



# η -Balance<sup>TM</sup> PWM Power Switch Fixed 65KHz Fsw

#### **FEATURES**

- Less than 75mW Standby Power
- Built-in 650V Power MOSFET
- ◆ Proprietary n -Balance<sup>™</sup> Control to Boost Light Load Efficiency
- Proprietary "Zero OCP/OPP Recovery Gap" Control
- Fixed 65KHz Switching Frequency
- Built-in Soft Start Function
- Very Low Startup Current
- Frequency Reduction and Burst Mode Control for Energy Saving
- Current Mode Control
- Built-in Frequency Shuffling
- Built-in Synchronous Slope Compensation
- Cycle-by-Cycle Current Limiting
- Built-in Leading Edge Blanking (LEB)
- Constant Power Limiting
- Pins Floating Protection
- Audio Noise Free Operation
- VDD OVP & Clamp
- VDD Under Voltage Lockout (UVLO)

#### APPLICATIONS

Offline AC/DC Flyback Converter for

- AC/DC Adaptors
- Open-frame SMPS

### **GENERAL DESCRIPTION**

SF5549 is a high performance, high efficiency, highly integrated current mode PWM power switch for offline flyback converter applications.

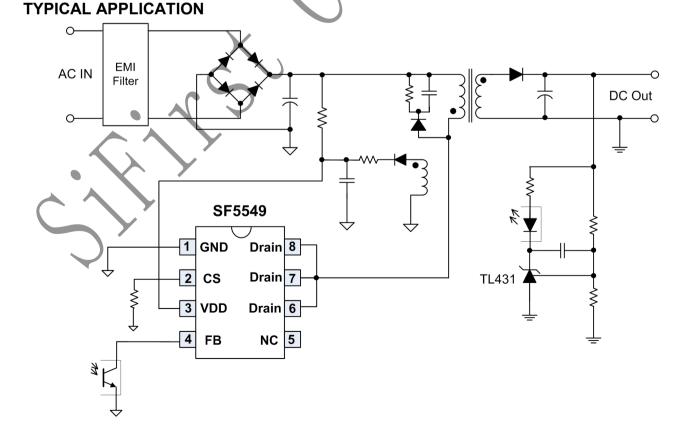
In SF5549, PWM switching frequency with shuffling is fixed to 65KHz and is trimmed to tight range. When the output power demands decrease, the IC decreases switching frequency based on the proprietary  $\eta$  -**Balance**<sup>TM</sup> control to boost power conversion efficiency at the light load. When output power falls below a given value, the IC enters into burst mode and provides excellent efficiency without audio noise.

The IC can achieve "**Zero OCP/OPP Recovery Gap**" using SiFirst's proprietary control algorithm. Meanwhile, the OCP/OPP variation versus universal line input is compensated.

The IC has built-in synchronized slope compensation to prevent sub-harmonic oscillation at high PWM duty output. The IC also has built-in soft start function to soften the stress on the MOSFET during power on period.

SF5549 integrates functions and protections of Under Voltage Lockout (UVLO), VCC Over Voltage Protection (OVP), Cycle-by-cycle Current Limiting (OCP), Pins Floating Protection, Over Load Protection (OLP), VCC Clamping, Leading Edge Blanking (LEB), etc.

SF5549 is available in DIP8 packages.





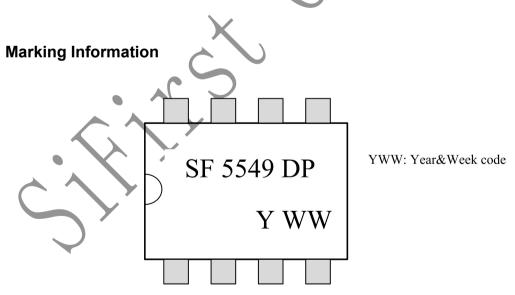
## **Pin Configuration**

	-		
	GND	1	8 Drain
	cs 🗌	DIP8	7 Drain
		3	6 Drain
	FB	4	5 NC
	L		
Ordering Informa			
Part Number	Top Mark		Package Tape & Reel
SF5549DP	SF5549DP	DIP8	RoHs
Output Power Ta	ble <sup>(1)</sup>		, XCY

output i onoi iu						
Part Number	230VAC ± 15% <sup>(2)</sup>			85-265VAC		
	Adapter <sup>(3)</sup>	Open Frame <sup>(4)</sup>		Adapter <sup>(3)</sup>	Open Frame <sup>(4)</sup>	
SF5549DP	21W	30W		16W	24W	
			-			

