LP3982 Micropower, Ultra-Low-Dropout, Low-Noise, 300-mA CMOS Regulator

1 Features

- 2.5-V to 6-V Input Range
- MAX8860 Pin, Package, and Specification Compatible
- 300-mA Output Current
- 120-mV Typical Dropout at 300 mA
- 90-μA Typical Quiescent Current
- 1-nA Typical Shutdown Mode
- 60-dB Typical PSRR
- 120-μs Typical Turnon Time
- Stable With Small Ceramic Output Capacitors
- 37-μV_{RMS} Output Voltage Noise (10 Hz to 100 kHz)
- Overtemperature/Overcurrent Protection
- ±2% Output Voltage Tolerance
- Create a Custom Design Using the LP3982 With the WEBENCH® Power Designer

2 Applications

- Wireless Handsets
- DSP Core Power
- Battery Powered Electronics
- Portable Information Appliances

3 Description

The LP3982 low-dropout (LDO) CMOS linear regulator is available in 1.8-V, 2.5-V, 2.82-V, 3-V, 3.3-V, and adjustable versions. They deliver 300 mA of output current. Packaged in an 8-pin VSSOP, the LP3982 is pin- and package-compatible with Maxim's MAX8860. The LM3982 is also available in the small footprint WSON package.

The LP3982 suits battery-powered applications because of its shutdown mode (1 nA typical), low quiescent current (90 μA typical), and LDO voltage (120 mV typical). The low dropout voltage allows for more utilization of a battery's available energy by operating closer to its end-of-life voltage. The LP3982 device's PMOS output transistor consumes relatively no drive current compared to PNP LDO regulators.

This PMOS regulator is stable with small ceramic capacitive loads (2.2 μF typical).

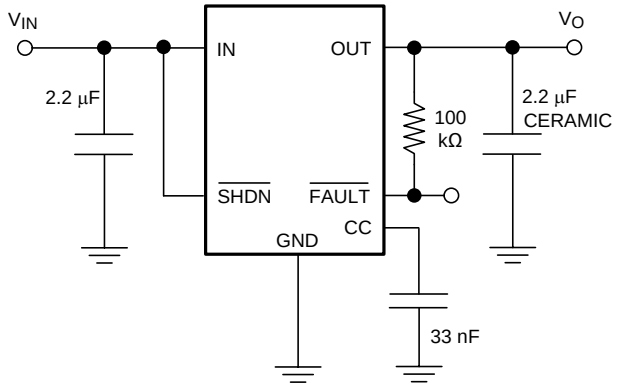
These devices also include regulation fault detection, a bandgap voltage reference, constant current limiting, and thermal-overload protection.

Device Information(1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
LP3982	WSON (8)	2.50 mm × 3.00 mm
	VSSOP (8)	3.00 mm × 3.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Application Circuit (Fixed V_{OUT} Version)



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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision E (October 2015) to Revision F — **Page**

- Added links for Webench and changed top navigator icon for TI Designs ... 1

Changes from Revision D (April 2013) to Revision E — **Page**

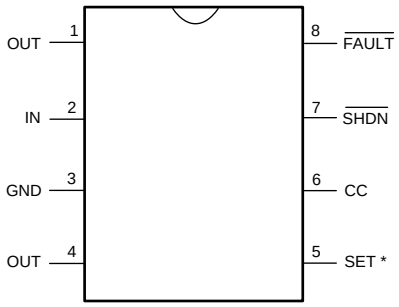
- Added *Device Information* and *Pin Configuration and Functions* sections, *ESD Ratings* table, *Feature Description*, *Device Functional Modes*, *Application and Implementation*, *Power Supply Recommendations*, *Layout*, *Device and Documentation Support*, and *Mechanical, Packaging, and Orderable Information* sections; update *Thermal Information* ... 1
- Deleted lead temperature from Abs Max table (in POA); revised wording for footnote 4 ... 4

Changes from Revision C (April 2013) to Revision D — **Page**

- Changed layout of National Data Sheet to TI format ... 9

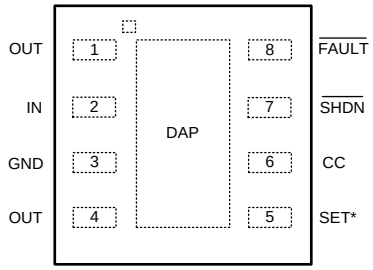
5 Pin Configuration and Functions

DGK Package
8-Pin VSSOP
Top View



The SET pin is internally disconnected for the fixed versions.

NGM Package
8-Pin WSON With Thermal Pad
Top View



The SET pin is internally disconnected for the fixed versions.

Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
CC	6	—	Connect a capacitor between CC pin and ground to reduce the output noise. The optimum value for CC is 33 nF.
$\overline{\text{FAULT}}$	8	Output	FAULT pin goes low during out of regulation conditions like current limit and thermal shutdown, or when it approaches dropout. Requires a pullup resistor because it is an active-low, open-drain output.
GND	3	Ground	Ground
IN	2	Input	This is the input supply voltage to the regulator.
OUT	1, 4	Output	Regulated output voltage
SET	5	Input	In the adjustable version a resistor divider connected to this pin sets the output voltage. The SET pin is internally disconnected for the fixed versions.
$\overline{\text{SHDN}}$	7	Input	The SHDN pin allows the part to be turned to an ON or OFF state by pulling SHDN pin high or low.
DAP	√	—	WSON Only - The DAP (Die Attached Pad) is an exposed pad that does not have an internal connection; it functions as a thermal relief when soldered to a copper plane. It is recommend that the DAP be connected to GND. See *WSON Mounting* section for more information.

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)(1)(2)(3)

	MIN	MAX	UNIT
V_{IN}, V_{OUT}, $V_{\overline{SHDN}}$, V_{SET}, V_{CC}, $V_{\overline{FAULT}}$	−0.3	6.5	V
Fault sink current		20	mA
Power dissipation	See(4)		
Junction temperature, T_J		150	°C
Storage temperature, T_{stg}	–65	160	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages are with respect to the potential at the GND pin.
(3) If Military/Aerospace-specified devices are required, contact Texas Instruments Sales Office/Distributors for availability and specifications.
(4) In applications where high power dissipation and/or poor thermal resistance is present, the maximum ambient temperature may have to be derated. Maximum ambient temperature ($T_{A(MAX)}$) is dependant on the maximum operating junction temperature ($T_{J(MAX-OP)}$), the maximum power dissipation ($P_{D(MAX)}$), and the junction-to-ambient thermal resistance in the application ($R_{\theta JA}$). This relationship is given by: $T_{A(MAX)} = T_{J(MAX-OP)} - (P_{D(MAX)} \times R_{\theta JA})$.The value of the $R_{\theta JA}$ for the WSON package is specifically dependent on the PCB trace area, trace material, and the number of layers and thermal vias. For improved thermal resistance and power dissipation for the WSON package, refer to TI Application Note AN-1187 *Leadless Leadframe Package (LLP)* (SNOA401).

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001(1)	±2000	V
		Machine model	±200	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions(1)(2)

	MIN	NOM	MAX	UNIT
Operating temperature	–40		85	°C
Supply voltage	2.5		6	V

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages are with respect to the potential at the GND pin.

6.4 Thermal Information

THERMAL METRIC(1)		LP3982		UNIT
		DGK (VSSOP)	NGM (WSON)(2)	
		8 PINS	8 PINS	
$R_{\theta JA}$(3)	Junction-to-ambient thermal resistance, High-K	175.2	52.6	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	66.0	66.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	95.6	16.7	°C/W
ψ_{JT}	Junction-to-top characterization parameter	9.7	1.9	°C/W
ψ_{JB}	Junction-to-board characterization parameter	94.2	16.7	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	n/a	11.1	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report.
(2) The PCB for the WSON/NGN package $R_{\theta JA}$ includes thermal vias under the exposed thermal pad per EIA/JEDEC JESD51-5.
(3) Thermal resistance value $R_{\theta JA}$ is based on the EIA/JEDEC High-K printed circuit board defined by: JESD51-7 - High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages.

6.5 Electrical Characteristics

Unless otherwise specified, all limits are specified for $V_{IN} = V_{OUT} + 0.5\ V^{(1)}$, $V_{\overline{SHDN}} = V_{IN}$, $C_{IN} = C_{OUT} = 2.2\ \mu F$, $C_{CC} = 33\ nF$, $T_J = 25°C$.

