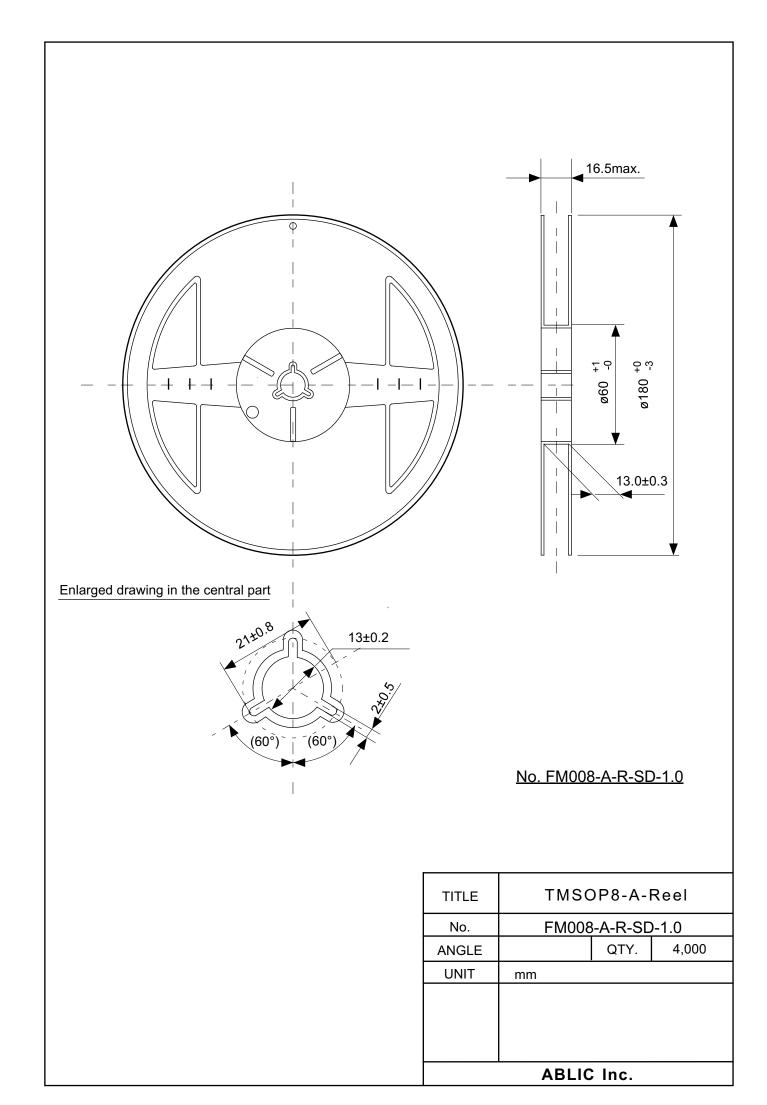
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No.	FM008-A-P-SD-1.2	
ANGLE	Q	
UNIT	mm	
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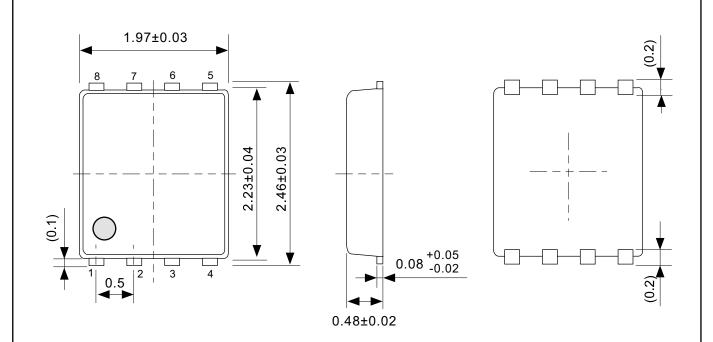


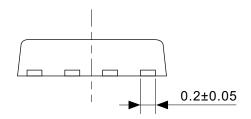


No. FM008-A-C-SD-2.0

TITLE	TMSOP8-A-Carrier Tape	
No.	FM008-A-C-SD-2.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		