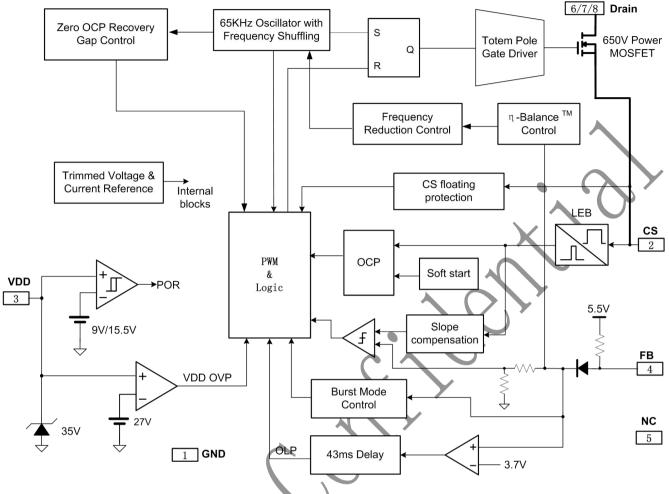
Note 1. The Max. output power is limited by junction temperature

- Note 2. 230VAC or 100/115VAC with doublers
- Note 3. Typical continuous power in a non-ventilated enclosed adapter with sufficient drain pattern as a heat sink at 50 °C ambient.
- Note 4. Max. practical continuous power in a open-frame design with sufficient drain pattern as a heat sink at 50 °C ambient.





# **Block Diagram**



# **Pin Description**

Pin Num	Pin Name	I/O	Description
1	GND	Р	IC Ground.
2	CS		Current sense input pin.
3	VDD	Р	IC power supply pin.
4	FB		Voltage feedback pin. The loop regulation is achieved by connecting a
			photo-coupler to this pin. PWM duty cycle is determined by this pin voltage
			and the current sense signal at Pin 2.
5	NC		No connect.
6/7/8	Drain	P	High voltage power MOSFET drain connection.

## Absolute Maximum Ratings (Note 5)

Parameter	Value	Unit
VDD DC Supply Voltage	35	V
VCC DC Clamp Current	10	mA
Drain pin	-0.3 to 650	V
FB, CS voltage range	-0.3 to 7	V
Package Thermal Resistance (DIP-8)	84	°C/W
Maximum Junction Temperature	150	Oo
Operating Temperature Range	-40 to 85	Oo
Storage Temperature Range	-65 to 150	Oo
Lead Temperature (Soldering, 10sec.)	260	°C
ESD Capability, HBM (Human Body Model)	3	kV
ESD Capability, MM (Machine Model)	250	V



# Recommended Operation Conditions (Note 6)

Parameter	Value	Unit
Supply Voltage, VDD	11 to 25	V
Operating Ambient Temperature	-40 to 85	٥C

