

bq24311 过压和过流保护 IC 以及 锂电池充电器前端保护 IC

1 特性

- 针对三个变量提供保护：
 - 输入过压、快速响应小于 1 μ s
 - 带有电流限制的用户可编程过流
 - 电池过压
- 30V 最大输入电压
- 支持高达 0.3A 的输入电流
- 防止由电流瞬变造成的错误触发
- 过热保护
- 使能输入
- 状态指示

2 应用

- 手机和智能电话
- 掌上电脑 (PDA)
- MP3 播放器
- 低功耗手持器件
- Bluetooth™ 耳机

4 应用信息

3 说明

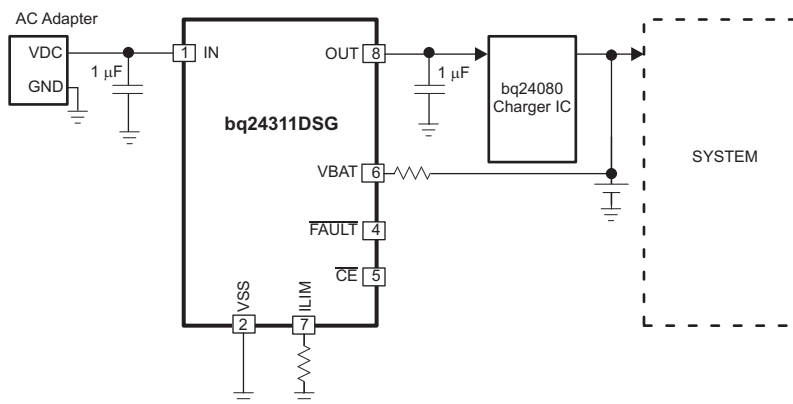
bq24311 是一个高度集成的电路，旨在保护锂离子电池免受充电电路故障的影响。该 IC 可持续监视输入电压、输入电流和电池电压。输入过压保护通过关断内部开关立即停止为充电电路供电。输入保护可将系统电流限制为用户可编程的值，如果过流情况仍存在，则在一个消隐周期之后关断导通元件。此外，该 IC 还会监控自身的芯片温度，并在过热时切断电源。

该 IC 可由一个处理器控制并且可为主机提供关于故障条件的状态信息。

器件信息

部件号	封装	封装尺寸 (标称值)
bq24311	WSON (8)	2.00mm x 2.00mm

(1) 如需了解所有可用封装，请见数据表末尾的可订购产品附录。



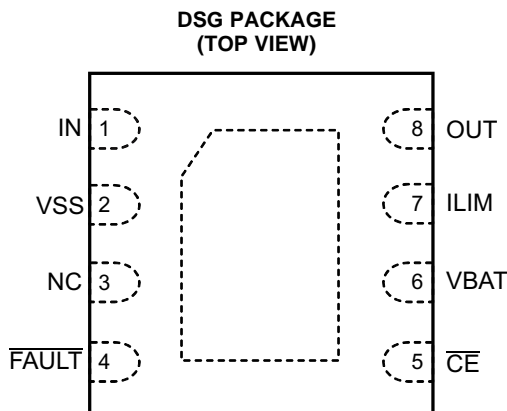
目录

1	特性	1	8.1	Overview	8
2	应用	1	8.2	Functional Block Diagram	8
3	说明	1	8.3	Feature Description	9
4	应用信息	1	8.4	Device Functional Modes	9
5	修订历史记录	2	9	Application and Implementation	12
6	Pin Configuration and Functions	3	9.1	Typical Application Circuit	12
7	Specifications	4	10	Power Supply Requirements	16
	7.1 Absolute Maximum Ratings	4	11	Layout	17
	7.2 Handling Ratings	4	11.1	Layout Guidelines	17
	7.3 Recommended Operating Conditions	4	11.2	Layout Example	17
	7.4 Thermal Information	4	12	器件和文档支持	18
	7.5 Electrical Characteristics	5	12.1	Trademarks	18
	7.6 Timing Requirements	5	12.2	Electrostatic Discharge Caution	18
	7.7 Typical Characteristics	6	12.3	术语表	18
8	Detailed Description	8	13	机械封装和可订购信息	18

5 修订历史记录

日期	修订版本	注释
6 月	*	最初发布。

6 Pin Configuration and Functions



Pin Functions

PIN		I/O	DESCRIPTION
NAME	DSG		
IN	1	I	Input power, connect to external DC supply. Connect external 1µF ceramic capacitor (minimum) to VSS.
OUT	8	O	Output pin to the charging system. Connect external 1 µF ceramic capacitor (minimum) to VSS.
VBAT	6	I	Battery voltage sense input. Connect to pack positive pin through a resistor.
ILIM	7	I/O	Input overcurrent threshold programming. Connect a resistor to VSS to set the overcurrent threshold.
\overline{CE}	5	I	Chip enable input. Active low. When \overline{CE} = High, the input FET is off. Internally pulled down.
\overline{FAULT}	4	O	Device status, open-drain output. \overline{FAULT} = Low indicates that the input FET Q1 has been turned on due to input overvoltage, input overcurrent, battery overvoltage, or thermal shutdown.
VSS	2	–	Ground pin
NC	3		This pin may have internal circuits used for test purposes. Do not make any external connections at these pins for normal operation.
Thermal PAD		–	There is an internal electrical connection between the exposed thermal pad and the VSS pin of the device. The thermal pad must be connected to the same potential as the VSS pin on the printed circuit board. Do not use the thermal pad as the primary ground input for the device. The VSS pin must be connected to ground at all times.

7 Specifications

7.1 Absolute Maximum Ratings⁽¹⁾

over operating free-air temperature range (unless otherwise noted)

PARAMETER		MIN	MAX	UNIT
Input voltage	IN (with respect to VSS)	-0.3	30	V
	OUT (with respect to VSS)	-0.3	12	
	ILIM, FAULT, CE, VBAT (with respect to VSS)	-0.3	7	
Input current	IN		0.5	A
Output current	OUT		0.5	A
Output sink current	FAULT		15	mA
Junction temperature, T _J		-40	150	°C

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

7.2 Handling Ratings

		MIN	MAX	UNIT	
T _{stg}	Storage temperature range	-65	150	°C	
V _{ESD}	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001, all pins ⁽¹⁾	-2000	2000	V
		Charged device model (CDM), per JEDEC specification JESD22-C101, all pins ⁽²⁾	-500	500	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V _{IN}	Input voltage range	3	26	V
I _{IN}	Input current, IN pin	50	300	mA
I _{OUT}	Output current, OUT pin	50	300	mA
R _{ILIM}	OCP Programming resistor	83.3	500	kΩ
T _J	Junction temperature	-40	125	°C

7.4 Thermal Information

THERMAL METRIC ⁽¹⁾	DSG	UNITS	
	8 PINS		
R _{θJA}	Junction-to-ambient thermal resistance	86.3	°C/W
R _{θJctop}	Junction-to-case (top) thermal resistance	116.9	
R _{θJB}	Junction-to-board thermal resistance	56.1	
Ψ _{JT}	Junction-to-top characterization parameter	8.1	
Ψ _{JB}	Junction-to-board characterization parameter	56.4	
R _{θJcbot}	Junction-to-case (bottom) thermal resistance	25.9	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

