

Single-chip Type with Built-in FET Switching Regulators

Simple Step-down Switching Regulator with Built-in Power MOSFET



BD9329AEFJ No.11027EAT56

Description

The BD9329AEFJ is a synchronous step-down switching regulator that integrates 2 low resistance N-channel MOSFETs. It achieves 3A continuous output current over a wide input supply range. Current mode operation provides fast transient response and easy phase compensation.

Features

- 1) Wide operating INPUT Range 4.2V~18.0V
- 2) 3A Output Current
- 3) Hi-side / Lo-side FET ON-resistance; $0.15 / 0.13\Omega$ Power Switch
- 4) Low ESR Output Ceramic Capacitors are Available
- 5) Low Standby Current during Shutdown Mode
- 6) 380 kHz Fixed Operating Frequency
- 7) Feedback voltage 0.9V ±1.5% Accuracy at room temp. (±2.0% guaranteed for -25°C to 85°C temperature range)
- 8) Protection Circuits

Under Voltage Lockout Protection

Thermal Shutdown

Over Current Protection

9) HTSOP-J8 Package with Exposed thermal PAD.

Applications

Distributed Power System Pre-Regulator for Linear Regulator

● Absolute maximum ratings (Ta = 25°C)

Parameter	Symbol	Ratings	Unit
Supply Voltage	VIN	20	V
Switch Voltage	Vsw	20	V
Power Dissipation for HTSOP-J8	Pd	3760 ^{*1}	mW
Package thermal resistance θ ja *2	θ ja	29.27	°C/W
Package thermal resistance θ jc *2	θjc	3.75	°C/W
Operating Temperature Range	Topr	-40~+85	°C
Storage Temperature Range	Tstg	-55~+150	°C
Junction Temperature	Tjmax	150	°C
BST Voltage	VBST	Vsw+7	V
EN Voltage	Ven	20	V
All other pins	Vотн	20	V

^{*1} Derating in done 30.08 mW/°C for operating above Ta≥25°C(Mount on 4-layer 70.0mm×10.0mm×1.6mm board)

^{*2} Mount on 4-layer 50mm x 30mm x 1.6mm application board

●Operation Range (Ta= -40~85°C)

Parameter	Symbol		Unit		
		Min	Тур	Max	Offic
Supply Voltage	VIN	4.2	12	18	V
SW Voltage	Vsw	-0.5	-	18	V
Output current	Isw3	-	-	3	А
Output voltage range	V _{RANGE}	0.9	-	VIN x 0.7	V

● Electrical characteristics (Unless otherwise specified VIN=12V Ta=25°C)

ectrical characteristics (Unless otherwise specified VIN=12V Ta=25°C)							
Parameter	Symbol		Limits	T.	Unit	Conditions	
1 diameter	Cynnoon	Min	Тур	Max	Offic	Conditions	
Error amplifier block							
FB input bias current	lfв	-	0.02	2	μΑ		
Feedback voltage1	VFB1	0.886	0.900	0.914	V	Voltage follower	
Feedback voltage2	VFB2	0.882	0.900	0.918	V	Ta=-25°C∼85°C	
SW block – SW							
Hi-side FET On-resistance	Ronh	-	0.15	-	Ω	Isw= -0.8A	
Lo-side FET On-resistance	RONL	-	0.13	-	Ω	Isw= 0.8A	
Hi/Lo-side FET Leak current	ILEAKN	-	0	10	μΑ	VIN= 18V, VSW = 0V / 18V	
Switch Current Limit	ILIMIT3	3.5	-	-	А		
Maximum duty cycle	MDUTY	-	90	-	%	VFB= 0V	
General							
Enable Sink current	len	90	180	270	μΑ	VEN= 12V	
Enable Threshold voltage	VEN	1.0	1.2	1.4	V		
Under Voltage Lockout threshold	Vuvlo	3.5	3.75	4.0	V	VIN rising	
Under Voltage Lockout Hysteresis	VHYS	-	0.3	-	V		
Soft Start Current	Iss	5	10	15	μA	Vss= 0 V	
Soft Start Time	Tss	-	22	-	ms	Css= 0.1 μF	
Operating Frequency	Fosc	300	380	460	kHz		
Circuit Current	Icc	-	1.2	3	mA	VFB= 1.5V, VEN= 12V	
Quiescent Current	IQUI	-	15	27	μA	VEN= 0V	