# **ELECTRICAL CHARACTERISTICS**

(T <sub>A</sub> = 25 <sup>o</sup> C, VDD=18V, if not otherwise no	ted)
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Symbol	Parameter	Test Conditions	Min	Тур	Max	Unit
	Section (VDD Pin)			- 7 6		
UVLO(ON)	VDD Under Voltage		14.5	15.5	16.5	V
	Lockout Exit (Startup)		1-7.0	10.0	10.0	
UVLO(OFF)	VDD Under Voltage		8	9	9.8	V
0120(011)	Lockout Enter			Ŭ	0.0	XX
I_Startup	VDD Start up Current	VDD =UVLO(ON)-1V,		3	15	uA
		Measure current into VDD				
I_VDD_Op	Operation Current	V <sub>FB</sub> =3V		2.0	3.5	/ mA
VDD OVP	VDD Over Voltage		25	27	29	V
	Protection trigger			$\frown$	$\mathbf{O}$	
V <sub>DD</sub> _Clamp	VDD Zener Clamp	I(V <sub>DD</sub> ) = 10mA		35.5		V
	Voltage				- -	
T Softstart	Soft Start Time			4		mSec
Feedback Input			$\checkmark$	)	1	
V <sub>FB</sub> Open	FB Open Voltage			4.5		V
				•		
I <sub>FB</sub> _Short	FB short circuit	Short FB pin to GND,	0.22	0.33	0.45	mA
	current	measure current	0	0.00	0.10	
Avcs	PWM Input Gain	$\Delta V_{FB} / \Delta V_{cs}$		1.6		V/V
VFB min duty	FB under voltage gate			1.0		V
	clock is off.					
V <sub>TH</sub> _PL	Power Limiting FB			3.7		V
	Threshold Voltage			-		
$T_D_PL_min$	Minimum Power	CT is floating		43		mSec
	limiting Debounce	Ŭ				
	Time					
Z <sub>FB</sub> IN	Input Impedance			14		Kohm
	nput Section (CS Pin)		•		•	•
Vth OC min	Internal current	Zero duty cycle	0.70	0.75	0.80	V
	limiting threshold	5 5 -	-	-		
T_blanking	CS Input Leading			250		nSec
_ •	Edge Blanking Time					
T <sub>D</sub> OC	Over Current			90		nSec
	Detection and Control					
	Delay					
<b>Oscillator Section</b>	on					
Fosc	Normal Oscillation		60	65	70	KHZ
	Frequency					
∆F(shuffle)/Fosc	Frequency shuffling	Note 8	-4		4	%
	range					
∆f_Temp	Frequency	-20°C to 100 °C (Note 7)		5		%
-	Temperature Stability					
$\Delta f_VDD$	Frequency Voltage	VDD = 12-25V,		5		%
	Stability					
Duty_max	Maximum Duty cycle		75	80	85	%
F_BM	Burst Mode Base			22		KHZ
	Frequency					
<b>Power MOSFET</b>	Section <sup>(8)</sup>					
BVdss	Power MOSFET		650			V
	Drain Source					



SF5549

	Breakdown Voltage				
Rdson	Static Drain-Source On	I(Drain)=2A	2.6	3	Ω
	Resistance	× ,			
ldss	Zero Gate Voltage			1	uA
	Drain Current				

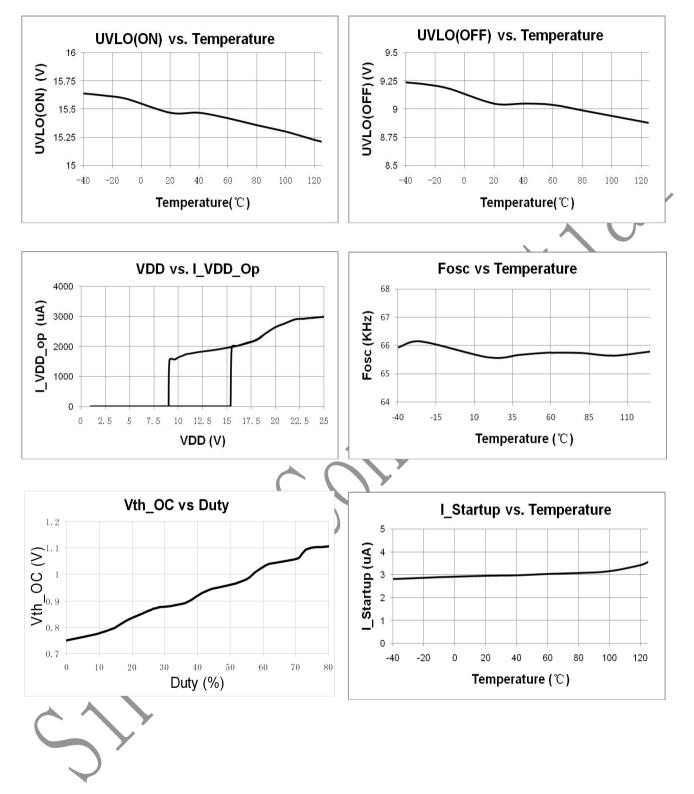
**Note 5.** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Note 6. The device is not guaranteed to function outside its operating conditions.

- Note 7. Guaranteed by design.
- Note 8. These parameters, although guaranteed, are not 100% tested in production



## **CHARACTERIZATION PLOTS**



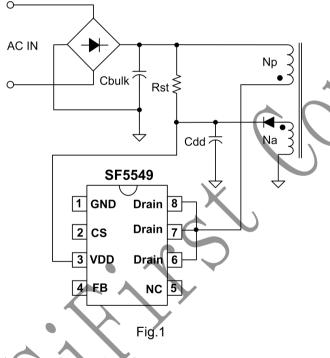


# **OPERATION DESCRIPTION**

SF5549 is a high performance, high efficiency, highly integrated current mode PWM power switch for offline flyback converter applications. The builtin advanced energy saving with high level protection features improves the SMPS reliability and performance without increasing the system cost.