PARAMETER		TEST CONDITIONS	MIN(2)	TYP(3)	MAX(2)	UNIT
V_{IN}	Input voltage	For operating temperature extremes: −40°C to 85°C	2.5		6	V
ΔV_{OUT}	Output voltage tolerance	100 µA ≤ I_{OUT} ≤ 300 mA $V_{IN} = V_{OUT} + 0.5\ V^{(1)}$ SET = OUT for the ADJ Versions	−2		2	% of V_{OUT} (NOM)
		For operating temperature extremes: −40°C to 85°C	−3		3	
V_{OUT}	Output adjust range	ADJ version only; for operating temperature extremes: −40°C to 85°C	1.25		6	V
I_{OUT}	Maximum output current	Average DC current rating; For operating temperature extremes: −40°C and 85°C	300			mA
I_{LIMIT}	Output current limit			770		mA
		For operating temperature extremes: −40°C to 85°C	330			
I_Q	Supply current	I_{OUT} = 0 mA		90		µA
		I_{OUT} = 0 mA; for operating temperature extremes: −40°C to 85°C			270	
		I_{OUT} = 300 mA		225		
	Shutdown supply current	V_O = 0 V, $\overline{SHDN}$ = GND		0.001	1	µA
V_{DO}	Dropout voltage(1)(4)	I_{OUT} = 1 mA		0.4		mV
		I_{OUT} = 200 mA		80		
		I_{OUT} = 200 mA; for operating temperature extremes: −40°C to 85°C			220	
		I_{OUT} = 300 mA		120		
ΔV_{OUT}	Line regulation	I_{OUT} = 1 mA, $(V_{OUT} + 0.5\ V) \le V_I \le 6\ V^{(1)}$		0.01		%/V
		I_{OUT} = 1 mA, $(V_{OUT} + 0.5\ V) \le V_I \le 6\ V^{(1)}$; for operating temperature extremes: −40°C to 85°C	−0.1		0.1	
	Load regulation	100 µA ≤ I_{OUT} ≤ 300 mA		0.002		%/mA
e_n	Output voltage noise	I_{OUT} = 10 mA, 10 Hz ≤ f ≤ 100 kHz		37		μV_{RMS}
	Output voltage noise density	10 Hz ≤ f ≤ 100 kHz, C_{OUT} = 10 µF		190		$nV/\sqrt{Hz}$
$V_{\overline{SHDN}}$	$\overline{SHDN}$ input threshold	V_{IH}, $(V_{OUT} + 0.5\ V) \le V_{IN} \le 6\ V^{(1)}$; for operating temperature extremes: −40°C to 85°C	2			V
		V_{IL}, $(V_{OUT} + 0.5\ V) \le V_{IN} \le 6\ V^{(1)}$; for operating temperature extremes:−40°C to 85°C			0.4	
$I_{\overline{SHDN}}$	$\overline{SHDN}$ input bias current	$\overline{SHDN}$ = GND or IN		0.1	100	nA
I_{SET}	SET input leakage	SET = 1.3 V, ADJ version only(5)		0.1	2.5	nA

(1) Condition does not apply to input voltages below 2.5 V because this is the minimum input operating voltage.
(2) All limits are verified by testing or statistical analysis.
(3) Typical values represent the most likely parametric norm.
(4) Dropout voltage is measured by reducing V_{IN} until V_{OUT} drops 100 mV from its nominal value at $V_{IN} - V_{OUT} = 0.5$ V. Dropout voltage does not apply to the 1.8-V version.
(5) The SET pin is not externally connected for the fixed versions.

Electrical Characteristics (continued)

Unless otherwise specified, all limits are specified for $V_{IN} = V_{OUT} + 0.5\ V$[1], $V_{\overline{SHDN}} = V_{IN}$, $C_{IN} = C_{OUT} = 2.2\ \mu F$, $C_{CC} = 33\ nF$, $T_J = 25°C$.

PARAMETER		TEST CONDITIONS	MIN[2]	TYP[3]	MAX[2]	UNIT
$V_{\overline{FAULT}}$	$\overline{FAULT}$ detection voltage	$V_O \geq 2.5\ V$, $I_{OUT} = 200\ mA$[6]		120		mV
		$V_{OUT} \geq 2.5\ V$, $I_{OUT} = 200\ mA$[6]; for operating temperature extremes: −40°C to 85°C			280	
	$\overline{FAULT}$ output low voltage	$I_{SINK} = 2\ mA$		0.115		V
		$I_{SINK} = 2\ mA$; for operating temperature extremes: −40°C to 85°C			0.25	
$I_{\overline{FAULT}}$	$\overline{FAULT}$ off-leakage current	$\overline{FAULT} = 3.6\ V$, $\overline{SHDN} = 0\ V$		0.1	100	nA
T_{SD}	Thermal shutdown temperature			160		°C
	Thermal shutdown hysteresis			10		
T_{ON}	Start-up time	$C_{OUT} = 10\ \mu F$, V_{OUT} at 90% of final value		120		µs

(6) The $\overline{FAULT}$ detection voltage is specified for the input-to-output voltage differential at which the $\overline{FAULT}$ pin goes active low.

6.6 Typical Characteristics

Unless otherwise specified, $V_{IN} = V_O + 0.5$ V, $C_{IN} = C_{OUT} = 2.2$ µF, $C_{CC} = 33$ nF, $T_J = 25°C$, $V_{\overline{SHDN}} = V_{IN}$.

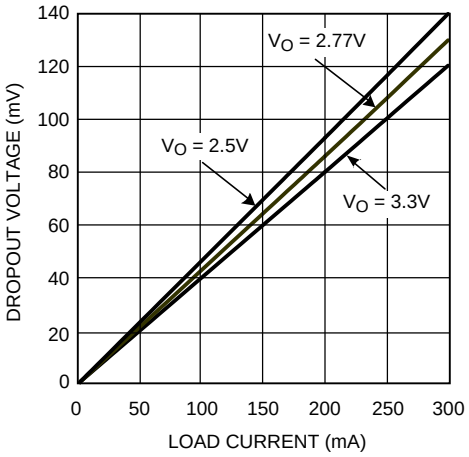


Figure 1. Dropout Voltage vs Load Current (for Different Output Voltages)

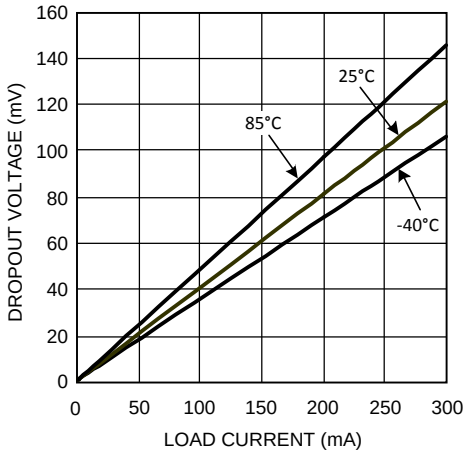


Figure 2. Dropout Voltage vs Load Current (for Different Output Temperatures)

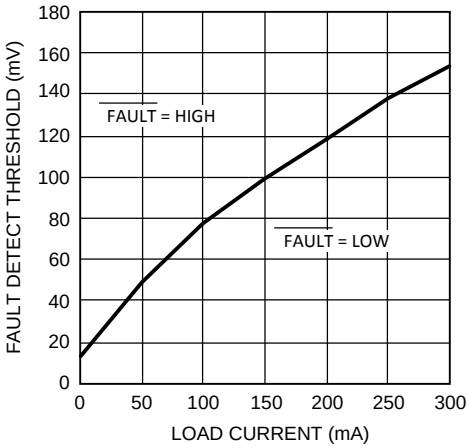


Figure 3. $\overline{FAULT}$ Detect Threshold vs Load Current

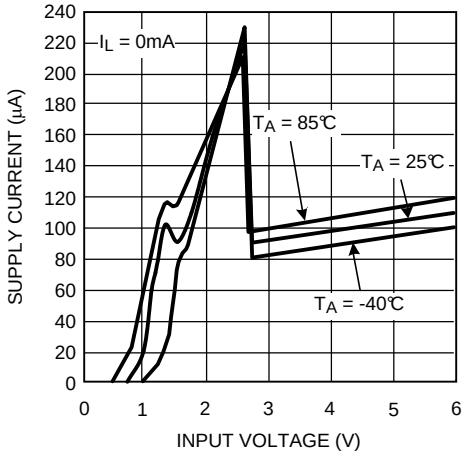


Figure 4. Supply Current vs Input Voltage

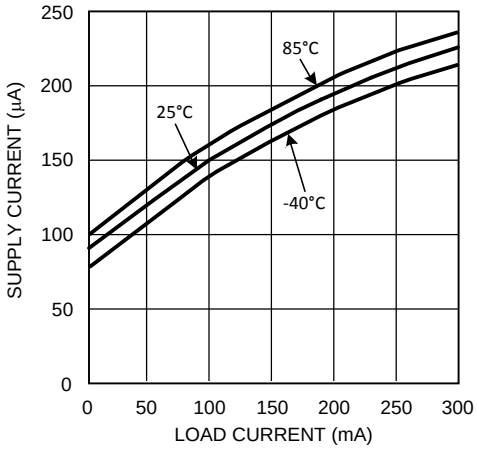


Figure 5. Supply Current vs Load Current

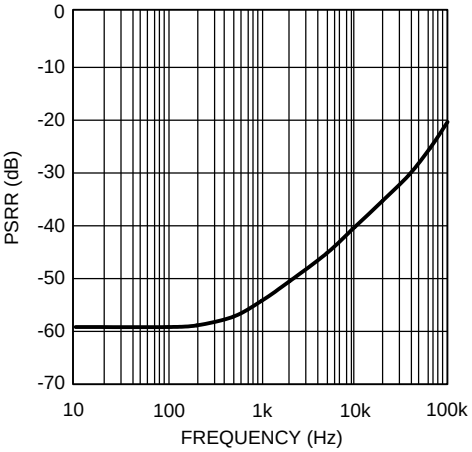


Figure 6. Power Supply Rejection Ratio vs Frequency

Typical Characteristics (continued)

Unless otherwise specified, $V_{IN} = V_O + 0.5$ V, $C_{IN} = C_{OUT} = 2.2$ µF, $C_{CC} = 33$ nF, $T_J = 25°C$, $V_{\overline{SHDN}} = V_{IN}$.

