

# **3-Pin Microprocessor Reset Monitors**

The HG809 and HG810 are cost–effective system supervisor circuits designed to monitor  $V_{CC}$  in digital systems and provide a reset signal to the host processor when necessary. No external components are required.

The reset output is driven active within 20  $\mu$ sec of V<sub>CC</sub> falling through the reset voltage threshold. Reset is maintained active for a minimum of 140msec after V<sub>CC</sub> rises above the reset threshold. The

HG810 has an <u>active-high RESET</u> output while the HG809 has an active-low RESET output. The output of the HG809 is guaranteed valid down to  $V_{CC} = 1V$ . Both devices are available in a SOT-23 package.

The HG809/810 are optimized to reject fast transient glitches on the  $V_{CC}$  line. Low supply current of 17µA ( $V_{CC}$  = 3.3V) makes these devices suitable for battery powered applications.

#### **Features**

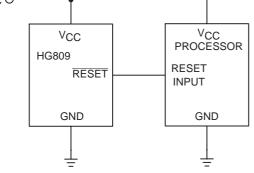
- Precision V<sub>CC</sub> Monitor for 3.0V, 3.3V, and 5.0V Supplies
- 140msec Guaranteed Minimum RESET, RESET Output Duration
- **RESET** Output Guaranteed to  $V_{CC} = 1.0V$  (HG809)
- Low 17µA Supply Current
- V<sub>CC</sub> Transient Immunity
- Small SOT–23 Package
- No External Components
- Wide Operating Temperature: -40°C to 85°C

## **Typical Applications**

- Computers
- Embedded Systems
- Battery Powered Equipment
- Critical µP Power Supply Monitoring

#### **TYPICAL APPLICATION DIAGRAM**

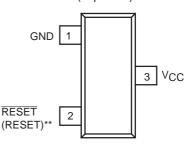






SOT-23 (TO-236) CASE 318





SOT-23\*

NOTE: \*SOT-23 is equivalent to JEDEC (TO-236) \*\* RESET is for HG809 \*\* RESET is for HG810

NOTE: The "x" denotes a suffix for  $\mathsf{V}_{CC}$  threshold – see table below

Suffix	Reset V <sub>CC</sub> Threshold (V)	
L	4.63	
M	4.38	
J*	4.00	
Т	3.08	
S	2.93	
R	2.63	

NOTE: \*J version is available for HG809 only



## **ABSOLUTE MAXIMUM RATINGS\***

Symbol	Parameter	Value	Unit
	Supply Voltage (V <sub>CC</sub> to GND)	6.0	V
	RESET, RESET	–0.3 to (V <sub>CC</sub> + 0.3)	V
	Input Current, V <sub>CC</sub>	20	mA
	Output Current, RESET, RESET	20	mA
	dV/dt (V <sub>CC</sub> )	100	V/µsec
PD	Power Dissipation (T <sub>A</sub> $\leq$ 70°C) SOT–23 (derate 4mW/°C above +70°C)	230 mW	
ТА	Operating Temperature Range	-40 to +85 °C	
T <sub>stg</sub>	Storage Temperature Range	-65 to +150 °C	
T <sub>sol</sub>	Lead Temperature (Soldering, 10 Seconds)	+260 °C	

\* Maximum Ratings are those values beyond which damage to the device may occur.