●Block Diagram

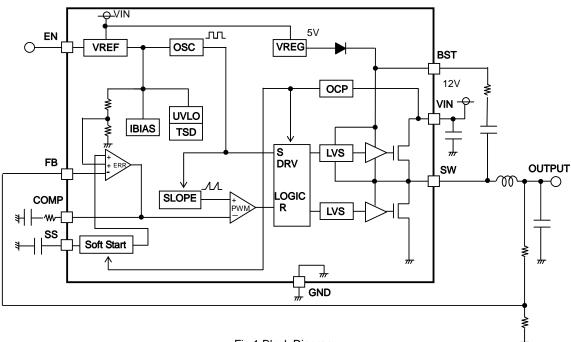
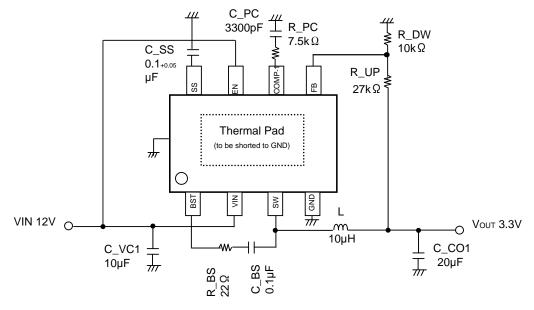


Fig.1 Block Diagram

● Typical Application Circuit



**R_BS protect from VIN-BST short destruction.

Fig.2 Application Circuit

	Symbol	Maker	Part No	
Input capacitor	C_VC1	TDK	C3225JB1E106K	10μF/25V
Output capacitor	C_CO1	TDK	C3216JB1C106M	10μF/16V
Inductor	L	TDK	SLF10165-100M3R8	10μH/3.8A

Block Operation

VREG

A block to generate constant-voltage for DC/DC boosting.

VREF

A block that generates internal reference voltage of 5.1 V (Typ.).

· TSD/UVLO

TSD (Thermal shutdown)/UVLO (Under Voltage Lockout) protection block.

The TSD circuit shuts down IC at high temperature.

The UVLO circuit shuts down the IC when the VCC is Low Voltage.

· Error amp block (ERR)

This is the circuit to compare the reference voltage and the feedback voltage of output voltage. The COMP pin voltage resulting from this comparison determines the switching duty. At the time of startup, since the soft start is operated by the SS pin voltage, the COMP pin voltage is limited to the SS pin voltage.

· Oscillator block (OSC)

This block generates the oscillating frequency.

SLOPE block

This block generates the triangular waveform from the clock created by OSC. Generated triangular waveform is sent to the PWM comparator.

PWM block

The COMP pin voltage output by the error amp is compared to the SLOPE block's triangular waveform to determine the switching duty. Since the switching duty is limited by the maximum duty ratio which is determined internally, it does not become 100%.

DRV block

A DC/DC driver block. A signal from the PWM is input to drive the power FETs.

· Soft start circuit

Since the output voltage rises gradually while restricting the current at the time of startup, it is possible to prevent the output voltage overshoot or the rush current.

Outward form

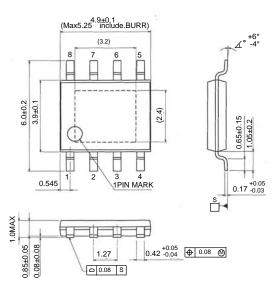


Fig.3 HTSOP-J8 Package (Unit:mm)

●Pin Assignment and Pin Function

Pin No.	Pin name	Function					
1	BST	High-Side Gate Drive Boost Input					
2	VIN	Power Input					
3	SW	Power Switching Output					
4	GND	Ground					
5	FB	Feed Back Input					
6	COMP	Compensation Node					
7	EN	Enable Input					
8	SS	Soft Start Control Input					

● Typical Performance Characteristics (Unless otherwise specified, VIN= 12V Ta = 25°C)

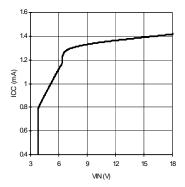


Fig.4 Circuit Current (No switching)