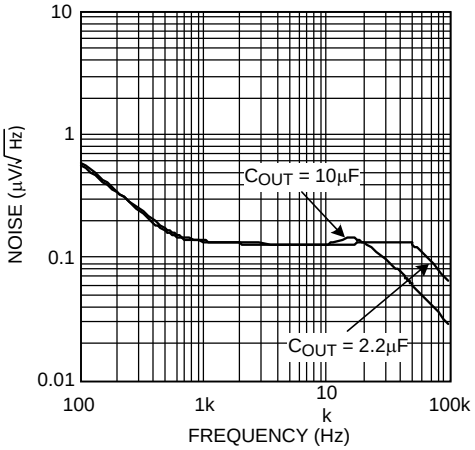


Figure 7. Output Noise Spectral Density

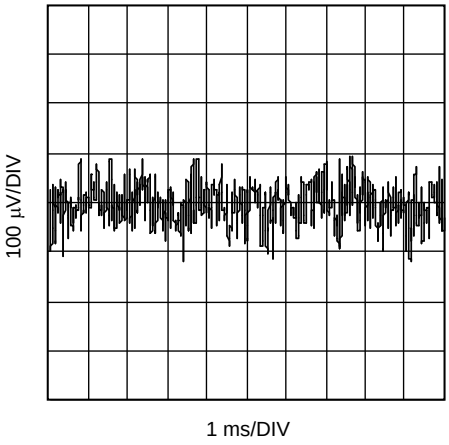


Figure 8. Output Noise (10 Hz to 100 kHz)

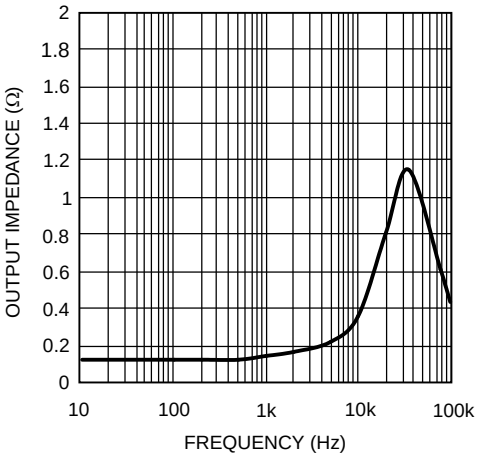


Figure 9. Output Impedance vs Frequency

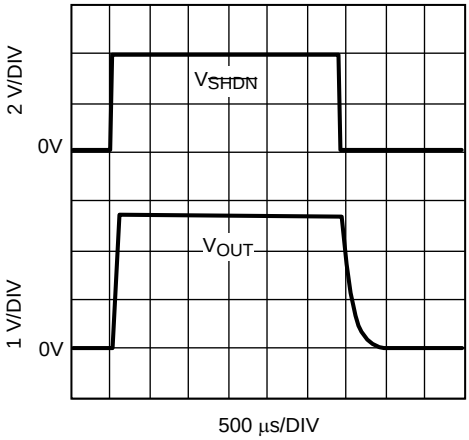


Figure 10. Shutdown Response

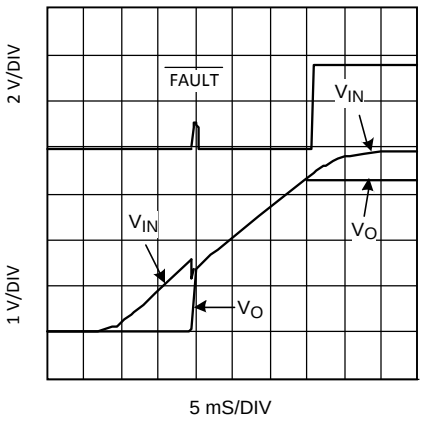


Figure 11. Power-Up Response

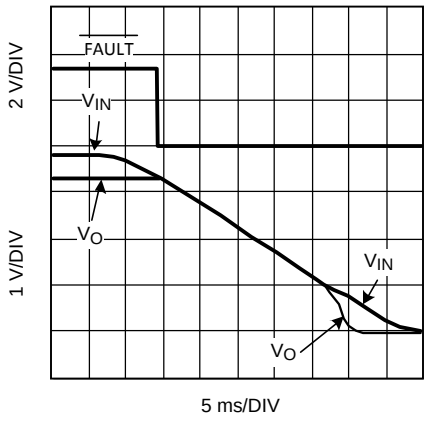


Figure 12. Power-Down Response

7 Detailed Description

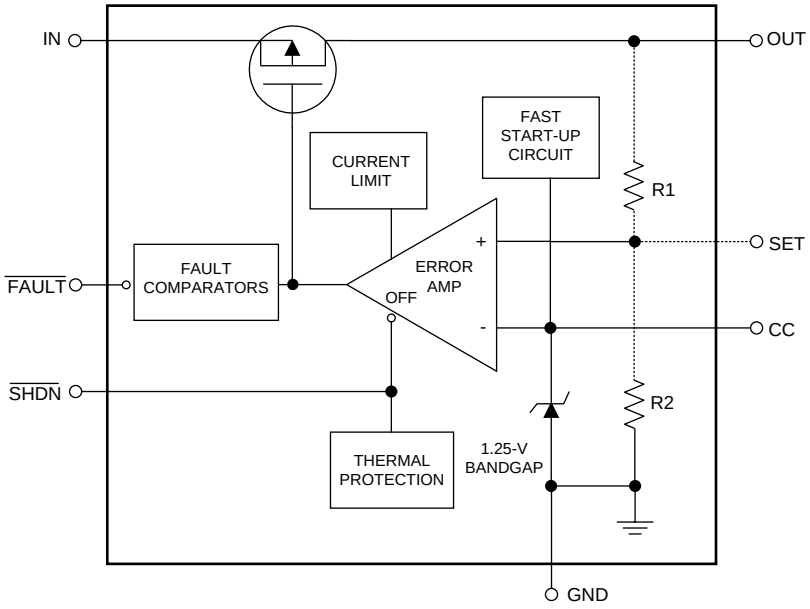
7.1 Overview

The LP3982 is package, pin, and performance compatible with Maxim's MAX8860, excluding reverse battery protection and dual-mode function (fixed and adjustable combined).

A 1.25-V bandgap reference, an error amplifier, and a PMOS pass transistor perform voltage regulation while being supported by shutdown, fault, and the usual temperature and current protection circuitry (see *Functional Block Diagram*).

The regulator topology is the classic type with negative feedback from the output to one of the inputs of the error amplifier. Feedback resistors R1 and R2 are either internal or external to the device, depending on whether it is the fixed-voltage version or the adjustable version. The negative feedback and high open loop gain of the error amplifier cause the two inputs of the error amplifier to be virtually equal in voltage. If the output voltage changes due to load changes, the error amplifier provides the appropriate drive to the pass transistor to maintain the error amplifier's inputs as virtually equal. In short, the error amplifier keeps the output voltage constant in order to keep its inputs equal.

7.2 Functional Block Diagram



7.3 Feature Description

7.3.1 No-Load Stability

The LP3982 remains stable during no-load conditions, a necessary feature for CMOS RAM keep-alive applications.

7.3.2 Fast Start-Up

The LP3982 provides fast start-up time for better system efficiency. The start-up speed is maintained when using the optional noise bypass capacitor. An internal 500-μA current source charges the capacitor until it reaches about 90% of its final value.

7.4 Device Functional Modes

7.4.1 Shutdown

The LP3982 goes into sleep mode when the $\overline{\text{SHDN}}$ pin is in a logic low condition. During this condition, the pass transistor, error amplifier, and bandgap are turned off, reducing the supply current to 1 nA typical. The maximum voltage for a logic low at the $\overline{\text{SHDN}}$ pin is 0.4 V. A minimum voltage of 2 V at the $\overline{\text{SHDN}}$ pin turns the LP3982 back on. The $\overline{\text{SHDN}}$ pin may be directly tied to V_{IN} to keep the part on. The $\overline{\text{SHDN}}$ pin may exceed V_{IN} but not the maximum of 6.5 V.