Symbol	Characteristic	Min	Тур	Max	Unit
	$V_{CC} Range$ $T_{A} = 0^{\circ}C \text{ to } +70^{\circ}C$ $T_{A} = -40^{\circ}C \text{ to } +85^{\circ}C$	1.0 1.2		5.5 5.5	V
ICC	Supply Current HG8xxL/M/J: V <sub>CC</sub> < 5.5V HG8xxR/S/T: V <sub>CC</sub> < 3.6V		24 17	60 50	μΑ
VTH	$ \begin{array}{l} \mbox{Reset Threshold (Note NO TAG)} \\ \mbox{HG8xxL: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{T}_{A} = -40^{\circ}\mbox{C} \ to +85^{\circ}\mbox{C} \\ \mbox{HG809J: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG809J: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG80xT: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xxT: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xxS: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xxS: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xxR: } T_{A} = -40^{\circ}\mbox{C} \ to +85^{\circ}\mbox{C} \\ \mbox{HG8xxR: } T_{A} = -40^{\circ}\mbox{C} \ to +85^{\circ}\mbox{C} \\ \mbox{HG8xxR: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xR: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xR: } T_{A} = 25^{\circ}\mbox{C} \\ \mbox{HG8xR: } T_$	4.56 4.50 4.31 4.25 3.93 3.89 3.04 3.00 2.89 2.85 2.59 2.55	4.63 — 4.38 — 4.00 — 3.08 — 2.93 — 2.63 —	4.70 4.75 4.45 4.50 4.06 4.10 3.11 3.15 2.96 3.00 2.66 2.70	V
	Reset Threshold Temperature Coefficient	—	30	-	ppm/°C
	V <sub>CC</sub> to Reset Delay V <sub>CC</sub> = V <sub>TH</sub> to (V <sub>TH</sub> – 100mV)	—	20	- 1	μsec
	Reset Active Timeout Period	140	240	560	msec
V <sub>OL</sub>	RESET Output Voltage Low (HG809) HG809R/S/T: V <sub>CC</sub> = V <sub>TH</sub> min, I <sub>SINK</sub> = 1.2mA HG809L/M/J: V <sub>CC</sub> = V <sub>TH</sub> min, I <sub>SINK</sub> = 3.2mA V <sub>CC</sub> = V <sub>TH</sub> min, I <sub>SINK</sub> = 3.2mA V <sub>CC</sub> > 1.0V, I <sub>SINK</sub> = 50µA I <sub>SINK</sub> = 50µA I <sub>SINK</sub> = 50µA I <sub>SINK</sub>			0.3 0.4 0.3	V
VOH	RESET Output Voltage High (HG809) HG809R/S/T: V <sub>CC</sub> > V <sub>TH</sub> max, I <sub>SOURCE</sub> = 500μA HG809L/M/J: V <sub>CC</sub> > V <sub>TH</sub> max, I <sub>SOURCE</sub> = 800μA	0.8 V <sub>CC</sub> V <sub>CC</sub> – 1.5			V
V <sub>OL</sub>	RESET Output Voltage Low (HG810) HG810R/S/T: V <sub>CC</sub> = V <sub>TH</sub> max, I <sub>SINK</sub> = 1.2mA HG810L/M/J: V <sub>CC</sub> = V <sub>TH</sub> max, I <sub>SINK</sub> = 3.2mA			0.3 0.4	V
VOH	RESET Output Voltage High (HG810) 1.8 < V <sub>CC</sub> < V <sub>TH</sub> min, I <sub>SOURCE</sub> = 150μA	0.8 V <sub>CC</sub>	_	_	V

**ELECTRICAL CHARACTERISTICS** ( $V_{CC}$  = Full Range,  $T_A$  = -40°C to +85°C unless otherwise noted. typical values are at  $T_A$  = +25C,  $V_{CC}$  = 5V for L/M/J, 3.3V for T/S, 3.0V for R) (Note NO TAG)

1. Production testing done at  $T_A = 25^{\circ}C$ , over temperature limits guaranteed by design.

# **PIN DESCRIPTION**

Pin No.	Symbol	Description
1	GND	Ground
2	RESET (HG809)	$\overline{\text{RESET}}$ output remains low while V_CC is below the reset voltage threshold, and for 240msec (typ.) after V_CC rises above reset threshold
2	RESET (HG810)	RESET output remains high while V <sub>CC</sub> is below the reset voltage threshold, and for 240msec (typ.) after V <sub>CC</sub> rises above reset threshold
3	VCC	Supply Voltage (typ.)



### **APPLICATIONS INFORMATION**

#### V<sub>CC</sub> Transient Rejection

The HG809/810 provides accurate  $V_{CC}$  monitoring and reset timing during power–up, power–down, and brownout/sag conditions, and rejects negative–going transients (glitches) on the power supply line. Figure 1 shows the maximum transient duration vs. maximum negative excursion (overdrive) for glitch rejection. Any combination of duration and overdrive which lies **under** the curve will **not** generate a reset signal. Combinations above the curve are detected as a brownout or power–down. Transient immunity can be improved by adding a capacitor in close proximity to the V<sub>CC</sub> pin of the HG809/810.