7.5 Electrical Characteristics

over junction temperature range -40°C to 125°C and recommended supply voltage (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
IN						
$V_{(UVLO)}$	Undervoltage lock-out, input power detected threshold	$\overline{CE} = \text{Low}$, V_{IN} increasing from 0 V to 3 V	2.6	2.7	2.8	V
$V_{(UVLO_HYS)}$	Hysteresis on UVLO	$\overline{CE} = \text{Low}$, V_{IN} decreasing from 3 V to 0 V	200	260	300	mV
I_{DD}	Operating current	$\overline{CE} = \text{Low}$, No load on OUT pin, $V_{IN} = 5\text{ V}$, $R_{(ILIM)} = 200\text{ k}\Omega$		400	500	μA
$I_{(STDBY)}$	Standby current	$\overline{CE} = \text{High}$, $V_{IN} = 5\text{ V}$		65	95	μA
INPUT TO OUTPUT CHARACTERISTICS						
$V_{(DO)}$	Drop-out voltage IN to OUT	$\overline{CE} = \text{Low}$, $V_{IN} = 5\text{ V}$, $I_{OUT} = 0.125\text{ A}$		21	35	mV
INPUT OVERVOLTAGE PROTECTION						
$V_{(OVP)}$	Input overvoltage protection threshold	$\overline{CE} = \text{Low}$, V_{IN} increasing from 5V to 7.5 V	5.71	5.85	6.00	V
$V_{(HYS-OVP)}$	Hysteresis on OVP	$\overline{CE} = \text{Low}$, V_{IN} decreasing from 7.5 V to 5 V	20	60	110	mV
INPUT OVERCURRENT PROTECTION						
$I_{(OCP)}$	Input overcurrent protection threshold range		50		300	mA
	Input overcurrent protection threshold	$\overline{CE} = \text{Low}$, $R_{(ILIM)} = 200\text{ k}\Omega$, $3\text{ V} \leq V_{IN} < V_{OVP}$	$T_J = 0^{\circ}\text{C}$ to 85°C	110	125	135
		$T_J = 0^{\circ}\text{C}$ to 125°C	110	125	140	
BATTERY OVERVOLTAGE PROTECTION						
$V_{(BOVP)}$	Battery overvoltage protection threshold	$\overline{CE} = \text{Low}$, $V_{IN} > 4.4\text{ V}$	4.30	4.35	4.4	V
$V_{(HYS-BOVP)}$	Hysteresis on $V_{(BOVP)}$	$\overline{CE} = \text{Low}$, $V_{IN} > 4.4\text{ V}$	200	275	320	mV
$I_{(VBAT)}$	Input bias current on VBAT pin	$V_{BAT} = 4.4\text{ V}$, $T_J = 25^{\circ}\text{C}$			10	nA
THERMAL PROTECTION						
$T_{J(OFF)}$	Thermal shutdown temperature			140	150	$^{\circ}\text{C}$
$T_{J(OFF-HYS)}$	Thermal shutdown hysteresis			20		$^{\circ}\text{C}$
LOGIC LEVELS ON \overline{CE}						
V_{IL}	Low-level input voltage		0		0.4	V
V_{IH}	High-level input voltage		1.4			V
I_{IL}	Low-level input current	$V_{CE} = 0\text{ V}$			1	μA
I_{IH}	High-level input current	$V_{CE} = 1.8\text{ V}$			15	μA
LOGIC LEVELS ON $\overline{\text{FAULT}}$						
V_{OL}	Output low voltage	$I_{(SINK)} = 5\text{ mA}$			0.2	V
$I_{(HI-Z)}$	Leakage current, $\overline{\text{FAULT}}$ pin HI-Z	$V_{(\text{FAULT})} = 5\text{ V}$			10	μA

7.6 Timing Requirements

			MIN	TYP	MAX	UNIT
$t_{DGL(PGOOD)}$	Deglintch time, input power detected status	$\overline{CE} = \text{Low}$. Time measured from V_{IN} 0 V \rightarrow 5 V 1 μs rise-time, to output turning ON		8		ms
$t_{PD(OVP)}$	Input OV propagation delay ⁽¹⁾	$\overline{CE} = \text{Low}$			1	μs
$t_{ON(OVP)}$	Recovery time from input overvoltage condition	$\overline{CE} = \text{Low}$, Time measured from V_{IN} 7.5 V \rightarrow 5 V, 1 μs fall-time		8		ms
$t_{BLANK(OCP)}$	Blanking time, input overcurrent detected			176		μs
$t_{REC(OCP)}$	Recovery time from input overcurrent condition			64		ms
$t_{DGL(BOVP)}$	Deglintch time, battery overvoltage detected	$\overline{CE} = \text{Low}$, $V_{IN} > 4.4\text{ V}$. Time measured from $V_{(VBAT)}$ rising from 4.1 V to 4.4 V to $\overline{\text{FAULT}}$ going low.		176		μs

(1) Not tested in production. Specified by design.

7.7 Typical Characteristics

Test conditions (unless otherwise noted) for typical operating performance: $V_{IN} = 5\text{ V}$, $C_{IN} = 1\ \mu\text{F}$, $C_{OUT} = 1\ \mu\text{F}$, $R_{(LIM)} = 200\ \text{k}\Omega$, $R_{(BAT)} = 100\ \text{k}\Omega$, $T_A = 25^\circ\text{C}$, $V_{(PU)} = 3.3\ \text{V}$ (see Figure 11 for the Typical Application Circuit)