Figure 13 shows an application that uses the $\overline{\text{SHDN}}$ pin. It detects when the battery is too low and disconnects the load by turning off the regulator. A micropower comparator (LMC7215) and reference (LM385) are combined with resistors to set the minimum battery voltage. At the minimum battery voltage, the comparator output goes low and tuns off the LP3982 and corresponding load. Hysteresis is added to the minimum battery threshold to prevent the battery's recovery voltage from falsely indicating an above minimum condition. When the load is disconnected from the battery, it automatically increases in terminal voltage because of the reduced IR drop across its internal resistance. The minimum battery detector of Figure 13 has a low detection threshold (V_{LT}) of 3.6 V that corresponds to the minimum battery voltage. The upper threshold (V_{UT}) is set for 4.6 V to exceed the recovery voltage of the battery.

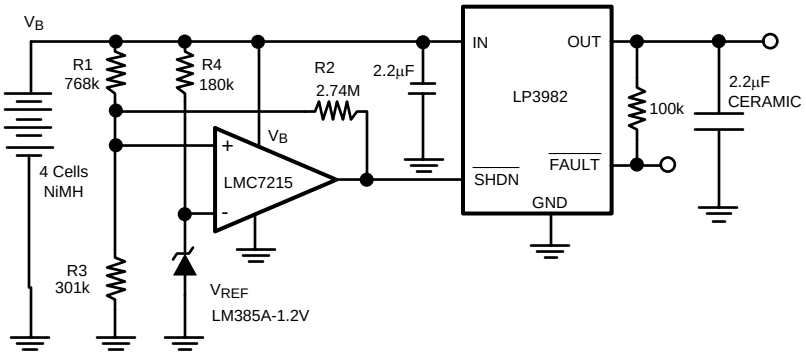


Figure 13. Minimum Battery Detector that Disconnects the Load Via the $\overline{\text{SHDN}}$ Pin of the LP3982

Resistor value for V_{UT} and V_{LT} are determined as follows:

$$G_T = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}$$

$$V_{UT} = R_1 (V_{REF})\, G_T$$

$$V_{LT} = R_1 // R_2 (V_{REF})\, G_T \quad (1)$$

(The application of Figure 13 used a G_T of 5 µ mho.)

$$R_1 = \frac{V_{UT1}}{V_{REF}(G_T)} \quad (2)$$

$$R_2 = \frac{1}{\frac{V_{REF}(G_T)}{V_{LT}} - \frac{1}{R_1}} \quad (3)$$

$$R_3 = \frac{1}{G_T - \left[\frac{1}{R_1} + \frac{1}{R_2}\right]} \quad (4)$$

The above procedure assumes a rail-to-rail output comparator. Essentially, R_2 is in parallel with R_1 prior to reaching the lower threshold, then R_2 becomes parallel with R_3 for the upper threshold. Note that the application requires rail-to-rail input as well.

The resistor values shown in Figure 13 are the closest practical to calculated values.

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LP3982 can provide 300-mA output current with 2.5-V to 6-V input. It is stable with a 2.2-µF ceramic output capacitor. An optional external bypass capacitor reduces the output noise without slowing down the load transient response. Typical output noise is 37 μV_{RMS} at frequencies from 10 Hz to 100 kHz. Typical PSSR is 60 dB at 1 kHz.

8.2 Typical Application

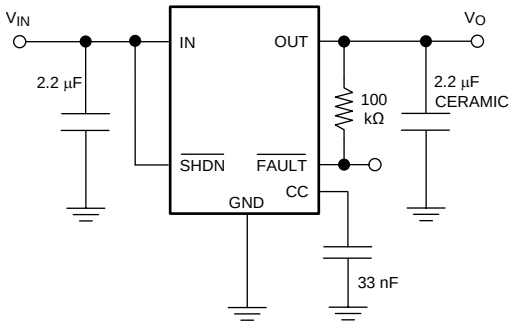


Figure 14. LP3982 Typical Application (Fixed V_{OUT} Version)

8.2.1 Design Requirements

For typical ultra low-dropout CMOS-regulator applications, use the parameters listed in Table 1.

Table 1. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Minimum input voltage	V_{OUT} + 0.5 V
Nominal output voltage	3.3 V
Maximum output current	300 mA
RMS noise, 10 Hz to 100 kHz	37 μV_{RMS}
PSRR at 1 kHz	60 dB

8.2.2 Detailed Design Procedure

8.2.2.1 Custom Design With WEBENCH® Tools

Click here to create a custom design using the LP3982 device with the WEBENCH® Power Designer.

1. Start by entering the input voltage (V_{IN}), output voltage (V_{OUT}), and output current (I_{OUT}) requirements.
2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance
- Run thermal simulations to understand board thermal performance
- Export customized schematic and layout into popular CAD formats
- Print PDF reports for the design, and share the design with colleagues

Get more information about WEBENCH tools at www.ti.com/WEBENCH.

8.2.2.2 Output Voltage Setting (ADJ Version Only)

The output voltage is set according to the amount of negative feedback (the pass transistor inverts the feedback signal.) Figure 15 simplifies the topology of the LP3982. This type of regulator can be represented as an op amp configured as non-inverting amplifier and a fixed DC Voltage (V_{REF}) for its input signal. The special characteristic of this op amp is its extra-large output transistor that only sources current. In terms of its non-inverting configuration, the output voltage equals V_{REF} times the closed loop gain:

$$V_O = V_{REF}\left[\frac{R_1}{R_2} + 1\right] \tag{5}$$

Use Equation 6 for adjusting the output to a particular voltage:

$$R_1 = R_2\left[\frac{V_O}{1.25V} - 1\right] \tag{6}$$

Choose R_2 = 100 kΩ to optimize accuracy, power supply rejection, noise, and power consumption.

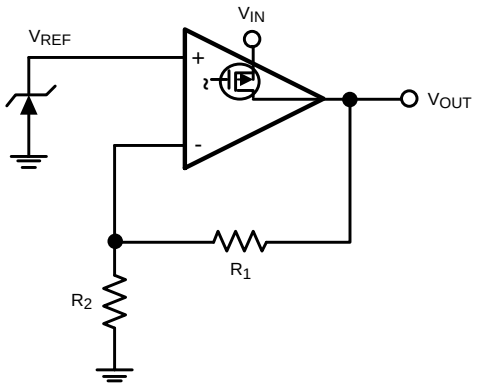


Figure 15. Regulator Topology Simplified

Similarity in the output capabilities exists between op amps and linear regulators. Just as rail-to-rail output op amps allow their output voltage to approach the supply voltage, low dropout regulators (LDOs) allow their output voltage to operate close to the input voltage. Both achieve this by the configuration of their output transistors. Standard operational amplifiers and regulator outputs are at the source (or emitter) of the output transistor. Rail-to-rail op amp and LDO regulator outputs are at the drain (or collector) of the output transistor. This replaces the threshold (or diode drop) limitations on the output with the less restrictive source-to-drain (or V_{SAT}) limitations. There is a trade-off; the output impedance become significantly higher, thus providing a critically lower pole when combined with the capacitive load. That is why rail-to-rail operational amplifiers are usually poor at driving capacitive loads and a series output resistor recommended when doing so. LDOs require the same series resistance except that the internal resistance of the output capacitor will usually suffice. Refer to the *Output Capacitance* section for more information.

8.2.2.3 Output Capacitance

The LP3982 is specifically designed to employ ceramic output capacitors as low as 2.2 µF. Ceramic capacitors below 10 µF offer significant cost and space savings, along with high frequency noise filtering. Higher values and other types and of capacitor may be used, but their equivalent series resistance (ESR) must be maintained below 0.5 Ω.

Ceramic capacitor of the value required by the LP3982 are available in the following dielectric types: Z5U, Y5V, X5R, and X7R. The Z5U and Y5V types exhibit a 50% or more drop in capacitance value as their temperature increases from 25°C, an important consideration. The X5R generally maintain their capacitance value within ±20%. The X7R type are desirable for their tighter tolerance of 10% over temperature.

Ceramic capacitors pose a challenge because of their relatively low ESR. Like most other LDOs, the LP3982 relies on a zero in the frequency response to compensate against excessive phase shift in the feedback loop of the regulator. If the phase shift reaches 360° (that is, becomes positive), the regulator oscillates. This compensation usually resides in the zero generated by the combination of the output capacitor with its ESR. The zero is intended to cancel the effects of the pole generated by the load capacitance (C_L) combined with the parallel combination of the load resistance (R_L) and the output resistance (R_O) of the regulator. The challenge posed by low ESR capacitors is that the zero it generates can be too high in frequency for the pole it is intended to compensate. The LP3982 overcomes this challenge by internally generating a strategically placed zero.