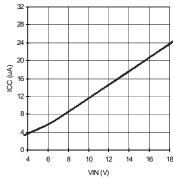


Fig.5 Stand by current (IC not active)

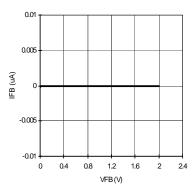


Fig.6 Input Bias Current

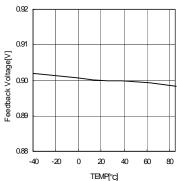


Fig.7 Feedback voltage

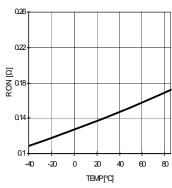


Fig.8 Hi,Low-Side On-resistance

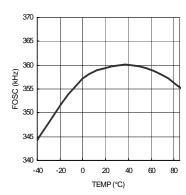


Fig.9 Operating Frequency

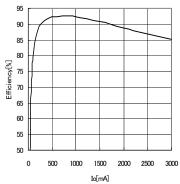


Fig.10 STEP Down Efficiency (VIN= 12V Vout= 3.3V L=10µH)

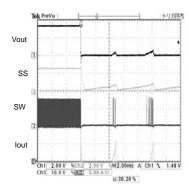


Fig.11 OverCurrent Protection (Vout is shorted to GND)

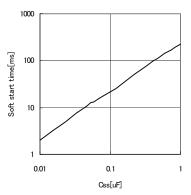


Fig.12 Soft Start Time

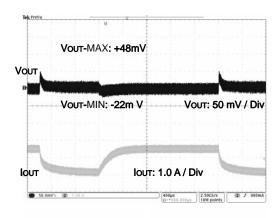


Fig.13 Transient Response (VIN= 12V Vout= 3.3V L= 10µH Cout = 22µF lout= 0.2-1.0A)

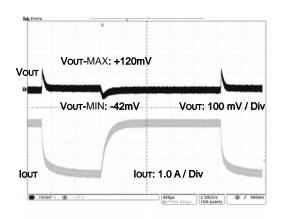


Fig.15 Transient Response (VIN= 12V Vout= 3.3V L= 10µH Cout =22µF lout= 0.2-3.0A)

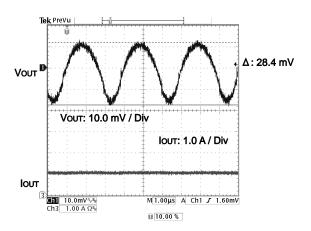


Fig.14 Output Ripple Voltage
(VIN= 12V Vout= 3.3V L= 10µH Cout = 22µF I out= 1.0A)

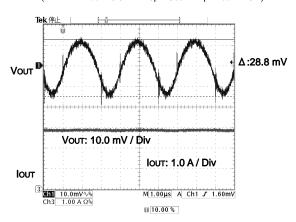


Fig.16 Output Ripple Voltage $(VIN=12V\ Vou\tau=3.3V\ L=10\mu H\ Cout=22\mu F\ I\ out=\ 3.0A\)$

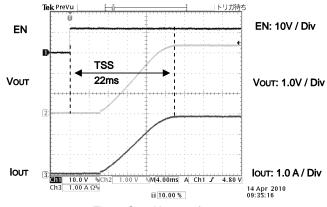
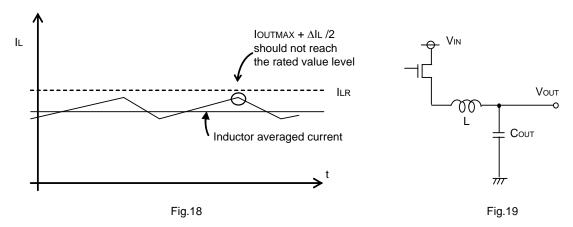


Fig.17 Start Up waveform (VIN= 12V Vout= 3.3V L= 10μ H Css= 0.1μ F)

Selecting Application Components

(1) Output LC filter constant selection (Buck Converter)

The Output LC filter is required to supply constant current to the output load. A larger value inductance at this filter results in less inductor ripple current ($\Delta I L$) and less output ripple voltage. However, the larger value inductors tend to have less fast load transient-response, a larger physical size, a lower saturation current and higher series resistance. A smaller value inductance has almost opposite characteristics above. So Choosing the Inductor ripple current ($\Delta I L$) between 20 to 40% of the averaged inductor current (equivalent to the output load current) is a good compromise.