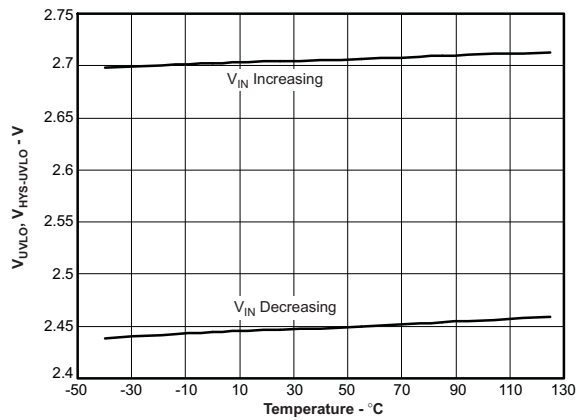


Figure 1. Undervoltage Lockout vs Free-Air Temperature

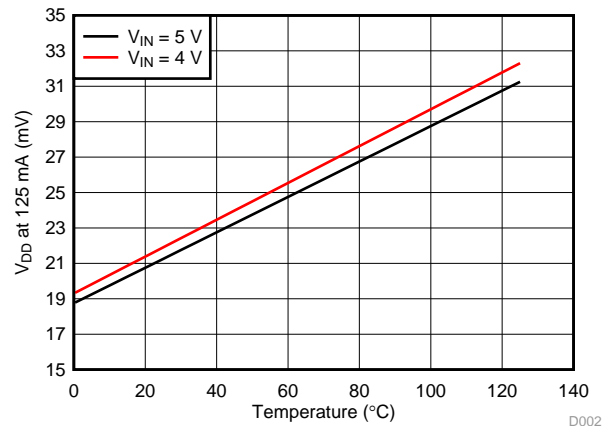


Figure 2. Dropout Voltage (In to Out) vs Free-Air Temperature

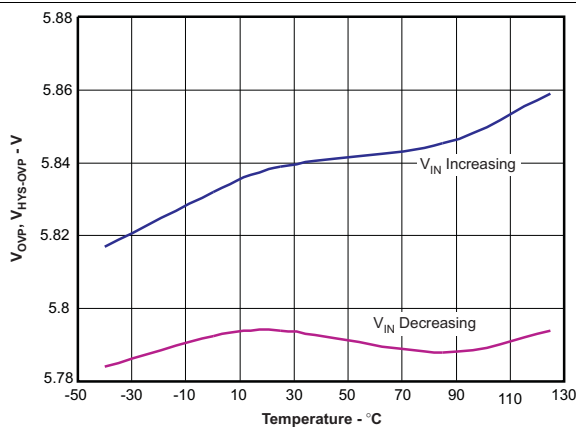


Figure 3. Overvoltage Threshold Protection vs Free-Air Temperature

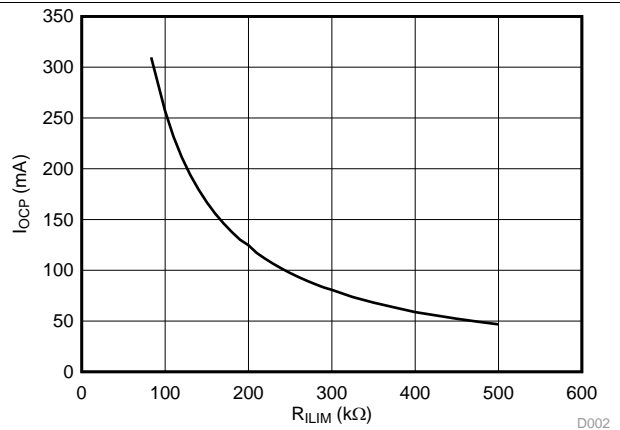


Figure 4. Input Overcurrent Protection vs ILIM Resistance

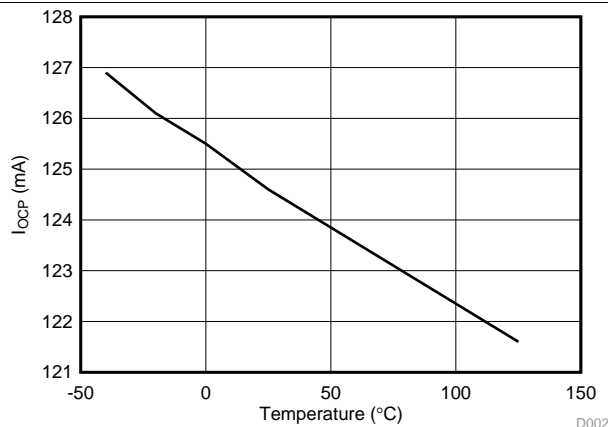


Figure 5. Input Overcurrent Protection vs Free-Air Temperature

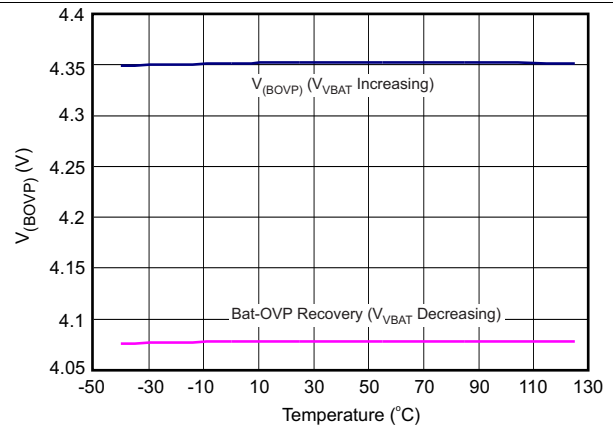
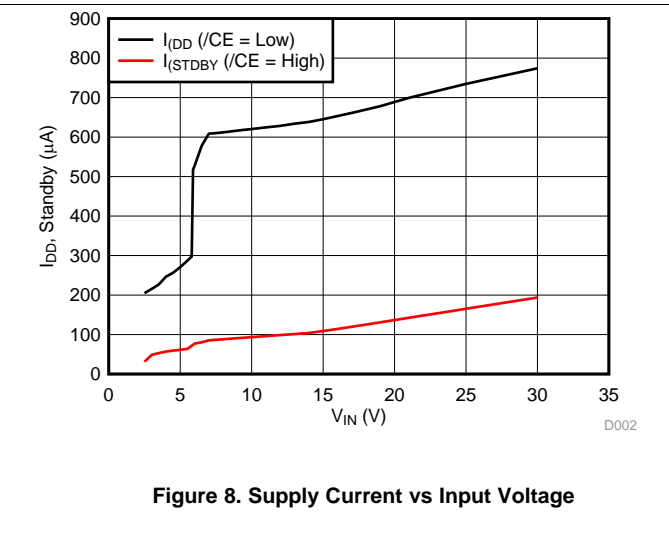
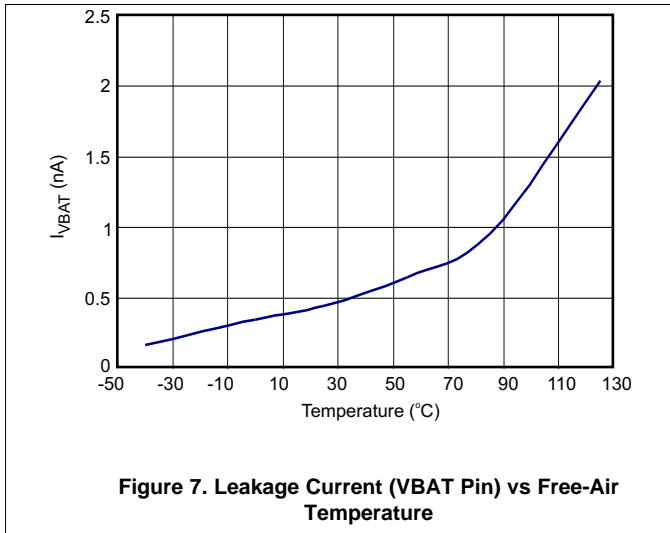


Figure 6. Battery Overvoltage Protection vs Free-Air Temperature

Typical Characteristics (continued)



8 Detailed Description

8.1 Overview

The bq24311 is a highly integrated circuit designed to protect Li-ion batteries from charging circuit failures. The IC continuously monitors the input voltage, input current, and battery voltage. The input overvoltage protection immediately removes power from the charging circuit by turning off an internal switch. The input protection limits the system current at the user-programmable value, and if the overcurrent persists, switches the pass element OFF after a blanking period. Additionally, the IC also monitors its own die temperature and switches off if it becomes too hot.

8.2 Functional Block Diagram

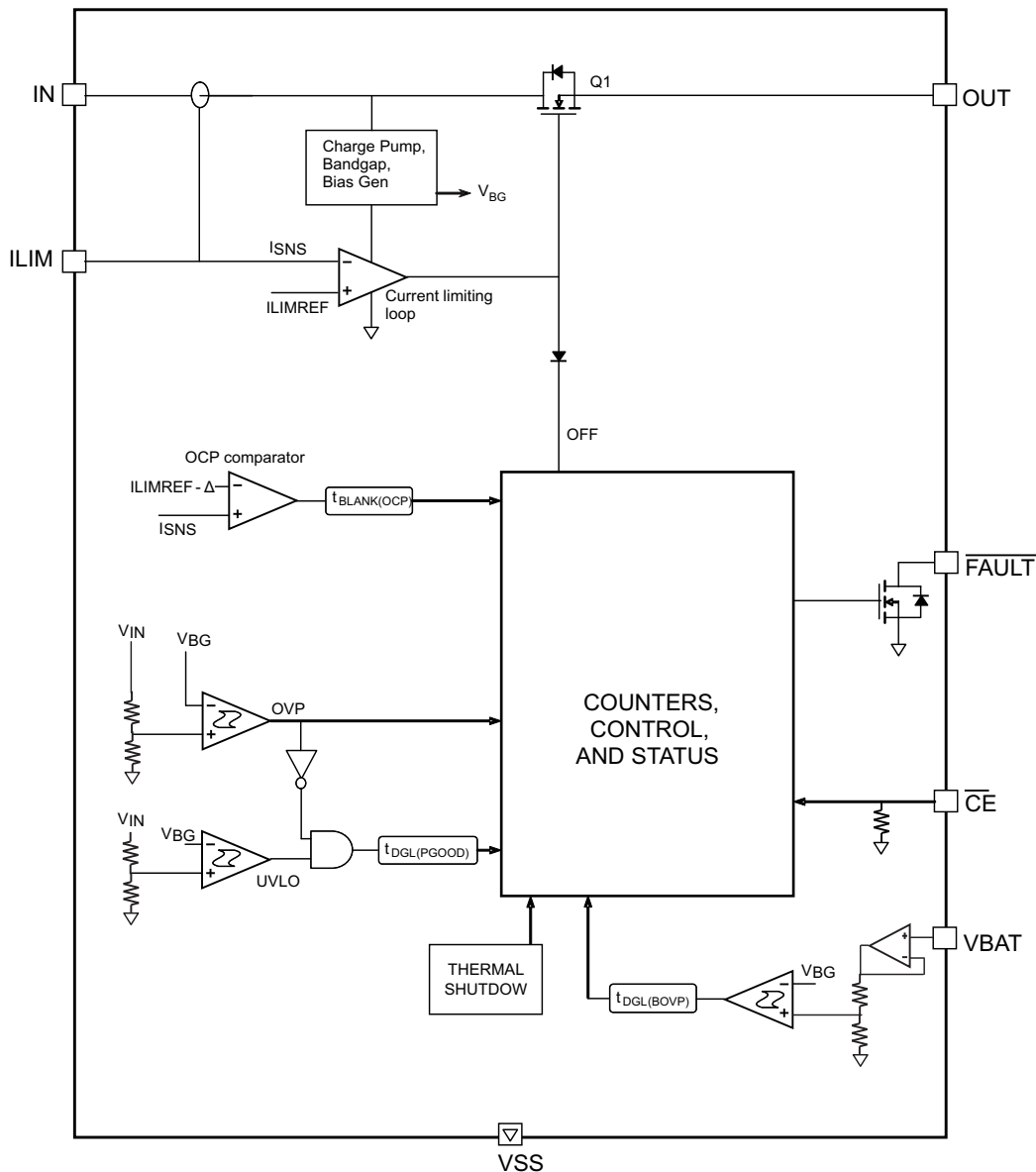


Figure 9. Simplified Block Diagram

8.3 Feature Description

8.3.1 Power Down

The device remains in power down mode when the voltage at the IN pin is below the undervoltage threshold V_{UVLO} . The FET Q1 connected between IN and OUT pins is off, and the status output, \overline{FAULT} , is set to Hi-Z.

8.3.2 Power-On Reset

The device resets when the voltage at the IN pin exceeds the UVLO threshold. All internal counters and other circuit blocks are reset. The IC then waits for duration $t_{DGL(PGOOD)}$ for the input voltage to stabilize. If, after $t_{DGL(PGOOD)}$, the input voltage and battery voltage are safe, FET Q1 is turned ON. The IC has a soft-start feature to control the inrush current which minimizes the ringing at input during power up, as shown in [Figure 15](#) (ringing occurs because the parasitic inductance of the adapter cable and the input bypass capacitor form a resonant circuit). Because of the deglitch time at power-on, if the input voltage rises rapidly to beyond the OVP threshold, the device will not switch on at all, instead it will go into protection mode and indicate a fault on the \overline{FAULT} pin, as shown in [Figure 16](#).