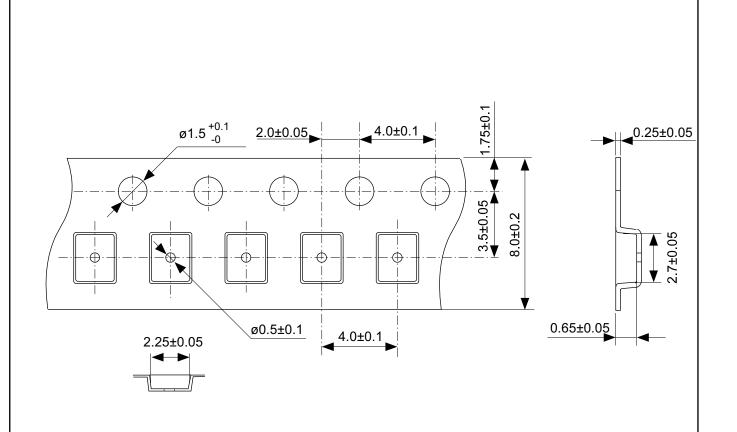


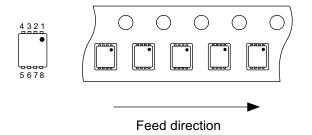




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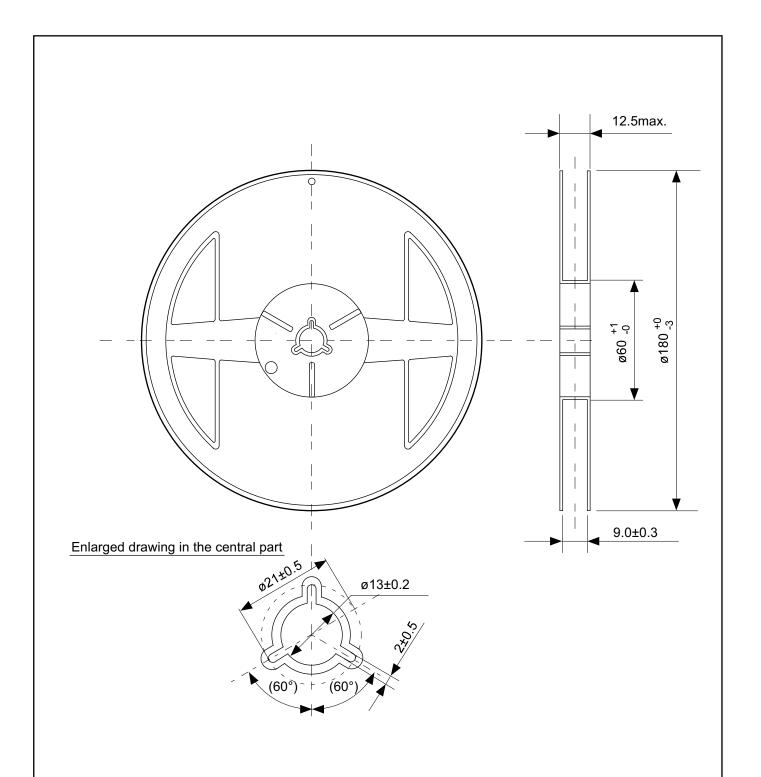
TITLE	SNT-8A-A-PKG Dimensions	
No.	PH008-A-P-SD-2.1	
ANGLE	$\Phi \Box$	
UNIT	mm	
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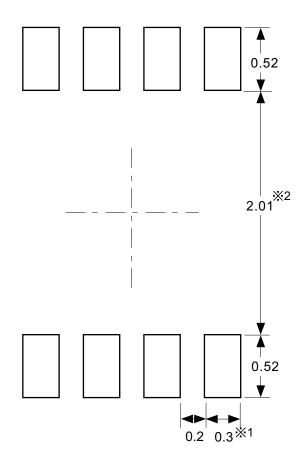
No. PH008-A-C-SD-2.0

TITLE	SNT-8A-A-Carrier Tape	
No.	PH008-A-C-SD-2.0	
ANGLE		
UNIT	mm	
ABLIC Inc.		



No. PH008-A-R-SD-1.0

TITLE	SNT-8A-A-Reel			
No.	PH008-A-R-SD-1.0			
ANGLE		QTY.	5,000	
UNIT	mm			
ABLIC Inc.				



- %1. ランドパターンの幅に注意してください (0.25 mm min. / 0.30 mm typ.)。 %2. パッケージ中央にランドパターンを広げないでください (1.96 mm \sim 2.06 mm)。
- 注意 1. パッケージのモールド樹脂下にシルク印刷やハンダ印刷などしないでください。
 - 2. パッケージ下の配線上のソルダーレジストなどの厚みをランドパターン表面から0.03 mm 以下にしてください。
 - 3. マスク開口サイズと開口位置はランドパターンと合わせてください。
 - 4. 詳細は "SNTパッケージ活用の手引き"を参照してください。
- X1. Pay attention to the land pattern width (0.25 mm min. / 0.30 mm typ.).
- X2. Do not widen the land pattern to the center of the package (1.96 mm to 2.06mm).
- Caution 1. Do not do silkscreen printing and solder printing under the mold resin of the package.
 - 2. The thickness of the solder resist on the wire pattern under the package should be 0.03 mm or less from the land pattern surface.
 - 3. Match the mask aperture size and aperture position with the land pattern.
 - 4. Refer to "SNT Package User's Guide" for details.
- ※1. 请注意焊盘模式的宽度 (0.25 mm min. / 0.30 mm typ.)。
- ※2. 请勿向封装中间扩展焊盘模式 (1.96 mm~2.06 mm)。
- 注意 1. 请勿在树脂型封装的下面印刷丝网、焊锡。
 - 2. 在封装下、布线上的阻焊膜厚度 (从焊盘模式表面起) 请控制在 0.03 mm 以下。
 - 3. 钢网的开口尺寸和开口位置请与焊盘模式对齐。
 - 4. 详细内容请参阅 "SNT 封装的应用指南"。

No. PH008-A-L-SD-4.1

TITLE	SNT-8A-A -Land Recommendation	
No.	PH008-A-L-SD-4.1	
ANGLE		
UNIT	mm	
ABLIC Inc.		

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