Setting $\Delta I_L = 30\%$ x Averaged Inductor current (2A) = 0.6 [A]

L = Vout × (Vin - Vout) x
$$\frac{1}{\text{Vin x Fosc x } \Delta \text{lL}}$$
 = 10μ [H] Where Vin= 12V, Vout= 3.3V, Fosc= 380 kHz, ; Fosc is a switching frequency

Also the inductor should have the higher saturation current than $IOUTMAX + \Delta IL / 2$.

The output capacitor Cout affects the output ripple-voltage. Choose the large capacitor to achieve the small ripple-voltage enough to meet the application requirement.

Output ripple voltage $\Delta VRPL$ is calculated by the following equation.

$$\Delta VRPL = \Delta IL \times (RESR + \frac{1}{8x Cout x Fosc}) [V]$$

Where Resr is a parasitic series resistance in output capacitor.

Setting Cout = $20\mu F$, Resr = $10m \Omega$

$$\Delta VRPL = 0.6 x (10m + 1 / (8 x 20u x 380k)) = 15.8mV$$

(2) Loop Compensation

Choosing compensation capacitor CCMP and resistor RCMP

The current-mode buck converter has 2-poles and 1-zero system. Choosing the compensation resistor and capacitor is important for a good load-transient response and good stability.

The example of DC/DC converter application bode plot is shown below.

The compensation resistor RCMP will decides the cross over frequency FcRs (the frequency that the total DC-DC loop-gain falls to 0dB).

Setting the higher cross over frequency achieves good response speed, however less stability. While setting the lower cross over frequency shows good stability but worse response speed.

The 1/10 of switching frequency for the cross over frequency shows a good performance at most applications.

(i) Choosing phase compensation resistor RCMP

The compensation resistor RCMP can be on following formula.

RCMP =
$$\frac{2 \pi x \text{ Vout x Fcrs x Cout}}{\text{Vfb x GMp x GMA}} [\Omega]$$

Where

 $\textbf{Vout} \; ; \; \textbf{Output} \; \text{voltage}, \; \textbf{Fcrs} \; ; \; \textbf{Cross over frequency}, \; \textbf{Cout} \; ; \; \textbf{Output} \; \textbf{Capacitor},$

VFB; internal feedback voltage (0.9V(TYP)), GMP; Current Sense Gain (7.8AV(TYP)),

GMA; Error Amplifier Trans-conductance (300µA/V(TYP))

Setting Vout= 3.3V, Fcrs= 38kHz, Cout= 20µF;

RCMP =
$$\frac{2 \pi \times 3.3 \times 38k \times 20u}{0.9 \times 7.8 \times 300u} = 7.48k \approx 7.5k \quad [\Omega]$$

(ii) Choosing phase compensation capacitor CCMP

For the stability of DC/DC converter, canceling the phase delay that derives from output capacitor Cout and resistive load Rout by inserting the phase advance.

The phase advance can be added by the zero on compensation resistor Rcmp and capacitor Ccmp.

Making Fz= Fcrs / 6 gives a first-order estimate of Ccmp.

Compensation Capacitor
$$C_{CMP} = \frac{1}{2\pi \times R_{CMP} \times Fz}$$
 [F]

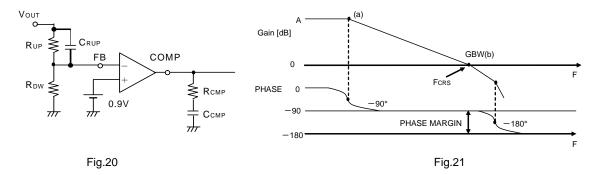
Setting Fz = Fcrs/6 = 6.3kHz;

Compensation Capacitor
$$C_{CMP} = \frac{1}{2\pi \times 7.5 k \times 6.3 k} = 3.54n \approx 3.54n \approx 3.3n$$
 [F]

(iii) The condition of the loop compensation stability

The stability of DC/DC converter is important. To secure the operating stability, please check the loop compensation has the enough phase-margin. For the condition of loop compensation stability, the phase-delay must be less than 150 degree where Gain is 0 dB.