8.4 Device Functional Modes

8.4.1 Operation

The device continuously monitors the input voltage, input current, and battery voltage as described in detail in the following sections.

8.4.1.1 Input Overvoltage Protection

If the input voltage rises above V_{OVP} , the internal FET Q1 is turned off, removing power from the circuit. As shown in [Figure 17](#), the response is rapid, with the FET turning off in less than a microsecond. The \overline{FAULT} pin is driven low. When the input voltage returns below $V_{OVP} - V_{HYS-OVP}$ (but is still above V_{UVLO}), the FET Q1 is turned on again after a deglitch time of $t_{ON(OVP)}$ to ensure that the input supply has stabilized. [Figure 18](#) shows the recovery from input OVP.

8.4.1.2 Input Overcurrent Protection

If the load current tries to exceed the I_{OCP} threshold, the device limits the current for a blanking period, $t_{BLANK(OCP)}$. If the load current returns to less than I_{OCP} before $t_{BLANK(OCP)}$ times out, the device continues to operate. However, if the overcurrent situation persists for $t_{BLANK(OCP)}$, the FET Q1 is turned off for a duration of $t_{REC(OCP)}$, and the \overline{FAULT} pin is driven low. The FET is then turned on again after $t_{REC(OCP)}$ and the current is monitored all over again. Each time an OCP fault occurs, an internal counter is incremented. If 15 OCP faults occur in one charge cycle, the FET is turned off permanently, as shown in [Figure 19](#). The counter is cleared either by removing and re-applying input power, or by disabling and re-enabling the device with the \overline{CE} pin. [Figure 19](#) and [Figure 20](#) show what happens in an overcurrent fault.

To prevent the input voltage from spiking up due to the inductance of the input cable, Q1 is turned off slowly, resulting in a “soft-stop”, as shown in [Figure 22](#).

8.4.1.3 Battery Overvoltage Protection

The battery overvoltage threshold $V_{(BOVP)}$ is internally set to 4.35V. If the battery voltage exceeds the $V_{(BOVP)}$ threshold, the FET Q1 is turned off, and the \overline{FAULT} pin is driven low. The FET is turned back on once the battery voltage drops to $V_{(BOVP)} - V_{HYS-BOVP}$ (see [Figure 22](#) and [Figure 23](#)). Each time a battery overvoltage fault occurs, an internal counter is incremented. If 15 such faults occur in one charge cycle, the FET is turned off permanently, as shown in [Figure 23](#). The counter is cleared either by removing and re-applying input power, or by disabling and re-enabling the device with the \overline{CE} pin. In the case of a battery overvoltage fault, Q1 is switched OFF gradually, resulting in a soft-stop (see [Figure 22](#)).

Device Functional Modes (continued)

8.4.1.4 Thermal Protection

If the junction temperature of the device exceeds $T_{J(OFF)}$, the FET Q1 is turned off, and the \overline{FAULT} pin is driven low. The FET is turned back on when the junction temperature falls below $T_{J(OFF)} - T_{J(OFF-HYS)}$.

8.4.1.5 Enable Function

The IC has an enable pin which can be used to enable or disable the device. When the \overline{CE} pin is driven high, the internal FET is turned off. When the \overline{CE} pin is low, the FET is turned on if other conditions are safe. The OCP counter and the Bat-OVP counter are both reset when the device is disabled and re-enabled. The \overline{CE} pin has an internal pulldown resistor and can be left floating. Note that the \overline{FAULT} pin functionality is also disabled when the \overline{CE} pin is high.

8.4.1.6 Fault Indication

The \overline{FAULT} pin is an active-low open-drain output. It is in a high-impedance state when operating conditions are safe, or when the device is disabled by setting \overline{CE} high. With \overline{CE} low, the \overline{FAULT} pin goes low whenever any of these events occurs:

- Input overvoltage
- Input overcurrent
- Battery overvoltage
- IC Overtemperature

Device Functional Modes (continued)

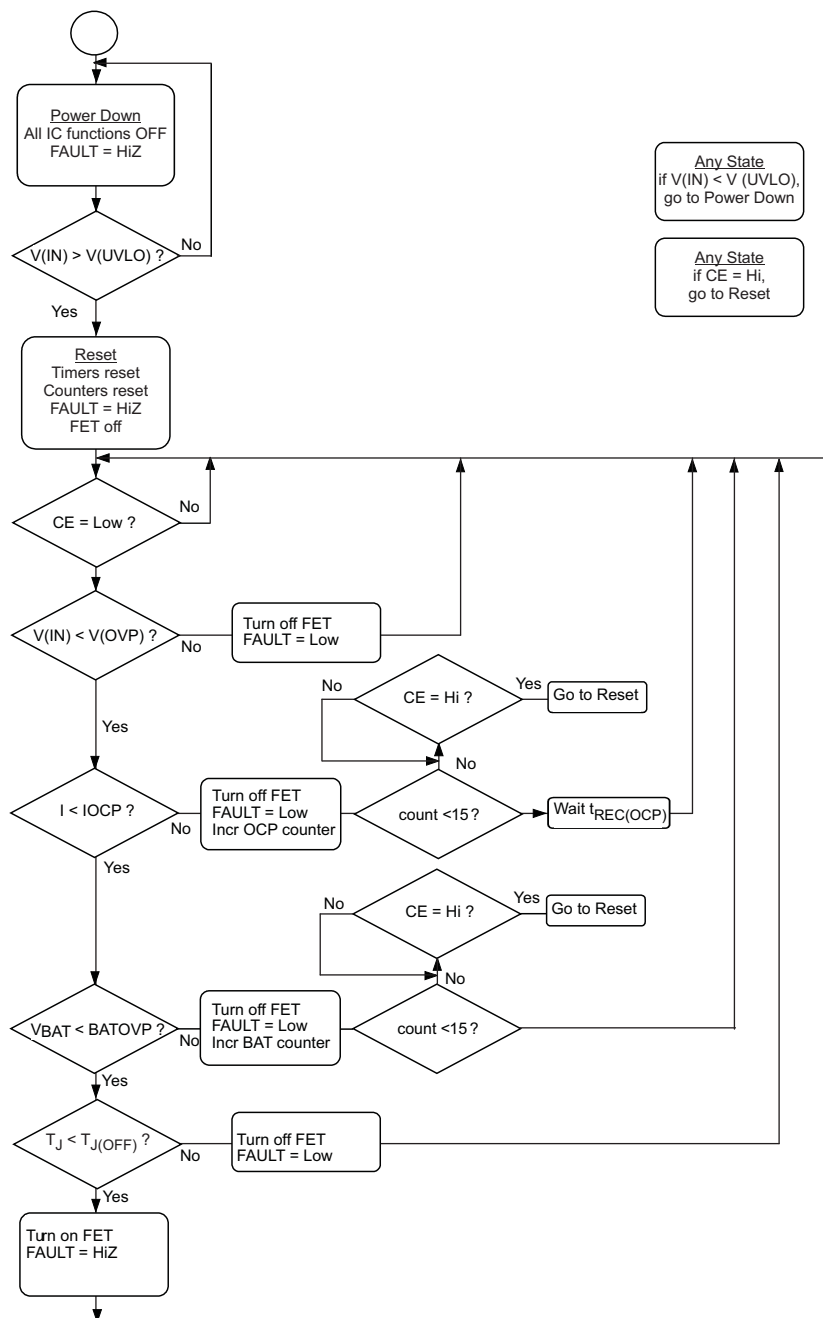


Figure 10. Flow Diagram

9 Application and Implementation

9.1 Typical Application Circuit

$V_{OVP} = 5.85\text{ V}$, $I_{OCP} = 125\text{ mA}$, $V_{(BOVP)} = 4.35\text{ V}$.