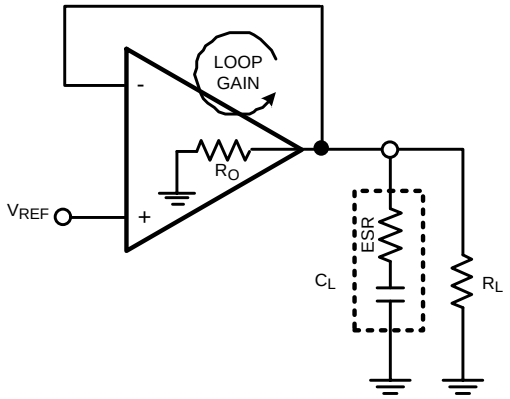


Figure 16. Simplified Model of Regulator Loop Gain Components

Figure 16 shows a basic model for the linear regulator that helps describe what happens to the output signal as it is processed through its feedback loop; that is, describe its loop gain (LG). The LG includes two main transfer functions: the error amplifier and the load. The error amplifier provides voltage gain and a dominant pole, while the load provides a zero and a pole. The LG of the model in Figure 16 is described by Equation 7:

$$LG(j\omega) = \frac{A_O}{1 + j\left[\frac{\omega}{\omega_{POLE}}\right]} * \frac{1 + j\omega (ESR \times C_L)}{1 + j\omega ((ESR + R_O // R_L) C_L)} \quad (7)$$

The first term of Equation 7 expresses the voltage gain (numerator) and a single pole role-off (denominator) of the error amplifier. The second term expresses the zero (numerator) and pole (denominator) of the load in combination with the R_O of the regulator.

Figure 17 shows a Bode plot that represents a case where the zero contributed by the load is too high to cancel the effect of the pole contributed by the load and R_O. The solid line represents the loop gain while the dashed line represents the corresponding phase shift. Notice that the phase shift at unity gain is a total 360°, the criteria for oscillation.

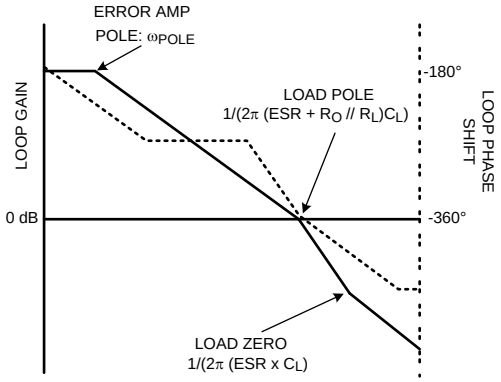


Figure 17. Loop Gain Bode Plot Illustrating Inadequately High Zero for Stability Compensation

The LP3982 generates an internal zero that makes up for the inadequately high zero of the low ESR ceramic output capacitor. This internally generated zero is strategically placed to provide positive phase shift near unity gain, thus providing a stable phase margin.

8.2.2.4 Input Capacitor

The LP3982 requires a minimum input capacitance of about 1 μF. The value may be increased indefinitely. The type is not critical to stability. However, instability may occur with bench set-ups where long supply leads are used, particularly at near dropout and high current conditions. This is attributed to the lead inductance coupling to the output through the gate oxide of the pass transistor; thus, forming a pseudo LCR network within the loop gain. A 10-μF tantalum input capacitor remedies this non-situ condition; its larger ESR acts to dampen the pseudo-LCR network. This may only be necessary for some bench setups. A 1-μF ceramic input capacitor are fine for most end-use applications.

If a tantalum input capacitor is intended for the final application, it is important to consider their tendency to fail in short circuit mode, thus potentially damaging the part.

8.2.2.5 Noise Bypass Capacitor

The noise bypass capacitor (CC) significantly reduces output noise of the LP3982. It connects between pin 6 and ground. The optimum value for CC is 33 nF.

Pin 6 directly connects to the high impedance output of the bandgap. The DC leakage of the CC capacitor must be considered; loading down the reference reduces the output voltage. NPO and COG ceramic capacitors typically offer very low leakage. Polypropylene and polycarbonate film carbonate capacitor offer even lower leakage currents.

CC does not affect the transient response; however, it does affect turnon time. The smaller the CC value, the faster the turnon time.

8.2.2.6 Fault Detection

The LP3982 provides a $\overline{\text{FAULT}}$ pin that goes low during out of regulation conditions like current limit and thermal shutdown, or when it approaches dropout. The latter monitors the input-to-output voltage differential and compares it against a threshold that is slightly above the dropout voltage. This threshold also tracks the dropout voltage as it varies with load current. Refer to Figure 3 in the *Typical Characteristics* section.

The $\overline{\text{FAULT}}$ pin requires a pullup resistor because it is an open-drain output. This resistor must be large in value to reduce energy drain. A 100-kΩ pullup resistor works well for most applications.

Figure 18 shows the LP3982 with delay added to the $\overline{\text{FAULT}}$ pin for the reset pin of a microprocessor. The output of the comparator stays low for a preset amount of time after the regulator comes out of a fault condition.

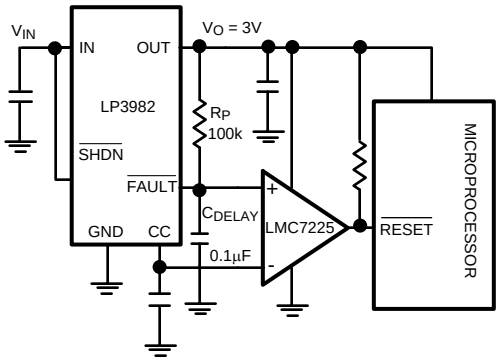


Figure 18. Power-On Delayed Reset Application

The delay time for the application of Figure 18 is set by Equation 8:

$$C_{DELAY} = \frac{-t}{R_P \ln\left[1 - \frac{V_{REF}}{V_O}\right]} \quad (8)$$

The application is set for a reset delay time of 8.8 ms. The comparator must have high impedance inputs so as to not load down the V_{REF} at the CC pin of the LP3982.

8.2.2.7 Power Dissipation

Knowing the device power dissipation and proper sizing of the thermal plane connected to the tab or pad is critical to ensuring reliable operation. Device power dissipation depends on input voltage, output voltage, and load conditions and can be calculated with Equation 9:

$$P_{D(MAX)} = (V_{IN(MAX)} - V_{OUT}) \times I_{OUT(MAX)} \quad (9)$$

Power dissipation can be minimized, and greater efficiency can be achieved, by using the lowest available voltage drop option that would still be greater than the dropout voltage (V_{DO}). However, keep in mind that higher voltage drops result in better dynamic (that is, PSRR and transient) performance. On the WSON (NGM) package, the primary conduction path for heat is through the exposed power pad to the PCB. To ensure the device does not overheat, connect the exposed pad, through thermal vias, to an internal ground plane with an appropriate amount of copper PCB area. On the VSSOP (DGK) package, the primary conduction path for heat is through the pins to the PCB. The maximum allowable junction temperature ($T_{J(MAX)}$) determines maximum power dissipation allowed ($P_{D(MAX)}$) for the device package. Power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance ($R_{\theta JA}$) of the combined PCB and device package and the temperature of the ambient air (T_A), according to Equation 10 or Equation 11:

$$(T_{J(MAX)} = T_{A(MAX)} + (R_{\theta JA} \times P_{D(MAX)}) \quad (10)$$

$$P_{D(MAX)} = (T_{J(MAX)} - T_{A(MAX)}) / R_{\theta JA} \quad (11)$$

$R_{\theta JA}$ is highly dependent on the heat-spreading capability of the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The $R_{\theta JA}$ recorded in *Thermal Information* is determined by the specific EIA/JEDEC JESD51-7 standard for PCB and copper-spreading area, and is to be used only as a relative measure of package thermal performance. For a well-designed thermal layout, $R_{\theta JA}$ is the sum of the package junction-to-case (bottom) thermal resistance ($R_{\theta JCbot}$) plus the thermal resistance contribution by the PCB copper area acting as a heat sink.

Improvements and absolute measurements of the $R_{\theta JA}$ can be estimated by utilizing the thermal shutdown circuitry that is internal to the device. The thermal shutdown turns off the pass transistor of the device when its junction temperature reaches 160°C (typical). The pass transistor does not turn on again until the junction temperature drops about 10°C (hysteresis).

Using the thermal shutdown circuit to estimate, $R_{\theta JA}$ can be as follows: with a low input-to-output voltage differential, set the load current to 300 mA. Increase the input voltage until the thermal shutdown begins to cycle on and off. Then slowly decrease V_{IN} (100-mV increments) until the device stays on. Record the resulting voltage differential (V_D) and use it in Equation 12:

$$R_{\theta JA} = \frac{(160 - T_A)}{(0.300 \times V_D)} \quad (12)$$

8.2.2.8 Estimating Junction Temperature

The EIA/JEDEC standard recommends the use of psi (Ψ) thermal characteristics to estimate the junction temperatures of surface mount devices on a typical PCB board application. These characteristics are not true thermal resistance values, but rather package specific thermal characteristics that offer practical and relative means of estimating junction temperatures. These psi metrics are determined to be significantly independent of copper-spreading area. The key thermal characteristics (Ψ_{JT} and Ψ_{JB}) are given in *Thermal Information* and are used in accordance with Equation 13 or Equation 14.

$$T_{J(MAX)} = T_{TOP} + (\Psi_{JT} \times P_{D(MAX)}) \quad (13)$$

where

- $P_{D(MAX)}$ is explained in Equation 9.
- T_{TOP} is the temperature measured at the center-top of the device package.

$$T_{J(MAX)} = T_{BOARD} + (\Psi_{JB} \times P_{D(MAX)}) \quad (14)$$

where

- $P_{D(MAX)}$ is explained in Equation 9.
- T_{BOARD} is the PCB surface temperature measured 1-mm from the device package and centered on the package edge.