Feed forward capacitor C_{RUP} boosts phase margin over a limited frequency range and is sometimes used to improve loop response. C_{RUP} will be more effective if $R_{UP} >> R_{UP} || R_{DW}$



(3) Design of Feedback Resistance constant Set the feedback resistance as shown below.

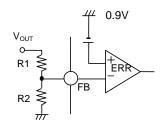


Fig.22

Soft Start Function

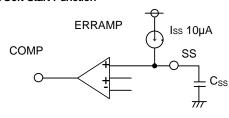


Fig.23

$VOUT = \frac{R1 + R2}{R2} \times 0.9 \quad [V]$

An adjustable soft-start function to prevent high inrush current during start-up is available.

The soft-start time is set by the external capacitor connected to SS pin.

The soft start time is given by:

Tss
$$[s] = 2.2 \times Css$$
 / Iss

Setting Css= 0.1μ F; Tss= $2.2 \times 0.1\mu / 10\mu = 22$ [ms]

Please confirm the overshoot of the output voltage and inrush

●EN Function

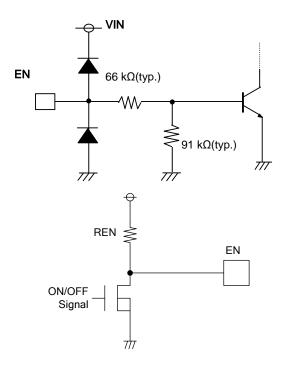


Fig.24

The EN terminal control IC's shut down.

Leaving EN terminal open makes IC shutdown.

current when deciding the SS capacitor value.

To start the IC, EN terminal should be connected to VIN or the other power source output.

When the EN voltage exceed 1.2V (typ.), the IC start operating.

(Attention)

Chattering happens if standing lowering speed is slow when standing of EN pin is lowered.

The reverse current in which the input side and the pressure operation are done from the output side is generated when chattering operates with the output voltage remained, and there is a case to destruction.

Please set to stand within 100us when you control ON/OFF by the EN signal.

This necessity doesn't exist when EN pin is connected with VIN and EN is not controlled.

The control by open drain MOSFET shown in a left chart is recommended.

● Layout Pattern Consideration

Two high pulsing current flowing loops exist in the buck regulator system. The first loop, when FET is ON, starts from the input capacitors, to the VIN terminal, to the SW terminal, to the inductor, to the output capacitors, and then returns to the input capacitor through GND. The second loop, when FET is OFF, starts from the low FET, to the inductor, to the output capacitor, and then returns to the low FET through GND. To reduce the noise and improve the efficiency, please minimize these two loop area. Especially input capacitor, output capacitor and low FET should be connected to GND plain. PCB Layout may affect the thermal performance, noise and efficiency greatly. So please take extra care when designing PCB Layout patterns.

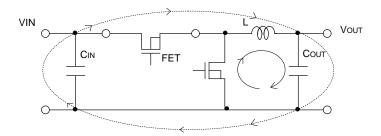


Fig.25 Current loop in Buck regulator system

- The thermal Pad on the back side of IC has the great thermal conduction to the chip. So using the GND plain as broad and wide as possible can help thermal dissipation. And a lot of thermal via for helping the spread of heat to the different layer is also effective.
- The input capacitors should be connected as close as possible to the VIN terminal.
- Keep sensitive signal traces such as trace connected FB and COMP away from SW pin.
- The inductor and the output capacitors should be placed close to SW pin as much as possible.