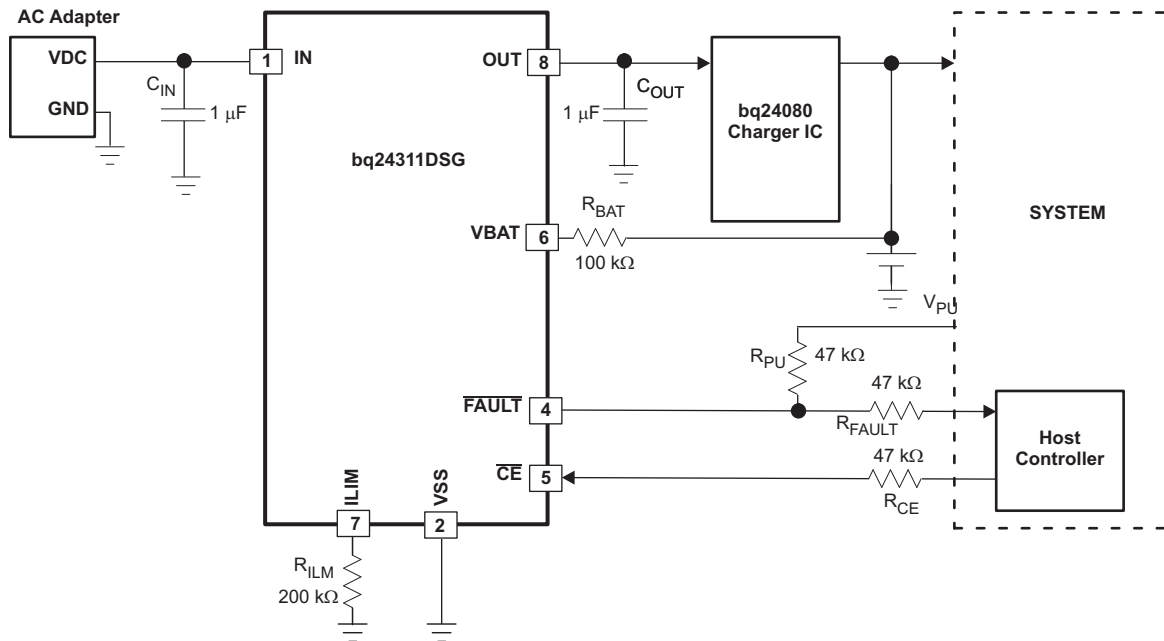


Figure 11.

9.1.1 Design Requirements

9.1.1.1 Selection of R_{ILIM}

The overcurrent threshold is programmed by a resistor, R_{ILIM} , connected from the ILIM pin to VSS. Figure 4 shows the OCP threshold as a function of R_{ILIM} , and may be approximated by the following equation:

$$I_{OCP} = 25 \div R_{ILIM} \text{ (current in A, resistance in k}\Omega\text{)} \quad (1)$$

Choose a I_{OCP} between 50 mA and 300 mA and apply the above equation to select a R_{ILIM} resistor value from 500 k Ω to 83.3 k Ω respectively. However, at lower OCP limits, approaching 50 mA, the precision of the current protection circuit decreases the achievable accuracy of the OCP threshold.

9.1.1.2 Selection of R_{BAT}

It is strongly recommended that the battery not be tied directly to the VBAT pin of the device, as under some failure modes of the IC, the voltage at the IN pin may appear on the VBAT pin. This voltage can be as high as 30 V, and applying 30 V to the battery in case of the failure of the bq24311 can be hazardous. Connecting the VBAT pin through $R_{(BAT)}$ prevents a large current from flowing into the battery in case of a failure of the IC. In the interests of safety, $R_{(BAT)}$ should have a high value. The problem with a large $R_{(BAT)}$ is that the voltage drop across this resistor because of the VBAT bias current $I_{(VBAT)}$ causes an error in the $V_{(BOVP)}$ threshold. This error is over and above the tolerance on the nominal 4.35V $V_{(BOVP)}$ threshold.

Choosing R_{BAT} in the range 100 k Ω to 470 k Ω is a good compromise. In the case of an IC failure, with R_{BAT} equal to 100k Ω , the maximum current flowing into the battery would be $(30\text{ V} - 3\text{ V}) \div 100\text{ k}\Omega = 246\text{ }\mu\text{A}$, which is low enough to be absorbed by the bias currents of the system components. $R_{(BAT)}$ equal to 100 k Ω would result in a worst-case voltage drop of $R_{(BAT)} \times I_{(VBAT)} = 1\text{ mV}$. This is negligible compared to the internal tolerance of 50mV on $V_{(BOVP)}$ threshold.

If the Bat-OVP function is not required, the VBAT pin should be connected to VSS.

Typical Application Circuit (continued)

9.1.1.3 Selection of $R_{(CE)}$, $R_{(FAULT)}$, and $R_{(PU)}$

The \overline{CE} pin can be used to enable and disable the IC. If host control is not required, the \overline{CE} pin can be tied to ground or left un-connected, permanently enabling the device.

In applications where external control is required, the \overline{CE} pin can be controlled by a host processor. As in the case of the VBAT pin (see [Selection of Rbat](#)), the \overline{CE} pin should be connected to the host GPIO pin through as large a resistor as possible. The limitation on the resistor value is that the minimum V_{OH} of the host GPIO pin less the drop across the resistor should be greater than V_{IH} of the bq24311 \overline{CE} pin. The drop across the resistor is given by $R_{(CE)} \times I_{IH}$.

The \overline{FAULT} pin is an open-drain output that goes low during OV, OC, battery-OV, and OT events. If the application does not require monitoring of the \overline{FAULT} pin, it can be left unconnected. But if the \overline{FAULT} pin has to be monitored, it should be pulled high externally through $R_{(PU)}$, and connected through $R_{(FAULT)}$ to the host. $R_{(FAULT)}$ prevents damage to the host controller if the bq24311 fails (see [Selection of Rbat](#)). The resistors should be of high value, in practice values between 22 k Ω and 100 k Ω should be sufficient.

9.1.1.4 Selection of Input and Output Bypass Capacitors

The input capacitor C_{IN} in [Figure 11](#) is for decoupling, and serves an important purpose. Whenever there is a step change downwards in the system load current, the inductance of the input cable causes the input voltage to spike up. C_{IN} prevents the input voltage from overshooting to dangerous levels. It is strongly recommended that a ceramic capacitor of at least 1 μF be used at the input of the device. It should be located in close proximity to the IN pin.

C_{OUT} in [Figure 11](#) is also important: If a fast (< 1 μs rise time) overvoltage transient occurs at the input, the current that charges C_{OUT} causes the device's current-limiting loop to kick in, reducing the gate-drive to FET Q1. This results in improved performance for input overvoltage protection. C_{OUT} should also be a ceramic capacitor of at least 1 μF , located close to the OUT pin. C_{OUT} also serves as the input decoupling capacitor for the charging circuit downstream of the protection IC.

9.1.2 Detailed Design Procedures

9.1.2.1 Powering Accessories

In some applications, the equipment that the protection IC resides in may be required to provide power to an accessory (that is, a cellphone may power a headset or an external memory card) through the same connector pins that are used by the adapter for charging. [Figure 12](#) and [Figure 13](#) illustrate typical charging and accessory-powering scenarios:

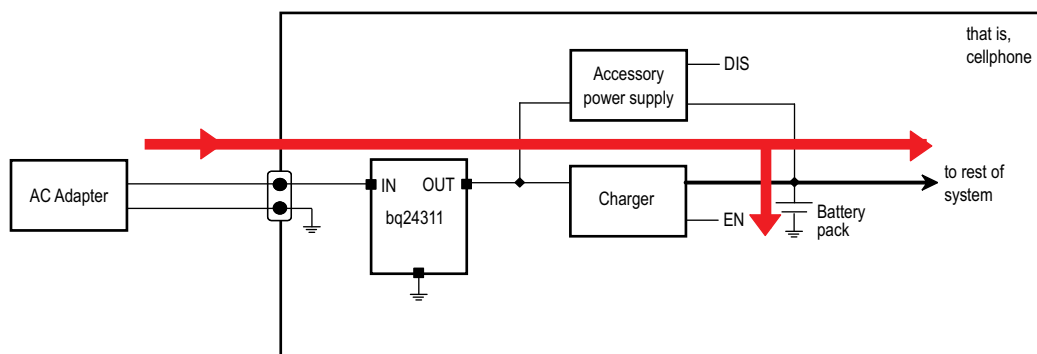


Figure 12. Charging - The Red Arrows Show the Direction of Current Flow

Typical Application Circuit (continued)

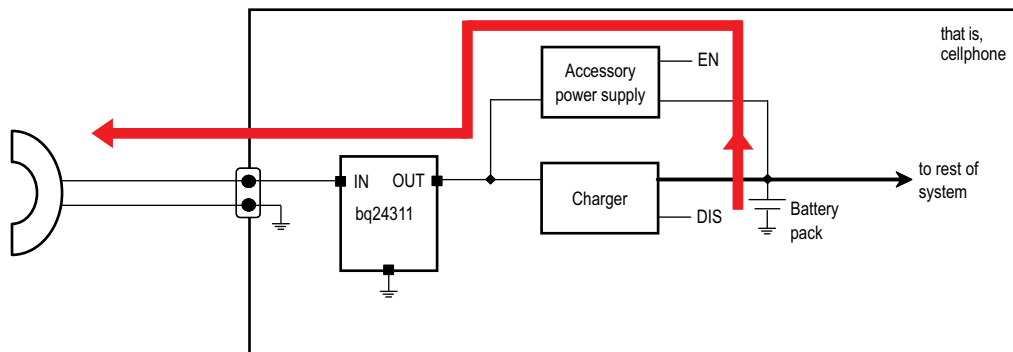


Figure 13. Powering an Accessory - The Red Arrows Show the Direction of Current Flow

In the second case, when power is being delivered to an accessory, the bq24311 device is required to support current flow from the OUT pin to the IN pin.