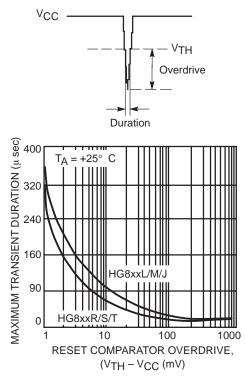


Figure 1. Maximum Transient Duration vs. Overdrive for Glitch Rejection at 25° C

## **RESET** Signal Integrity During Power–Down

The HG809 RESET output is valid to  $V_{CC} = 1.0V$ . Below this voltage the output becomes an "open circuit" and does not sink current. This means CMOS logic inputs to the  $\mu$ P will be floating at an undetermined voltage. Most digital systems are completely shutdown well above this voltage. However, in situations where RESET must be maintained valid to  $V_{CC} = 0V$ , a pull-down resistor must be connected from RESET to ground to discharge stray capacitances and hold the output low (Figure 2). This resistor value, though not critical, should be chosen such that it does not appreciably load  $\overline{\text{RESET}}$  under normal operation (100k $\Omega$ will be suitable for most applications). Similarly, a pull–up resistor to V<sub>CC</sub> is required for the HG810 to ensure a valid high  $\overline{\text{RESET}}$  for V<sub>CC</sub> below 1.0V.

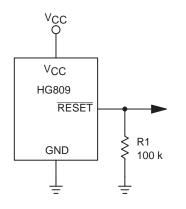


Figure 2. Ensuring RESET Valid to V<sub>CC</sub> = 0 V

### **Processors With Bidirectional I/O Pins**

Some  $\mu$ P's (such as Motorola 68HC11) have bi–directional reset pins. Depending on the current drive capability of the processor pin, an indeterminate logic level may result if there is a logic conflict. This can be avoided by adding a 4.7k $\Omega$  resistor in series with the output of the HG809/810 (Figure 3). If there are other components in the system which require a reset signal, they should be buffered so as not to load the reset line. If the other components are required to follow the reset I/O of the  $\mu$ P, the buffer should be connected as shown with the solid line.

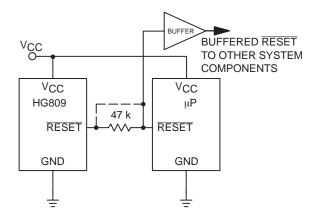


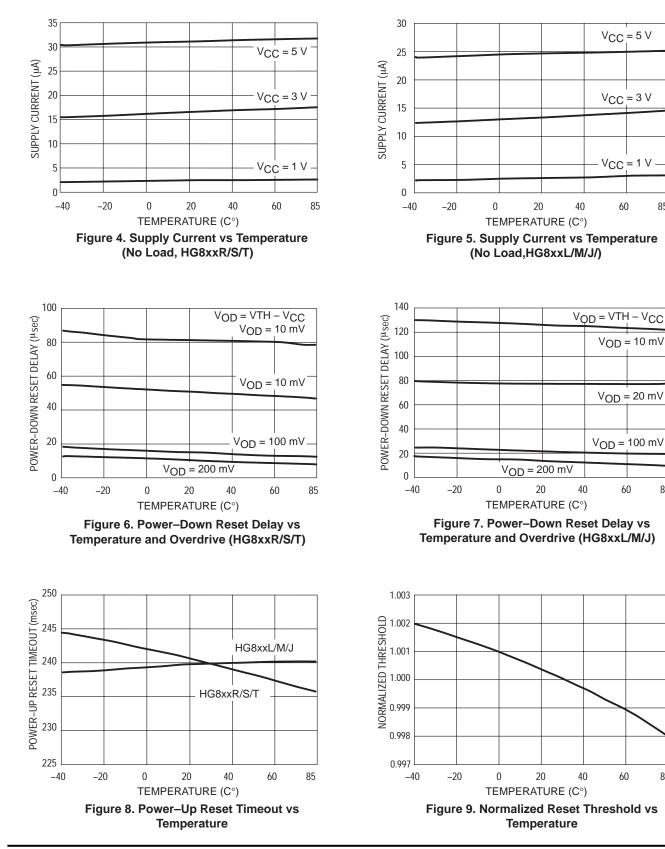
Figure 3. Interfacing to Bidirectional Reset I/O

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# **TYPICAL CHARACTERISTICS**



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