For more information about the thermal characteristics Ψ_{JT} and Ψ_{JB}, see the TI Application Report *Semiconductor and IC Package Thermal Metrics* (SPRA953), available for download at www.ti.com.

For more information about measuring T_{TOP} and T_{BOARD}, see the TI Application Report *Using New Thermal Metrics* (SBVA025), available for download at www.ti.com.

For more information about the EIA/JEDEC JESD51 PCB used for validating $R_{\theta JA}$, see the TI Application Report *Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs* (SZZA017), available for download at www.ti.com.

8.2.3 Application Curves

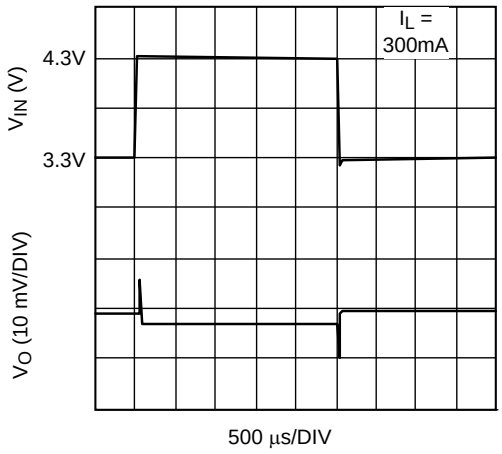


Figure 19. Line Transient Response

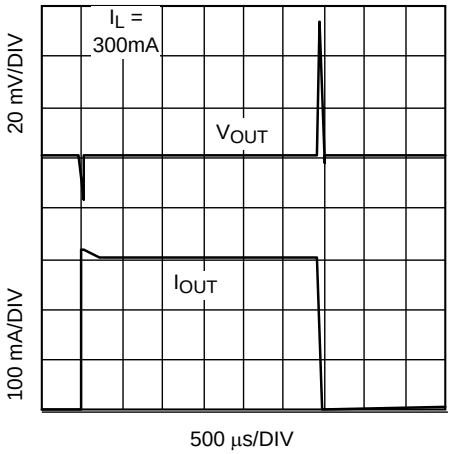


Figure 20. Load Transient

9 Power Supply Recommendations

The LP3982 is designed to operate from an input voltage supply range between 2.5 V and 6 V. The input voltage range provides adequate headroom in order for the device to have a regulated output. This input supply must be well regulated. If the input supply is noisy, additional input capacitors with low ESR can help to improve the output noise performance.

10 Layout

10.1 Layout Guidelines

Best performance is achieved by placing C_{IN}, C_{OUT}, and C_{CC} on the same side of the PCB as the LP3982 device, and as close as is practical to the package. The ground connections for C_{IN} and C_{OUT} must be back to the LP3982 device GND pin using as wide and as short of a copper trace as is practical.

Avoid connections using long trace lengths and narrow trace widths. These add parasitic inductances and resistance that results in inferior performance especially during transient conditions.

10.2 Layout Example

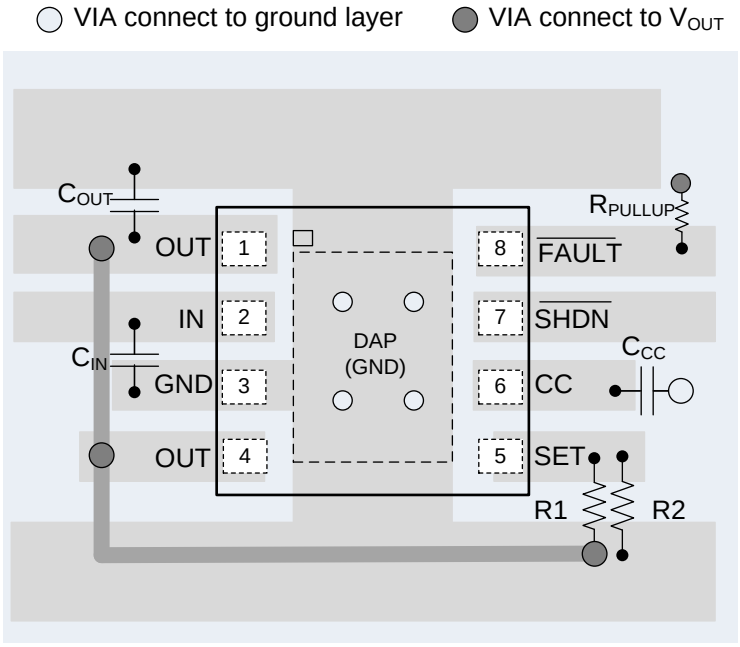


Figure 21. WSON Package Adjustable Version (Not to Scale)

10.3 WSON Mounting

The WSON package requires specific mounting techniques which are detailed in TI Application Report *Leadless Leadframe Package (LLP)* (SNOA401). Referring to the section *PCB Design Recommendations*, the pad style which must be used with the WSON package is the NSMD (non-solder mask defined) type. Additionally, it is recommended the PCB terminal pads be 0.2 mm longer than the package pads to create a solder fillet to improve reliability and inspection. The thermal dissipation of the WSON package is directly related to the printed circuit board construction and the amount of additional copper area connected to the DAP. The DAP (exposed pad) on the bottom of the WSON package is connected to the die substrate with a conductive die attach adhesive. The DAP has no direct electrical (wire) connection to any of the pins. There is a parasitic PN junction between the die substrate and the device ground. As such, it is strongly recommend that the DAP be connected directly to the ground at device pin 3 (GND). Alternately, but not recommended, the DAP may be left floating (no electrical connection). The DAP must not be connected to any potential other than ground.

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For additional information, see the following:

- *AN-1187 Leadless Leadframe Package (LLP)*
- *Semiconductor and IC Package Thermal Metrics*
- *Using New Thermal Metrics*
- *Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs*

11.1.2 Development Support

11.1.2.1 Custom Design With WEBENCH® Tools

Click here to create a custom design using the LP3982 device with the WEBENCH® Power Designer.

1. Start by entering the input voltage (V_{IN}), output voltage (V_{OUT}), and output current (I_{OUT}) requirements.
2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance
- Run thermal simulations to understand board thermal performance
- Export customized schematic and layout into popular CAD formats
- Print PDF reports for the design, and share the design with colleagues

Get more information about WEBENCH tools at www.ti.com/WEBENCH.

11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community ***TI's Engineer-to-Engineer (E2E) Community.*** Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support ***TI's Design Support*** Quickly find helpful E2E forums along with design support tools and contact information for technical support.

11.4 Trademarks

E2E is a trademark of Texas Instruments.
WEBENCH is a registered trademark of Texas Instruments.
All other trademarks are the property of their respective owners.

11.5 Electrostatic Discharge Caution

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.6 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LP3982ILD-1.8/NOPB	ACTIVE	WSON	NGM	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	LNB	Samples
LP3982ILD-2.5/NOPB	ACTIVE	WSON	NGM	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	LPB	Samples
LP3982ILD-3.0/NOPB	ACTIVE	WSON	NGM	8	1000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU SN	Level-3-260C-168 HR	-40 to 85	LTB	Samples
LP3982ILD-3.3/NOPB	ACTIVE	WSON	NGM	8	1000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU SN	Level-3-260C-168 HR	-40 to 85	LUB	Samples
LP3982ILD-ADJ/NOPB	ACTIVE	WSON	NGM	8	1000	Green (RoHS & no Sb/Br)	CU NIPDAU | CU SN	Level-3-260C-168 HR	-40 to 85	LVB	Samples
LP3982ILDX-1.8/NOPB	ACTIVE	WSON	NGM	8	4500	Green (RoHS & no Sb/Br)	CU SN	Level-3-260C-168 HR	-40 to 85	LNB	Samples
LP3982ILDX-3.3/NOPB	ACTIVE	WSON	NGM	8	4500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU SN	Level-3-260C-168 HR	-40 to 85	LUB	Samples
LP3982ILDX-ADJ/NOPB	ACTIVE	WSON	NGM	8	4500	Green (RoHS & no Sb/Br)	CU NIPDAU | CU SN	Level-3-260C-168 HR	-40 to 85	LVB	Samples
LP3982IMM-1.8	NRND	VSSOP	DGK	8	1000	TBD	Call TI	Call TI	-40 to 85	LENB	
LP3982IMM-1.8/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LENB	Samples
LP3982IMM-2.5/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LEPB	Samples
LP3982IMM-3.0	NRND	VSSOP	DGK	8	1000	TBD	Call TI	Call TI	-40 to 85	LETB	
LP3982IMM-3.0/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LETB	Samples
LP3982IMM-3.3	NRND	VSSOP	DGK	8	1000	TBD	Call TI	Call TI	-40 to 85	LEUB	
LP3982IMM-3.3/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LEUB	Samples
LP3982IMM-ADJ	ACTIVE	VSSOP	DGK	8	1000	TBD	Call TI	Call TI	-40 to 85	LEVB	Samples
LP3982IMM-ADJ/NOPB	ACTIVE	VSSOP	DGK	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LEVB	Samples
LP3982IMMX-1.8/NOPB	ACTIVE	VSSOP	DGK	8	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LENB	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LP3982IMMX-2.5/NOPB	ACTIVE	VSSOP	DGK	8	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LEPB	Samples
LP3982IMMX-2.82/NOPB	ACTIVE	VSSOP	DGK	8	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LESB	Samples
LP3982IMMX-ADJ/NOPB	ACTIVE	VSSOP	DGK	8	3500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	LEVB	Samples