If $V_{OUT} > V_{(UVLO)} + 0.7 \text{ V}$, FET Q1 is turned on, and the reverse current does not flow through the diode but through Q1. Q1 will then remain ON as long as $V_{OUT} > V_{(UVLO)} - V_{(HYS-UVLO)} + R_{DS(on)} \times I_{(ACCESSORY)}$. Within this voltage range, the reverse current capability is the same as the forward capability, 0.5 A. It should be noted that there is no overcurrent protection in this direction.

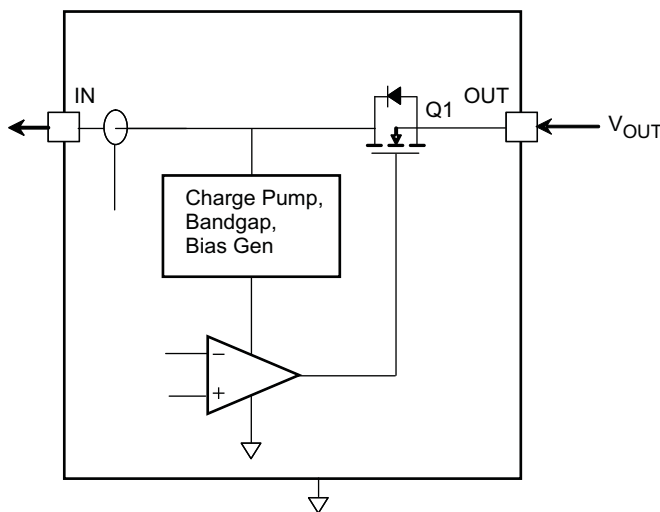
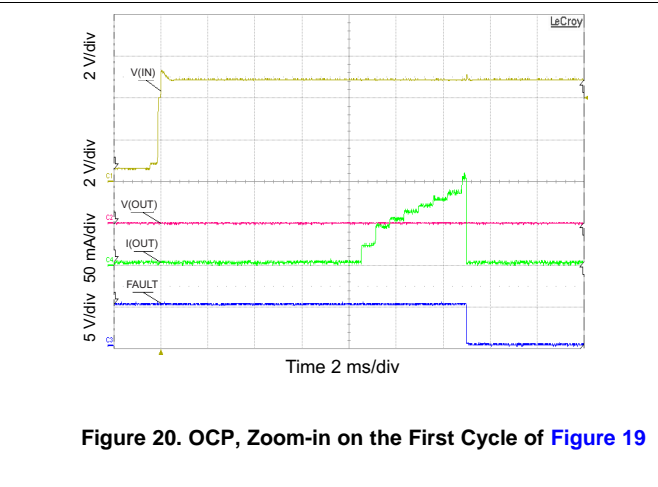
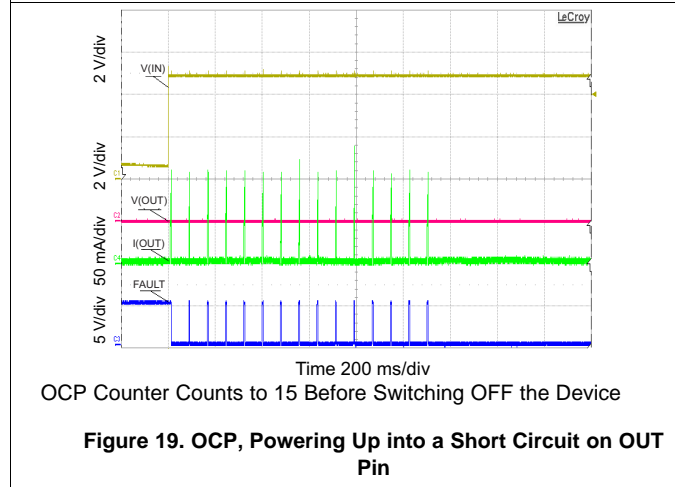
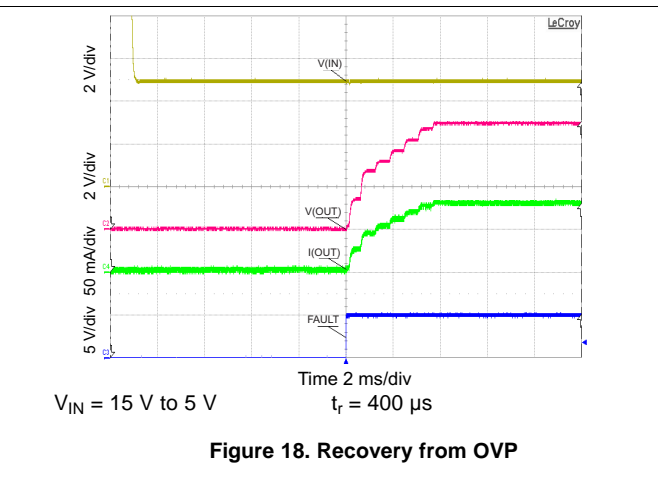
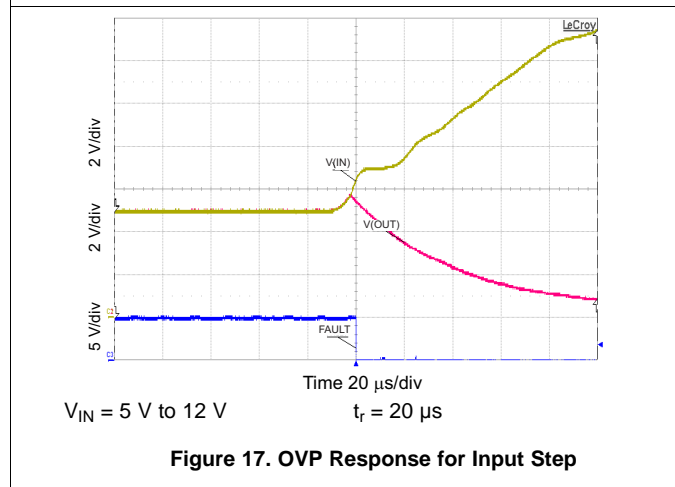
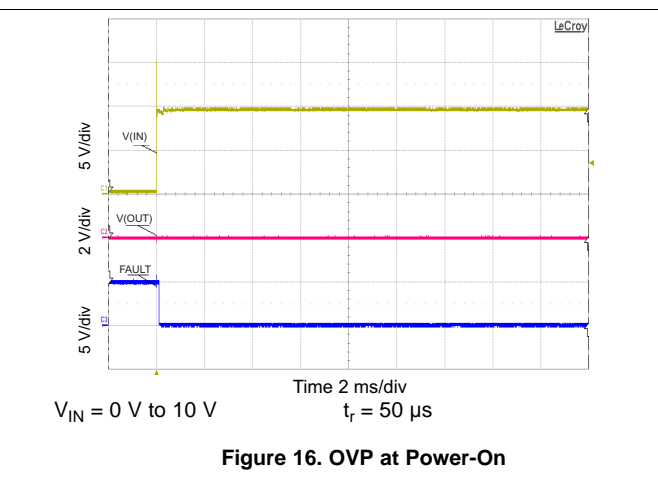
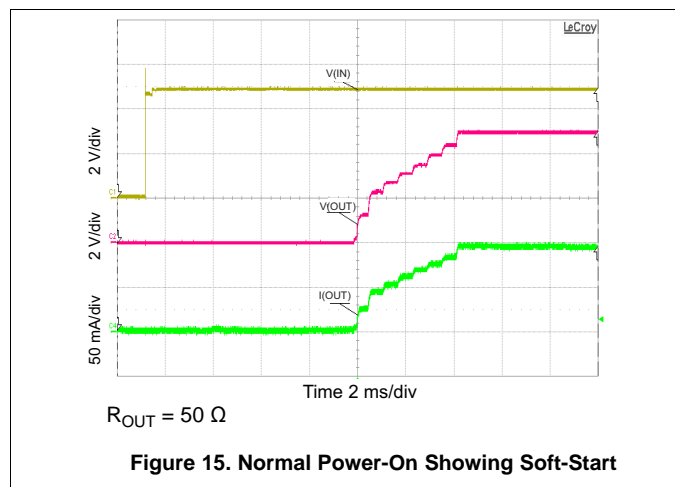


Figure 14.