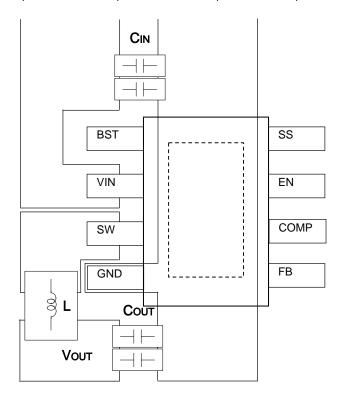
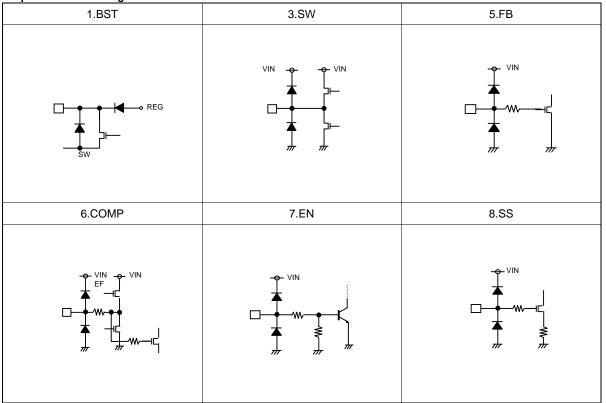
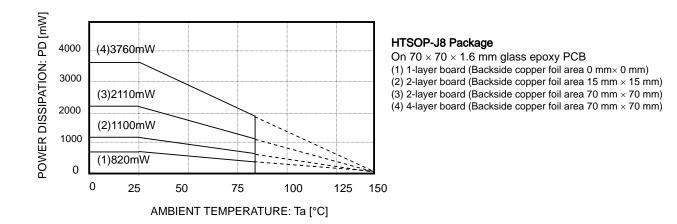


Fig.26 The example of PCB layout pattern

●I/O Equivalent Circuit Diagram



Power Dissipation



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Notes for use

1) Absolute maximum ratings

Use of the IC in excess of absolute maximum ratings such as the applied voltage or operating temperature range may result in IC damage. Assumptions should not be made regarding the state of the IC (short mode or open mode) when such damage is suffered. A physical safety measure such as a fuse should be implemented when use of the IC in a special mode where the absolute maximum ratings may be exceeded is anticipated.

2) GND potential

Ensure a minimum GND pin potential in all operating conditions.

3) Setting of heat

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) in actual operating conditions.

4) Pin short and mistake fitting

Use caution when orienting and positioning the IC for mounting on printed circuit boards. Improper mounting may result in damage to the IC. Shorts between output pins or between output pins and the power supply and GND pins caused by the presence of a foreign object may result in damage to the IC.

5) Actions in strong magnetic field

Use caution when using the IC in the presence of a strong magnetic field as doing so may cause the IC to malfunction.

6) Testing on application boards

When testing the IC on an application board, connecting a capacitor to a pin with low impedance subjects the IC to stress. Always discharge capacitors after each process or step. Ground the IC during assembly steps as an antistatic measure, and use similar caution when transporting or storing the IC. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process.

7) Ground wiring patterns

When using both small signal and large current GND patterns, it is recommended to isolate the two ground patterns, placing a single ground point at the application's reference point so that the pattern wiring resistance and voltage variations caused by large currents do not cause variations in the small signal ground voltage. Be careful not to change the GND wiring patterns of any external components.

8) Regarding input pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P/N junctions are formed at the intersection of these P layers with the N layers of other elements to create a variety of parasitic elements.

For example, when the resistors and transistors are connected to the pins as shown in Fig.27, a parasitic diode or a transistor operates by inverting the pin voltage and GND voltage.

The formation of parasitic elements as a result of the relationships of the potentials of different pins is an inevitable result of the IC's architecture. The operation of parasitic elements can cause interference with circuit operation as well as IC malfunction and damage. For these reasons, it is necessary to use caution so that the IC is not used in a way that will trigger the operation of parasitic elements such as by the application of voltages lower than the GND (P substrate) voltage to input and output pins.

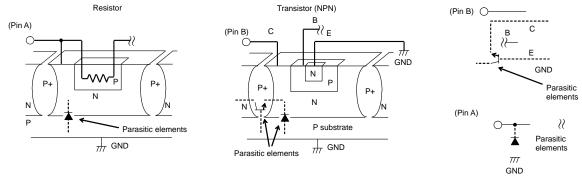


Fig.27 Example of a Simple Monolithic IC Architecture

9) Overcurrent protection circuits

An overcurrent protection circuit designed according to the output current is incorporated for the prevention of IC damage that may result in the event of load shorting. This protection circuit is effective in preventing damage due to sudden and unexpected accidents. However, the IC should not be used in applications characterized by the continuous operation or transitioning of the protection circuits. At the time of thermal designing, keep in mind that the current capacity has negative characteristics to temperatures.