(1) The marketing status values are defined as follows:
ACTIVE: Product device recommended for new designs.
LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.
NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.
PREVIEW: Device has been announced but is not in production. Samples may or may not be available.
OBSOLETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".
RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.
Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

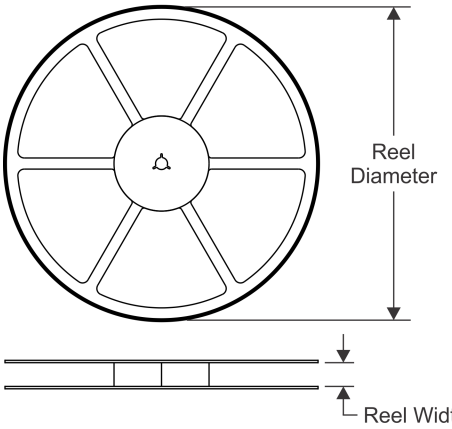
(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

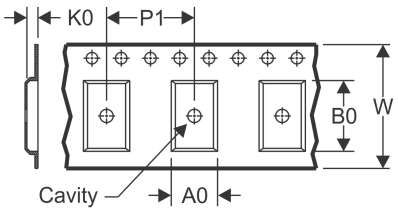
(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

TAPE AND REEL INFORMATION

REEL DIMENSIONS

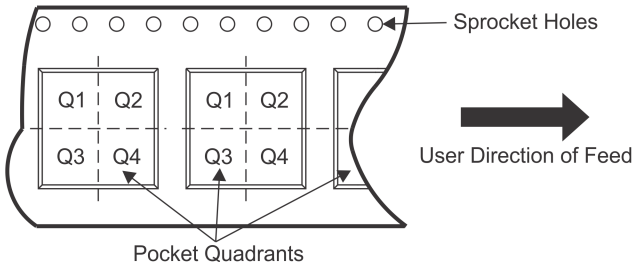


TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

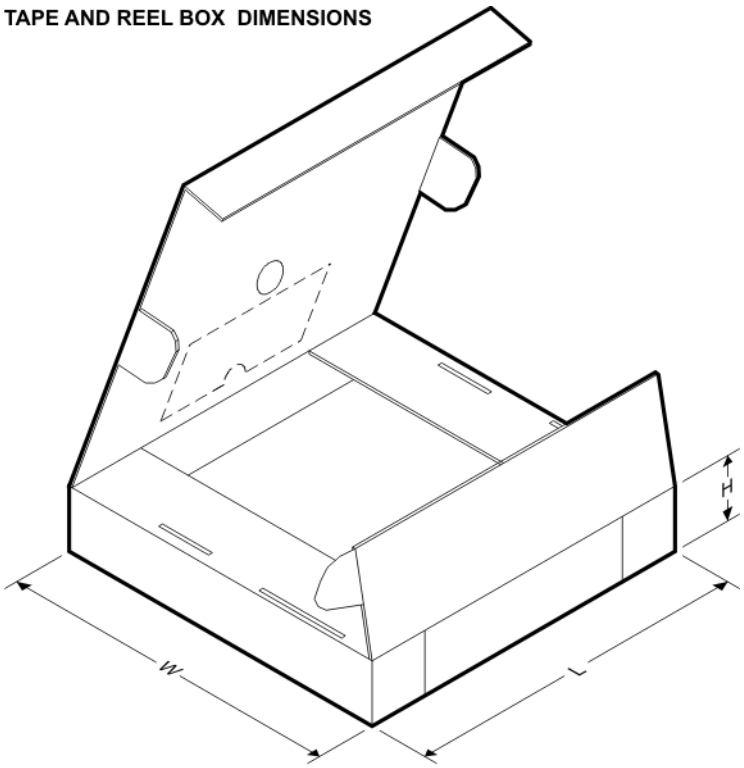


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP3982ILD-1.8/NOPB	WSON	NGM	8	1000	178.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILD-2.5/NOPB	WSON	NGM	8	1000	178.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILD-3.0/NOPB	WSON	NGM	8	1000	180.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILD-3.3/NOPB	WSON	NGM	8	1000	180.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILD-ADJ/NOPB	WSON	NGM	8	1000	180.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILDX-1.8/NOPB	WSON	NGM	8	4500	330.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILDX-3.3/NOPB	WSON	NGM	8	4500	330.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982ILDX-ADJ/NOPB	WSON	NGM	8	4500	330.0	12.4	3.3	2.8	1.0	8.0	12.0	Q1
LP3982IMM-1.8	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-1.8/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-2.5/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-3.0	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-3.0/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-3.3	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-3.3/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-ADJ	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMM-ADJ/NOPB	VSSOP	DGK	8	1000	178.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMMX-1.8/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LP3982IMMX-2.5/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMMX-2.82/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
LP3982IMMX-ADJ/NOPB	VSSOP	DGK	8	3500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

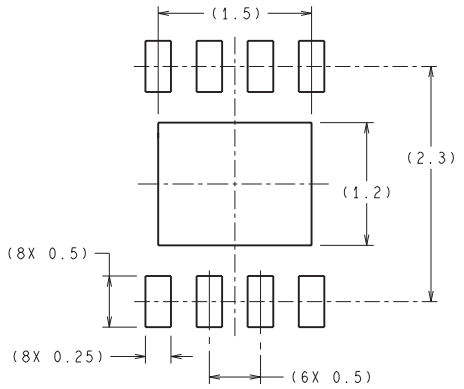


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP3982ILD-1.8/NOPB	WSON	NGM	8	1000	210.0	185.0	35.0
LP3982ILD-2.5/NOPB	WSON	NGM	8	1000	210.0	185.0	35.0
LP3982ILD-3.0/NOPB	WSON	NGM	8	1000	195.0	200.0	45.0
LP3982ILD-3.3/NOPB	WSON	NGM	8	1000	195.0	200.0	45.0
LP3982ILD-ADJ/NOPB	WSON	NGM	8	1000	195.0	200.0	45.0
LP3982ILDX-1.8/NOPB	WSON	NGM	8	4500	367.0	367.0	35.0
LP3982ILDX-3.3/NOPB	WSON	NGM	8	4500	370.0	355.0	55.0
LP3982ILDX-ADJ/NOPB	WSON	NGM	8	4500	370.0	355.0	55.0
LP3982IMM-1.8	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-1.8/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-2.5/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-3.0	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-3.0/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-3.3	VSSOP	DGK	8	1000	210.0	185.0	35.0