#### • UVLO and Startup Operation

Fig.1 shows a typical startup circuit. Before the IC begins switching operation, it consumes only startup current (typically 3uA) and current supplied through the startup resistor Rst charges the VDD hold-up capacitor Cdd. When VDD reaches UVLO turn-on voltage of 15.5V(typical), SF5549 begins switching and the IC current consumed increased to 2mA (typical). The hold-up capacitor Cdd continues to supply VDD before the energy can be delivered from auxiliary winding Na. During this process, VDD must not drop below UVLO turn-off voltage (typical 9V). The selection of Rst and Cdd should be a trade off between the power loss and startup time.



#### Low Operating Current

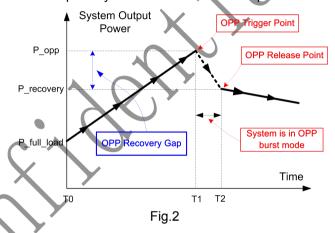
The operating current in SF5549 is as small as 2mA (typical). The small operating current results in higher efficiency and reduces the VDD hold-up capacitance requirement.

#### • Soft Start

SF5549 features an internal 4ms (typical) soft start that slowly increases the threshold of cycle-bycycle current limiting comparator during startup sequence. It helps to prevent transformer saturation and reduce the stress on the secondary diode during startup. Every restart attempt is followed by a soft start activation.

#### "Zero OCP/OPP Recovery Gap" Control

The definition of OCP or OPP recovery gap of a power adaptor is illustrated in Fig.2. At TO. assuming an adaptor is at full loading mode. If the loading keeps increasing, then the system will output maximum power P\_opp, which will trigger OPP protection at the same time. After the OPP protection is triggered, usually the system will enter into the auto-recovery mode, in burst manner. If the system power demand decreases below P recovery, then system will enter into normal mode again, as shown in Fig.2. The difference between P opp and P recovery is defined as "OPP Recovery Gap", which can cause system startup failure especially in 90VAC full load startup.



SF5549 can achieve "Zero OCP/OPP Recovery Gap" in the whole universal AC input range using SiFirst's proprietary control algorithm.

#### Oscillator with Frequency Shuffling

PWM switching frequency in SF5549 is fixed to 65KHz and is trimmed to tight range. To improve system EMI performance, SF5549 operates the system with  $\pm$ 4% frequency shuffling around setting frequency.

#### Synchronous Slope Compensation

InSF5549, the synchronous slope compensation circuit is integrated by adding voltage ramp onto the current sense input voltage for PWM generation. This greatly improves the close loop stability at CCM and prevents the sub-harmonic oscillation and thus reduces the output ripple voltage.

#### Leading Edge Blanking (LEB)

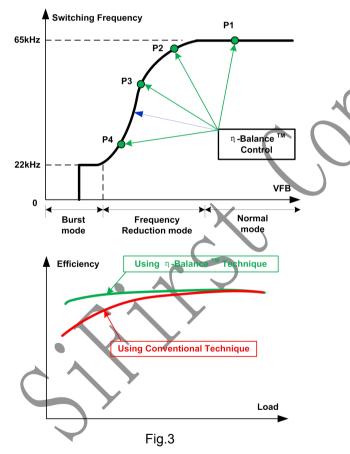
Each time the power MOSFET is switched on, a turn-on spike occurs across the sensing resistor. The spike is caused by primary side capacitance and secondary side rectifier reverse recovery. To avoid premature termination of the switching pulse, an internal leading edge blanking circuit is built in. During this blanking period (250ns, typical), the PWM comparator is disabled and cannot switch off

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the gate driver. Thus, external RC filter with a small time constant is enough for current sensing.