10) Thermal shutdown circuit (TSD)

This IC incorporates a built-in TSD circuit for the protection from thermal destruction. The IC should be used within the specified power dissipation range. However, in the event that the IC continues to be operated in excess of its power dissipation limits, the attendant rise in the chip's junction temperature Tj will trigger the TSD circuit to turn off all output power elements. Operation of the TSD circuit presumes that the IC's absolute maximum ratings have been exceeded. Application designs should never make use of the TSD circuit.

11) Testing on application boards

At the time of inspection of the installation boards, when the capacitor is connected to the pin with low impedance, be sure to discharge electricity per process because it may load stresses to the IC. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process. Ground the IC during assembly steps as an antistatic measure, and use similar caution when transporting or storing the IC.

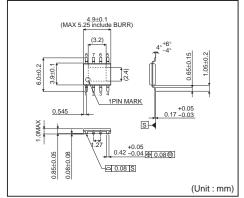
12) EN control speed

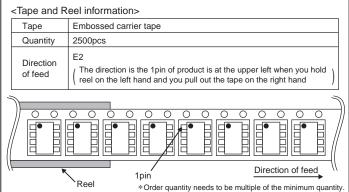
Chattering happens if standing lowering speed is slow when standing of EN pin is lowered. The reverse current in which the input side and the pressure operation are done from the output side is generated when chattering operates with the output voltage remained, and there is a case to destruction. Please set to stand within 100µs when you control ON/OFF by the EN signal.

Ordering part number



HTSOP-J8





Notice

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1. Our Products are designed and manufactured for application in ordinary electronic equipments (such as AV equipment, OA equipment, telecommunication equipment, home electronic appliances, amusement equipment, etc.). If you intend to use our Products in devices requiring extremely high reliability (such as medical equipment (Note 1), transport equipment, traffic equipment, aircraft/spacecraft, nuclear power controllers, fuel controllers, car equipment including car accessories, safety devices, etc.) and whose malfunction or failure may cause loss of human life, bodily injury or serious damage to property ("Specific Applications"), please consult with the ROHM sales representative in advance. Unless otherwise agreed in writing by ROHM in advance, ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of any ROHM's Products for Specific Applications.

(Note1) Medical Equipment Classification of the Specific Applications

JÁPAN	USA	EU	CHINA
CLASSⅢ	CLASSII	CLASS II b	СГУССШ
CLASSIV		CLASSIII	CLASSⅢ

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 - [a] Installation of protection circuits or other protective devices to improve system safety
 - [b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure
- 3. Our Products are designed and manufactured for use under standard conditions and not under any special or extraordinary environments or conditions, as exemplified below. Accordingly, ROHM shall not be in any way responsible or liable for any damages, expenses or losses arising from the use of any ROHM's Products under any special or extraordinary environments or conditions. If you intend to use our Products under any special or extraordinary environments or conditions (as exemplified below), your independent verification and confirmation of product performance, reliability, etc, prior to use, must be necessary:
 - [a] Use of our Products in any types of liquid, including water, oils, chemicals, and organic solvents
 - [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
 - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

Precautions Regarding Application Examples and External Circuits

- If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
- 2. You agree that application notes, reference designs, and associated data and information contained in this document are presented only as guidance for Products use. Therefore, in case you use such information, you are solely responsible for it and you must exercise your own independent verification and judgment in the use of such information contained in this document. ROHM shall not be in any way responsible or liable for any damages, expenses or losses incurred by you or third parties arising from the use of such information.

Precaution for Electrostatic

This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

Precaution for Storage / Transportation

- 1. Product performance and soldered connections may deteriorate if the Products are stored in the places where:
 - [a] the Products are exposed to sea winds or corrosive gases, including Cl2, H2S, NH3, SO2, and NO2
 - [b] the temperature or humidity exceeds those recommended by ROHM
 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

Precaution for Product Label

QR code printed on ROHM Products label is for ROHM's internal use only.

Precaution for Disposition

When disposing Products please dispose them properly using an authorized industry waste company.

Precaution for Foreign Exchange and Foreign Trade act

Since our Products might fall under controlled goods prescribed by the applicable foreign exchange and foreign trade act, please consult with ROHM representative in case of export.

Precaution Regarding Intellectual Property Rights

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General Precaution

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