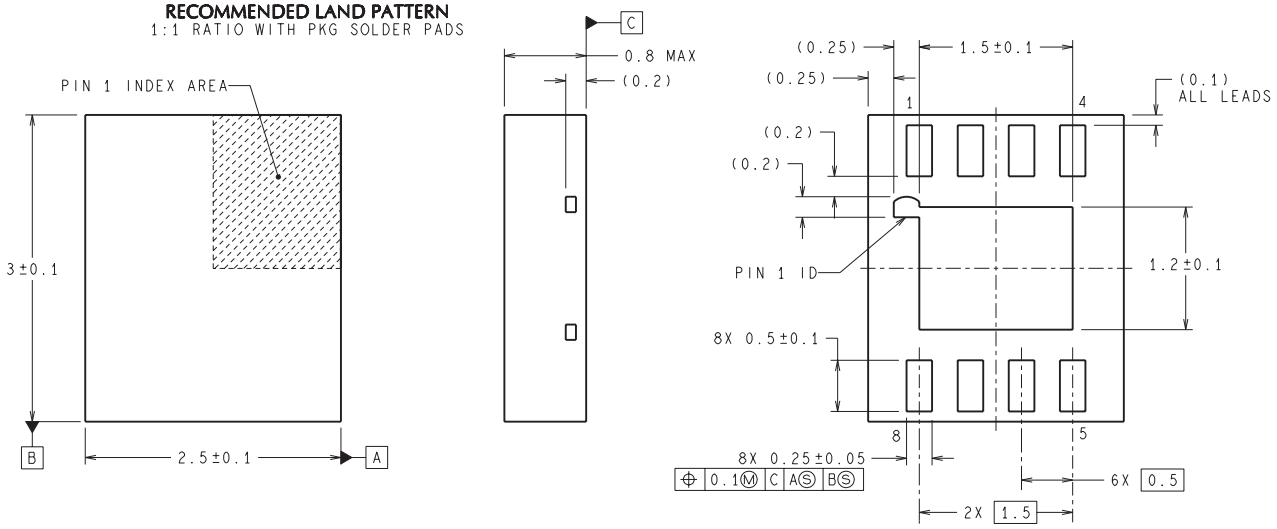
Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LP3982IMM-3.3/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-ADJ	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMM-ADJ/NOPB	VSSOP	DGK	8	1000	210.0	185.0	35.0
LP3982IMMX-1.8/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0
LP3982IMMX-2.5/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0
LP3982IMMX-2.82/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0
LP3982IMMX-ADJ/NOPB	VSSOP	DGK	8	3500	367.0	367.0	35.0

NGM0008C



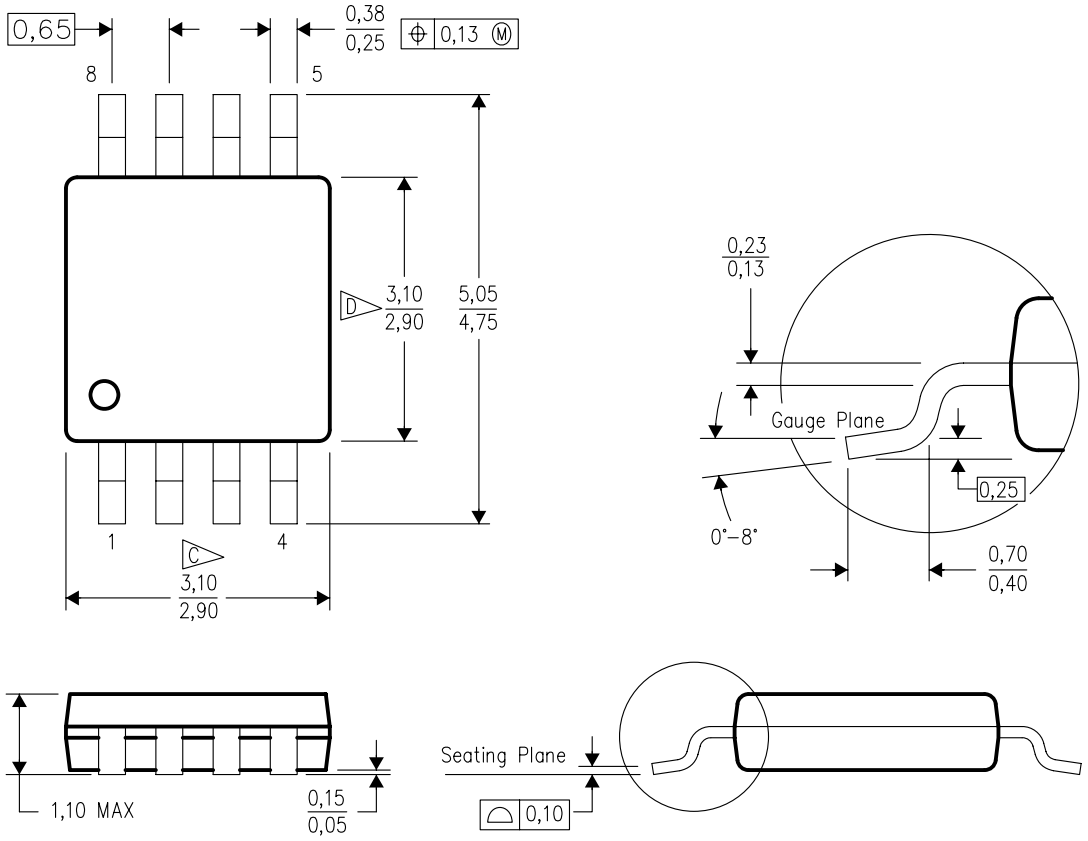
DIMENSIONS ARE IN MILLIMETERS
DIMENSIONS IN ( ) FOR REFERENCE ONLY

RECOMMENDED LAND PATTERN
1:1 RATIO WITH PKG SOLDER PADS



LDA08C (Rev B)

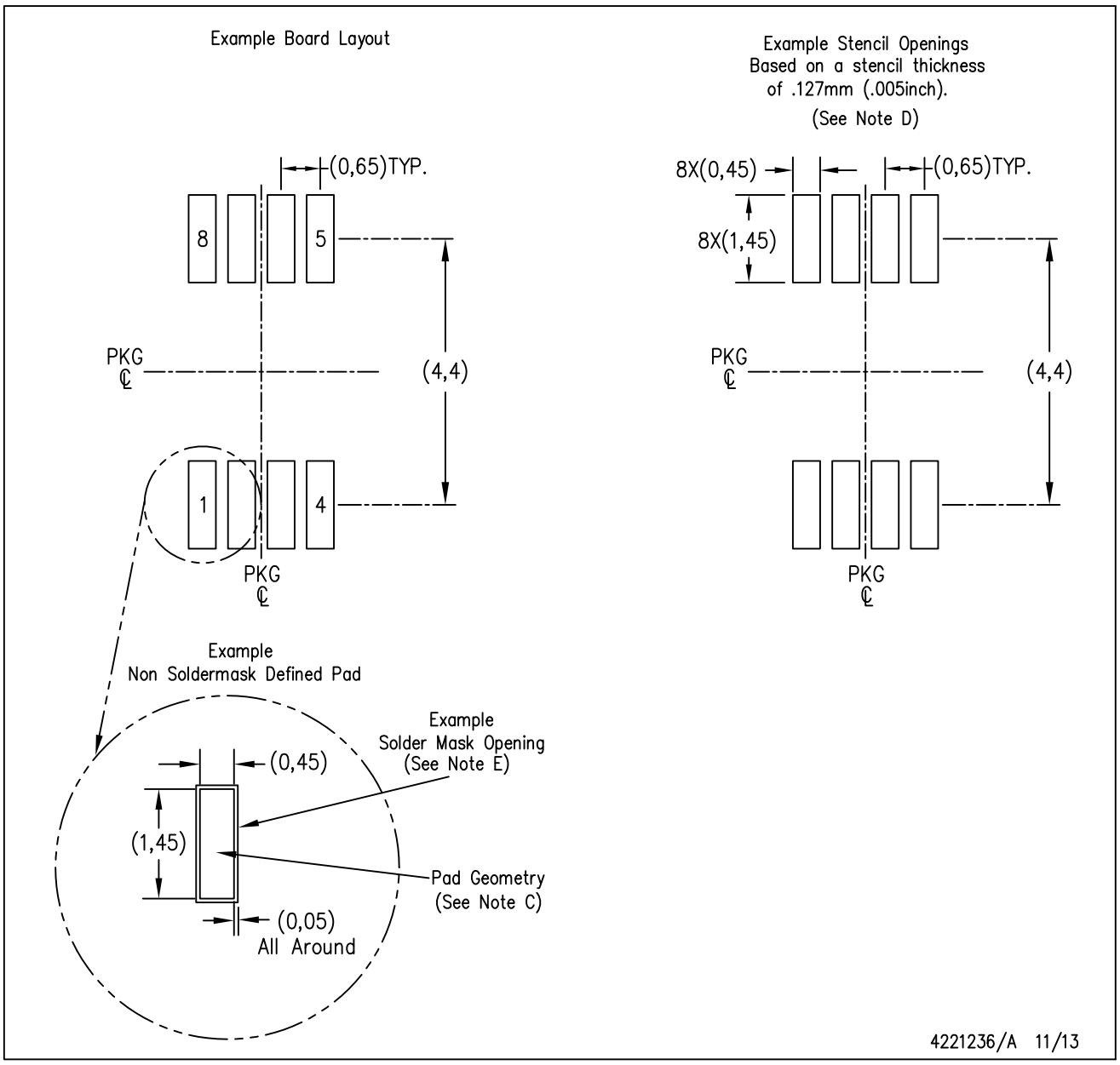
DGK (S–PDSO–G8) PLASTIC SMALL–OUTLINE PACKAGE



4073329/E 05/06

NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
D. Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
E. Falls within JEDEC MO–187 variation AA, except interlead flash.

DGK (S–PDSO–G8) PLASTIC SMALL OUTLINE PACKAGE



NOTES:
A. All linear dimensions are in millimeters.
B. This drawing is subject to change without notice.
C. Publication IPC–7351 is recommended for alternate designs.
D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC–7525 for other stencil recommendations.
E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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