Typical Application Circuit (continued)

9.1.3 Application Curves



Typical Application Circuit (continued)

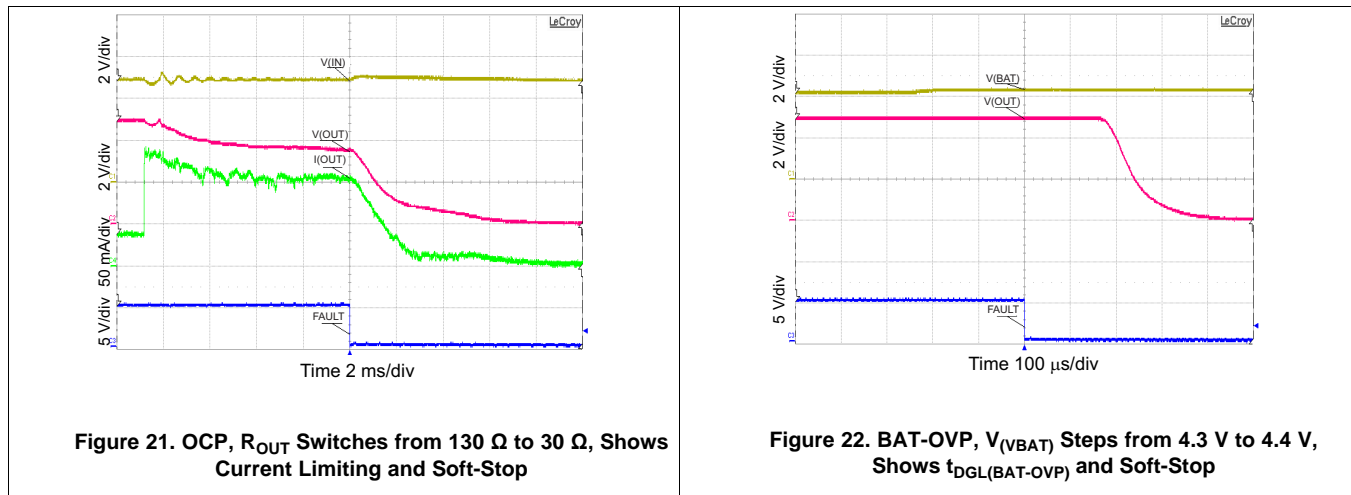


Figure 21. OCP, R_{OUT} Switches from 130 Ω to 30 Ω , Shows Current Limiting and Soft-Stop

Figure 22. BAT-OVP, $V_{(VBAT)}$ Steps from 4.3 V to 4.4 V, Shows $t_{DGL(BAT-OVP)}$ and Soft-Stop

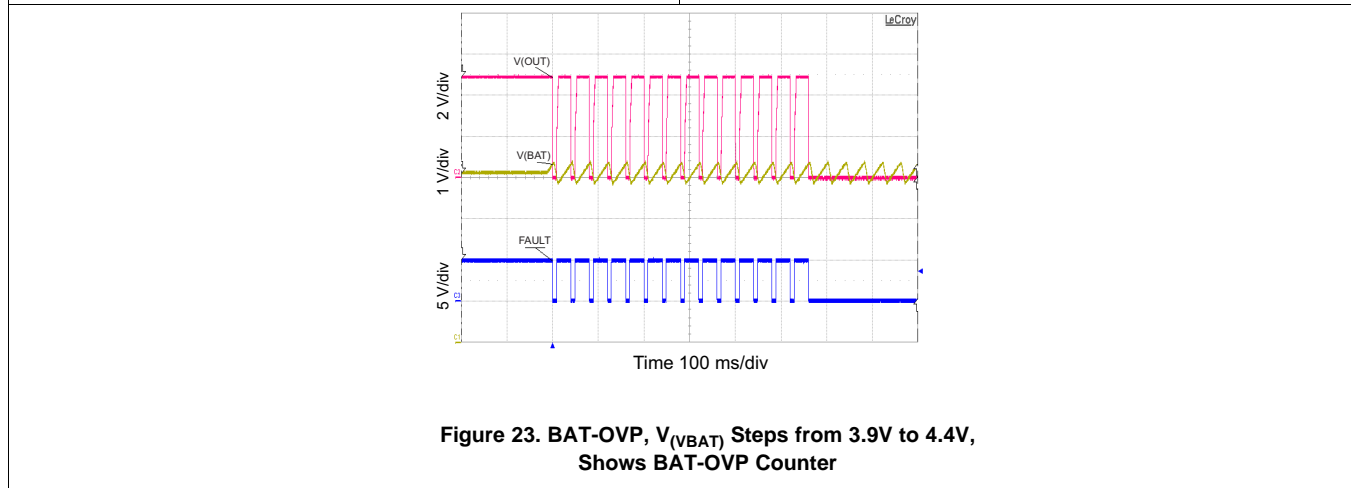


Figure 23. BAT-OVP, $V_{(VBAT)}$ Steps from 3.9V to 4.4V, Shows BAT-OVP Counter

10 Power Supply Requirements

In a typical application, the system is powered by a USB port or USB wall adapter.

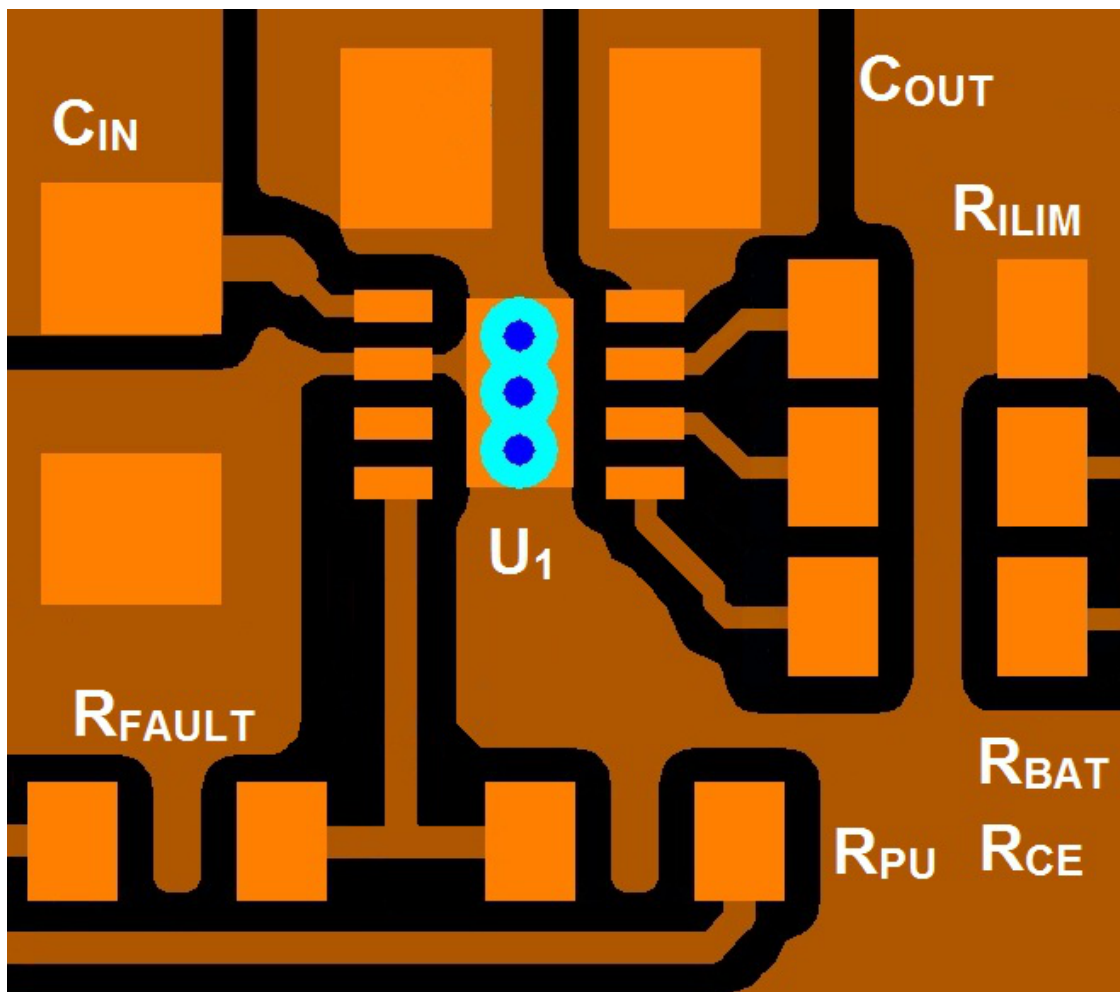
The minimum input voltage, where the protector starts to pass current assuming V_{BAT} is acceptable, could be 2.7 V. The maximum supported input voltage is up to 5.85 V; the overvoltage protection kicks in at 5.85 V and the maximum input voltage rating is 30 V input rating.