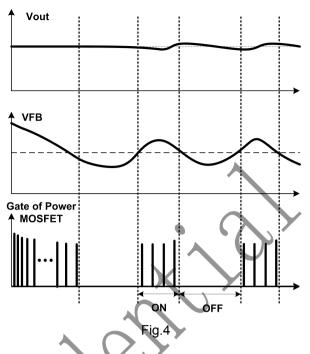
## ◆ Proprietary ŋ-Balance<sup>™</sup> Control

The efficiency requirement of power conversion is becoming tighter than before. These new energy standards focus on the average efficiency of the whole loading range. Therefore, the light load efficiency is becoming more and more important. In SF5549, a proprietary  $\eta$  -Balance<sup>TM</sup> control is integrated to boost the light load efficiency. As shown in Fig.3, when the loading becomes light, the IC will reduce the PWM switching frequency according to an optimized frequency reduction curve. The specific frequency reduction curve and the power at a frequency are determined by the output of  $\eta$  -Balance<sup>TM</sup> control. For example, P1 is at full load, P2 is at 75% full load, P3 and P4 are 50% and 25% full load respectively. The  $\eta$  -Balance<sup>™</sup> control can provide higher average efficiency than conventional frequency reduction technique, as illustrated in Fig.3



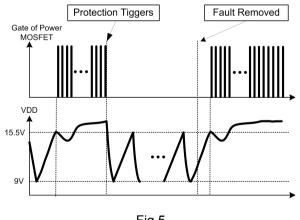
#### Burst Mode Control

When the loading is very small, the system enters into burst mode. When VFB drops below Vskip, SF5549 will stop switching and output voltage starts to drop, which causes the VFB to rise. Once VFB rises above Vskip, switching resumes. Burst mode control alternately enables and disables switching, thereby reducing switching loss in standby mode.



#### Auto Recovery Mode Protection

As shown in Fig.5, once a fault condition is detected, switching will stop. This will cause VDD to fall because no power is delivered form the auxiliary winding. When VDD falls to UVLO(off) (typical 9V), the protection is reset and the operating current reduces to the startup current, which causes VDD to rise, as shown in Fig.4. However, if the fault still exists, the system will experience the above mentioned process. If the fault has gone, the system resumes normal operation. In this manner, the auto restart can alternatively enable and disable the switching until the fault condition is disappeared.





#### VDD OVP(Over Voltage Protection)

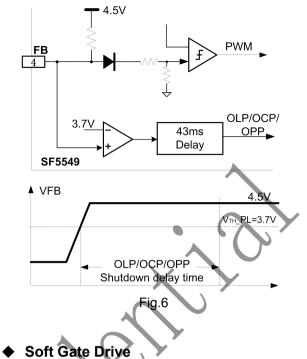
VDD OVP (Over Voltage Protection) is implemented in SF5549 and it is a protection of auto-recovery mode.

 Over Load Protection (OLP) / Over Current Protection (OCP) / Over Power



#### Protection (OPP) / Open Loop Protection (OLP)

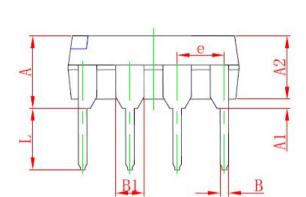
When OLP/OCP/OPP/Open Loop occurs, a fault is detected. If this fault is present for more than 42ms (typical), the protection will be triggered, the IC will experience an auto-recovery mode protection as mentioned above, as shown in Fig.6. The 42ms delay time is to prevent the false trigger from the power-on and turn-off transient.

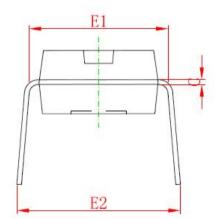


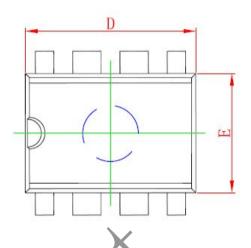
The driving stage of SF5549 is a soft totem-pole gate driver to minimize EMI. Cross conduction has been avoided to minimize heat dissipation, increase efficiency, and enhance reliability.

# PACKAGE MECHANICAL DATA

# DIP8 PACKAGE OUTLINE DIMENSIONS







	Cumb al	Dimensions	In Millimeters	Dimensions In Inches		
	Symbol	Min	Max	Min	Max	
	A	3.710	5.334	0.146	0.210	
	A1	0.381		0.015		
	A2	3.175	3.600	0.125	0.142	
	В	0.350	0.650	0.014	0.026	
	B1	1.524	(BSC)	0.06 (	BSC)	
$\rightarrow$	C	0.200	0.360	0.008	0.014	
	D	9.000	10.160	0.354	0.400	
	E	6.200	6.600	0.244	0.260	
	<b>E</b> 1	7.320	7.920	0.288	0.312	
	е	2.540	(BSC)	0.1 (E	BSC)	
	L	2.921	3.810	0.115	0.150	
	E2	8.200	9.525	0.323	0.375	



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