11 Layout

11.1 Layout Guidelines

- This device is a protection device, and is meant to protect down-stream circuitry from hazardous voltages. Potentially, high voltages may be applied to this IC. It has to be ensured that the edge-to-edge clearances of PCB traces satisfy the design rules for high voltages.
- The device uses SON packages with a PowerPAD™. For good thermal performance, the PowerPAD should be thermally coupled with the PCB ground plane. In most applications, this will require a copper pad directly under the IC. This copper pad should be connected to the ground plane with an array of thermal vias.
- C_{IN} and C_{OUT} should be located close to the IC. Other components like R_{ILIM} and R_{BAT} should also be located close to the IC.

11.2 Layout Example



12 器件和文档支持

12.1 Trademarks

PowerPAD is a trademark of Texas Instruments.

Bluetooth is a trademark of Bluetooth SIG, Inc.

12.2 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

12.3 术语表

[SLYZ022](#) — *TI* 术语表。

这份术语表列出并解释术语、首字母缩略词和定义。

13 机械封装和可订购信息

以下页中包括机械封装和可订购信息。 这些信息是针对指定器件可提供的最新数据。 这些数据会在无通知且不对本文档进行修订的情况下发生改变。 欲获得该数据表的浏览器版本，请查阅左侧的导航栏。

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数据转换器	www.ti.com.cn/dataconverters	消费电子	www.ti.com.cn/consumer-apps
DLP® 产品	www.dlp.com	能源	www.ti.com.cn/energy
DSP - 数字信号处理器	www.ti.com.cn/dsp	工业应用	www.ti.com.cn/industrial
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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
BQ24311DSGR	ACTIVE	WSON	DSG	8	3000	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	SHN	Samples
BQ24311DSGT	ACTIVE	WSON	DSG	8	250	Green (RoHS & no Sb/Br)	NIPDAU	Level-2-260C-1 YEAR		SHN	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "-" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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GENERIC PACKAGE VIEW

DSG 8

WSON - 0.8 mm max height

2 x 2, 0.5 mm pitch

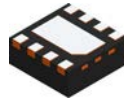
PLASTIC SMALL OUTLINE - NO LEAD

This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224783/A

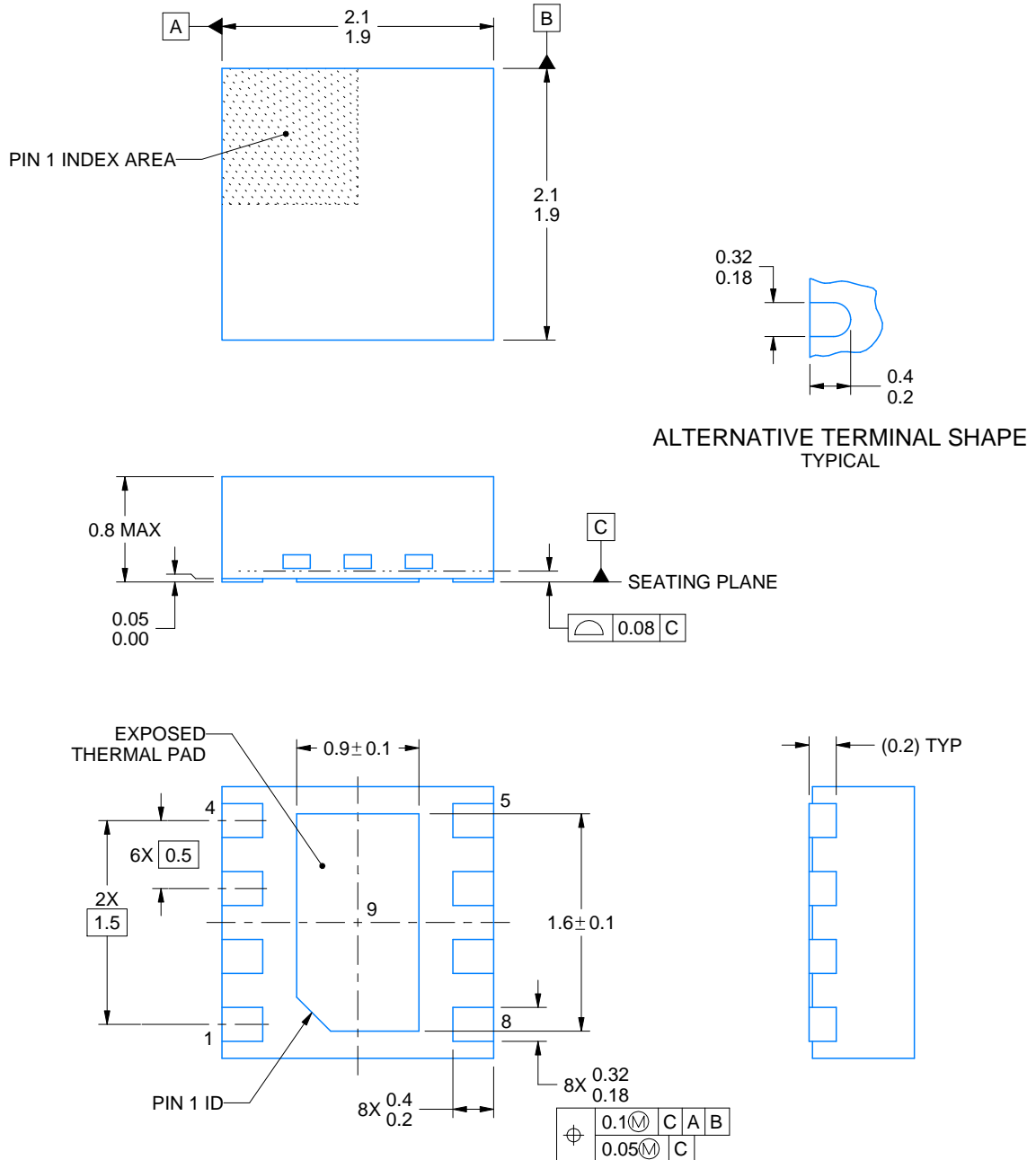
DSG0008A



PACKAGE OUTLINE

WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



4218900/D 04/2020

NOTES:

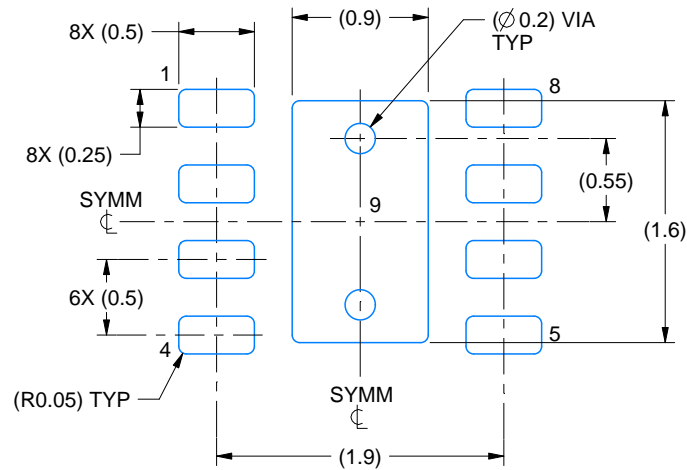
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

EXAMPLE BOARD LAYOUT

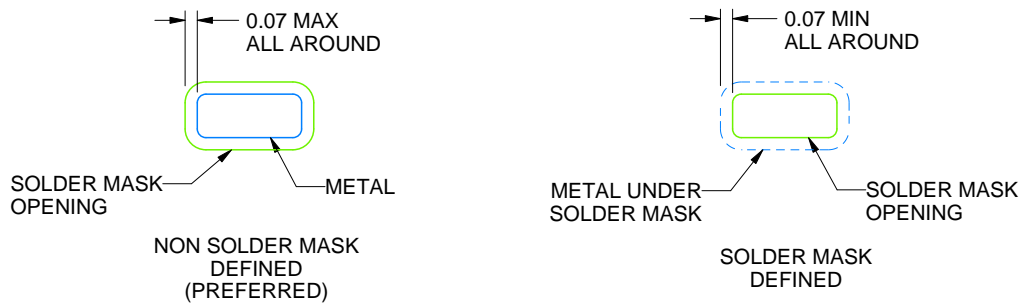
DSG0008A

WSON - 0.8 mm max height

PLASTIC SMALL OUTLINE - NO LEAD



LAND PATTERN EXAMPLE
SCALE:20X



SOLDER MASK DETAILS

4218900/D 04/2020

NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

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